

# UNISONIC TECHNOLOGIES CO., LTD

# **TEA2025D**

### LINEAR INTEGRATED CIRCUIT

# STEREO AUDIO AMPLIFIER

#### ■ DESCRIPTION

The UTC **TEA2025D** is a monolithic integrated circuit that intended for use as dual or bridge power audio amplifier portable radio cassette players.

#### ■ FEATURES

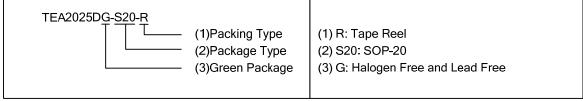
- \* Dual or bridge connection modes.
- \* Few external components.
- \* Supply voltage down to 3V.
- \* High channel separation.
- \* Very low switch on\off noise.
- \* Max gain of 45dB with adjust external resistor.
- \* Soft clipping.
- \* Thermal protection.
- \* 3V<V<sub>CC</sub><15V
- \* P=2\*1W, $V_{CC}$ =6V, $R_L$ =4 $\Omega$
- \* P=2\*2.3W, $V_{CC}$ =9V, $R_L$ =4 $\Omega$
- \* P=2\*0.1W, $V_{CC}$ =3V, $R_L$ =4 $\Omega$

#### ORDERING INFORMATION

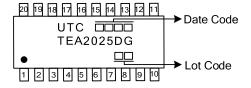
Order Number

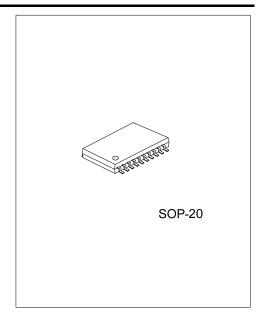
TEA2025DG-S20-R	SOP-20	Tape Reel
TEA202EDC 620 D		

Package



#### MARKING

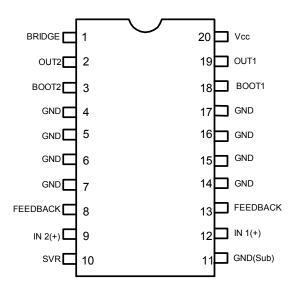




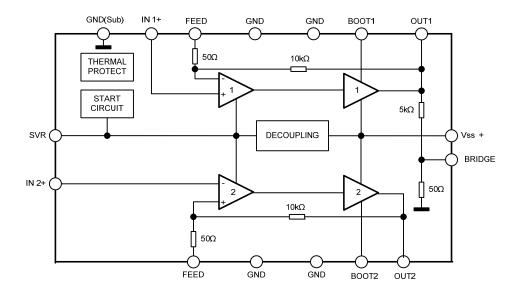
Packing

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#### ■ PIN CONFIGURATION



#### ■ BLOCK DIAGRAM



#### ABSOLUTE MAXIMUM RATINGS

PARAMETER	SYMBOL	RATINGS	UNIT
Supply Voltage	$V_{SS}$	15	V
Output Peak Current	I <sub>O(PEAK)</sub>	1.5	Α
Junction Temperature	T <sub>J</sub>	150	°C
Storage Temperature	T <sub>STG</sub>	150	°C

Note Absolute maximum ratings are those values beyond which the device could be permanently damaged. Absolute maximum ratings are stress ratings only and functional device operation is not implied.

#### ■ THERMAL DATA

PARAMETER	SYMBOL	RATINGS	UNIT
Thermal Resistance Junction-case	$\theta_{JC}$	15	°C/W
Thermal Resistance Junction-ambient	$\theta_{JA}$	65	°C/W

Note: The  $\theta_{JA}$  is measured with 4 cm<sup>2</sup> copper area heatsink.

## ■ ELECTRICAL CHARACTERISTICS (Ta=25°C, V<sub>CC</sub>=9V, Stereo unless otherwise specified)

PARAMETER		SYMBOL	TEST CONDITIONS		MIN	TYP	MAX	UNIT
Supply Voltage		$V_{SS}$			3		12	V
Quiescent Current		IQ				35	50	mA
Quiescent Output Voltage	)	$V_{OUT}$			4.5		V	
Voltage Gain		Gu	Stereo		43	45	47	dB
		G <sub>V</sub>	Bridge		49	51	53	dB
Voltage Gain Difference		$\Delta G_{V(DIFF)}$					±1	dB
Input Impedance	_	Rı				30		kΩ
	$R_L=4\Omega$			V <sub>SS</sub> =3V		0.1		
	$R_L$ =32 $\Omega$			V SS-3V		0.02		
	$R_L=4\Omega$				0.7	1		
	$R_L=8\Omega$		Stereo 8 (per channel)  Bridge	V <sub>SS</sub> =6V		0.6		W
	$R_L=16\Omega$	P <sub>OUT</sub>				0.25		
	$R_L=32\Omega$					0.13		
Output Power (d=10%)	$R_L=4\Omega$			V <sub>SS</sub> =9V	1.7	2.3		
Output Fower (u=1070)	$R_L=8\Omega$					1.3		
	$R_L=8\Omega$			V <sub>SS</sub> =12V		2.4		
	$R_L=16\Omega$			V <sub>SS</sub> =3V		0.18		
	$R_L$ =32 $\Omega$					0.06		
	$R_L=4\Omega$	Pout		V <sub>SS</sub> =6V		2.8		
	$R_L=8\Omega$					1.5		
	$R_L=8\Omega$			V <sub>SS</sub> =9V		4.7		
Distortion	Stereo	d	$V_{SS}=9V; R_{I}=4\Omega$			0.3	1.5	%
Bridge		u	VSS-3V, INL-452			0.5		%
Supply Voltage Rejection		SVR	f=100Hz,V <sub>R</sub> =0.5V,Rg=0		40	46		dB
Input Noise Voltage		eN <sub>(IN)</sub>	Rg=0			1.5	3	mV
		CIA(IN)	Rg=104Ω			3	6	mV
Cross-Talk		CT	f=1KHz, Rg=10KΩ		40	52		dB

#### ■ APPLICATION INFORMATION

#### **Input Capacitor**

Input capacitor is PNP type allowing source to be referenced to ground.

