## FEATURES:

- 500V, 50 Amp Capability at $110^{\circ} \mathrm{C}$
- Fully Isolated Bridge
- Ultra Low Thermal Resistance
- Integral Free Wheeling Fast Recovery Epitaxial Diode (FRED)
- Self-Contained, Smart Lowside/Highside Drive Circuitry and Isolated Supply
- Adjustable Deadtime
- Capable of Switching Frequencies to 20 KHz
- Isolated Case Allows Direct Heat Sinking; On Board Temp Sensor
- Bolt-down Design Allows Superior Heat Dissipation



## DESCRIPTION:

The MSK4351 is a 50 Amp, 3 Phase Isolated Bridge Smart Power Motor Drive Hybrid with a 500 volt rating. The output switches are Insulated Gate Bipolar Transistors (IGBT's) tailored for high switching speeds. The free-wheeling diodes are Fast Recovery Epitaxial Diodes (FRED's) to provide matched current capabilities with the IGBT's and are specified with excellent reverse recovery times at high current ratings. The bridge is optically isolated from the control circuitry. This new smart power motor drive hybrid is compatible with 5 V CMOS logic levels. The internal circuitry prevents simultaneous turn-on of the in-line half bridge transistors with adjustable deadtime to prevent shoot-through. Undervoltage lockout shuts down the bridge when the supply voltage gets to a point of incomplete turn-on of the output switches. The isolated internal high-side power supply derived from the +15 volt supply completely eliminates the need for 3 floating independent power supplies for the high-side drive.

## EQUIVALENT SCHEMATIC



## TYPICAL APPLICATIONS



3 PHASE SIX STEP DC BRUSHLESS MOTOR DRIVE OR 3 PHASE SINUSOIDAL INDUCTION MOTOR DRIVE

## PIN-OUT INFORMATION

| 1 | +15V | 13 | $\overline{\text { RESET }}$ | 25 | RKELVIN + | 37 | BV- |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 2 | GND | 14 | R/C | 26 | RKELVIN- | 38 | BV- |
| 3 | AHI | 15 | +15V | 27 | RSENSE- | 39 | $B \varnothing$ |
| 4 | ALO | 16 | N/C | 28 | RSENSE- | 40 | BØ |
| 5 | BHI | 17 | OSCOUT | 29 | RSENSE + | 41 | BV + |
| 6 | BLO | 18 | GND | 30 | RSENSE + | 42 | BV + |
| 7 | +15V | 19 | N/C | 31 | CV- | 43 | AV- |
| 8 | GND | 20 | N/C | 32 | CV- | 44 | AV- |
| 9 | CHI | 21 | N/C | 33 | CØ | 45 | AØ |
| 10 | CLO | 22 | N/C | 34 | CØ | 46 | AØ |
| 11 | +15V | 23 | N/C | 35 | $\mathrm{CV}+$ | 47 | AV + |
| 12 | GND | 24 | TEMP SENSE | 36 | CV + | 48 | $A V+$ |


| V+ | High Voltage Supply . ${ }^{(8)}$ | 500V |
| :---: | :---: | :---: |
| +15V | Logic Supply | 18 V |
| Iout | Continuous Output Current | 50A |
| IPK | Peak Output Current ( 1 pulse, $10 \mu \mathrm{Sec}$ ) | 100A |
| $\theta \mathrm{Jc}$ | Thermal Resistance <br> (IGBT - Junction to Case) | $0.38^{\circ} \mathrm{C} / \mathrm{W}$ |
|  | Thermal Resistance (Diode - Junction to Case) | $0.45{ }^{\circ} \mathrm{C} / \mathrm{W}$ |

TsT Storage Temperature Range ${ }^{(9)}-65^{\circ} \mathrm{C}$ to $+150^{\circ} \mathrm{C}$
TLD Lead Temperature Range(10 Seconds) . . . $300^{\circ} \mathrm{C}$
Tc Case Operating Temperature
MSK4351 . . . . . . . . . . . $-40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$
MSK4351H . . . . . . . . . $-55^{\circ} \mathrm{C}$ to $+125^{\circ} \mathrm{C}$
TJ Junction Temperature . . . . . . . . . . $+150^{\circ} \mathrm{C}$

