## LA71076SM

## Monolithic Linear IC

For VHS VTR Video Signal Processor
(Y/CIA single-chip)

## Overview

The LA71076SM is a video signal processing system IC that handles VHS VCR format. In addition to conventional video signal processing circuits, it integrates normal audio processing and record/playback FM-EQ circuits on a chip. The LA71076SM is combined with a CCD to create a 2-chip-1-package semiconductor device. Chip internal trimming is used to make this IC adjustment free, further the automatically adjustable comb filter makes the IC fully adjustment free. These features significantly reduces the number of external components, thus streamlining the design of the signal processing board and reducing in the production cost.

## Functions

- Fully adjustment free.
- Built-in normal audio processing.
- Built-in and record/playback FM-EQ circuits.
- Built-in NTSC delay-line (LC89961 equivalent).


## Specifications

Maximum Ratings at $\mathrm{Ta}=25^{\circ} \mathrm{C}$

| Parameter | Symbol | Conditions | Ratings | Unit |
| :--- | :---: | :---: | :---: | :---: |
| Maximum supply voltage | $\mathrm{V}_{\mathrm{CC}} \max$ |  | 7.0 | V |
| Allowable power dissipation | $\mathrm{Pd} \max$ | $\mathrm{Ta} \leq 75^{\circ} \mathrm{C} *$ | 1040 | mW |
| Operating temperature | Topg |  | -10 to +75 | ${ }^{\circ} \mathrm{C}$ |
| Storage temperature | Tstg |  | -40 to +150 | ${ }^{\circ} \mathrm{C}$ |

*: When mounted on a $114.3 \times 76.1 \times 1.6 \mathrm{~mm}^{3}$,glass epoxy board.

Recommended Operating Conditions at $\mathrm{Ta}=25^{\circ} \mathrm{C}$

| Parameter | Symbol | Conditions | Ratings | Unit |
| :--- | :---: | :---: | :---: | :---: |
| Supply voltage | $V_{C C}$ |  | 5.0 | V |
| Allowable operating voltage range | $\mathrm{V}_{\mathrm{CC}}$ op |  | 4.8 to 5.2 | V |

Any and all SANYO Semiconductor products described or contained herein do not have specifications that can handle applications that require extremely high levels of reliability, such as life-support systems, aircraft's control systems, or other applications whose failure can be reasonably expected to result in serious physical and/or material damage. Consult with your SANYO Semiconductor representative nearest you before using any SANYO Semiconductor products described or contained herein in such applications.
SANYO Semiconductor assumes no responsibility for equipment failures that result from using products at values that exceed, even momentarily, rated values (such as maximum ratings, operating condition ranges, or other parameters) listed in products specifications of any and all SANYO Semiconductor products described or contained herein.

Electrical Characteristics at $\mathrm{Ta}=25^{\circ} \mathrm{C}$
Recording mode Head Amplifier (T87:5.0V T11:5.0V)

| Parameter | Symbol | In | Out | Conditions | T13 | T15 | Ratings |  |  | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  |  |  |  |  | min | typ | max |  |
| Rec AGC Amp output level | $\begin{aligned} & \mathrm{V}_{\mathrm{R}} \mathrm{SP} \\ & \mathrm{v}_{\mathrm{R}} \mathrm{EP} \end{aligned}$ | T66A | $\begin{aligned} & \text { T83A } \\ & \text { T89A } \end{aligned}$ | Output level when $\mathrm{V}_{\mathrm{IN}}=300 \mathrm{mVp}-\mathrm{p}, \mathrm{f}=4 \mathrm{MHz}$ Enter by applying DC 3.5 V or more to pin 66. |  | $\begin{array}{r} 0 \\ 5.0 \end{array}$ | $\begin{aligned} & 127 \\ & 104 \end{aligned}$ | $\begin{aligned} & 135 \\ & 111 \end{aligned}$ | $\begin{aligned} & 143 \\ & 119 \end{aligned}$ | mVp-p |
| Difference of gain between mode | $\Delta \mathrm{GVR}$ |  |  | $\mathrm{V}_{\mathrm{R}} \mathrm{SP} / \mathrm{V}_{\mathrm{R}} \mathrm{EP}$ |  |  | 1.40 | 1.7 | 2.00 | dB |
| REC AGC AMP control characteristics 1 | $\begin{aligned} & \Delta \mathrm{V}_{\mathrm{AGC}} 1-\mathrm{SP} \\ & \Delta \mathrm{~V}_{\mathrm{AGC}}{ }^{1-\mathrm{EP}} \end{aligned}$ | T66A | $\begin{aligned} & \text { T83A } \\ & \text { T89A } \end{aligned}$ | Output level/ $\mathrm{V}_{\mathrm{R}} \mathrm{SP}$, EP with $\mathrm{f}=4 \mathrm{MHz}$ and $\mathrm{V}_{\mathrm{IN}}=700 \mathrm{mVp}-\mathrm{p}$ |  | $\begin{array}{r} 0 \\ 5.0 \end{array}$ |  | 0.5 | 1.0 | dB |
| REC AGC AMP control characteristics 2 | $\begin{aligned} & \Delta \mathrm{V}_{\mathrm{AGC}}{ }^{2-S P} \\ & \Delta \mathrm{~V}_{\mathrm{AGC}}{ }^{2-E P} \end{aligned}$ | T66A | $\begin{aligned} & \text { T83A } \\ & \text { T89A } \end{aligned}$ | Output level/VRSP, EP with $\mathrm{f}=4 \mathrm{MHz}$ and $\mathrm{V}_{\mathrm{IN}}=100 \mathrm{mVp}$-p. |  | $\begin{array}{r} 0 \\ 5.0 \end{array}$ | -1.0 | -0.5 |  | dB |
| REC AGC AMP <br> frequency characteristics | $\begin{aligned} & \Delta V_{F} R S \\ & \Delta V_{F} R E \end{aligned}$ | T66A | $\begin{aligned} & \text { T83A } \\ & \text { T89A } \end{aligned}$ | The output ratio when f is $1 \mathrm{M}, 7 \mathrm{MHz}$ as $\mathrm{V}_{\mathrm{IN}}=300 \mathrm{mVp}$-p. <br> $7 \mathrm{MHz} / 1 \mathrm{MHz}$ (Note 1) |  | $\begin{array}{r} 0 \\ 5.0 \end{array}$ | -1.0 | 0.0 | +1.0 | dB |
| REC AGC AMP <br> second harmonic distortion | $\Delta \mathrm{V}_{\mathrm{HD}} \mathrm{RS}$ <br> $\Delta \mathrm{V}_{\mathrm{HD}} \mathrm{RE}$ | T13A | $\begin{aligned} & \text { T83A } \\ & \text { T89A } \end{aligned}$ | The ratio of the 8 Mz (second component) and 4 Mz (first component) of the output with $V_{I N}=300 \mathrm{mVp}-\mathrm{p}$ and $\mathrm{f}=4 \mathrm{MHz}$ |  | $\begin{array}{r} 0 \\ 5.0 \end{array}$ |  | -45 | -40 | dB |
| REC AGC AMP maximum output level | $\Delta \mathrm{V}_{\mathrm{HD}} \mathrm{RS}$ $\Delta V_{\text {HD }}$ RE | T66A | $\begin{aligned} & \text { T83A } \\ & \text { T89A } \end{aligned}$ | The output level at which the second distortion with $f=4 \mathrm{MHz}$ is -35 dB . |  | $\begin{array}{r} 0 \\ 5.0 \end{array}$ | 20 | 22 |  | mVp-p |
| REC AGC AMP attenuate volume of mute | $\Delta \mathrm{V}_{\mathrm{M}} \mathrm{RS}$ <br> $\Delta \mathrm{V}_{\mathrm{M}} \mathrm{RE}$ | T66A | $\begin{aligned} & \text { T83A } \\ & \text { T89A } \end{aligned}$ | Output level/VRSP, EP with $\mathrm{V}_{\mathrm{IN}}=300 \mathrm{mVp}-\mathrm{p}$ and $\mathrm{f}=4 \mathrm{MHz}$ |  | $\begin{array}{r} 0 \\ 5.0 \end{array}$ |  | -45 | -40 | dB |
| REC AGC AMP <br> mixed modulation relative level | $\Delta \mathrm{VCYS}$ <br> $\triangle$ VCYE | T66A | $\begin{aligned} & \text { T83A } \\ & \text { T89A } \end{aligned}$ | $\begin{aligned} & \text { Vin1 }=300 \mathrm{mVp}-\mathrm{p}, \mathrm{f}=4 \mathrm{MHz} \\ & \text { Vin2 }=300 \mathrm{Vp}-\mathrm{p}, \mathrm{f}=629 \mathrm{kHz}(4 \mathrm{M} \pm 629 \mathrm{k}) / 4 \mathrm{M} \end{aligned}$ <br> ratio of output |  | $\begin{array}{r} 0 \\ 5.0 \end{array}$ |  | -45 | -40 | dB |

