# 24-channels LED driver with error detection and gain control 



## Features

- 24 constant current output channels
- Output current: from 4 mA to 72 mA
- $8 \times 3$ independently controlled channels (RGB)
- Current programmable through external resistor
- 7-bit global current gain adjustment in two ranges
- Error detection mode (both open and shorted LED)
- Programmable shorted LED detection thresholds
- Auto power-saving / auto wakeup
- Gradual output delay (selectable)
- Supply voltage: 3 V to 5.5 V
- Thermal shutdown and thermal flag
- Up to 30 MHz CLK 4 wires interface
- 20 V current generators rated voltage


## Applications

- Full color large displays
- LED signage
- LED screens for indoor and outdoor billboards


## Description

The LED2472G is a monolithic, low voltage, low current power 24-bit shift register designed for LED panel displays with particular features oriented to indoor and outdoor LED screen billboards. The LED2472G guarantees 20 V of output driving capability, allowing several LEDs to be connected in series. The device is configured in 3 groups (red, green and blue) of 8 independently-controlled channels. The LED current can be separately regulated for each color within the range from 4 mA to 72 mA . This range is divided into two sub-ranges and the current can be adjusted within each range in 64 steps of resolution ( 6 bits per color). A single external resistor is required. All the controls and the shift register data are accessible via serial interface. A single 24-bit configuration register is used to choose features and settings to fit the application. The LED failure detection circuit checks 3 different conditions that can occur at the output line: short to GND, short to LED power supply rail or open channel. The auto power shutdown and auto power-on feature (selectable) allows the device to save power without any external intervention. Thermal management includes overtemperature flag and the output thermal shutdown $\left(170^{\circ} \mathrm{C}\right)$. The high clock frequency of up to 30 MHz makes the device suitable for high data rate transmission. A selectable gradual output delay reduces the inrush current. The supply voltage ranges from 3 V and 5.5 V .

Table 1. Device summary

| Order code | Package | Packaging |
| :---: | :---: | :---: |
| LED2472GBTR | TQFP48-EP | Tape and reel |
| LED2472GQTR | MLPQ40-EP $5 \times 5$ |  |

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## 1

 Pin descriptionFigure 1. Pinout for TQFP48EP (top view)


Figure 2. Pinout for MLPQ40 (top view)


Table 2. Pin description

| Pin |  | Symbol | Name and function |
| :---: | :---: | :---: | :---: |
| TQFP48 | MLPQ40 |  |  |
| 29, 30 | 23, 24 | GND | Ground |
| 2 | 2 | SDI | serial data input |
| 3 | 3 | LE | Latch enable |
| 8,9 | 8, 9 | Reserved | Not used in applications |
| 4 | 7 | CLK | Clock |
| 1, 5, 6, 7, 10, 11, 12, 25, 26, 28, 36 | 4, 5, 6 | NC | Not connected |
| 31 | 25 | VDD | Power supply voltage |
| 13 | 10 | R1 | Red output 1 |
| 14 | 11 | G1 | Green output 1 |
| 15 | 12 | B1 | Blue output 1 |
| 16 | 13 | R2 | Red output 2 |
| 17 | 14 | G2 | Green output 2 |
| 18 | 15 | B2 | Blue output 2 |
| 19 | 16 | R3 | Red output 3 |
| 20 | 17 | G3 | Green output 3 |
| 21 | 18 | B3 | Blue output 3 |
| 22 | 19 | R4 | Red output 4 |
| 23 | 20 | G4 | Green output 4 |
| 24 | 21 | B4 | Blue output 4 |
| 27 | 22 | ISET | Current setup |
| 32 | 26 | BOE | Blue output enable |
| 33 | 27 | GOE | Green output enable |
| 34 | 28 | ROE | Red output enable |
| 35 | 29 | SDO | Serial data output |
| 37 | 30 | B5 | Red output 5 |
| 38 | 31 | G5 | Green output 5 |
| 39 | 32 | R5 | Blue output 5 |
| 40 | 33 | B6 | Red output 6 |
| 41 | 34 | G6 | Green output 6 |
| 42 | 35 | R6 | Blue output 6 |
| 43 | 36 | B7 | Red output 7 |
| 44 | 37 | G7 | Green output 7 |
| 45 | 38 | R7 | Blue output 7 |
| 46 | 39 | B8 | Red output 8 |

Table 2. Pin description (continued)

| Pin |  | Symbol | Name and function |
| :---: | :---: | :---: | :---: |
| TQFP48 | MLPQ40 |  |  |
| 47 | 40 | G8 | Green output 8 |
| 48 | 1 | R8 | Blue output 8 |

## 2 Absolute maximum ratings

Stressing the device above the ratings listed in the Table 3 may cause permanent damage to the device. These are stress ratings only and operation of the device at these or any other conditions above those indicated in the operating sections of this specification is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

Table 3. Absolute maximum ratings

| Symbol | Parameter | Value | Unit |
| :---: | :--- | :---: | :---: |
| $\mathrm{V}_{\mathrm{DD}}$ | Supply voltage | -0.4 to 7 | V |
| $\mathrm{~V}_{\mathrm{I}}$ | Digital inputs voltage | -0.4 to $\mathrm{Vdd}+0.4$ | V |
| OUT | Driver outputs voltage $(\mathrm{R}<1: 8>, \mathrm{G}<1: 8>, \mathrm{B}<1: 8>)$ | 20 | V |
| Io | Output current | 80 | mA |
| IGND | GND terminal current | 1.9 | A |
| ESD | Electrostatic discharge protection HBM human body model | $\pm 2$ | KV |
|  | Electrostatic discharge protection MM machine model | $\pm 200$ | V |