## ELECTRICAL SPECIFICATIONS

| All Ratings: $\mathrm{Tc}=+25^{\circ} \mathrm{C}$ Unless Otherwise Specified |  |  |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Parameters Test Conditions |  | Group A Subgroup (5) | MSK4351H |  |  | $\begin{array}{cc} & \text { MSK4351 } \\ \text { Min. } & \text { Typ. } \\ \text { Ty }\end{array}$ |  |  | UNITS |
| OUTPUT CHARACTERISTICS |  |  |  |  |  |  |  |  |  |
| VC-E On Voltage (Each IGBT) (6) | $I C=45 A$ | 1 | - | 1.95 | 2.5 | - | 1.95 | 2.6 | V |
|  |  | 2 | - | 2.0 | 2.4 | - | - | - | V |
|  |  | 3 | - | 2.3 | 2.75 | - | - | - | V |
| Instantaneous Forward Voltage (6) (FRED Flyback Diode) | $I D=45 A$ | 1 | - | 1.4 | 1.8 | - | 1.4 | 1.9 | V |
|  |  | 2 | - | 1.3 | 1.5 | - | - | - | V |
|  |  | 3 | - | 1.7 | 2.2 | - | - | - | V |
| Turn On Switching Energy (1) | $\mathrm{Vt}=270 \mathrm{~V}, 45 \mathrm{Amps} @ 125^{\circ} \mathrm{C}$ | - | - | - | 4.1 | - | - | 4.5 | mJ |
| Turn Off Switching Energy (1) | $\mathrm{L}=100 \mu \mathrm{H}, \mathrm{R}=0.2 \Omega$ | - | - | - | 2.3 | - | - | 2.5 | mJ |
| Reverse Recovery Time (1) | $\mathrm{ID}=45 \mathrm{~A}, \mathrm{di} / \mathrm{dt}=400 \mathrm{~A} / \mathrm{uS}, \mathrm{Vr}=350 \mathrm{~V}$ | - | - | - | 180 | - | - | 180 | nS |
| Leakage Current (Each IGBT/Diode) | $\mathrm{V}+=500 \mathrm{~V}$ | 1 | - | 25 | 150 | - | 25 | 150 | uA |
|  | $\mathrm{V}+=400 \mathrm{~V}$ | 2 | - | 1.6 | 2.5 | - | - | - | mA |
|  | $\mathrm{V}+=500 \mathrm{~V}$ (1) | 3 | - | 15 | 150 | - | - | - | uA |
| BIAS SUPPLY CHARACTERISTICS |  |  |  |  |  |  |  |  |  |
| Quiescent Bias Current | $\mathrm{Vcc}=15 \mathrm{~V}$ | 1 | - | 230 | 300 | - | 230 | 330 | mA |
|  |  | 2 | - | 170 | 250 | - | - | - | mA |
|  |  | 3 | - | 230 | 300 | - | - | - | mA |
| Supply Voltage (1) |  | - | 14.75 | 15.00 | 15.25 | 14.75 | 15.00 | 15.25 | V |
| INPUT SIGNALS CHARACTERISTICS (1) |  |  |  |  |  |  |  |  |  |
| Positive Trigger Threshold Voltage |  | 1,2,3 | 3.65 | - | - | 3.65 | - | - | V |
| Negative Trigger Threshold Voltage |  | 1,2,3 | - | - | 0.8 | - | - | 0.8 | V |
| SWITCHING CHARACTERISTICS (1) |  |  |  |  |  |  |  |  |  |
| Upper Drive: Deadtime R/C = 10K/47pF V $\quad$ = 270V, IC = 45A |  |  |  |  |  |  |  |  |  |
| Turn-On Propagation Delay |  | 4 | - | 6.0 | 6.2 | - | 6.0 | 6.3 | $\mu \mathrm{S}$ |
| Turn-Off Propagation Delay |  | 4 | - | 1.1 | 1.2 | - | 1.1 | 1.3 | $\mu \mathrm{S}$ |
| Turn-On |  | 4 | - | 50 | 60 | - | 50 | 60 | nS |
| Turn-Off |  | 4 | - | 80 | 90 | - | 80 | 90 | nS |
| Lower Drive: Deadtime R/C = 10K/47pF V $+=270 \mathrm{~V}$, IC $=45 \mathrm{~A}$ |  |  |  |  |  |  |  |  |  |
| Turn-On Propagation Delay |  | 4 | - | 2.3 | 2.7 | - | 2.3 | 2.7 | $\mu \mathrm{S}$ |
| Turn-Off Propagation Delay |  | 4 | - | 3.1 | 3.4 | - | 3.1 | 3.4 | $\mu \mathrm{S}$ |
| Turn-On |  | 4 | - | 50 | 70 | - | 50 | 70 | nS |
| Turn-Off |  | 4 | - | 66 | 75 | - | 66 | 75 | nS |
| TEMPERATURE SENSOR |  |  |  |  |  |  |  |  |  |
| \|litial Accuracy (1) | $\mathrm{TC}=25^{\circ} \mathrm{C}$ | 1 | - | $\pm 2.0$ | $\pm 5.0$ | - | $\pm 3.0$ | $\pm 6.0$ | ${ }^{\circ} \mathrm{C}$ |
|  | TMIN $\leq$ TC $\leq$ TMAX | 2,3 | - | $\pm 5.0$ | $\pm 8.0$ | - | - | - | ${ }^{\circ} \mathrm{C}$ |
| SENSE RESISTOR |  |  |  |  |  |  |  |  |  |
| Resistance | @2 Amps | 1 | 2.91 | 3.0 | 3.09 | 2.9 | 3 | 3.1 | $\mathrm{m} \Omega$ |
|  |  | 2 | 2.95 | 3.05 | 3.14 | - | - | - | $\mathrm{m} \Omega$ |
|  |  | 3 | 2.92 | 3.01 | 3.10 | - | - | - | $\mathrm{m} \Omega$ |

