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June 2013

FL7701 Smart LED Lamp Driver IC with PFC Function

Features

- Digitally Implemented Active PFC Function (No Additional Circuit Necessary for High PF)
- Built-in HV Supplying Circuit: Self Biasing
- Application Input Range: 80 V_{AC} ~ 308 V_{AC}
- AOCP Function with Auto-Restart Mode
- Built-in Over-Temperature Protection Function
- Cycle-by-Cycle Current Limit
- Current-Sense Pin Open Protection
- Low Operating Current: 0.85 mA (Typical)
- Under-Voltage Lockout with 5 V Hysteresis
- Programmable Oscillation Frequency
- Programmable LED Current
- Analog Dimming Function
- Soft-Start Function
- Precise Internal Reference: ±3%

Applications

- LED Lamp for Decorative Lighting
- LED Lamp for Low-Power Lighting

Ordering Information

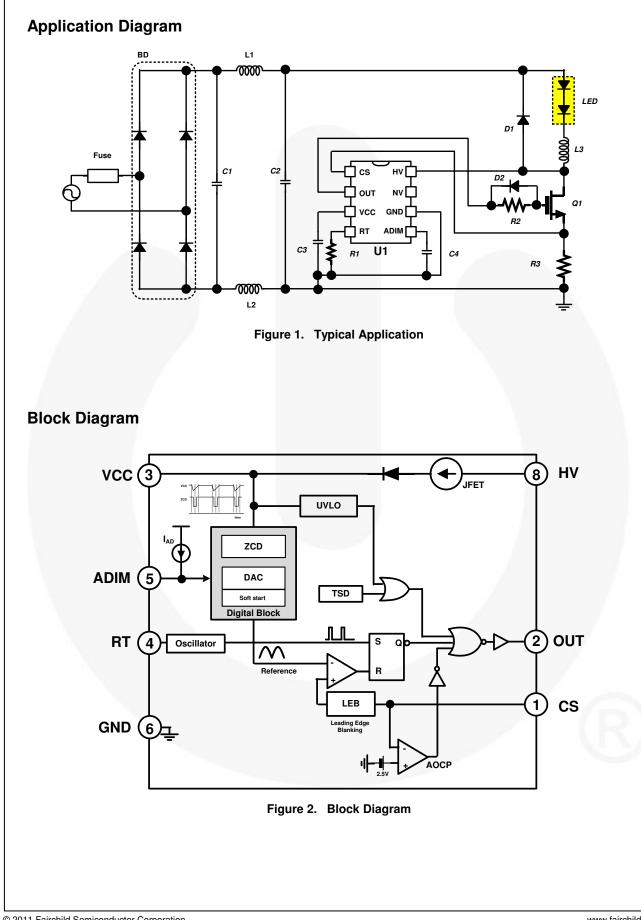
Description

The FL7701 LED lamp driver is a simple IC with PFC function. The special "adopted digital" technique of the IC can automatically detect input voltage condition and send an internal reference signal for achieving high power factor. When AC input is applied to the IC, PFC function is automatically enabled. Otherwise, when DC input is applied to the IC, PFC function is automatically disabled. The FL7701 does not need a bulk capacitor (electrolytic capacitor) for supply rail stability, which can significantly affect to LED lamp system.

Related Resources

Evaluation Board: FEBFL7701 L30U003A

Part Number	Operating Temperature Range	Package	Packing Method	
FL7701MX	-40°C to +125°C	8-Lead, Small Outline Integrated Circuit (SOIC), JEDEC MS-012, .150-inch Narrow Body	Tape & Reel	



Pin Configuration

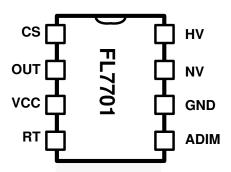


Figure 3. Pin Configuration

Pin Definitions

Pin #	Name	Description
1	CS	Current Sense . Limits output current depending on the sensing resistor voltage. The CS pin is also used to set the LED current regulation
2	OUT	OUT. Connects to the MOSFET gate.
3	VCC	VCC. Supply pin for stable IC operation ZCD signal detection used for accurate PFC function.
4	RT	RT . Programmable operating frequency using external resistor and the IC has pre-fixed frequency when this pin is open or floating.
5	ADIM	Analog Dimming . Connect to the internal current source and can change the output current using an external resistor. If ADIM is not used, it is recommended to connect a 0.1 μ F bypass capacitor between the ADIM and GND.
6	GND	GROUND. Ground for the IC.
7	NC	No Connection
8	HV	High Voltage. Connect to the high voltage line and supply current to the IC