Note1: Apply DC of about 1.6 V to AGC detection filter terminal (Pin92), and fix the AGC amplifier gain.
Use a resistor with a tolerance of $\pm 1.0 \%$ between Pin 93and GND.
PB mode Head Amplifier ( $\mathrm{T} 87=5.0 \mathrm{~V}$ T11 = 0V)

| Parameter |  | Symbol | In | Out | Conditions | T13 | T15 | Ratings |  |  | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | min |  |  |  |  |  | typ | max |  |
| Voltage SP-H | CH1 |  | Gvp1 | T82A | T74 | $\mathrm{V}_{\mathrm{IN}}=38 \mathrm{mVp}-\mathrm{p}, \mathrm{f}=1 \mathrm{MHz}$ | 0 | 0 | 56.0 | 59.0 | 62.0 | dB |
| gain SP-H | CH 2 | Gvp2 | T85A | 5.0 |  |  | 0 | 56.0 | 59.0 | 62.0 |  |  |
| EP-L | CH3 | Gvp3 | T88A | 0 |  |  | 5.0 | 56.0 | 59.0 | 62.0 |  |  |
| EP-H | CH4 | Gvp4 | T91A | 5.0 |  |  | 5.0 | 56.0 | 59.0 | 62.0 |  |  |
| Difference of voltage gain 1 |  | $\Delta \mathrm{Gvp1}$ |  |  | Gvp1 - Gvp2 |  |  | -1 | 0 | +1 | dB |  |
| Difference of voltage gain 2 |  | $\Delta \mathrm{Gvp} 2$ |  |  | Gvp3 - Gvp4 |  |  | -1 | 0 | +1 | dB |  |
| Difference of gain between mode |  | $\Delta \mathrm{Gvp} 3$ |  |  | Gvp3 - Gvp1 |  |  | -1 | 0 | +1 | dB |  |
| Input calculation noise voltage | $\begin{aligned} & \mathrm{CH} 1 \\ & \mathrm{CH} 2 \\ & \mathrm{CH} 3 \\ & \mathrm{CH} 4 \end{aligned}$ | $\mathrm{V}_{\mathrm{N}}$ IN1 <br> $\mathrm{V}_{\mathrm{N}} \mathrm{IN} 2$ <br> $\mathrm{V}_{\mathrm{N}} \mathrm{IN} 3$ <br> $\mathrm{V}_{\mathrm{N}}$ IN4 | $\begin{aligned} & \text { T82A } \\ & \text { T85A } \\ & \text { T88A } \\ & \text { T91A } \end{aligned}$ | T74 | The ratio of the output which has passed the 1.1 MHz LPF and the output without input under the same input conditions as the voltage gain. | $\begin{array}{r} 0 \\ 5.0 \\ 0 \\ 5.0 \end{array}$ | $\begin{array}{r} 0 \\ 0 \\ 5.0 \\ 5.0 \end{array}$ |  | 0.7 | 1.0 | $\mu \mathrm{Vrms}$ |  |
| Frequency characteristics | $\begin{aligned} & \mathrm{CH} 1 \\ & \mathrm{CH} 2 \\ & \mathrm{CH} 3 \\ & \mathrm{CH} 4 \end{aligned}$ | $\Delta \mathrm{Vfp} 1$ <br> $\Delta \mathrm{Vfp} 2$ <br> $\Delta \mathrm{Vfp} 3$ <br> $\Delta \mathrm{Vfp} 4$ | $\begin{aligned} & \text { T82A } \\ & \text { T85A } \\ & \text { T88A } \\ & \text { T91A } \end{aligned}$ | T74 | The ratio of the $V_{I N}=38 \mathrm{mVp}-\mathrm{p}$, $\mathrm{f}=7 \mathrm{MHz}$ output and Gvp1, 2, 3, 4. | $\begin{array}{r} 0 \\ 5.0 \\ 0 \\ 5.0 \end{array}$ | $\begin{array}{r} 0 \\ 0 \\ 5.0 \\ 5.0 \end{array}$ | -2.5 | 0 |  | dB |  |
| Secondary harmonic distortion | $\begin{aligned} & \mathrm{CH} 1 \\ & \mathrm{CH} 2 \\ & \mathrm{CH} 3 \\ & \mathrm{CH} 4 \end{aligned}$ | $\begin{aligned} & \mathrm{V}_{\mathrm{H}} \mathrm{DP} 1 \\ & \mathrm{v}_{\mathrm{H}} \mathrm{DP} 2 \\ & \mathrm{v}_{\mathrm{H}} \mathrm{DP} 3 \\ & \mathrm{v}_{\mathrm{H}} \mathrm{DP} 4 \end{aligned}$ | $\begin{aligned} & \text { T82A } \\ & \text { T85A } \\ & \text { T88A } \\ & \text { T91A } \end{aligned}$ | T74 | The ratio of 8 MHz (second component) and 4 MHz (first component) of output with $\mathrm{V}_{\mathrm{IN}}=38 \mathrm{mVp}-\mathrm{p}$ and $\mathrm{f}=4 \mathrm{MHz}$. | $\begin{array}{r} 0 \\ 5.0 \\ 0 \\ 5.0 \end{array}$ | $\begin{array}{r} 0 \\ 0 \\ 5.0 \\ 5.0 \end{array}$ |  | -40 | -35 | dB |  |
| Maximum output level | $\begin{aligned} & \mathrm{CH} 1 \\ & \mathrm{CH} 2 \\ & \mathrm{CH} 3 \\ & \mathrm{CH} 4 \end{aligned}$ | VOMP1 <br> $V_{\mathrm{O}} \mathrm{MP} 2$ <br> $V_{0}$ MP3 <br> $V_{0}$ MP4 | $\begin{aligned} & \text { T82A } \\ & \text { T85A } \\ & \text { T88A } \\ & \text { T91A } \end{aligned}$ | T74 | The output level at which the ratio of 3 MHz (third component) and 1MHz (first component) of the output with $f=1 \mathrm{MHz}$. | $\begin{array}{r} 0 \\ 5.0 \\ 0 \\ 5.0 \end{array}$ | $\begin{array}{r} 0 \\ 0 \\ 5.0 \\ 5.0 \end{array}$ | 1.0 | 1.2 |  | Vp-p |  |

LA71076SM
Continued from preceding page.

| Parameter | Symbol | In | Out | Conditions | T13 | T15 | Ratings |  |  | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  |  |  |  |  | min | typ | max |  |
| Cross talk SP1 CH1 | $\mathrm{v}_{\mathrm{C}} \mathrm{R} 1$ | $\begin{aligned} & \hline \text { T85A } \\ & \text { T88A } \\ & \text { T91A } \\ & \hline \end{aligned}$ | T74 | The ratio of output of $\mathrm{V}_{\mathrm{IN}}=38 \mathrm{mVp}-\mathrm{p}, \mathrm{f}=4 \mathrm{MHz}$ and Gvp1. | 0 | 0 |  | -40 | -35 | dB |
| Cross talk SP2 CH2 | $\mathrm{V}_{\mathrm{C}} \mathrm{R} 2$ | $\begin{aligned} & \text { T82A } \\ & \text { T88A } \\ & \text { T91A } \end{aligned}$ | T74 | The ratio of output of $\mathrm{V}_{\mathrm{IN}}=38 \mathrm{mVp}-\mathrm{p}, \mathrm{f}=4 \mathrm{MHz}$ and Gvp2. | 5.0 | 0 |  | -40 | -35 | dB |
| Cross talk EP1 CH3 | $\mathrm{V}_{\mathrm{C}} \mathrm{R} 3$ | $\begin{aligned} & \hline \text { T82A } \\ & \text { T85A } \\ & \text { T88A } \\ & \hline \end{aligned}$ | T74 | The ratio of output of $\mathrm{V}_{\mathrm{IN}}=38 \mathrm{mVp}-\mathrm{p}, \mathrm{f}=4 \mathrm{MHz}$ and Gvp3. | 0 | 5.0 |  | -40 | -35 | dB |
| Cross talk EP2 CH4 | $\mathrm{V}_{\mathrm{C}} \mathrm{R} 4$ | $\begin{aligned} & \text { T82A } \\ & \text { T85A } \\ & \text { T91A } \end{aligned}$ | T74 | The ratio of output of $\mathrm{V}_{\mathrm{IN}}=38 \mathrm{mVp}-\mathrm{p}, \mathrm{f}=4 \mathrm{MHz}$ and Gvp4. | 5.0 | 5.0 |  | -40 | -35 | dB |
| Output DC offset | $\begin{aligned} & \Delta \mathrm{v}_{\mathrm{O}} \mathrm{DC1} \\ & \Delta \mathrm{v}_{\mathrm{O}} \mathrm{DC} 2 \\ & \Delta \mathrm{v}_{\mathrm{O}} \mathrm{DC} 3 \\ & \Delta \mathrm{v}_{\mathrm{O}} \mathrm{DC4} \\ & \Delta \mathrm{v}_{\mathrm{O}} \mathrm{DC5} \\ & \Delta \mathrm{v}_{\mathrm{O}} \mathrm{DC6} \end{aligned}$ |  | T74 | $\begin{gathered} \mathrm{CH}-\mathrm{CH}_{2} \\ \mathrm{CH}-\mathrm{CH}_{4} \\ \mathrm{CH} 1- \\ \mathrm{CH} 3 \\ \mathrm{CH} 2- \\ \mathrm{CH} 4 \\ \mathrm{CH}- \\ \mathrm{CH} 4 \\ \mathrm{CH} 2- \\ \mathrm{CH} 3 \end{gathered}$ | $\begin{array}{r} 0 \\ 5.0 \\ 0 \\ 5.0 \\ 0 \\ 0 \\ 0.0 \\ 5.0 \\ 5.0 \\ 0 \\ 5.0 \\ 0 \\ 5.0 \end{array}$ | $\begin{array}{r} 0 \\ 0 \\ 0.0 \\ 5.0 \\ 0 \\ 5.0 \\ 0 \\ 5.0 \\ 0 \\ 5.0 \\ 0 \\ 5.0 \end{array}$ | -150 | 0 | +150 | mV |
| Envelope detection output terminal voltage | $\mathrm{V}_{\text {ENV }}$ |  | T93 | T93DC when no input is provided. | $\begin{array}{r} 0 \\ 5.0 \\ 0 \\ 5.0 \end{array}$ | $\begin{array}{r} 0 \\ 0 \\ 5.0 \\ 5.0 \end{array}$ | 0 | 0.8 | 1.3 | V |
| Envelope detection output terminal voltage SP1 | $\mathrm{V}_{\text {ENV }} \mathrm{SP} 1$ | T82A | T93 | When input $\mathrm{f}=4 \mathrm{MHz}$, T93 DC as becomes 175mVp-p, for T74 output level. | 0 | 0 | 2.0 | 2.5 | 3.0 | V |
| Envelope detection output terminal voltage SP2 | $\mathrm{V}_{\mathrm{ENV}} \mathrm{SP} 2$ | T82A | T93 | When input $\mathrm{f}=4 \mathrm{MHz}$, T 93 DC as becomes 400 mVp -p, for T74 output level. | 0 | 0 | 4.0 | 4.5 | 5.0 | V |
| Envelope detection output terminal voltage EP1 | VENVEP1 | T89A | T93 | When input $\mathrm{f}=4 \mathrm{MHz}$, T93 DC as becomes 125 mV p-p, for T 74 output level. | 0 | 5.0 | 2.0 | 2.5 | 3.0 | V |
| Envelope detection output terminal voltage EP2 | $\mathrm{V}_{\text {ENV }}$ EP2 | T89A | T93 | When input $\mathrm{f}=4 \mathrm{MHz}$, T93 DC as becomes 300 mV p-p, for T74 output level. | 0 | 5.0 | 4.0 | 4.5 | 5.0 | V |
| Comparator output voltage 1 | $\mathrm{V}_{\text {COMP }}{ }^{1}$ | T82A | T94 | T94 DC voltage when $V_{I N}=38 \mathrm{mVp}-\mathrm{p}, \mathrm{f}=4 \mathrm{MHz}$. | 0 | 0 |  | 0.4 | 0.7 | V |
| Comparator output voltage 2 | $\mathrm{V}_{\text {COMP }}{ }^{2}$ | T89A | T94 | T94 DC voltage when $V_{I N}=38 \mathrm{mVp}-\mathrm{p}, \mathrm{f}=4 \mathrm{MHz}$. | 5.0 | 0 | 4.5 | 4.8 |  | V |