## NOTES:

(1) Guaranteed by design but not tested. Typical parameters are representative of actual device performance but are for reference only.
(2) Industrial devices shall be tested to subgroups 1 and 4 unless otherwise specified.
(3) Military grade devices ("H" suffix) shall be $100 \%$ tested to subgroups $1,2,3$ and 4 .
(4) Subgroups 5 and 6 testing available upon request.
(5) Subgroup 1,4 TA $=\mathrm{TC}=+25^{\circ} \mathrm{C}$
$2,5 \mathrm{TA}=\mathrm{TC}=+125^{\circ} \mathrm{C}$
$3,6 \quad \mathrm{TA}=\mathrm{TC}=-55^{\circ} \mathrm{C}$
(6) Measurements are made by forcing current through one pin and measuring on the other for determining thermal dissipation on the IGBT/diode. When measuring on the pins very close to the package, add approximately 0.3 V to the limits to account for pin resistance.
(7) Continuous operation at or above absolute maximum ratings may adversely effect the device performance and/or life cycle.
(8) When applying power to the device, apply the low voltage followed by the high voltage or alternatively, apply both at the same time. Do not apply high voltage without low voltage present.
(9) Internal solder reflow temperature is $180^{\circ} \mathrm{C}$, do not exceed.

## MSK4351 PIN DESCRIPTION

+15 V - is the low voltage supply for all the internal logic and isolated supplies which provide power to the gate drivers. A $0.1 \mu \mathrm{~F}$ ceramic capacitor in parallel with a $22 \mu \mathrm{~F}$ tantalum capacitor is recommended for bypassing the low voltage supply to GND.

GND - is the low voltage supply return for the +15 V . All bypassing of the +15 V should return here. Since the output section of the hybrid is completely isolated, there are no restrictions for potential differences between this GND and any hivoltage returns, up to 500 V .
$\mathbf{A H I}, \mathrm{BHI}, \mathrm{CHI}$ - are the logic inputs for controlling the switching of the corresponding hi-side bridge outputs. A logic high will turn on the corresponding hi-side output. The input levels are 5 V CMOS compatible. If one of these inputs are active at the same time as the corresponding low-side bridge outputs, neither output will be allowed to turn on until one of the inputs is switched low. There will be a deadtime inserted before the corresponding bridge output is switched in all cases. This prevents simultaneous conduction of the output, shorting high voltage supply and destroying the bridge.

ALO,BLO,CLO - are the logic inputs for controlling the switching of the corresponding low-side bridge outputs. A logic high will turn on the corresponding low-side output. The input levels are 5 V CMOS compatible. If one of these inputs are active at the same time as the corresponding hi-side bridge outputs, neither output will be allowed to turn on until one of the inputs is switched low. There will be a deadtime inserted before the corresponding bridge output is switched in all cases. This prevents simultaneous conduction of the output, shorting the high voltage supply and destroying the bridge.
$\overline{\text { RESET }}$ - is an active low logic input for causing all switching to cease. The input level is 5 V CMOS compatible. Upon releasing $\overline{\text { RESET, the outputs will resume after the dead time. }}$

R/C - is the input pin for setting the deadtime of the bridge. Connecting a resistor between this input and OSC OUT, and a capacitor to ground will create the time for an internal oscillator.