## 3 Thermal characteristics

Table 4. Thermal characteristics

| Symbol | Parameter | Value | Unit |
| :---: | :--- | :---: | :---: |
| Ta | Operative free-air temperature range ${ }^{(1)}$ | -40 to +85 |  |
| TJ-OPR | Operative thermal junction temperature range | -40 to +125 | C |
| Tstg | Storage temperature range | -55 to +150 |  |
| Rth $_{\text {ja }}$ | Junction-ambient thermal resistance; QFN40-EP ${ }^{(2)}$ | 25.3 | ${ }^{\circ} \mathrm{C} / \mathrm{W}$ |
|  | Junction-ambient thermal resistance; TSSOP48-EP ${ }^{(1)}$ | 33 | ${ }^{\circ}{ }^{\circ} \mathrm{C} / \mathrm{W}$ |

1. This data must be considered in adequate power dissipation conditions. The junction temperature must be maintained below $150^{\circ} \mathrm{C}$.
2. In accordance with JEDEC standard 51-7B. The exposed pad should be soldered directly to the PCB to obtain the thermal benefits.

## 4 Electrical characteristics

$\mathrm{Vdd}=3.3 \mathrm{~V}, \mathrm{Tj}=25^{\circ} \mathrm{C}, \mathrm{GRG}=" 1 "$ (gain reg), $\mathrm{R}_{\text {ext }}=13 \mathrm{k} \Omega$, unless otherwise specified.

Table 5. Electrical characteristics

| Symbol | Parameter | Test conditions | Min. | Typ. | Max. | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Vdd | Supply voltage |  | 3 |  | 5.5 | V |
| $\mathrm{V}_{\text {OUT }}$ | Output voltage | For all outputs | - | - | 19 |  |
| $\mathrm{V}_{\mathrm{IH}}$ | Input voltage |  | $0.7 \cdot \mathrm{Vdd}$ | - | Vdd |  |
| $\mathrm{V}_{\text {IL }}$ |  |  | GND | - | 0.3-Vdd |  |
| $\mathrm{V}_{\text {OL }}$ | Serial data output voltage (SDO) | $\begin{gathered} \mathrm{V}_{\mathrm{DD}}=3 \text { to } 5.5 \mathrm{~V} \\ \mathrm{I}=+/-1 \mathrm{~mA} \end{gathered}$ | - | - | 0.4 |  |
| $\mathrm{V}_{\mathrm{OH}}$ |  |  | $\mathrm{V}_{\mathrm{DD}}-0.4$ | - | - |  |
| loleak | Output leakage current | $\mathrm{Vo}=19 \mathrm{~V}$, all outputs OFF | - | - | 0.5 | uA |
| $\mathrm{V}_{\text {uvlo } 1}$ | UVLO threshold voltage (rising) |  |  | 2.7 | 2.9 | V |
| $\mathrm{V}_{\text {uvio1 }}$ | UVLO threshold voltage (falling) |  | 2.2 | 2.3 |  | V |
| Hyuvlo | UVLO hysteresis |  |  | 400 |  | mV |
| ${ }^{\text {l }} \mathrm{OL1}$ | Output current precision channel-to-channel per each color group$\text { (all outputs } \mathrm{ON})^{(1)(2)}$ | $\begin{gathered} \mathrm{V}_{\mathrm{o}}=0.3 \mathrm{~V} ;(\mathrm{lo}=5 \mathrm{~mA}) \\ \mathrm{CFG}-0=\mathrm{CFG}-1=\mathrm{CFG}-2=" 0 " \\ \mathrm{GRG}=" 0 " \end{gathered}$ | - | - | $\pm 4$ | \% |
| ${ }^{\Delta} \mathrm{OL3}$ |  | $\begin{gathered} \mathrm{V}_{\mathrm{o}}=0.6 \mathrm{~V} ;(\mathrm{lo}=21 \mathrm{~mA}) \\ \text { CFG-0 }=\text { CFG-1 }=\mathrm{CFG}-2=" 0 " \end{gathered}$ | - | - | $\pm 3$ |  |
| ${ }^{\Delta} \mathrm{OLL}$ |  | $\begin{aligned} & \mathrm{V}_{\mathrm{o}}= 0.5 \mathrm{~V} ;(\mathrm{lo}=15 \mathrm{~mA}) \\ & \mathrm{CFG}-0= \text { CFG-1 }=\mathrm{CFG}-2=" 1 " \\ & \text { GRG }=" 0 " \end{aligned}$ | - | - | $\pm 3$ |  |
| ${ }^{\text {l }} \mathrm{OL4}$ |  | $\begin{aligned} \mathrm{Vo} & =1.2 \mathrm{~V} ;(\mathrm{lo}=61 \mathrm{~mA}) \\ \text { CFG-0 } & =\text { CFG-1 }=\mathrm{CFG}-2=" 1 " \end{aligned}$ | - | - | $\pm 3$ |  |
| $\Delta^{\text {l }}$ OLIa | Output current error device-todevice per each color group (all outputs ON ) ${ }^{(1)}$ | $\begin{gathered} \mathrm{V}_{\mathrm{o}}=0.3 \mathrm{~V} ;(\mathrm{lo}=5 \mathrm{~mA}) \\ \mathrm{CFG}-0=\mathrm{CFG}-1=\mathrm{CFG}-2=" 0 " \\ \mathrm{GRG}=" 0 " \end{gathered}$ | - | - | $\pm 6$ |  |
| $\Delta^{\text {l }}$ OL3a |  | $\begin{aligned} \mathrm{V}_{\mathrm{o}} & =0.6 \mathrm{~V} ;(\mathrm{Io}=21 \mathrm{~mA}) \\ \text { CFG-0 } & =\text { CFG-1 }=\mathrm{CFG}-2=" 0 " \end{aligned}$ | - | - | $\pm 6$ |  |
| ${ }^{\text {a }}$ OL2a |  | $\begin{gathered} \mathrm{V}_{\mathrm{o}}=0.5 \mathrm{~V} ;(\mathrm{lo}=15 \mathrm{~mA}) \\ \mathrm{CFG}-0=\mathrm{CFG}-1=\mathrm{CFG}-2=" 1 " \\ \mathrm{GRG}=" 0 " \end{gathered}$ | - | - | $\pm 6$ |  |
| $\Delta^{\text {OL4a }}$ |  | $\begin{gathered} \mathrm{Vo}=1.2 \mathrm{~V} ;(\mathrm{lo}=61 \mathrm{~mA}) \\ \text { CFG-0 }=\text { CFG-1 }=\mathrm{CFG}-2=" 1 " \end{gathered}$ | - | - | $\pm 6$ |  |