## LA71076SM

REC Mode Y

| Parameter | Symbol | In | Out | Conditions | Ratings |  |  | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  |  |  | min | typ | max |  |
| Current dissipation (REC) | ${ }^{1} \mathrm{CCR}$ |  |  | Measure the sum of currents flowing into pins $21,54,56,57$, 75, 87. | 135 | 160 | 185 | mA |
| EE output level 2 | $V_{E E}{ }^{2}$ | T42A | T35 | With $\mathrm{V}_{\mathrm{IN}}$ being a 1 V p-p video signal (PAL), measure the output level on T35. | 2.00 | 2.10 | 2.20 | Vp-p |
| AGC characteristics 1 | AGC1 | T42A | T35 | With $\mathrm{V}_{\text {IN }}$ being a 2.0 Vp -p video signal, measure the ratio of the output level on T 35 and $\mathrm{V}_{\mathrm{EE}}{ }^{1}$. | 0 | 0.6 | 1.2 | dB |
| AGC characteristics 2 | AGC2 | T42A | T35 | With $\mathrm{V}_{\mathrm{IN}}$ being a 0.5 Vp -p video signal, measure the ratio of the output level on T 35 and $\mathrm{V}_{\mathrm{EE}} 1$. | -1.2 | -0.2 | 0.0 | dB |
| AGC characteristics 3 | AGC3 | T42A | T35 | With $\mathrm{V}_{\mathrm{IN}}$ being a 700 mVp -p luminance, 600 mVp -p sync, measure the sync level on T35. | 550 | 650 | 750 | mVp-p |
| AGC characteristics 4 | AGC4 | T42A | T35 | With $\mathrm{V}_{\mathrm{IN}}$ being a 700 mVp -p luminance, 150 mVp -p sync, measure the sync level on T35. | 330 | 380 | 430 | mVp-p |
| Sync separator output level | $\mathrm{V}_{\text {SYR }}$ | T42A | T34 | With $\mathrm{V}_{\mathrm{IN}}$ being a 1.0 V p-p video signal, measure the output pulse wave height on T34. | 4.0 | 4.2 | 4.4 | Vp-p |
| Sync separator output pulse width | PWSYR | T42A | T34 | With $\mathrm{V}_{\text {IN }}$ being a 1.0 Vp -p video signal, measure the output pulse width on T34. | 4.1 | 4.4 | 4.7 | $\mu \mathrm{S}$ |
| Sync separator threshold level | THSYR | T42A | T34 | Gradually reduce the input level, and measure the input level at which the output pulse width is $1 \mu$ s or more wider than PWSYR. |  | -20 | -15 | dB |
| H-Sync output level | VHSYR | T42A | T33 | With $\mathrm{V}_{\mathrm{IN}}$ being a 1.0 V p-p video signal, measure the output pulse wave height on T33. | 4.0 | 4.2 | 4.4 | Vp-p |
| H-Sync output pulse width | $\mathrm{PWH}_{\text {SYR }}$ | T42A | T33 | $\mathrm{V}_{\text {IN }}=1.0 \mathrm{Vp}$-p video signal, Measure the output pulse on T33. | 4.4 | 4.7 | 5.0 | $\mu \mathrm{s}$ |
| Sync tip level <br> Pedestal level <br> White level | LVOR | T42A | T35 | With $\mathrm{V}_{\mathrm{IN}}$ being a 1.0 Vp -p video signal, measure the sync tip and pedestal and white level on T35 video output, and take these as LSYN LPED LWHT, respectively. |  |  |  |  |
| Quasi-V insertion level | $\Delta \mathrm{VDR}$ | T42A | T35 | Measure the T35 DC voltage with 4.0V applied to T31, and take this as LVDR, and calculate the difference from LSYN measured above. $\Delta \mathrm{WHR}=\mathrm{LSYN}-\mathrm{LVDR}$ | -100 | 0 | 100 | mV |
| Quasi-V insertion level | $\Delta \mathrm{HDR}$ | T42A | T35 | Measure the T35 DC voltage with 3.0 V applied T31, and takes this as LHDP, and calculates the difference from LSYN measured above. $\Delta H D R=\text { LPED-LHDR }$ | -500 | -400 | -300 | mV |
| White insertion level | $\Delta \mathrm{WHR}$ | T42A | T35 | Measure the T35 DC voltage with 2.0 V applied to T31. and take this as LWHP, and calculate the difference from LWHT measured above. $\Delta \mathrm{WHR}=\mathrm{LWHT}-\mathrm{LWHR}$ | 500 | 600 | 700 | mV |
| Edge insertion level | $\Delta \mathrm{EGR}$ | T42A | T35 | Measure the T35 DC voltage with 1.2 V applied toT31.and take this as LWHP, and calculate the difference from LPED measured above. $\Delta \mathrm{WHR}=\text { LPED-LEGR }$ | -500 | -400 | -300 | mV |
| Y LPF frequency characteristics (1) | Y ${ }_{\text {LPF }}{ }^{1}$ | T42A | T26 | With $\mathrm{V}_{\mathrm{IN}}$ being a standard multi-burst signal ( 1 Vp -p), measure the 1 MHz response to a 500 kHz signal on T26. | -0.6 | -0.1 | 0.4 | dB |
| Y LPF frequency characteristics (2) | Y LPF $^{2}$ | T42A | T26 | With $\mathrm{V}_{\mathrm{IN}}$ being a standard multi-burst signal ( 1 Vp -p), measure the 2 MHz response to a 500 kHz signal on T26. | -1.3 | -0.3 | 0.7 | dB |
| Y LPF frequency characteristics (3) | Y ${ }_{\text {LPF }}{ }^{3}$ | T42A | T26 | With $\mathrm{V}_{\mathrm{IN}}$ being a standard multi-burst signal ( $1 \mathrm{Vp}-\mathrm{p}$ ), measure the 3 MHz response to a 500 kHz signal on T 26 . | -8.0 | -6.0 | -4.0 | dB |
| Y LPF frequency characteristics (5) | YLPF5 | T42A | T26 | With $\mathrm{V}_{\mathrm{IN}}$ being a standard multi-burst signal ( 1 Vp -p), measure the 3.58 MHz response to a 500 kHz signal on T 26 |  |  | -25 | dB |

Continued on next page

## LA71076SM

| Parameter | Symbol | In | Out | Conditions | Ratings |  |  | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  |  |  | min | typ | max |  |
| REC-FM modulator output level | $\mathrm{V}_{\mathrm{FM}}$ |  | T66 | Measure the T66 output level with no input. | 220 | 300 | 360 | mVp-p |
| Carrier frequency | $\mathrm{FFM}^{2}$ |  | T66 | Measure the output frequency on T66 with no input. | 3.36 | 3.46 | 3.56 | MHz |
| REC-FM output second distortion | HMOD |  | T66 | Measure the second distortion with the above state. |  | -40 | -35 | dB |
| Deviation 2 | DEV2 | T42A | T66 | With $\mathrm{V}_{\mathrm{IN}}$ being a $100 \%$ white 1 Vp -p signal, measure the deviation on $T 66$. | 0.95 | 1.00 | 1.05 | MHz |
| FM modulator linearity | LMOD | T25 | T66 | Assume that f 2.85 is the output frequency when 3.85 VDC is applied to T25. $L M O D=\frac{f 2.85-(f 3.1+f 2.6) / 2}{f 3.1-f 2.6} \times 100$ | -2 | 0 | 2 | \% |
| 1/2 fH carrier shift | CS | T13 | T66 | The output frequency change | 6.5 | 8.2 | 9.5 | kHz |
| Emphasis gain | GEMP | T25A | T23 | With $\mathrm{V}_{\mathrm{IN}}$ being a 500 mVp -p 10 kHz sine wave, measure the ratio of the levels on T25A and T23. | -0.75 | -0.25 | 0.25 | dB |
| Detail enhancer characteristics (1) | $\mathrm{GENH}^{1}$ | T25A | T23 | With $\mathrm{V}_{\mathrm{IN}}$ being a 158 mVp -p 2 MHz sine wave, measure the ratio of the levels on T25A and T23, and calculate the difference from GEMP. | 1.5 | 2.0 | 2.5 | dB |
| Detail enhancer characteristics (2) | $\mathrm{GENH}^{2}$ | T25A | T23 | With $\mathrm{V}_{\mathrm{IN}}$ being a $50 \mathrm{mVp}-\mathrm{p} 2 \mathrm{MHz}$ sine wave, measure the ratio of the levels on T25A and T23, and calculate the difference from GEMP. | 3.5 | 4.5 | 5.5 | dB |
| Detail enhancer characteristics (3) | $\mathrm{GENH}^{3}$ | T25A | T23 | With $\mathrm{V}_{\mathrm{IN}}$ being a $15.8 \mathrm{mVp}-\mathrm{p} 2 \mathrm{MHz}$ sine wave, measure the ratio of the levels on T25A and T23, and calculate the difference from GEMP. | 4.3 | 5.8 | 7.3 | dB |
| Nonlinear emphasis characteristics (1) | $\mathrm{G}_{\mathrm{NL}} \mathrm{EMP1}$ | T25A | T23 | With $\mathrm{V}_{\text {IN }}$ being a $500 \mathrm{mVp}-\mathrm{p} 2 \mathrm{MHz}$ signal measure the ratio of the levels on T25A and T23, and calculate the difference from GEMP. | -3.0 | -2.0 | -1.0 | dB |
| Nonlinear emphasis characteristics (2) | $\mathrm{G}_{\mathrm{NL}} \mathrm{EMP} 2$ | T25A | T23 | With $\mathrm{V}_{\mathrm{IN}}$ being a 158 mVp -p 2 MHz signal measure the ratio of the levels on T25A and T23, and calculate the difference from GEMP. | 2.5 | 4.0 | 5.5 | dB |
| Nonlinear emphasis characteristics (3) | $\mathrm{G}_{\mathrm{NL}} \mathrm{EMP3}$ | T25A | T23 | With $\mathrm{V}_{\mathrm{IN}}$ being a $50 \mathrm{mVp}-\mathrm{p} 2 \mathrm{MHz}$ signal measure the ratio of the levels on T25A and T23, and calculate the difference from GEMP. | 5.0 | 6.5 | 8.0 | dB |
| Main linear emphasis characteristics (1) | $\mathrm{G}_{\mathrm{M}} \mathrm{E}^{1}$ | T25A | T23 | With $\mathrm{V}_{\mathrm{IN}}$ being a $50 \mathrm{mVp}-\mathrm{p} 500 \mathrm{kHz}$ sine wave, measure the ratio of the levels on T25 and T23, and calculate the difference from GEMP. | 10.0 | 11.5 | 12.0 | dB |
| Main linear emphasis characteristics (2) | $\mathrm{G}_{\mathrm{M}} \mathrm{E}^{2}$ | T25A | T23 | With $\mathrm{V}_{\text {IN }}$ being a $50 \mathrm{mVp}-\mathrm{p} 2 \mathrm{MHz}$ signal measure the ratio of the levels on T25A and T23, and calculate the difference from GEMP. | 17.0 | 18.5 | 20.0 | dB |
| White clipping level | ${ }^{\text {L W }}$ | T42A | T23 | With $\mathrm{V}_{\mathrm{IN}}$ being a $1.0 \mathrm{Vp}-\mathrm{p} 100 \%$ white video signal, measure the white clipping level on T23. | 180 | 190 | 200 | \% |
| Dark clipping level | ${ }^{\text {DC }}$ | T42A | T23 | With $\mathrm{V}_{\mathrm{IN}}$ being a $1.0 \mathrm{Vp}-\mathrm{p} 100 \%$ white video signal, measure the dark clipping level on T23. | -52 | -50 | -47 | \% |