OSC OUT - is a pin that brings the deadtime oscillator out to be connected through the timing resistor to R/C. This is not an output to be used externally, but just for the timing circuit.
$\mathbf{A V}+, \mathbf{B V}+, \mathbf{C V}+-$ are pins for connecting the tops of each half bridge to the high voltage supply. Each pin must be connected individually, as there is no internal connection across the three half bridges. Proper power supply bypassing must be connected to these pins and the V - pins as close to the hybrid as possible for proper filtering.

AV-,BV-,CV- - are pins for connecting the bottoms of each half bridge to the high voltage supply return. Each pin must be connected individually, as there is no internal connection across the three half bridges. Proper power supply bypassing must be connected to these pins and the $\mathrm{V}+$ pins as close to the hybrid as possible for proper filtering.
$\mathbf{A} \varnothing, \mathbf{B} \varnothing, \mathbf{C}$ - - are the pins connecting the 3 phase bridge switch outputs.

TEMP SENSE - is a pin for measuring the output of a temperature sensor IC. The case temperature is depicted as a voltage corresponding to $10 \mathrm{mV} /{ }^{\circ} \mathrm{C}$ with 0 volts equating to absolute zero, $0^{\circ} \mathrm{K}$ or $-273^{\circ} \mathrm{C}$. This temperature sensor IC is pulled up to +15 V through a $10 \mathrm{~K} \Omega$ resistor internally. Buffering of the output will be necessary if it needs to be connected to a low impedance load.

RSENSE + - is the pin for connecting to the internal sense resistor. It has a value of 0.003 ohms, 20 watts. AV-,BV- and CV- should connect to this point for sensing the current at the bottom of the bridge.

RSENSE- - is the pin for connecting the internal sense resistor to the high voltage return.

RKELVIN + - is the pin for connecting to the sense resistor + KELVIN connection. This is on the same side of the resistor as RSENSE + .

RKELVIN- - is the pin for connecting to the sense resistor -KELVIN connection. This is on the same side of the resistor as RSENSE-.

## DEADTIME SELECTION

The amount of deadtime required is based on the propagation delay of the input to actual completion of switching of the output transistors. Not taking all this into account can possibly allow the opposite transistor in a half bridge to turn on before the active transistor can turn off. Excessive current will flow through the half bridge because this creates a momentary short across the power supply.

Once all these factors are taken into account, the deadtime can be determined. Allow sufficient safety factor for changes in components over temperature, and variations from system to system in production.

Deadtime is exactly 8 R/C clock periods. Use the formula:
Max. Clock $=8 /$ Min. Deadtime


For clock operation below 1 MHz :
Clock Frequency $=\quad \frac{0.95}{\operatorname{Cosc} \times \text { Rosc }}$
For clock operation above 1 MHz :
Clock Frequency $=\frac{0.95}{\operatorname{Cosc}(\text { Rosc }+30)+3 \times 10^{-8}}$

As an alternative, the R/C pin can be driven directly with an HCMOS compatible clock up to 24 MHz .


TYPICAL PERFORMANCE CURVES




The MSK4351 is designed to be used with a +270 volt high voltage bus, +15 volt low power bus, and +5 volt logic signals. Proper derating should be applied when designing the MSK4351 into a system. High frequency layout techniques with ground planes on a printed circuit board is the only method that should be used for circuit construction. This will prevent pulse jitter caused by excessive noise pickup on the current sense signal or the error amp signal.

Ground planes for the low power circuitry and high power circuitry should be kept separate. The two sections of the hybrid are completely isolated, and can float relative to each other without referencing one to the other. An RC filter will filter out the current spikes and keep the detected noise for that circuit down to a minimum.

The logic signals coming from the typical motor controller IC are set up for driving N channel low side and P channel high side switches directly, and are usually 15 volt levels. Provision should be made for getting 5 volt logic signals to the MSK4351 of the correct assertion levels. Typically, the low side signals out of the controller are high active and the high side are low active. Inverters are shown in the system schematic for the high side controller output.


ESD TRIANGLE INDICATES PIN 1
WEIGHT $=200$ GRAMS TYPICAL

## ORDERING INFORMATION



NOTE: THE CENTRIFUGE LEVEL FOR THE CLASS H DEVICE IS 1000G'S

REVISION HISTORY

| REV | STATUS | DATE | DESCRIPTION |
| :---: | :--- | :---: | :--- |
| $R$ | Released | $06 / 14$ | Clarify tested parameters, add new note for solder reflow and clarify mechanical specifications. |

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