Table 5. Electrical characteristics (continued)

| Symbol | Parameter | Test conditions | Min. | Typ. | Max. | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\% / \Delta \mathrm{V}_{\text {OUT }}$ | Output current vs. output voltage regulation ${ }^{(3)}$ | $\begin{gathered} \begin{array}{r} \mathrm{V}_{0} \text { from } 1.2 \mathrm{~V} \text { to } 3 \mathrm{~V} ;(\mathrm{lo}=61 \\ \mathrm{mA}) \end{array} \\ \text { CFG-0 }=\text { CFG-1 }=\text { CFG-2 }=" 1 " \end{gathered}$ | - | $\pm 0.2$ | - | \%/V |
| $\% / \Delta \mathrm{V}_{\mathrm{DD}}$ | Output current vs. supply voltage regulation ${ }^{(4)}$ | Vdd from 3 V to 5.5 V $\begin{gathered} \mathrm{V}_{0}=1.2 \mathrm{~V} ;(\mathrm{lo}=61 \mathrm{~mA}) \\ \mathrm{CFG}-0=\text { CFG-1 }=\mathrm{CFG}-2=" 1 " \end{gathered}$ | - | $\pm 1$ | - |  |
| Rup | Pull-up resistor for OE pin |  |  |  |  | k $\Omega$ |
| Rdw | Pull-down resistor for LE pin |  |  |  |  |  |
| $\mathrm{R}_{\text {ext }}$ | External current setup resistance |  |  |  | 100 |  |
| $l_{\text {DD1 }}$ | Supply current (OFF) | No data transfers, all outputs OFF, $\begin{gathered} \text { CFG-0 }=\text { CFG- } 1=\text { CFG- } 2=" 0 " \\ \text { GRG }=" 0 " ; \text { CFG- } 6=" 0 " \end{gathered}$ |  |  | 8 | mA |
| ${ }^{\text {DD2 }}$ |  | No data transfers, all outputs OFF, $\begin{gathered} \text { CFG- } 0=\text { CFG- } 1=\mathrm{CFG}-2=" 1 " \\ \text { CFG- } 6=" 0 " \end{gathered}$ |  |  | 16 |  |
| $\mathrm{I}_{\mathrm{DD} 1}$ | Supply current (ON) | No data transfers, all outputs ON, $\begin{gathered} \text { CFG- } 0=\mathrm{CFG}-1=\mathrm{CFG}-2=" 0 " \\ \text { GRG }=0 " 0 \end{gathered}$ | - |  | 8 | mA |
| $\mathrm{I}_{\mathrm{DD} 2}$ |  | No data transfers, all outputs ON, CFG-0 = CFG-1 = CFG-2 = " 1 " | - |  | 15 |  |
| $\begin{gathered} \text { IDD } \\ \text { (AutoOFF) } \end{gathered}$ | Supply current (autoOFF) | All output OFF $\text { CFG-6 = " } 1 \text { " }$ | - | 200 | 500 | $\mu \mathrm{A}$ |
| $\mathrm{SDE}_{1}$ | LED short detection voltage | CFG-3 = CFG-4 = CFG-5 = "0" |  | 2.0 |  | V |
| $\mathrm{SDE}_{2}$ |  | CFG-3 = CFG-4 = CFG-5 = "1" |  | 3.0 |  |  |
| ODC | LED open detection current |  |  | $0.5 \mathrm{ILL}^{\text {a }}$ |  |  |

Table 5. Electrical characteristics (continued)

| Symbol | Parameter | Test conditions | Min. | Typ. | Max. | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Tflg | Thermal flag |  |  | 150 |  |  |
| Tsd | Thermal shutdown $^{(5)}$ |  |  | 170 |  |  |
| Tsd-hy | Thermal shutdown <br> hysteresis${ }^{(5)}$ |  |  |  |  |  |$\quad$| $\circ$ |
| :---: | :---: | :---: | :---: |

1. Tested with just one output loaded
2. $\left(\left(\right.\right.$ lout $_{n}-$ lout $\left._{\text {avg } 1-15}\right) /$ lout $\left._{\text {avg } 1-15}\right) \times 100$
3. 

$$
\Delta(\% / \mathrm{V})=\frac{(\text { Ioutn @ Voutn }=3.0 \mathrm{~V})-(\text { Ioutn @ Voutn }=1.0 \mathrm{~V})}{(\text { Ioutn } @ \text { Voutn }=1.0 \mathrm{~V})} \times \frac{100}{3-1}
$$

4. 

$$
\Delta(\% / \mathrm{V})=\frac{(\text { Ioutn @ Vdd }=5.5 \mathrm{~V})-(\text { Ioutn @ } \mathrm{Vdd}=3.0 \mathrm{~V})}{(\text { Ioutn } @ \mathrm{Vdd}=3.0 \mathrm{~V})} \times \frac{100}{5.5-3}
$$

