

Description

The ACE7215C is a high-efficiency monolithic synchronous buck regulator using a constant frequency, current mode architecture. The device is available in an adjustable version. Supply current with no load is 40uA and drops to <1uA in shutdown. The 2.5V to 6V input voltage range makes the ACE7215C ideally suited for single Li-Ion battery powered applications. 100% duty cycle provides low dropout operation, extending battery life in portable systems. PWM/PFM mode operation provides very low output ripple voltage for noise sensitive applications.

Switching frequency is internally set at 1.5MHz, allowing the use of small surface mount inductors and capacitors. Low output voltages are easily supported with the 0.6V feedback reference voltage. The ACE7215C is available in SOT23-5 and DFN2*2-6L package.

Features

- High Efficiency: Up to 96%
- 2.5V to 6V Input Voltage Range
- 1.5MHz Constant Frequency Operation
- No Schottky Diode Required
- Low Dropout Operation:100% Duty Cycle
- PFM Mode for High Efficiency in Light Load
- Over temperature Protected
- Low Quiescent Current: 40µA
- Short Circuit Protection
- Inrush Current Limit and Soft Start
- Available in SOT23-5 and DFN2*2-6L

Application

- Cellular and Smart Phones
- Wireless and DSL Modems
- PDAs
- Portable Instruments
- Digital Still and Video Cameras
- PC Cards



ACE7215C 1.5A 1.5MHz Synchronous Buck Converter

Absolute Maximum Ratings

Paramete	er	Value		
Max Input Vo	ltage	7V		
Max Voltage A	II Pins	7V		
Max Operating Junction	Temperature(Tj)	125 ℃		
Ambient Temperature(Ta)		-40 °C ~85 °C		
Maximum Dowar Dissinction	SOT23-5	400mW		
Maximum Power Dissipation	DFN2*2-6L	600mW		
Storage Temperature(Ts)		-40 °C ~50 °C		
Lead Temperature & Time		260°C, 10S		

Note: Exceed these limits to damage to the device. Exposure to absolute maximum rating conditions may affect device reliability.

Packaging Type





PIN NO		Description	Description		
SOT23-5	DFN2*2-6L	Description	Description		
1	2	EN	Chip Enable Pin. Drive EN above 1.5V to turn on the part. Drive EN below 0.3V to turn it off. Do not leave EN floating.		
2	1/5/7	GND	Ground		
3	4	SW	Power Switch Output. It is the switch node connection to Inductor. This pin connects to the drains of the internal P-ch and N-ch MOSFET switches.		
4	3	VIN	Supply voltage. Must be closely decoupled to GND with a 10μ F or greater ceramic capacitor.		
5	6	FB	Output Voltage Feedback Pin. An internal resistive divider divides the output voltage down for comparison to the internal reference voltage.		

Ordering information





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Typical Application



Block Diagram





Electrical Characteristics

Symbol	Parameter	Conditions	Min	Тур	Мах	Unit	
Vin	Input Voltage Range		2.5		6	V	
Vref	Feedback Voltage		0.588	0.6	0.612	V	
lq		FB=90%, Vin=5.0V, Iload=0mA		150	300		
	Quiescent Current	FB= 05%, Vin=5.0V, Iload=0mA		40	70	uA	
		V _{EN} =0V, Vin=4.2V		0.1	1.0		
LnReg	Line Regulation	Vin=2.5V to 6V		0.1	0.2	%/V	
LdReg	Load Regulation			0.5		%	
Fsoc	Switching Frequency			1.5		MHz	
RdsonP	PMOS Rdson	I _{sw} =100mA		300		mohm	
RdsonN	NMOS Rdson	I _{SW} =-100mA		200		mohm	
llimit	Peak Current Limit	Vin=5V, Vout=3.3V		1.5		А	
Inoload*		Vin=5V, Vout=3.3V, Iout=0		43		uA	
Iswlk	SW Leakage Current	V _{EN} =0V,Vin=Vsw=5V			1	uA	
Ienlk	EN Leakage Current				1	uA	
Vh_en	EN Input High Voltage		1.04			V	
VI_en	EN Input Low Voltage				0.98	V	

Note: *When Dutycycle >90%, Inoload will increase. e.g. Vin=3.5V/Vout=3.3V, Inoload=800uA.