## REC Mode EQ

| Parameter | Symbol | In | Out | Conditions | Ratings |  |  | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  |  |  | min | typ | max |  |
| REC EQ <br> characteristics 1 | $\mathrm{GREQ}^{1}$ | T36A | T66 | With $\mathrm{V}_{\mathrm{IN}}$ being a CW $2 \mathrm{MHz}, 400 \mathrm{mVp}$-p signal, measure the input/output response. | -5.7 | -4.5 | -3.3 | dB |
| REC EQ <br> characteristics 2 | $\mathrm{GREQ}^{2}$ | T36A | T66 | With $\mathrm{V}_{\mathrm{IN}}$ being a $\mathrm{CW} 4 \mathrm{MHz}, 400 \mathrm{mVp}$-p signal, measure the input/output response. | -4.0 | -2.7 | -1.4 | dB |
| REC EQ <br> characteristics 3 | $\mathrm{GREQ}^{3}$ | T36A | T66 | With $\mathrm{V}_{\mathrm{IN}}$ being a $\mathrm{CW} 750 \mathrm{kHz}, 400 \mathrm{mVp}$-p signal, measure the input/output response. |  |  | -20 | dB |
| REC EQ <br> 2'nd distortion | $\mathrm{H}_{\text {REQ }}$ | T36A | T66 | Measure the second harmonic in the above conditions. |  | -40 | -15 | dB |

LA71076SM
PB Mode Y

| Parameter | Symbol | In | Out | Conditions | Ratings |  |  | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  |  |  | min | typ | max |  |
| Current dissipation (PB) | ${ }^{1} \mathrm{CCP}$ |  |  | Measure the sum of the currents flowing into pins $21,54,56$, 57, 75, and 87. | 160 | 170 | 180 | mA |
| Dropout compensation Period 1H for one horizontal synchronization period | TDOC | $\begin{aligned} & \text { T74A } \\ & \text { T25A } \end{aligned}$ | T35 | T74A: 4MHz $300 \mathrm{mVp}-\mathrm{p}$ sine wave +3VDC <br> T25A: 0.5Vp-p video signal <br> The I/O response 5 H after the T74A input is set to 0 . | 10.0 | 13.0 | 15.0 | H |
| DOC characteristics | GDOC | $\begin{aligned} & \text { T74A } \\ & \text { T25A } \end{aligned}$ | T35 | T74A: 4MHz 300mVp-p sine wave +3VDC <br> T25A: 0.5Vp-p video signal <br> The time from the instant when the T74A input level is set to 0 to the time point when the T35 output is restored. | -1.5 | 0 | 1.5 | dB |
| PB Y level | V-YOUT | T74A | T35 | DEV $=1.0 \mathrm{MHz}$ PB Y level when input FM signal is input. | 2.00 | 2.10 | 2.20 | Vp-p |
| Self R/P, PB-Y level | R/P-OUT | T74A | T35 | Self R/P-Y, PB-Y level | 1.93 | 2.10 | 2.27 | Vp-p |
| FM demodulator linearity | LDEM | T74A | T26 | $\begin{aligned} L E D M & =\frac{V D E M 4-(\text { VDEM } 3+V D E M 5) / 2}{V D E M 5-V D E M 3} \times 100 \\ * V D E M 4 & =\text { DC: } \mathrm{T} 26(\text { Input } 4 \mathrm{MHz}, 300 \mathrm{mVp}-\mathrm{p}) \end{aligned}$ | -3.5 | 0 | +3.5 | \% |
| Carrier leakage | CL | T74 | T26 | Measure the ratio of the 4MHz component on T26 and SDEM. |  |  | -35 | dB |
| PB YNR characteristics | PYNR | T25A | T35 | Measure the ratio of 32 fH component and 32.5 fH . | -8.5 | -7.5 | -6.5 | dB |
| Nonlinear de-emphasis characteristics (1) *Serial-control | $\mathrm{G}_{\mathrm{NL}}$ DE1 | T25A | T35 | With $\mathrm{V}_{\mathrm{IN}}$ being a $50 \%$ white video $\mathrm{f}=2 \mathrm{MHz}, 158 \mathrm{mVp}$-p sine wave, measure the I/O response. | -3.5 | -2.5 | -1.5 | dB |
| Nonlinear de-emphasis characteristics (2) | $\mathrm{G}_{\text {NL }} \mathrm{DE} 2$ | T25A | T35 | $\mathrm{f}=2 \mathrm{MHz}, 50 \mathrm{mVp}-\mathrm{p}$ | -6.0 | -4.5 | -3.0 | dB |
| Double noisecanceller characteristics (1) | $\mathrm{GWNC}^{1}$ | T25A | T35 | $\mathrm{f}=1.4 \mathrm{MHz}, 158 \mathrm{mVp}-\mathrm{p}$ | 3.5 | -2.5 | -1.5 | dB |
| Double noisecanceller characteristics (2) | $\mathrm{G}_{W N}{ }^{2}$ | T25A | T35 | $\mathrm{f}=1.4 \mathrm{MHz}, 50 \mathrm{mVp}-\mathrm{p}$ | -12 | -10 | -8 | dB |
| Double noisecanceller characteristics (3) | $\mathrm{G}_{\mathrm{WNC}}{ }^{3}$ | T25A | T35 | $\mathrm{f}=1.4 \mathrm{MHz}, 15.8 \mathrm{mVp}-\mathrm{p}$ | -15 | -13 | -11 | dB |
| Sync separator output level | VSYP | T25A | T34 | With $\mathrm{V}_{\text {IN }}$ being a 0.5 Vp -p video signal, measure the output pulse wave height on $T 34$. | 4.0 | 4.2 | 4.4 | Vp-p |
| Sync separator output pulse width | PWSYP | T25A | T34 | With $\mathrm{V}_{\mathrm{IN}}$ being a $0.5 \mathrm{~V} \mathrm{p}-\mathrm{p}$ video signal, measure the output pulse width on T34. | 4.1 | 4.4 | 4.7 | $\mu \mathrm{s}$ |
| H-Sync output level | $\mathrm{VH}_{\text {SYP }}$ | T25A | T33 | With $\mathrm{V}_{\text {IN }}$ being a 0.5 V p-p video signal, measure the output pulse wave height on T33. | 4.0 | 4.2 | 4.4 | Vp-p |
| H-Sync output pulse width | $\mathrm{PWH}_{\text {SYP }}$ | T41 | T33 | With $\mathrm{V}_{\text {IN }}$ being a 0.5 V p-p video signal, measure the output pulse width on T33. | 4.4 | 4.7 | 5.0 | $\mu \mathrm{s}$ |
| Sync tip level <br> Pedestal level <br> White level | LVOR | T25A | T35 | With $\mathrm{V}_{\text {IN }}$ being a $100 \%$ white $0.5 \mathrm{Vp}-\mathrm{p}$ signal, measure the sync tip and pedestal and white levels on T35 video output, and take these as LSYN, LPED, and LWHT, respectively. |  |  |  |  |
| Quasi-V insertion level | $\Delta \mathrm{VDP}$ | T25A | T35 | Measure the T35 DC voltage with 4.0 V applied to T31, and take this as LVDP, and calculate the difference from LSYN measured above. $\triangle \mathrm{VDP}=\mathrm{LSYN}-\mathrm{LVDP}$ | -100 | 0 | 100 | mV |
| Quasi-H insertion level | $\Delta \mathrm{HDP}$ | T25A | T35 | Measure the T35 DC voltage with 3.0 V applied to T31, and take this as LHDP, and calculate the difference from LPED measured above. $\triangle H D P=\text { LPED-LHDP }$ | -500 | -400 | -300 | mV |
| White insertion level | $\Delta \mathrm{WHP}$ | T25A | T35 | Measure the T35 DC voltage with 2.0 V applied to T31, and take this as LWHP, and calculate the difference from LWHT measured above. <br> $\Delta$ WHP = LWHT-LWHP | 500 | 600 | 700 | mV |
| Edge insertion level | $\Delta \mathrm{EGP}$ | T25A | T35 | Measure the T35 DC voltage with 1.2 V applied to T31, and take this as LEGP, and calculate the difference from LPED measured above. <br> $\Delta$ WHP = LPED-LEGP | -500 | -400 | -300 | mV |
| 4V regulator | $V_{\text {REG }}$ |  | T65 | Measure the T65 DC level. | 3.9 | 4.0 | 4.3 | V |