5. Not tested, guaranteed by design.

### 4.1 Typical application circuit

Figure 3. Typical application circuit


## 5 Switching characteristics

Vdd $=3.3 \mathrm{~V}, \mathrm{Tj}=25^{\circ} \mathrm{C}, \mathrm{GRG}=" 1 "$ (gain reg), $\mathrm{R}_{\mathrm{ext}}=13 \mathrm{k} \Omega$, unless otherwise specified.
Table 6. Switching characteristics ${ }^{(1)(2)}$

| Symbol | Parameter |  | Conditions | Min. | Typ. | Max. | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\mathrm{f}_{\mathrm{clk}}$ | Clock frequency |  | Cascade operation | - | - | 30 | MHz |
| $\mathrm{tr}_{(\mathrm{SDO}}$ | SDO rise time |  | $\begin{aligned} \mathrm{R}_{\text {ext }}= & 13 \mathrm{k} \Omega ; \text { lout }=21 \mathrm{~mA} \\ & \text { Vout }=0.6 \mathrm{~V} \\ \mathrm{VIH} & =\mathrm{VDD} ; \mathrm{VIL}=\mathrm{GND} \\ \mathrm{RL} & =56 \Omega ; \mathrm{CL}=10 \mathrm{pF} \\ \text { CFG- } 0 & =\mathrm{CFG}-1=\mathrm{CFG}-2=" 0 " \end{aligned}$ | - | 5 | - | ns |
| $\mathrm{tf}_{(S D O)}$ | SDO fall time |  |  | - | 5 | - |  |
| tPLH2 | LE-OUTn ${ }^{(3)}$ | Propagation delay time ("L" to "H") |  | - | 70 | - |  |
| tPLH3 | $\overline{\mathrm{OE}-O U T n}{ }^{(3)}$ |  |  | - | 100 | - |  |
| tPLH | $\begin{aligned} & \text { CLK-SDO } \\ & \text { CFG-7 = '0' } \end{aligned}$ |  |  | 8 | 15 | 25 |  |
| tPHL2 | LE-OUTn ${ }^{(3)}$ | Propagation delay time ("H" to "L") |  | - | 70 | - |  |
| tPHL3 | $\overline{\mathrm{OE}}$-OUTn ${ }^{(3)}$ |  |  | - | 100 | - |  |
| tPHL | $\begin{aligned} & \text { CLK-SDO } \\ & \text { CFG- } 7=0 \text { ' } \end{aligned}$ |  |  | 8 | 15 | 25 |  |
| tw(CLK) | CLK | Pulse width |  | 20 | - | - |  |
| $\mathrm{t}_{\mathrm{w}}(\overline{\mathrm{OE}})$ | $\overline{\mathrm{OE}}$ |  |  | $150^{(4)}$ | - | - |  |
| tw(L) | LE |  |  | 20 | - | - |  |
| $\mathrm{t}_{\mathrm{gr-d}}$ | Gradual delay Ch to Ch |  |  |  | 10 |  |  |
| $\mathrm{t}_{\text {su(L) }}$ | Setup time for LE |  |  | 5 | - | - |  |
| $\mathrm{th}_{\mathrm{h}}(\mathrm{L})$ | Hold time for LE |  |  | 5 | - | - |  |
| $\mathrm{t}_{\text {su( }{ }^{\text {D }} \text { ) }}$ | Setup time for SDI |  |  | 5 | - | - |  |
| $\mathrm{th}_{\mathrm{h}(\mathrm{D})}$ | Hold time for SDI |  |  | 10 | - | - |  |
| tor ${ }^{(5)}$ | Maximum CLK rise time |  |  | - | - | 5 | $\mu \mathrm{s}$ |
| tof ${ }^{(5)}$ | Maximum CLK fall time |  |  | - | - | 5 |  |
| $\mathrm{I}_{\text {out-ov }}$ | Output current turn-on overshoot |  | $\begin{gathered} \text { Vout }=0.3 \text { to } 3 \mathrm{~V} \\ \mathrm{CL}=10 \mathrm{pF} ; \text { lout }=5 \text { to } 61 \mathrm{~mA} \end{gathered}$ | - | - | 10 | \% |
| $t_{\text {n-err }}$ | Normal error detection minimum output ON time |  |  | - | - | 1 | $\mu \mathrm{s}$ |

Table 6. Switching characteristics ${ }^{(1)(2)}$

| Symbol | Parameter | Conditions | Min. | Typ. | Max. | Unit |
| :--- | :---: | :---: | :---: | :---: | :---: | :---: |
| $t_{\text {shut-down }}$ | Auto power shutdown time (autoOFF) | From LE falling edge to $R_{\text {ext }}$ <br> voltage reference at $-10 \%$ | - | 75 | - | ns |
| $\mathrm{t}_{\text {wake-up }}$ | Auto power wakeup time | From LE falling edge to $R_{\text {ext }}$ <br> voltage reference at $90 \%$ | - | 1 | - | $\mu \mathrm{s}$ |

1. All table limits are guaranteed by design.
2. Not tested in production.
3. $\mathrm{CFG}-8=$ " 1 " (no output gradual delay)
4. In normal error detection mode must be longer than $1 \mu \mathrm{~s}$
5. If devices are connected in cascade and $t_{o r}$ or $t_{\text {of }}$ is large, it may be critical to achieve the timing required for data transfer between two cascaded devices