Absolute Maximum Ratings

Stresses exceeding the absolute maximum ratings may damage the device. The device may not function or be operable above the recommended operating conditions and stressing the parts to these levels is not recommended. In addition, extended exposure to stresses above the recommended operating conditions may affect device reliability. The absolute maximum ratings are stress ratings only.

Symbol	Parameter		Min.	Max.	Unit
V _{CC}	IC Supply Voltage			20	V
HV	High Voltage Sensing			500	V
I _{O+} /I _{O-}	Peak Drive Output Current (Sourcing / Sinking)			250	mA
V _{ADIM}	Analog Dimming			5	V
V _{RT}	RT Pin Voltage			5	V
V _{CS}	Allowable Current Sensing Detection Voltage			5	V
T _A	Operating Ambient Temperature Range		-40	+125	°C
TJ	Operating Junction Temperature		-40	+150	°C
T _{STG}	Storage Temperature Range		-65	+150	°C
θ _{JA}	Thermal Resistance Junction-Air ^(1,2)			135	°C/W
PD	Power Dissipation			660	mW
ESD	Electrostatic Discharge Capability	Human Body Model, JESD22-A114		2000	v
		Charged Device Model, JESD22-C101		1000	

Notes:

1. Thermal resistance test board. Size: 76.2 mm x 114.3 mm x 1.6 mm (1S0P); JEDEC standard: JESD51-2, JESD51-3.

2. Assume no ambient airflow.

Electrical Characteristics

Typical values are at $T_A = +25^{\circ}$ C. Specifications to -40° C $\sim 125^{\circ}$ C are guaranteed by design based on final characterization results.

Symbol	Parameter	Conditions	Min.	Тур.	Max.	Unit
V _{cc} Bias Se	ection					
Vcc	VCC Regulator Output Voltage	V _{HV} =100 V _{DC}	14.0	15.5	17.0	V
V _{CCST+}	UVLO Positive Going Threshold	V _{CC} Increasing	12	13	14	V
V _{CCST-}	UVLO Negative Going Threshold	V _{CC} Decreasing	7	8	9	V
V _{CCHY}	UVLO Hysteresis		4	5	6	V
I _{HV}	HV Pin Current	V_{HV} =100 V_{DC} , C _L =150 pF, RT=Open		0.85	1.10	mA
I _{ST}	Startup Current			120	150	μA
Switching	Section					
	Operating Frequency	R _T =5.95 kΩ	200	250	300	kHz
fosc		R _T =87 kΩ	16	20	24	kHz
		R⊤ Open	40.5	45.0	49.5	kHz
t _{MIN}	Minimum On Time			400		ns
D _{MAX}	Maximum Duty Cycle			50		%
t _{LEB}	Leading Edge Blanking Time ⁽³⁾			350		ns
V _{RT}	Voltage Reference of RT pin			1.5		V
Soft-Start S	Section					
	Soft-Start Time ⁽³⁾	DC Mode	48	60	72	ms
t _{ss}		AC Mode		7		Periods
Reference	Section					
V _{CS1}	Internal Deference Maltana of CC Dia	DC Mode	0.354	0.365	0.376	v
V _{CS2}	Internal Reference Voltage of CS Pin	AC Mode ⁽³⁾	0.485	0.500	0.515	
Protection	Section				-	
OVPvcc	Over-Voltage Protection on VCC Pin		17.7	18.7	19.7	V
VAOCP	Abnormal OCP Level at CS Pin ⁽³⁾			2.5		V
t _{AOCP}	Abnormal Detection Time ⁽³⁾			70		ns
t _{TSDH}	Thermal Shutdown Threshold ⁽³⁾		140	150	1	°C
t _{TSDHY}	Thermal Shutdown Threshold Hysteresis ⁽³⁾			50	1	°C
Dimming S	ection			1		
V _{ADIM(ST+)}	Analog Dimming Positive Going Threshold ⁽³⁾		3.15	3.50	3.85	V
V _{ADIM(ST-)}	Analog Dimming Negative Going Threshold ⁽³⁾			0.50	0.75	V
I _{AD}	Internal Current Source for ADIM Pin		9	12	15	μA