## LA71076SM

PB mode EQ T72 = 5V

| Parameter | Symbol | In | Out | Conditions | Ratings |  |  | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  |  |  | min | typ | max |  |
| PB EQ characteristics 1 *Serial-control | GpEQ1 | T28A | T23 | With $\mathrm{V}_{\mathrm{IN}}$ being a $\mathrm{CW} 4 \mathrm{MHz}, 300 \mathrm{mVp}$-p signal, measure the input / output response. | 1.5 | 3.0 | 4.5 | dB |
| PB EQ <br> 2'nd distortion | HpEQ | T28A | T23 | Measure the second harmonic in the above condition. |  | -40 | -30 | dB |
| PB EQ characteristics 2 | GpEQ2 | T28A | T23 | With $\mathrm{V}_{\mathrm{IN}}$ being a $\mathrm{CW} 4 \mathrm{MHz}, 300 \mathrm{mVp}$-p signal, measure the input / output response |  |  | -30 | Vp-p |
| PB EQ Trap characteristics | $f_{p} E Q$ | T28A | T23 | With $\mathrm{V}_{\mathrm{IN}}$ being a 300 mVp -p signal, measure high-band trap frequency and gain. <br> (Using network analyzer) |  |  | -25 | dB |

## PB mode S discrimination

| Parameter | Symbol | In | Out | Conditions | Ratings |  |  | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  |  |  | min | typ | max |  |
| Output voltage when normal VHS | $\mathrm{VN}_{\text {DET }} \mathrm{N}$ | T74 | T32 | $\mathrm{V}_{\mathrm{IN}}=300 \mathrm{mVp}-\mathrm{pf}: 4 \mathrm{MHz}+3 \mathrm{VDC}$ | 0 | 0.2 | 0.5 | V |
| Output voltage when S-VHS | $V S_{\text {DET }} \mathrm{S}$ | T74 | T32 | $\mathrm{V}_{\text {IN }}=300 \mathrm{mVp}-\mathrm{p} \mathrm{f:} 6 \mathrm{MHz}+3 \mathrm{VDC}$ | 4.4 | 4.7 | 5.0 | V |
| S-discrimination input level | $\mathrm{VS}_{\text {DET }}$ | T74 | T32 | Input level at which no mis-discrmination occurs while changing T74 input level. | 50 |  |  | mVp-p |
| Normal-discrimination input level | VN ${ }_{\text {DET }}$ | T74 | T32 | Input level at which no mis-discrmination occurs while changing T74 input level. | 50 |  |  | mVp-p |
| Normal $\rightarrow$ S-discrimination threshold level | $\mathrm{FS}_{\text {DET }} \mathrm{NS}$ | T74 | T32 | The frequency at which T 32 becomes H when the sine wave input to T 74 is increased from $\mathrm{f}=4 \mathrm{MHz}$. | 5.5 | 6.0 | 6.5 | MHz |
| $\mathrm{S} \rightarrow$ normal discrimination threshold level | $\mathrm{FS}_{\text {DET }} \mathrm{SN}$ | T74 | T32 | The frequency at which T 32 becomes L when the sine wave input to T 74 is increased from $\mathrm{f}=5 \mathrm{MHz}$. | 5.0 | 5.5 | 6.0 | MHz |

LA71076SM
REC mode chroma T16 $=5 \mathrm{~V}, \mathrm{~T} 11=5 \mathrm{~V}, \mathrm{~T} 55=0 \mathrm{~V}, \mathrm{~T} 59=0 \mathrm{~V}, \mathrm{~T} 72=5 \mathrm{~V}$

| Parameter | Symbol | In | Out | Conditions |  | Ratings |  |  | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  |  |  |  | min | typ | max |  |
| REC chroma low frequency conversion output level | VOR-66 | T42A | T66 | With $\mathrm{V}_{\text {IN }}$ being the standard color bar signal (1Vp-p), measure the burst level on T66. |  | 600 | 750 | 900 | mVp-p |
| Burst emphasis | GBE | T42A | T66 | With $\mathrm{V}_{\text {IN }}$ being the standard color bar signal ( $1 \mathrm{Vp}-\mathrm{p}$ ) calculate the ratio of the T 66 burst levels for $\mathrm{SP} / E \mathrm{P}(\mathrm{T} 59=0 \mathrm{~V} / 5 \mathrm{~V})$ and $\mathrm{LP}(\mathrm{T} 59=2.5 \mathrm{~V})$ modes. |  | 5.5 | 6.0 | 6.5 | dB |
| VXO oscillation level | $\mathrm{V}_{\mathrm{VXO}}$-RN | T42A | T62 | With $\mathrm{V}_{\text {IN }}$ being the standard color bar signal (1Vp-p), measure the T62 output amplitude with an FET probe. |  | 290 | 430 | 690 | mVp-p |
| REC ACC <br> characteristics (1) | $\mathrm{ACC}_{\mathrm{R}}{ }^{1}$ | T42A | T66 | With $\mathrm{V}_{\text {IN }}$ being the standard color bar signal (1Vp-p), increase only the chroma signal level by +6 dB , measure the T66 burst level, and calculate its ratio with VOR-66. |  |  | +0.2 | +0.5 | dB |
| REC ACC <br> characteristics (2) | $\mathrm{ACC}_{\mathrm{R}}{ }^{2}$ | T42A | T66 | With $\mathrm{V}_{\text {IN }}$ being the standard color bar signal (1Vp-p), decrease only the chroma signal level by -6dB, measure the T66 burst level, and calculate its ratio with VOR-66. |  | -0.5 | -0.1 |  | dB |
| REC ACC killer-on input level | $\mathrm{V}_{\text {ACCK }}{ }^{-O N}$ | T42A | T66 | With $\mathrm{V}_{\text {IN }}$ being the standard color bar signal (1Vp-p), decrease the chroma signal and measure the input burst level at which T66 output ceases. Calculate the ratio of this value with the standard input level. |  |  | -26 |  | dB |
| REC ACC killer-on output level | $\mathrm{V}_{0} \mathrm{ACCK}$ | T42A | T66 | Measure the T66 output level with a spectrum analyzer in the killer state of the above item and calculate its ratio with VOR-66. |  |  | -60 | -50 | dB |
| REC ACC killer restored input level | $\mathrm{V}_{\text {ACCK }}{ }^{-O F F}$ | T42A | T66 | From the killer state of the above item gradually increase the input chroma level and measure the input burst level at which T66 output reappears. Calculate its ratio with the standard input level. |  |  | -20 |  | dB |
| REC APC pull-in range (1) | $\Delta^{\text {f }} \mathrm{APC}^{1}$ | T42A | T66 | Input a signal consisting of a 3.5795 MHz 300 mVp -p CW added to a $50 \%$ white signal. After confirming that a signal is output from T66, increase the CW frequency until T66 output ceases. Now slowly reduce the CW frequency, and determine f1 frequency at which T66 output reappears.$\Delta f^{A P C}{ }^{1}=\mathrm{f} 1-3579545(\mathrm{~Hz})$ |  | 350 |  |  | Hz |
| REC APC pull-in range (2) | ${ }^{\text {f }} \mathrm{APC}^{2}$ | T42A | T66 | As in the previous item, decrease the CW frequency until T66 output ceases. <br> Now slowly increase the CW frequency and determine f2 frequency at which T66 output reappears. $\Delta \mathrm{f}_{\mathrm{APC}}{ }^{2}=\mathrm{f} 2-3579545(\mathrm{~Hz})$ |  |  |  | -350 | Hz |
| REC AFC pull-in range (1) | $\Delta^{\text {AFC }}{ }^{1}$ | T42A | T60 | Input a $300 \mathrm{mVp}-\mathrm{p}, 15.7 \mathrm{kHz}$, $5 \mu \mathrm{~s}$ width pulse train (negative polarity). <br> After increasing the pulse train frequency until the T60 wave form is disrupted, decrease the frequency to determine the f1 pulse train frequency at which the T 60 wave form returns to normal. $\Delta \mathrm{f} \text { AFC1 }=\mathrm{f} 1-15.734(\mathrm{kHz})$ |  | 1.0 |  |  | kHz |
| REC AFC pull-in range (2) | ${ }^{\text {f }} \mathrm{AFC}^{2}$ | T42A | T60 | As in the previous item, decrease the pulse train frequency until the T60 wave form is disrupted, then increase the frequency to determine the f 2 pulse train frequency at which the T60 wave form returns to normal.$\Delta \mathrm{f}_{\mathrm{AFC}}{ }^{2}=\mathrm{f} 2-15.734(\mathrm{kHz})$ |  |  |  | -1.0 | kHz |
| The ratio of the REC chroma level and FM modulator output level | C/FM2 | T42 | T66 | The ratio of $100 \%$ chroma's level which was converted to low band and FM modulator output level. $\mathrm{T} 72=0 \mathrm{~V}$ | $\mathrm{T} 68=0 \mathrm{~V}$ |  | 8.0 |  | dB |
|  |  |  |  |  | $\mathrm{T} 68=2.5 \mathrm{~V}$ |  | 6.7 |  | dB |
|  |  |  |  |  | $\mathrm{T} 68=5 \mathrm{~V}$ |  | 5.3 |  | dB |