In this way no input coupling capacitor is required. However, a series capacitor  $(0.22\mu F)$  to the input side can be useful in case of noise due to variable resistor contact.

#### **Bootstrap**

The bootstrap connection allows to increase the output swing.

The suggested value for the bootstrap capacitors ( $100\mu F$ ) avoids a reduction of the output signal also at low frequencies and low supply voltage.

#### Voltage Gain Adjust

#### STEREO MODE

The voltage gain is determined by on-chip resistors R1 and R2 together with the external RfC1 series connected between pin 8(13) and ground. The frequency response is given approximated by:

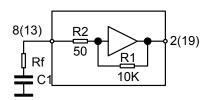
$$\frac{\text{Vout}}{\text{Vin}} = \frac{\text{R1}}{\text{Rf} + \text{R2} + \frac{1}{\text{JWC1}}}$$

With Rf=0,C1=100µF, the gain results 46dB with pole at f=32Hz

The purpose of Rfl is to reduce the gain. It is recommended to not reduce it under 36dB.

#### **BRIDGE MODE**

Figure 7

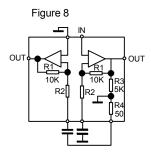


The bridge configuration is realized very easily thanks to an internal voltage divider which provides (at pin 1)the CH1 output signal after reduction. It is enough to connect pin8(inverting input of CH 2 )with a capacitor to pin 1 and to connect to ground the pin 9.

The total gain of the bridge is given by:

$$\frac{V_{OUT}}{V_{IN}} = \frac{R1}{Rf + R2 + \frac{1}{JWC1}} (1 + \frac{R3}{R4} \frac{R1}{R2 + R4 + \frac{1}{JWC1}})$$

and with the suggested values (C1=C2=100 $\mu$ F,Rf=0)means: Gv=52dB



With first pole at f=32Hz

#### ■ APPLICATION INFORMATION(Cont.)

#### **Output Capacitors.**

The low cut off frequency due to output capacitor depending on the load is given by:

$$F_L = \frac{1}{2 \pi COUT *RL}$$

With  $C_{OUT}$  470 $\mu$ F and  $R_L$ =4 ohm it means  $F_L$ =80Hz.

#### Stability

A good layout is recommended in order to avoid oscillations. In general, the designer must pay attention on the following points:

- -Short wires of components and short connections.
- -No ground loops.
- -Bypass of supply voltage with capacitors as nearest as possible to the supply I.C. pin. The low value (polyester) capacitors must have good temperature and frequency characteristics.
- -No sockets

The heatsink can have a smaller factor of safety compared with that of a conventional circuit. There is no device damage in the case of excessive junction temperature: all that happens is that  $P_{OUT}$  (and therefore  $P_D$ ) and Id are reduced.

#### **APPLICATION SUGGESTION**

The recommended values of the components are those shown on stereo application circuit of Fig.2 different values can be used, the following table can help the designer.

Component	Recommended	Purpose	Larger Than	Smaller Than
C1,C2	0.22µF	Input DC Decoupling in Case of Slider Contact Noise of Variable Resistor		
С3	100μF	Ripple Rejection		Degradation of SVR, Increase of THD at Low Frequency and Low Voltage
C4,C5	100μF	Bootstrap		
C6,C7	470μF	Output DC Decoupling		Increase of Low Frequency Cut-off
C8,C9	0.15µF	Frequency Stability		Danger of Oscillations
C10,C11	100μF	Inverting Input DC Decoupling		Increase of Low Frequency Cut-off

## ■ TYPICAL APPLICATION CIRCUIT

Figure 1:Bridge Application (Powerdip)

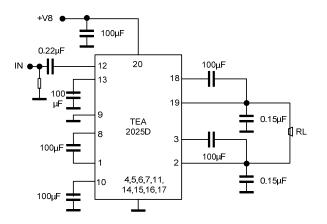
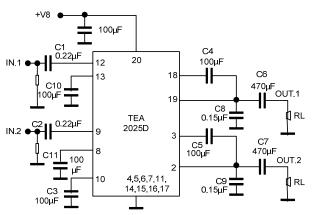
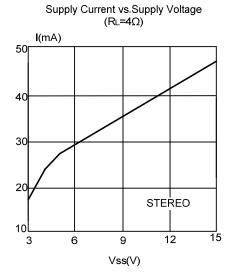
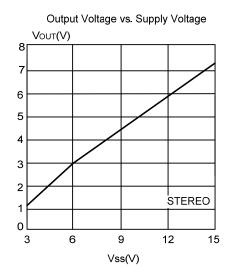


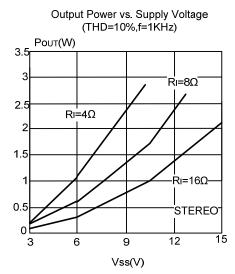
Figure 2:Stereo Application (Powerdip)

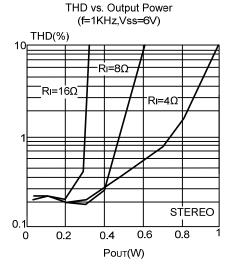


#### TYPICAL CHARACTERISTICS









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