Note:

3. These parameters, although guaranteed, are not 100% tested in production.

Functional Description

The FL7701 is a basic PWM controller for buck converter topology in Continuous Conduction Mode (CCM) with intelligent PFC function using a digital control algorithm. The FL7701 has an internal selfbiasing circuit that uses the high-voltage switching device. The IC does not need an auxiliary powering path to the VCC pin typical in flyback control ICs or PSR product family.

When the input voltage applied to the HV pin is over 25 V to 500 V, the FL7701 maintains a 15.5 V DC voltage at the VCC pin for stable operation. The FL7701 has UVLO block functions such that when the V_{CC} voltage rises higher than V_{CCST+} , the internal UVLO block releases and starts operation. Otherwise the VCC goes down to the V_{CCST-} and IC operation stops. Normally, the hysteresis function provides stable operation even if the input voltage is operating under the very noisy or unstable circumstances.

The FL7701 has a smart internal digital block for determining input condition: AC or DC. When an AC source with 50 Hz or 60 Hz is applied to the IC, the IC automatically changes its internal reference signal, which is similar to input signal, for creating high power factor. Otherwise, once the DC source connects to the IC, the internal reference immediately changes to DC.

1. Soft-Start Function

The FL7701 has an internal soft-start function to reduce inrush current at startup. When the IC starts operation following an internal sequence, the internal reference slowly increases for a pre-determined fixed time. After this transient period, the internal reference goes to a steady-state level. In this time, the IC continually tries to find phase information from the VCC pin. If the IC succeeds in getting phase information, it automatically follows a similar shape reference made during the transient times, 7 periods. If not, the IC has a DC reference level.

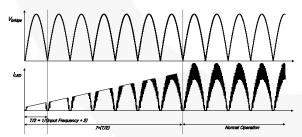


Figure 4. Soft Starting Function in AC Input Mode

2. Internal PFC Function: How to Achieve High Power Factor

The FL7701 has a simple, but smart, internal PFC function that does not require additional detection pins for detecting input phase information or an electrolytic capacitor for supply voltage stabilization. For achieving high PF, the FL7701 does not use the rectification capacitor after the bridge diode. This is important because the IC instead uses fluctuation in the signal on the VCC pin. Basically, the VCC pin, which is supplies

power for the IC, has voltage ripple as well as the rectification voltage after bridge, changing voltage level according to the VCC capacitor value. Using this kind of voltage fluctuation on the VCC pin, the IC can detect the time reference and create the internal ZCD signal.

For precise and reliable internal reference for input voltage signal, the FL7701 uses a digital technique (sigma/delta modulation) and creates a new internal signal (DAC_OUT) that has the same phase as the input voltage, as shown in Figure 5. This signal enters the final comparator and is compared with current information from the sensing resistor.

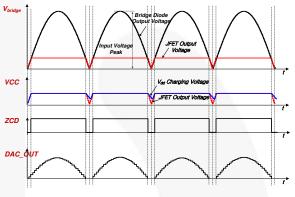


Figure 5. Internal PFC Function

3. Self-Biasing Function

The self-biasing function, using an HV device, can supply enough operating current to the IC and guarantee similar startup time across the whole input voltage range ($80 V \sim 308 V_{AC}$). However, self-biasing has a weakness in high-voltage condition. Normally, the HV device acts as constant current source, so the internal HV device has power loss when high input voltage connects to the HV pin. This power loss is proportional to input voltage. To reduce this power loss, one of the possible solutions is an additional resistor between the input voltage source and the HV pin, as shown in Figure 6.