PB mode chroma T16 $=0 \mathrm{~V}, \mathrm{~T} 11=0 \mathrm{~V}, \mathrm{~T} 55=0 \mathrm{~V}, \mathrm{~T} 59=0 \mathrm{~V}, \mathrm{SW} 25=2$

| Parameter | Symbol | In | Out | Conditions | Ratings |  |  | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  |  |  | min | typ | max |  |
| PB chroma video output level | NVop-35 | $\begin{aligned} & \text { T74A } \\ & \text { T25A } \end{aligned}$ | T35 | Apply a mixture of the SP mode chroma signal (SP mode, burst $100 \mathrm{mVp}-\mathrm{p}$ ) that was obtained by converting the T74A NTSC chroma noise test signal to the low band and the $4 \mathrm{MHz}, 300 \mathrm{mVp}-\mathrm{p}$ sine wave to T 74 through 3 V bias. Apply the $50 \%$ white signal ( $321.5 \mathrm{mVp}-\mathrm{p}$ ) from T25A. Measure the T35A burst level. | 510 | 600 | 690 | mVp-p |
| PB chroma pin 72 output level | Vop-72 | $\begin{aligned} & \text { T74A } \\ & \text { T25A } \end{aligned}$ | T72 | Measure the burst level with the same conditions as those for NVop-35. |  | 200 |  | mVp-p |
| PB ACC <br> characteristics (1) | ${ }^{\text {ACC }}{ }^{1} 1$ | $\begin{aligned} & \text { T74A } \\ & \text { T41A } \end{aligned}$ | T72 | With the conditions used for NVop-35, increase the input chroma level by +6 dB , measure the burst level on T72, and calculate the ratio with Vop-72. |  | +0.5 | +0.8 | dB |
| PB ACC <br> characteristics (2) | $A^{\prime} C_{P}{ }^{2}$ | $\begin{aligned} & \text { T74A } \\ & \text { T25A } \end{aligned}$ | T72 | With the conditions used for NVop-35, decrease the input chroma level by -6dB, measure the burst level on T72, and calculate the ratio with Vop-72. | -0.5 | -0.2 |  | dB |
| PB killer-on input level | $\mathrm{V}_{\mathrm{ACK}^{-}}{ }^{\text {P }}$ | $\begin{aligned} & \text { T74A } \\ & \text { T25A } \end{aligned}$ | T72 | With the conditions used for NVop-35, the input chroma level until output from T72 cease and measure the input burst level at that point. <br> (Calculate the ratio with the standard input $100 \mathrm{mVp}-\mathrm{p}$ signal) |  |  | -25 | dB |
| PB killer-on chroma output level | V ${ }_{\text {OACK-P }}$ | $\begin{aligned} & \text { T74A } \\ & \text { T25A } \end{aligned}$ | T35 | Measure the T35A chroma output with a spectrum analyzer in the killer state of the previous item. Calculate its ratio with NVop-35. |  | -44 | -40 | dB |
| PB main converter carrier leakage | CLP | $\begin{aligned} & \text { T74A } \\ & \text { T25A } \end{aligned}$ | T35 | With the conditions used for NVop-35, measure the T58A with a spectrum analyzer, and calculate the ratio of the 3.58 MHz component and the 4.21 MHz carrier leakage component. |  | -40 | -33 | dB |
| Burst de-emphasis | GBD | $\begin{aligned} & \text { T74A } \\ & \text { T25A } \end{aligned}$ | T72 | Apply the mixture of the low-band chroma signal of 629 kHz , burst $100 \mathrm{mVp}-\mathrm{p}$, and chroma $125 \mathrm{mV} \mathrm{p}-\mathrm{p}$ and the sine wave of 4 MHz and $300 \mathrm{mVp}-\mathrm{p}$. <br> Apply the 50\% white signal from T25A. <br> Measure burst and chroma amplitudes of T72 and accume them as $B$ and $C$ respectively. $\text { GBD }=20 \operatorname{LOG}(125 \times B) /(100 \times C)$ | -5.25 | -5.00 | -4.75 | dB |
| PB XO output level | $\mathrm{V}_{\mathrm{XO}}$-PN |  | T62 | Measure the output level on T62 with an FET probe. | 230 | 380 | 600 | mVp-p |
| PB XO oscillator frequency deviation | $\triangle_{\mathrm{XOO}}$ |  | T62 | In PB mode, let $f$ be the measured frequency on T62. $\Delta \mathrm{f}_{\mathrm{XO}} \mathrm{~N}=\mathrm{f}-3579545(\mathrm{~Hz})$ | -7 | 0 | +7 | Hz |
| PB Chroma 2'Fsc distortion | $\mathrm{P}_{\text {THD }}{ }^{2}$ | $\begin{aligned} & \text { T74A } \\ & \text { T25A } \end{aligned}$ | T35 | With the conditions used for NVOP-35, measure the T35 with a spectrum analyzer, and calculate the ratio of the 3.58 MHz component and the 7.16 MHz component. |  |  | -25 | dB |

LA71076SM
AUDIO REC mode $\mathrm{T} 55=0 \mathrm{~V}, \mathrm{~T} 59=0 \mathrm{~V}, \mathrm{~T} 17=5 \mathrm{~V}$

| Parameter | Symbol | In | Out | Conditions | Ratings |  |  | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  |  |  | min | typ | max |  |
| Voltage gain | $\mathrm{V}_{\mathrm{GR}}$ | T98 | T7 | $V_{\text {IN }}=-20 \mathrm{dBV}$ | 13.5 | 14.0 | 14.5 | dB |
| Distortion ratio | THD ${ }_{\text {R }}$ | T98 | T7 | $V_{\text {IN }}=-20 \mathrm{dBV}$ | 0.01 | 0.1 | 0.4 | \% |
| Maximum output voltage | $\mathrm{V}_{\mathrm{O}} \mathrm{MR}$ | T98 | T7 |  | 0.8 | 1.0 | 1.1 | Vrms |
| Voltage conversion recording bias current | $\mathrm{V}_{\text {BIAS }}$ |  | T6 | SW99 = ON | 270 | 300 | 330 | mVrms |
| Recording bias current control voltage | $\mathrm{V}_{\text {CTL }}$ |  | T6 | SW99 = ON | 2.9 | 3.2 | 3.5 | V |

AUDIO PB/EE mode T55 $=0 \mathrm{~V}, \mathrm{~T} 59=0 \mathrm{~V}, \mathrm{~T} 17=2.5 \mathrm{~V}$

| Parameter | Symbol | In | Out | Conditions | Ratings |  |  | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  |  |  | min | typ | max |  |
| LINE AMP <br> Voltage gain (PB) | $\mathrm{V}_{\mathrm{GLP}}$ | T100A | T96A | $\mathrm{V}_{\mathrm{IN}}=-30 \mathrm{dBV}$ T12 $=0 \mathrm{~V}, \mathrm{~T} 17=0 \mathrm{~V}$ | 22.5 | 23.0 | 23.5 | dB |
| LINE AMP <br> Voltage gain (A1) | $\mathrm{V}_{\mathrm{GLR}}$ | T76A | T96A | $\mathrm{V}_{\mathrm{IN}}=-30 \mathrm{dBV}$ T12 $=0 \mathrm{~V}, \mathrm{~T} 17=2.5 \mathrm{~V}$ | 22.5 | 23.0 | 23.5 | dB |
| LINE AMP <br> Distortion ratio (PB) | THD ${ }_{\text {L }}$ | T100A | T96A | $\mathrm{V}_{\mathrm{IN}}=-30 \mathrm{dBV}=0 \mathrm{~V}, \mathrm{~T} 17=0 \mathrm{~V}$ | 0.01 | 0.1 | 0.4 | \% |
| LINE AMP <br> Output noise voltage | $\mathrm{V}_{\mathrm{N}} \mathrm{OL}$ |  | T96A | $\mathrm{Rg}=1 \mathrm{~K} \Omega$, DIN Audio filter $\mathrm{T} 12=0 \mathrm{~V}, \mathrm{~T} 17=0 \mathrm{~V}$ SW76 = 2 | -80.0 | -74.0 | -70.5 | dBV |
| LINE AMP <br> Maximum output voltage | $\mathrm{V}_{\mathrm{O}} \mathrm{ML}$ | T100A | T96A | $\mathrm{THD}=1 \% \mathrm{~T} 12=0 \mathrm{~V}, \mathrm{~T} 17=0 \mathrm{~V}$ | 0.8 | 1.0 | 1.1 | dBV |
| LINE AMP Output voltage when ALC | $\mathrm{V}_{0} \mathrm{~A}$ | T76A | T96A | $\mathrm{T} 76 \mathrm{~A}=-28 \mathrm{dBV}$ T12 $=0 \mathrm{~V}, \mathrm{~T} 17=2.5 \mathrm{~V}$ | -7.0 | -6.0 | -5.0 | dBV |
| LINE AMP <br> Effect of ALC | ALC | T76A | T96A | $\mathrm{T} 76 \mathrm{~A}=-28$ to -8dBV $\mathrm{T} 12=0 \mathrm{~V}, \mathrm{~T} 17=2.5 \mathrm{~V}$ | 0.0 | 1.0 | 3.0 | dB |
| LINE AMP <br> Distortion ratio of when ALC | THDA | T76A | T96A | $\mathrm{T} 76 \mathrm{~A}=-28 \mathrm{dBV}$ T12 $=0 \mathrm{~V}, \mathrm{~T} 17=2.5 \mathrm{~V}$ | 0.01 | 0.1 | 0.5 | \% |
| MUTE attenuation(PB, A1, A2, A3) | MPB <br> $\mathrm{M}_{\mathrm{A} 1}$ | T100A | T96A | T100A $=-10 \mathrm{dBV}$ T12 $=5 \mathrm{~V}, \mathrm{~T} 17=0 \mathrm{~V}$ | 80 | 90 | 120 | dB |
|  |  | T76A |  | T76A $=-10 \mathrm{dBV}$ T12 $=5 \mathrm{~V}, \mathrm{~T} 17=2.5 \mathrm{~V}$ | 80 | 90 | 120 | dB |
| EQ AMP Open loop voltage gain | VGOE | T4A | T1 | $\mathrm{V}_{\mathrm{IN}}=-66 \mathrm{dBV}$ T12 $=0 \mathrm{~V}, \mathrm{~T} 17=0 \mathrm{~V}, \mathrm{SW} 3=\mathrm{ON}$ | 58.0 | 64.0 | 70.0 | dB |
| EQ AMP <br> Input conversion noise voltage | $\mathrm{V}_{\text {NIE }}$ |  | T1 | $\mathrm{Rg}=620 \Omega$, DIN Audio filter $\mathrm{T} 12=0 \mathrm{~V}, \mathrm{~T} 17=0 \mathrm{~V}$ | 0.1 | 1.0 | 1.8 | $\mu \mathrm{Vrms}$ |