## 6 Timing

Figure 4. Timing for clock, serial-in, serial-out, latch enable and outputs


Correct sampling of the data depends on the stability of the data at SDI on the rising edge of the clock signal and it is assured by a proper data setup and hold time ( $\mathrm{t}_{\mathrm{SU}(\mathrm{D})}$ and $\mathrm{t}_{\mathrm{h}(\mathrm{D})}$ ), as shown in Figure 4. The same figure shows the propagation delay from CLK to SDO ( $\mathrm{t}_{\mathrm{PLH}} / \mathrm{t}_{\mathrm{PHL}}$ ). Figure 4 describes also the minimum duration of CLK and LE pulses ( $\mathrm{t}_{\mathrm{W}(\mathrm{CLK})}$ and $t_{W(L)}$, respectively) and the propagation delay from LE to $\mathrm{OUT}_{\mathrm{n}}\left(\mathrm{t}_{\mathrm{PLH} 1} / \mathrm{t}_{\mathrm{PHL} 1}\right.$ and $t_{\text {PLH2 }} / \mathrm{t}_{\mathrm{PHL} 2}$, respectively). Finally, Figure 5 also defines the turn-on and turn-off time ( $\mathrm{t}_{\mathrm{of}}$ and $t_{\text {or }}$ ) of the output voltage.

Figure 5. OE and Outputs timing


## 7 Simplified Internal block diagram

Figure 6. LED2472G simplified block diagram


### 7.1 Equivalent circuits of inputs and outputs

Input terminals LE and /OE have pull-down and pull-up connections, respectively. CLK and SDI must be connected to external circuits to fix the logic level.

Figure 7. Input terminals


CLK, SDI terminal


LE terminal


SDO terminal


## 8 Digital blocks

The data inputs come in through the serial interface at each CLK rising edge and after 24 CLK cycles all data are loaded into the shift register. The LE signal is used to latch the loaded data and also to generate digital keys for CFG management, scrolling, thermal check and LED error detection. When one of the output enable signal (OER, OEB or OEG) is low, the corresponding data are transferred to the relative output drivers. The data flow is "first in, first out". To latch the data, the LE signal must be high during the last data loading CLK rising edge (Table 7). When one of the output enable signals (OER, OEB or OEG) is at low level, output terminals (R1-R8,G1-G8, B1-B8) respond to the data either ON or OFF. When one of the output enable signals (OER, OEB or OEG) goes to "1", all outputs switch off all the data on the output terminal. LE and /OE signals are asynchronous with respect to the CLK signal. The time diagram below refers to RGB flow setting.

Figure 8. Data input time diagram


### 8.1 Register access

Access to the different registers of the device (configuration register, gain register, etc.) is achieved by using different digital keys, defined as a number of CLK pulses during which the LE signal is asserted. The available digital keys are summarized in Table 7.

Table 7. Digital keys summary

| \#CLK rising edges with the LE asserted | Description |
| :---: | :---: |
| $1-2$ | Data latch |
| $3-4$ | Write configuration register |
| $5-6$ | Read configuration register |
| $7-8$ | Write gain |
| $9-10$ | Read gain |
| 11 | Open detection |
| 12 | Short detection |
| 13 | Open/short detection |
| 14 | Thermal alert reading |
| 15 | Reserved |
| 16 | Reserved |

Figure 9. Digital key timing diagram


### 8.2 Configuration register

The configuration register is used to enable or disable some device features, to program some parameters and to change other settings. Access to this register (read or write) is
managed as described in Table 8 where a description is provided for each bit. The default value of the configuration register (when the device is switched ON or after a reset) is all bits set to " 0 ". To change anything in the configuration register, a 24 -bit digital word must be sent (CFG-0 represents the LSB, CFG-23 the MSB).

Figure 10. Configuration register


Table 8. Configuration register

| BIT | Definition | Attribute <br> read/write | Configuration register function description | Default |
| :---: | :---: | :---: | :--- | :---: |
| CFG-0 | RED <br> current range | R/W | "0" low current range <br> "1" high current range | 0 |
| CFG-1 | GREEN current <br> range | R/W | "0" low current range <br> "1" high current range | 0 |
| CFG-2 | BLUE <br> current range | R/W | "0" low current range <br> "1" high current range | 0 |
| CFG-3 | RED <br> voltage det. thr. | R/W | "0" LED short-circuit detection threshold 2 V <br> "1" LED short-circuit detection threshold 3 V | 0 |
| CFG-4 | GREEN <br> voltage det. thr. | R/W | "0" LED short-circuit detection threshold 2 V <br> "1"LE D short-circuit detection threshold 3 V | 0 |
| CFG-5 | BLUE <br> voltage det. thr. | R/W | "0" LED short-circuit detection threshold 2 V <br> "1" LED short-circuit detection threshold 3 V | 0 |
| CFG-6 | Auto OFF | R/W | "0" device always ON <br> "1" auto power shutdown active (Auto OFF) | 0 |
| CFG-7 | SDO delay | R/W | "0" SDO half clock delay disabled <br> "1" SDO half clock delay enabled | 0 |
| CFG-8 | Gradual output <br> delay | R/W | "0" gradual outputs delay is applied <br> "1" all channels switch ON and OFF <br> simultaneously | 0 |

Table 8. Configuration register (continued)

| BIT | Definition | Attribute read/write | Configuration register function description |  |  |  |  | Default |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| CFG-9 | Data flow | R/W | Color data flow management | CFG9 | CFG10 | CFG11 |  | 0 |
|  |  |  |  | 0 | 0 | 0 | RGB |  |
|  |  |  |  | 0 | 0 | 1 | GBR |  |
| CFG-10 |  |  |  | 0 | 1 | 0 | GRB | 0 |
|  |  |  |  | 0 | 1 | 1 | BGR |  |
| CFG-11 |  |  |  | 1 | 0 | 0 | BRG | 0 |
|  |  |  |  | 1 | 0 | 1 | RBG |  |
| $\begin{gathered} \text { CFG } \\ 12 \div 23 \end{gathered}$ | Don't care |  |  |  |  |  |  |  |

### 8.3 Current ranges (CFG 0-CFG 2)

The output LED currents can be programmed using an external resistor connected to GND from the ISET pin and can be adjusted using 6 bits in a dedicated gain register with two possible current ranges selectable in the configuration register. Each range can be separately selected for each color by the bits CFG-0, CFG-1 and CFG-2, respectively, for the RED, GREEN and BLUE channels.