Detailed Description

The ACE7215C uses a constant frequency, current mode step-down architecture. Both the main (P-channel MOSFET) and synchronous (N-channel MOSFET) switches are internal. During normal operation, the internal top power MOSFET is turned on each cycle when the oscillator sets the RS latch, and turned off when the current comparator, ICOMP, resets the RS latch. The peak inductor current at which ICOMP resets the RS latch, is controlled by the output of error amplifier EA. When the load current increases, it causes a slight decrease in the feedback voltage, FB, relative to the 0.6V reference, which in turn, causes the EA amplifier's output voltage to increase until the average inductor current matches the new load current. While the top MOSFET is off, the bottom MOSFET is turned on until either the inductor current starts to reverse, as indicated by the current reversal comparator IRCMP, or the beginning of the next clock cycle.

Applications Information Setting the Output Voltage

In the adjustable version, the output voltage is set by a resistive divider according to the following formula:

$$R_2 = \frac{R_1}{V_{out} / V_{FB} - 1}$$

The external resistive divider is connected to the output, allowing remote voltage sensing as shown in on page 1.

Inductor Selection

For most designs, the ACE7215C operates with inductors of 1μ H to 4.7μ H. Low inductance values are physically smaller but require faster switching, which results in some efficiency loss. The inductor value can be derived from the following equation:

$$L = \frac{V_{OUT} \times (V_{I\!N} - V_{OUT})}{V_{I\!N} \times \Delta I_L \times f_{OSC}}$$

Where ΔI_L is inductor Ripple Current. Large value inductors result in lower ripple current and small value inductors result in high ripple current. For optimum voltage-positioning load transients, choose an inductor with DC series resistance in the 50m Ω to 150m Ω range.



Input Capacitor Selection

Higher values, lower cost ceramic capacitors are now becoming available in smaller case sizes. Their high ripple current, high voltage rating and low ESR make them ideal for switching regulator applications. Because the ACE7215C's control loop does not depend on the output capacitor's ESR for stable operation, ceramic capacitors can be used freely to achieve very low output ripple and small circuit size. However, care must be taken when ceramic capacitors are used at the input and the output. When a ceramic capacitor is used at the input and the power is supplied by a wall adapter through long wires, a load step at the output can induce ringing at the input, VIN. At best, this ringing can couple to the output and be mistaken as loop instability. At worst, a sudden inrush of current through the long wires can potentially cause a voltage spike at VIN, large enough to damage the part. When choosing the input and output ceramic capacitors, choose the X5R or X7R dielectric formulations. These dielectrics have the best temperature and voltage characteristics of all the ceramics for a given value and size.

PC Board Layout Checklist

When laying out the printed circuit board, the following checking should be used to ensure proper operation of the ACE7215C. Check the following in your layout:

- 1. The power traces, consisting of the GND trace, the SW trace and the VIN trace should be kept short, direct and wide.
- 2. Does the (+) plates of Cin connect to Vin as closely as possible? This capacitor provides the AC current to the internal power MOSFETs.
- 3. Keep the switching node, SW, away from the sensitive VOUT node.
- 4. Keep the (-) plates of Cin and Cout as close as possible



Electrical Performance

Tested under TA= 25° C, unless otherwise specified





Output Ripple and SW at 1.5A load Vin=5V / Vout=3.3V Ch1—Vin, Ch2—Vout, Ch3—Vsw, Ch4—IL





Packing Information

SOT23-5





Packing Information

DFN2*2-6L





Notes

ACE does not assume any responsibility for use as critical components in life support devices or systems without the express written approval of the president and general counsel of ACE Electronics Co., LTD. As sued herein:

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- 2. A critical component is any component of a life support device or system whose failure to perform can be reasonably expected to cause the failure of the life support device or system, or to affect its safety or effectiveness.

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