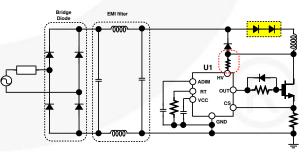


Figure 6. High-Voltage Application

4. Dimming Function

The FL7701 uses the ADIM pin for analog or 0 V to 10 V dimming by using a resistive divider. The peak voltage of internal reference, which is DAC_OUT signal in Figure 5, is changed by the V_{ADIM} level, as shown in Figure 7, and has different peak level according to the operating mode.

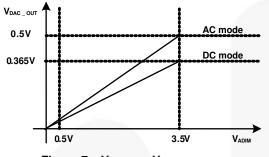


Figure 7. VADIM VS. VDAC_OUT(peak)

5. Inductor Design

The FL7701 prefixed internal duty ratio range is below 50%, or around 400 ns, from timing point of view. The range is dependent on the input voltage and LED numbers in its string.

Minimum duty is calculated as:

$$D_{\min} = \frac{n \cdot V_f}{\eta \cdot V_{in(\max)}} \tag{1}$$

where:

 η is efficiency of system; $V_{IN(max)}$ is maximum input voltage; V_f is forward drop voltage of LED; and n is LED number in series connection.

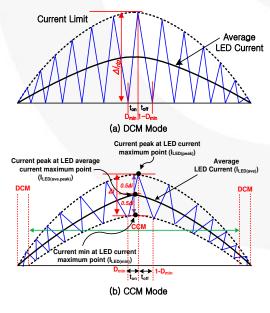


Figure 8. DCM and CCM Operation

In DCM Mode, inductance is:

$$L_m = \frac{n \cdot V_f \cdot (1 - D_{\min})}{f_s \cdot \Delta i_{rip}} [H]$$
⁽²⁾

If the peak current is fixed at 350 mApk, the formula for the peak current is:

$$I_{LED(ave.peak)} = \Delta i_{con} + \frac{\Delta i_{rip}}{2} \quad [A]$$
(3)

In FL7701, the LED RMS current determines the inductance parameter. To drive for CCM Mode, define LED RMS current first, as:

$$I_{LED(rms)} = \frac{I_{LED(ave.peak)}}{\sqrt{2}} \quad [A]$$
(4)

Substituting Equation (2) for Equation (4), the inductance of inductor is obtained.