### 8.4 Error detection conditions (CFG 3-CFG 5)

During error detection phases for each channel, the following are checked:

- output current for open circuit detection
- output voltage for short-circuit detection

The thresholds for the error diagnostics are summarized in the Table 9:

Table 9. Diagnostic thresholds

| Error detection | Checked malfunction | CFG-x ${ }^{(1)}$ | Thresholds |
| :---: | :---: | :---: | :---: |
| Open detection <br> combined mode | Open line or output short to GND | Don't care | lo $<0.5 \times$ lo_programmed |
| Short detection | Short on LED or short to VLED | 0 | Vo $>2 \mathrm{~V}$ |
|  |  | 1 | Vo $>3 \mathrm{~V}$ |

1. $x=3$ for RED, $x=4$ for GREEN, $x=5$ for BLUE

### 8.5 Auto power shutdown / wakeup (CFG 6)

This feature reduces the power consumption when all outputs are OFF. It is active when the bit CFG-6 of the configuration register is at "1". The auto power shutdown (auto OFF) starts when the data latched is " 0 " for all channels, and the device will be active again (wakeup) at the first latched data string including at least one bit equal to "1" (at least one channel ON).

Timings for shutdown and wakeup are present in the dynamic features table. While the auto power shutdown is active, the device ignores any other command except channel power-on.

### 8.6 SDO delay (CFG 7)

Normally, on SDO terminals data is shifted out at the rising edge of the CLK signal with a propagation delay of about 15 ns [signal (1) in Figure 11]. The device provides the possibility to shift data out also at the falling edge of the CLK signal with a propagation delay of few ns [signal (2) in Figure 11]. This feature can be activated by setting to "1" the bit CFG-7 of the configuration register. The default setting for this bit is " 0 ", hence the SDO delay is not activated by default. This feature is particularly useful when multiple devices are connected in daisy chain configuration with non-matched delays between the CLK and SDO data paths (board routing).

Figure 11. SDO Delay


### 8.7 Gradual output delay (CFG 8)

The gradual output delay consists in turning on gradually the current generators, avoiding turning on all channels at the same time. This feature prevents large inrush current and reduces the bypass capacitor values. The fixed delay time can be activated by bit CFG-8 of the configuration register, and the typical delay is 10 ns for each group of 8 outputs $R, G, B$ (e.g. R1, G1, B1 has no delay, R2, G2, B2 has 10 ns of delay and R3, G3, B3, has 20 ns delay, and so on), as described in Table 10.

Table 10. Gradual output delay values

| Delay time (ns) from the falling edge of $\overline{\text { xOE }}$ | R1 | R2 | R3 | R4 | R5 | R6 | R7 | R8 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | G1 | G3 | G4 | G5 | G6 | G7 | G8 |  |
| B3 | B4 | B5 | B6 | B7 | B8 |  |  |  |
| CFG-8 $=" 0 "$ | 0 | 10 | 20 | 30 | 40 | 50 | 60 | 70 |
| CFG-8 $=" 1 "$ | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |

Figure 12 shows an example of the effect of the output gradual delay on the RED color group outputs.

Figure 12. Gradual delay on first four channels of RED color group


### 8.8 Data flow management (CFG 9-CFG 11)

The $8 \times 3$ shift registers have a default RGB sequence serial data flow according to the table shown into the configuration register (bit CFG-9, CFG-10 and CFG-11). Figure 13 shows how serial data are loaded in accordance with the data flow sequence selected through the configuration register. The default sequence is RGB (first bit will be R8, last bit B1 then: R8R1, G8-G1, B8-B1).

Figure 13. Different color sequence in data loading


### 8.9 Gain register

The LED current can be programmed using an external resistor connected to GND from REXT pin and can be adjusted using the dedicated bits of the gain register (G-0 to G-17 defines the gain and CFG-0/1/2 the current range within the gain can be adjusted). The device can regulate the current up to 72 mA and down to 4 mA . To change anything in the gain register, a 24-bit digital word must be sent (CFG-0 represents the LSB, CFG-23 the MSB). The accuracy of the LED current depends on the selected range and it is assured only in the ranges indicated in the static electrical characteristics (see Table 5).

Table 11. Gain register

| BIT | Definition | Attribute read/write | Register function description | Default |
| :---: | :---: | :---: | :---: | :---: |
| G-0 | RED current gain adjustment | R/W | 6-bit DAC allows adjustment of the device output current in 64 steps for each range (defined by CFG-0). default: gain = 1 | 1 |
| G-1 |  |  |  | 1 |
| G-2 |  |  |  | 1 |
| G-3 |  |  |  | 1 |
| G-4 |  |  |  | 1 |
| G-5 |  |  |  | 1 |
| G-6 | GREEN current gain adjustment | R/W | 6-bit DAC allows adjustment of the device output current in 64 steps for each range (defined by CFG-1). default: gain = 1 | 1 |
| G-7 |  |  |  | 1 |
| G-8 |  |  |  | 1 |
| G-9 |  |  |  | 1 |
| G-10 |  |  |  | 1 |
| G-11 |  |  |  | 1 |