LA71076SM
CCD block SW47 = 2, SW49 = 2

| Parameter | Symbol | In | Out | Conditions | Ratings |  |  | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  |  |  | min | typ | max |  |
| Voltage Gain | Gv | T49 | T49A | T47:200kHz, 500 mVp -p SW49B=ON Ratio of level of T49A relative to T47A. | -1.5 | 0.5 | 2.5 | dB |
| Frequency Response | Gf | $\begin{gathered} \mathrm{T} 47 \\ \mathrm{~T} 47 \mathrm{~B} \end{gathered}$ | T49A | T47B: Adding 250 mV higher bias than pin 47 clamp level. Ratio of 3.58 MHz component toward to T49A 200 kHz . | -2 | -1 | 0 | dB |
| Deferential Gain | DG | T47 | T49B | T47: Stair step signal ( 500 mVp -p) | 0 | 5 | 8 | \% |
| Deferential Phase | DP | T47 | T49B | T47: Stair step signal ( 500 mVp -p) | 0 | 5 | 8 | deg |
| Linearity | LS | T47 | T49B | T47: Stair step signal only for $Y$ ( 500 mVp -p) V/S Ratio of T49B. | 37 | 40 | 43 | \% |
| Clock leakage | Lck |  | T49A | SW49B:ON <br> 4fsc component of T49A |  | 10 | 50 | mVrms |
| Noise | $\mathrm{N}_{\mathrm{O}}$ |  | T49A | SW49B:ON <br> Measure T49A using a Video noise meter. <br> Filter condition: HPF $=200 \mathrm{kHz}$, LPF $=4.2 \mathrm{MHz}$, <br> TRAP $=3.58 \mathrm{MHz}$ |  | 1 | 2 | mVrms |
| Output Impedance | $\mathrm{Z}_{\mathrm{O}}$ | T47 | T49A | T47:200kHz,500mVp-p, Assuming that the T49A amplitude under conditions of SW49B = ON/OFF is T49A (ON) and T49A (OFF) respectively, and calculate as follows: $z 0=\{\{\text { T49A(OFF)-T49A(ON) }\} / T 49 A(O N)\} \times 500 .$ | 80 | 230 | 480 | $\Omega$ |
| Delay time | TD | T47 | T49B | T47:100\% white ( 500 mVp -p) <br> Calculate T49B delay time toward to T47 input. <br> * Except for reversing Amp and delay time |  | 63.35 |  | $\mu \mathrm{S}$ |

## Package Dimensions

unit: mm
3252A


## Block Diagram and Sample Application Circuit



LA71076SM
Pin Functions


Continued on next page.

LA71076SM

\begin{tabular}{|c|c|c|c|c|}
\hline Pin No. \& Pin name \& DC voltage \& Signal wave form \& Input/Output form <br>
\hline 10 \& NC CTL \& REC: 2.5 V

PB: 2.5 V \& DC \&  <br>
\hline 11 \& HA R/P CTL \& REC: 1 to 5 V

PB: 0 to 1 V \& $$
\begin{array}{ll}
5 \mathrm{~V}: & \text { REC } \\
2.5 \mathrm{~V}: & \text { R-MUTE } \\
0: & \text { PB }
\end{array}
$$ \&  <br>

\hline 12 \& | AUDIO |
| :--- |
| MUTE CTL | \& REC: 0/5V

PB: 0/5V \& MUTE $=5 \mathrm{~V}$ \&  <br>
\hline 13 \& RF_SW_IN \& REC: 0/1V

PB: 0/1V \&  \&  <br>
\hline 14 \& C-ROT IN \& REC: 0/1V

PB: 0/1V \&  \&  <br>

\hline 15 \& HA SW IN \& | REC: 0/1V |
| :--- |
| PB: 0/1V | \&  \&  <br>

\hline
\end{tabular}

Continued on next page.

\begin{tabular}{|c|c|c|c|c|}
\hline Pin No. \& Pin name \& DC voltage \& Signal wave form \& Input/Output form \\
\hline 16 \& \begin{tabular}{l}
YC \\
R/P CTL
\end{tabular} \& REC: 5V

PB: 0V \& | 5 V : REC |
| :--- |
| OV: PB | \&  <br>

\hline 17 \& | AUDIO |
| :--- |
| R/P CTL | \& | REC: 5V |
| :--- |
| PB: OV | \& \[

$$
\begin{array}{ll}
5 \mathrm{~V}: & \mathrm{REC} \\
2.5 \mathrm{~V}: & \mathrm{EE} \\
\mathrm{VV}: & \mathrm{PB}
\end{array}
$$
\] \&  <br>

\hline 18 \& | COMB |
| :--- |
| THROUGH |
| CTL | \& | REC: 1.7 V |
| :--- |
| PB: 1.7V | \& \& |  |
| :--- |
| OMP05271 | <br>

\hline 19 \& DOC/XO CTL \& REC: $0 \mathrm{~V} / 5 \mathrm{~V}$

PB: $0 \mathrm{~V} / 5 \mathrm{~V}$ \& | 5V: XO MODE OV: NORMAL |
| :--- |
| 5V: DOC OFF OV: DOC AUTO | \&  <br>

\hline 20 \& FM FILT \& | REC: 1.8 V |
| :--- |
| PB: 1.8V | \& DC \&  <br>

\hline 21 \& Y-V ${ }_{\text {CC }}$ \& 5 V \& $\mathrm{V}_{\mathrm{CC}}$ \& <br>

\hline 22 \& \[
$$
\begin{aligned}
& \text { PHASE EQ } \\
& \text { Q-CTL }
\end{aligned}
$$

\] \& | REC: 1.0V |
| :--- |
| PB: 1.0V | \& DC \&  <br>

\hline
\end{tabular}

Continued on next page.

LA71076SM

| Continued from preceding page. |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: |
| Pin No. | Pin name | DC voltage | Signal wave form | Input/Output form |
| 23 | CTL-AMP OUT | REC: 2.5 V <br> PB: 2.5V |  |  |
| 24 | MAIN- <br> EMPHA <br> FILT | REC: 2.1 V <br> PB: 2.1V |  |  |
| 25 | CLAMP-IN | REC: 2.8 V <br> PB: 2.8V |  |  |
| 26 | MAIN <br> DEEMPHA <br> -OUT1 | REC: 2.1 V <br> PB: 2.1V |  |  |
| 27 | Y-GND | OV | GND |  |
| 28 | MAIN <br> DEEMPHA <br> -OUT2 | REC: 2.1V <br> PB: 2.1V |  |  |

Continued on next page.

LA71076SM

| Continued from preceding page. |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: |
| Pin No. | Pin name | DC voltage | Signal wave form | Input/Output form |
| 29 | PHASE EQ <br> $\mathrm{F}_{0} \mathrm{CTL}$ | REC: 1.0 V <br> PB: 1.0V | DC |  |
| 30 | TRICK /CG CTL | REC <br> PB | 5V: CG OFF <br> OV: CG ON <br> 5V: TRICK <br> OV: NORMAL PB |  |
| 31 | QV/QH-INS | REC: 0 to 5 V <br> PB: 0 to 5 V | OV : Through <br> 1.0V: 20IRE INS <br> 2.0V: 6OIRE INS <br> 3.0V: QH INS <br> $\mathrm{V}_{\mathrm{CC}}$ : QV INS |  |
| 32 | S-DET OUT | REC: 0 V <br>  <br> PB: $4.2 / 0.2 \mathrm{~V}$ | DC <br> DC <br> SVHS: 4.2V <br> VHS: 0.2 V |  |
| 33 | H-SYNC OUT | REC: OV PB: OV |  |  |
| 34 | C.SYNC OUT | REC: <br> PB: |  |  |

Continued on next page.

LA71076SM

| Continued from preceding page. |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: |
| Pin No. | Pin name | DC voltage | Signal wave form | Input/Output form |
| 35 | VIDEO-OUT | REC: 0.8 V |  |  |
| 36 | EQ CTL | REC: 1.0 V <br>  <br> PB: 1.0 V | DC |  |
| 37 | AGC TC2 | REC: 2.0 V <br>  <br> PB: 2.0 V | DC <br> DC |  |
| 38 | VIDEO-IN3 | REC: 1.8 V <br>  <br> PB: 1.8 V |  |  |
| 39 | AFC2-FILT | REC: 3.5 V <br> PB: 3.5V |  |  |

Continued on next page.

LA71076SM

| Pin No. | Pin name | DC voltage | Signal wave form | Input/Output form |
| :---: | :---: | :---: | :---: | :---: |
| P40 | VIDEO-IN2 | REC: 1.8 V <br>  <br> PB: 1.8 V |  |  |
| P41 | SYNC DET FILT | REC 4.9V NO-SIG: 0.3V <br> PB: 4.9V NO-SIG: 0.3V | DC |  |
| P42 | VIDEO-IN1 | REC: 1.8 V <br> PB: 1.8V |  |  |
| P43 | VCA-FILT <br> (Phase) | REC: 2.8 V <br> PB : 2.8V | DC |  |
| P44 | VCA-IN | REC: 2.7 V <br> PB: 2.7V |  |  |

Continued on next page.

| Pin No. | Pin name | DC voltage | Signal wave form | Input/Output form |
| :---: | :---: | :---: | :---: | :---: |
| 45 | VCA- <br> FILT(Gain) | REC: 2.9 V <br>  <br> PB: 2.9 V | DC |  |
| 46 | CCD-DRIVE | REC: 1.9V <br> PB: 1.9V |  |  |
| 47 | CCD INPUT | REC: <br> PB: |  |  |
| 48 | CCD-V ${ }_{\text {SS }}$ | OV | $\mathrm{V}_{\text {SS }}$ |  |
| 49 | DELEY OUT | REC: PB: |  | OMP05309 |
| 50 | CCD-V ${ }_{\text {SS }}$ | OV | $\mathrm{V}_{S S}$ |  |
| 51 | CLOCK IN | REC: <br> PB: |  | (51) |
| 52 | VCO FILT | REC: 2.5 V PB: 2.5 V | DC | OMP05312 |

Continued on next page.