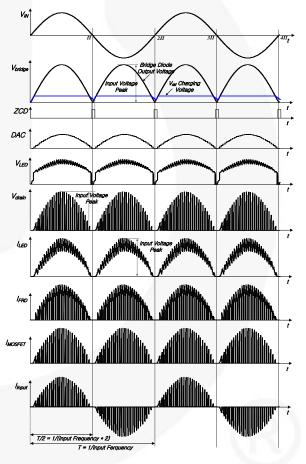
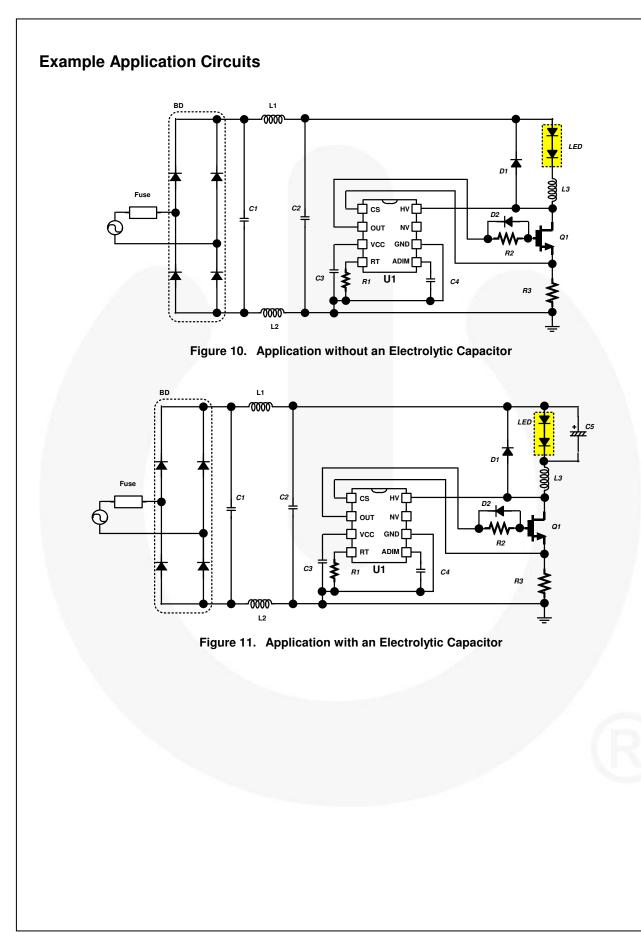
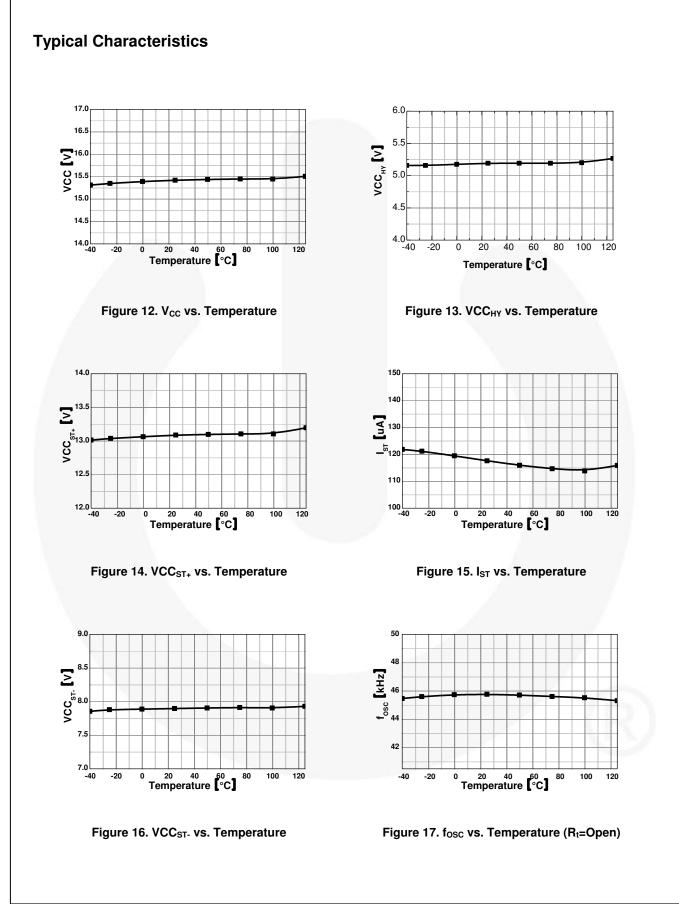
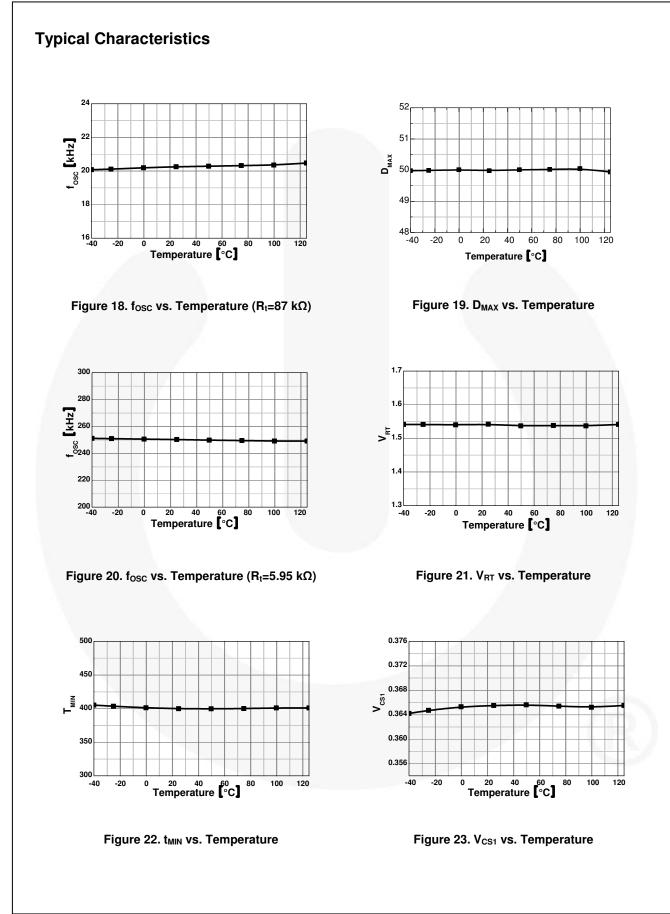
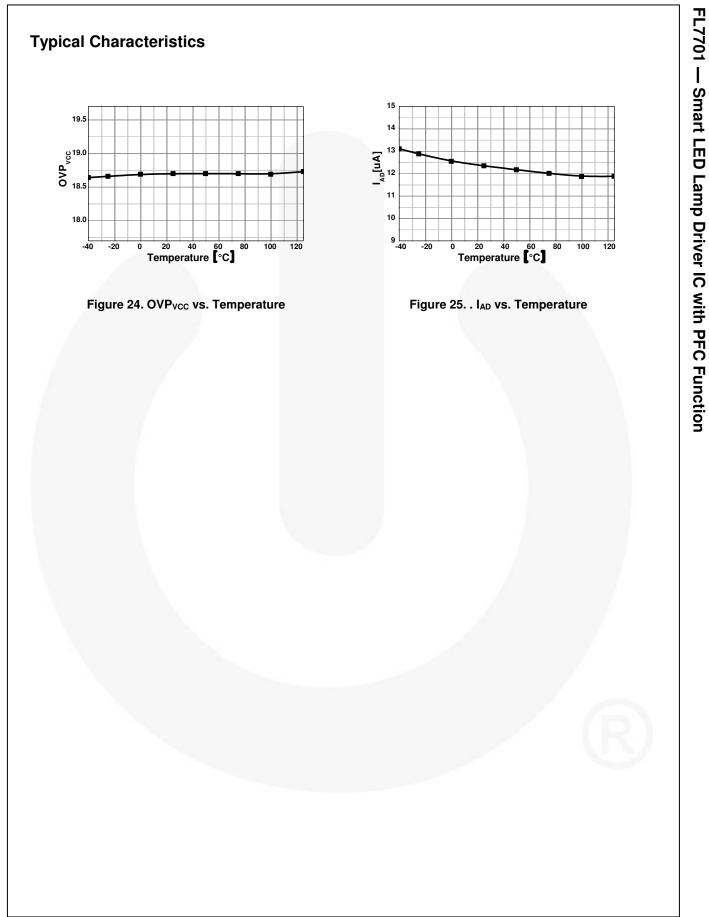