Table 11. Gain register (continued)

| BIT | Definition | Attribute read/write | Register function description | Default |
| :---: | :---: | :---: | :---: | :---: |
| G-12 | BLUE current gain adjustment | R/W | 6-bit DAC allows adjustment of the device output current in 64 steps for each range (defined by CFG-2). default: gain =1 | 1 |
| G-13 |  |  |  | 1 |
| G-14 |  |  |  | 1 |
| G-15 |  |  |  | 1 |
| G-16 |  |  |  | 1 |
| G-17 |  |  |  | 1 |
| $\begin{aligned} & \mathrm{G}-18 \\ & \text { to } \\ & \mathrm{G}-23 \end{aligned}$ |  |  | Don't care |  |

Figure 14. $\mathrm{I}_{\text {out }}$ vs. gain


## 9 Current adjustment

The LED2472G is designed to provide a current in the range between 4 mA and 72 mA per channel. The current is programmed for all color groups by connecting an external resistor (see Figure 15) to the pin ISET and then adjusted separately for each color by the gain register. The current ranges can be separately selected for each color by the bits CFG-0, CFG-1, CFG-2 of the configuration register (respectively, for RED, GREEN and BLUE). The current for each color can be adjusted in 64 steps using 6 bits (per color) contained in the gain register.

Figure 15. External resistor to connect to ISET pin


When the device is switched on, the default value of the gain register together with the bits of the configuration register selecting the current range set a current value that can be calculated as follows:

$$
I_{O L \_ \text {_deatut }}=\frac{V_{B E F}}{R_{E T T}} \cdot K
$$

Where $\mathrm{V}_{\mathrm{REF}} \approx 1.23 \mathrm{~V}$ is the voltage of the ISET pin and K is the mirroring current ratio, whose value depends on the selected current range:
$\mathrm{K}=55$ with low current range selected (CFG-0, CFG-1 or CFG-2 set to " 0 ")
$\mathrm{K}=160$ with high current range selected (CFG-0, CFG-1 or CFG-2 set to " 1 ")
The relationship between the programmed current and the current gain settings is the following:

$$
I_{O L}=\left(I_{O L-d e f u \mu t t}+G \cdot \Delta I_{\text {sepp }}\right)
$$

where $G$ is the current gain (decimal value) defined by the dedicated bits of the current gain register. $\Delta \mathrm{I}_{\text {step }}$ can be instead defined as follows:

$$
\Delta I_{\text {step }}=\frac{I_{O L_{-} \text {default }}}{21}
$$

The recommended resistor values to cover the above mentioned current range are 11$18 \mathrm{k} \Omega$, which respectively define the following ranges:

Table 12. Current adjustment example

| Range | R $_{\text {SET }}[\mathrm{k} \Omega]$ | CFG-x ${ }^{(1)}$ | G-y to G-z ${ }^{(\mathbf{2})}$ | LED current $^{(3)}{ }^{2}$ [mA] |
| :---: | :---: | :---: | :---: | :---: |
|  | 18 | 0 | 000000 | 4 |
|  | 18 | 0 | 111111 | $\mathbf{1 5}$ |
| High | 18 | 1 | 000000 | 11 |
|  | 18 | 1 | 111111 | 44 |
| Low | 13 | 0 | 000000 | 5 |
|  | 13 | 0 | 111111 | $\mathbf{2 1}$ |
| High | 13 | 1 | 000000 | 15 |
|  | 13 | 1 | 111111 | 61 |
| Low | 11 | 0 | 000000 | 6 |
|  | 11 | 0 | 111111 | $\mathbf{2 5}$ |
| High | 11 | 1 | 000000 | 18 |
|  | 11 | 1 | 111111 | 72 |

1. $x=0$ for RED, $x=1$ for GREEN, $x=2$ for BLUE
2. $y=0 \& z=5$ for RED, $y=6 \& z=11$ for GREEN, $y=12 \& z=17$ for BLUE
3. The indicated values may be slightly different on the actual device

The current values in bold in the Table 12 are the current default values. The above mentioned resistor values are not mandatory, but only suggested to cover precisely the current range indicated for this device. Due to internal power dissipation, it is important to highlight that, loading at the maximum current of all device channels simultaneously can cause a current shift for each output. In the worst case, with all channels loaded at the maximum current of about 72 mA , the regulated current could decrease in the range of about 3-5\%.

## 10 LED error detection

The LED error detection implemented by the LED2472G is performed in order to detect shorted LED and open LED conditions. The shorted LED condition is determined by checking the voltage across each current generator. If this voltage is higher than the threshold, programmed by bits CFG-3, CFG-4 and CFG-5 of the configuration register, the LED connected to that generator is considered shorted.

The open LED condition is determined by measuring the current flowing through each current generator. If this current is lower than half the expected current, the LED connected to that generator is considered open. The error detection request and the acquisition of the results of the detection are managed by the serial interface. The steps to follow for correct performance of error detection are summarized as follows (see Figure 16):

- Enter the error detection. To do this, the appropriate digital key must be provided (see Figure 9). There is both the possibility of "generic" error detection (open/short detection) and the possibility to specifically select the type of failure (short detection or open detection).
- Performing the error detection. When the ROE, GOE and BOE signals become low, error detection starts. These signals must be kept low for at least $1 \mu \mathrm{~s}$ in order to correctly complete the error detection process. After this time, at least one CLK pulse must be provided in order to make the detection result available at the SDO pin while the output enable signals are still low. The bit shifted out of SDO after this clock pulse represents the first bit of the detection result word.
- Detection results. To complete the detection result acquisition, at least another 23 CLK pulses must be provided after the xOE signals have been set high again ( 24 CLK pulses in total). The detection result will be always in RGB sequence regardless of any different programmed data flow (CFG9-CFG11).