Figure 9. Typical Performance Characteristics











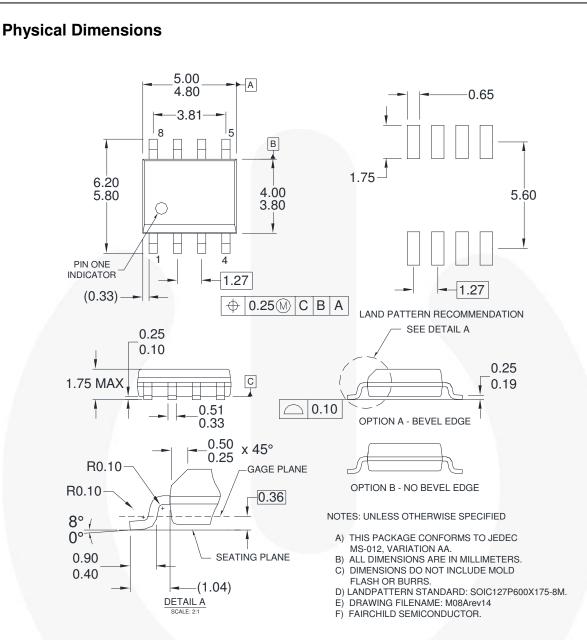


Figure 26. 8-Lead, Small-Outline Integrated Circuit (SOIC), JEDEC MS-012, .150-Inch Narrow Body

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