The detection result indicates " 1 " for each channel considered good, " 0 " for each channel that has a failure (shorted or open LED). To check the status of all channels and to obtain an accurate detection result, it is important to set all outputs to ON before starting the error detection process. If this is not done, it is worth noting that the detection result will indicate a " 0 " also for those channels not set to ON before the detection process, although they may not actually have any failure.

Figure 16. Error detection process


## 11 Thermal shutdown and thermal alert

The device can monitor the internal temperature. Based on the temperature value, the device can simply provide an alert (if the temperature exceeds $150^{\circ} \mathrm{C}$ ) through the serial interface, or trigger a thermal shutdown (if the temperature exceeds $170^{\circ} \mathrm{C}$ ). The effect of the thermal shutdown is to turn off all channels until the temperature falls (considering a hysteresis of around $15^{\circ} \mathrm{C}$ ). The thermal alert can be read by running the digital key "Thermal alert reading", holding the LE high for 14 CLK rising edges (see Figure 17). If thermal alert is asserted, a 24-bit string at " 1 " will be sent through SDO at the next 24 CLK rising edge.

Table 13. Thermal alert status summary

| Thermal alert status | Meaning |
| :---: | :---: |
| $" 000000000000000000000000 "$ | Device temperature under $150^{\circ} \mathrm{C}$ |
| $" 111111111111111111111111 "$ | Device temperature over $150^{\circ} \mathrm{C}$ |

Figure 17. LE high for 14 CLK


## 12 Dropout voltage

In order to correctly regulate the channel current, a minimum voltage ( $\mathrm{V}_{\mathrm{DROP}}$ ) across each current generator must be guaranteed. Figure 18 and Table 14 show the minimum $V_{\text {DROP }}$ related to the current to regulate. A $\mathrm{V}_{\mathrm{DROP}}$ lower than the minimum recommended implies the regulation of a current lower than that expected. However, an excess of $\mathrm{V}_{\text {DROP }}$ increases the power dissipation. When all outputs are loaded simultaneously, the minimum working drop rises. In full load condition at 61 mA per channel the minimum voltage to apply on each channel must be increased by about 400 mV ( 550 mV at 72 mA ).

Figure 18. Typical dropout voltage vs. output current (only one channel ON)


Table 14. Minimum dropout voltage for some current values (only one channel ON)

| Output nominal current [mA] | Minimum $\mathbf{V}_{\text {DROP }}[\mathrm{mV}]$ <br> $\mathbf{V}_{\mathrm{DD}}=3.3 \mathrm{~V}$ |
| :---: | :---: |
| 5 | 120 |
| 15 | 280 |
| 21 | 320 |
| 61 | 800 |
| 75 | 1050 |

## 13 Package mechanical data

In order to meet environmental requirements, ST offers these devices in different grades of ECOPACK ${ }^{\circledR}$ packages, depending on their level of environmental compliance. ECOPACK ${ }^{\circledR}$ specifications, grade definitions and product status are available at: www.st.com. ECOPACK ${ }^{\circledR}$ is an ST trademark.

Figure 19. TQFP48-EP package dimensions


Table 15. TQFP48-EP mechanical data

| Dim. | mm |  |  |
| :---: | :---: | :---: | :---: |
|  | Min. | Typ. | Max. |
| A |  |  | 1.20 |
| A1 | 0.05 |  | 0.15 |
| A2 | 0.95 | 1.00 | 1.05 |
| b | 0.17 | 0.22 | 0.27 |
| c | 0.09 | 9.00 | 0.20 |
| D | 8.80 | 7.00 | 9.20 |
| D1 | 6.80 | 5.50 | 7.20 |
| D2 | 2.00 | 9.00 | 9.20 |
| D3 | 8.80 | 7.00 | 7.20 |
| E | 6.80 |  |  |
| E1 | 2.00 | 5.50 |  |
| E2 |  | 0.50 | 0.75 |
| E3 | 0.45 | 0.60 |  |
| e |  | 1.00 | 0.08 |
| L | 0 | 3.5 |  |
| L1 |  |  |  |
| K |  |  |  |
| ccc |  |  |  |

Figure 20. TQFP48-EP recommended footprint


Figure 21. MLPQ40-EP 5x5 package dimensions


Table 16. MLPQ40-EP 5x5 mechanical data

| Dim. | mm |  |  |
| :---: | :---: | :---: | :---: |
|  | Min. | Typ. | Max. |
| A | 0.90 | 0.80 | 1.00 |
| A1 | 0.02 | 0.00 | 0.05 |
| b | 0.20 | 0.15 | 0.25 |
| D | 5 |  |  |
| E | 5 | 3.50 | 3.75 |
| D2 | 3.65 | 3.50 | 3.75 |
| E2 | 3.65 |  | 0.477 |
| e | 0.40 | 0.277 | 0.50 |
| L1 | 0.377 | 0.30 |  |
| L2 | 0.40 | 0.20 |  |
| K |  |  |  |

Figure 22. MLPQ40-EP 5x5 recommended footprint


## 14 Revision history

Table 17. Document revision history

| Date | Revision | Changes |
| :---: | :---: | :--- |
| 19-Aug-2013 | 1 | Initial release. |
| 05-Mar-2014 | 2 | Modified footnote1 in Table 6: Switching characteristics <br> Added footnote 2 in Table 6: Switching characteristics and footnote 5 <br> in Table 5: Electrical characteristics. |
| 22-Apr-2014 | 3 | Document status promoted from preliminary data to production data. |

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