LA71076SM

| Pin No. | Pin name | DC voltage | Signal wave form | Input/Output form |
| :---: | :---: | :---: | :---: | :---: |
| 53 | 4FSC OUT | REC: 1.5 V |  |  |
| 54 | CCD-V ${ }_{\text {DD }}$ | 5 V | $V_{\text {DD }}$ |  |
| 55 | VIDEO <br> INPUT CTL | REC: 2.5V <br> PB: 2.5 V | 0 to $1 \mathrm{~V}:$ VIDEO1 <br> 2 to $3 \mathrm{~V}:$ VIDEO2 <br> 4 to $\mathrm{V}_{\mathrm{CC}}:$ VIDEO3 |  |
| 56 | $\mathrm{C}-\mathrm{V}_{\mathrm{Cc}}$ | 5 V | $\mathrm{V}_{\mathrm{CC}}$ |  |
| 57 | ALWAYS $\mathrm{V}_{\mathrm{CC}}$ | 5 V | $\mathrm{V}_{\mathrm{CC}}$ |  |
| 58 | SLD-FILT | REC: <br> PB: | DC <br> DC |  |
| 59 | SP/LP/EP | REC:  <br> $5 \mathrm{~V}:$ EP <br> OPEN: LP <br> OV: SP <br> PB:  <br> $5 \mathrm{~V}:$ EP <br> OPEN: LP <br> OV: SP |  |  |

Continued on next page.

LA71076SM

| Continued from preceding page. |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: |
| Pin No. | Pin name | DC voltage | Signal wave form | Input/Output form |
| 60 | AFC/APC-FILT | REC: <br>  <br> PB: |  |  |
| 61 | XO IN | REC: 4.0V <br> PB: 4.0V |  |  |
| 62 | XO OUT |  |  |  |
| 63 | REC <br> APC-FILT | REC: 1.8 V <br>  <br> PB: 1.8 V | DC |  |

Continued on next page.

| Continued from preceding page. |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: |
| Pin No. | Pin name | DC voltage | Signal wave form | Input/Output form |
| 64 | AGC-TC1 <br> /BALANCER | REC: 1.9 V <br> PB: 2.3V | DC <br> DC |  |
| 65 | REG4.0V | REC: 4.0V <br> PB: 4.0V |  | OMP05328 |
| 66 | REQ monitor | REC: 1.7 V <br> PB: | $300 \mathrm{mVp}-\mathrm{p}$ <br> DC |  |
| 67 | C-GND | OV | GND |  |
| 68 | REC-C-CTL | REC: 2.3 V <br> PB: 2.3V | REC-C-LEVEL <br> 0 to 1.0 V : 270 mV p -p <br> 1.8 to $2.7 \mathrm{~V}: 320 \mathrm{mVp}-\mathrm{p}$ <br> 3.5 to $5.0 \mathrm{~V}: 380 \mathrm{mVp}-\mathrm{p}$ |  |
| 69 | KIL FILT | REC: <br> PB: 1.8V |  |  |

Continued on next page.

| Continued from preceding page. |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: |
| Pin No. | Pin name | DC voltage | Signal wave form | Input/Output form |
| 70 | ACC FILT | REC: 1.8 V <br>  <br>  <br> PB: 1.8 V | DC |  |
| 71 | REC <br> CURRENT-2 | REC: 2.25 <br>  <br> PB: 2.25 | DC |  |
| 72 | PB C MONI | REC: 2.5 V <br> PB: 2.5V | $150 \mathrm{mVp}-\mathrm{p}$ | OMP05335 |
| 73 | REC <br> CURRENT-1 | REC: 2.25 <br>  <br> PB: 2.25 | DC |  |
| 74 | HA monitor | REC: - <br> PB: 2.0V | $\text { FM } 300 \mathrm{mVp} \mathrm{p}-\mathrm{p}$ |  |
| 75 | AUDIO $\mathrm{V}_{\mathrm{CC}}$ | 5.0 V | $\mathrm{V}_{\mathrm{CC}}$ |  |
| 76 | AUDIO <br> IN-1 | REC: 3V <br> PB: 3V | CW 95mVp-p <br> CW 95mVp-p |  |

Continued on next page.

LA71076SM

| Continue | om preceding |  |  |  |
| :---: | :---: | :---: | :---: | :---: |
| Pin No. | Pin name | DC voltage | Signal wave form | Input/Output form |
| 77 | ALC DET | REC <br>  <br> PB: 0 V | DC |  |
| 78 | AUDIO <br> IN-2 | REC: 3V <br> PB: 3V | CW 95mVp-p |  |
| 79 | VREF | REC: 2.3 V <br> PB: 2.3 V | DC |  |
| 80 | AUDIO <br> IN-3 | REC: 3V <br> PB: 3V | CW 95mVp-p |  |
| 81 | HEAD AMP GND | OV | GND |  |
| 82 | PBEPL+ | REC: 4.1V <br> PB: 1.8V |  |  |
| 83 | PBEPL- | REC: 4.1V <br> PB: 1.8V | $\begin{aligned} & \text { SP } 13 \mathrm{mAp}-\mathrm{p} \\ & \text { EP } 10 \mathrm{mAp}-\mathrm{p} \end{aligned}$ $\text { FM 0.5mVp-p } \bigcap \int=$ |  |

LA71076SM

| Continued from preceding page. |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: |
| Pin No. | Pin name | DC voltage | Signal wave form | Input/Output form |
| 84 | PBEPH- | REC: 4.1 V <br>  <br> PB: 1.8 V | SP 13mAp-p <br> EP 10mAp-p <br> FM 0.5mVp-p | OMP05347 |
| 85 | PBEPH+ | REC: 4.1V <br> PB: 1.8V | FM 0.5mVp-p |  |
| 86 | HEAD AMP GND | OV | GND |  |
| 87 | HEAD AMP $\mathrm{V}_{\mathrm{CC}}$ | 5 V | $\mathrm{V}_{\mathrm{CC}}$ |  |
| 88 | PBSPL+ | REC: 4.1V <br> PB: 1.8V | $\begin{aligned} & \text { FM 0.5mVp-p } \\ & \text { ? } \end{aligned}$ |  |
| 89 | PBSPL- | REC <br> PB: | SP 13mAp-p <br> EP 10mAp-p <br> FM 0.5mVp-p |  |

Continued on next page.

LA71076SM

| Continued from preceding page. |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: |
| Pin No. | Pin name | DC voltage | Signal wave form | Input/Output form |
| 90 | PBSPH- | REC | $\begin{aligned} & \text { SP 13mAp-p } \\ & \text { EP 10mAp-p } \end{aligned}$ |  |
| 91 | PBSPH+ | REC: 4.1V <br> PB: 1.8V | FM 0.5mVp-p |  |
| 92 | AGC FILT | REC: 1.6 V <br> PB: - | DC |  |
| 93 | ENV DET OUT | REC <br> PB <br> 0.5 to 4.8 V | DC |  |
| 94 | COMP OUT | REC: 0.7 V <br> PB: - <br> TRICK PB: <br> 0.5V/4.5V | DC <br> TRICK MODE HA SW OUT |  |

Continued on next page.

LA71076SM

\begin{tabular}{|c|c|c|c|c|}
\hline Pin No. \& Pin name \& DC voltage \& Signal wave form \& Input/Output form \\
\hline 95 \& AUDIO GND \& OV \& GND \& \\
\hline 96 \& \begin{tabular}{l}
AUDIO \\
LINE OUT
\end{tabular} \& REC: 2.3 V

PB: 2.3 V \& CW 1.4Vp-p
CW 1.4Vpp \&  <br>
\hline 97 \& ALC DET IN \& REC: OV
PB: OV \& CW 600mVp-p
GND \& OMP05357 <br>

\hline 98 \& AUDIO REC IN \& | REC: 0V |
| :--- |
| PB: OV | \& CW 280mVp-p

GND \&  <br>

\hline 99 \& AUTO BIAS OUT \& | REC: 4.3V |
| :--- |
| PB: 5V | \& DC \&  <br>


\hline 100 \& AUDIO PB IN \& | REC: 2.3 V |
| :--- |
| PB: 2.3V | \& Half-wave rectified wave form ( 70 kHz )

CW 95mVp-p \&  <br>
\hline
\end{tabular}

Function Control Table

| Pin No. | Function | L (to 0.5V) |  |  |  | H (4.0V to ) |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 12 | AUDIO MUTE | OFF |  |  |  | ON |  |
| 19 | REC: XO MODE | NORMAL |  |  |  | XO MODE |  |
|  | PB: DOC OFF | DOC AUTO |  |  |  | DOC OFF |  |
| 30 | PB: TRICK | NORMAL |  |  |  | TRICK |  |
|  | REC: COPY GUARD | ON |  |  |  | OFF |  |
| Pin No. | Function | L (to 0.5V) |  |  |  | H (1.0V to ) |  |
| 13 | RF SW PULSE | L |  |  |  | H |  |
| 14 | C ROT PULSE | L |  |  |  | H |  |
| 15 | HA PULSE | L |  |  |  | H |  |
| Pin No. | Function | to 0.7V | 1.1 V to 3.3 V |  | 1.7 V to 2.3 V | 2.7V to 3.3 V | 3.7 V to |
| 31 | CHARACTER INSERT | THROUGH | $\begin{aligned} & \text { EDGH } \\ & 25 I R E \end{aligned}$ |  | CHARA 4OIRE | $\begin{gathered} \mathrm{QH} \\ 25 \mathrm{IRE} \end{gathered}$ | QV SYNC |
| Pin No. | Function | L (to 0.5V) |  | M (2.0 to 2.8V OR OPEN) |  | H (3.6V to ) |  |
| 16 | REC/EE/PB (Y/C) | PB |  | REC |  |  |  |
| 17 | REC/EE/PB (Audio) | PB |  | EE |  | REC |  |
| 55 | INPUT SELECT | IN-1 |  | IN-2 |  | IN-3 |  |
| 59 | TAPE SPEED | SP |  | LP |  | EP |  |
| 68 | REC C-LEVEL | +1.5dB |  | STANDARD |  | -1.5dB |  |
| 71 | REC: REC-CURRENT-2 | LOW |  | STANDARD |  | MAX |  |
| 73 | REC: REC-CURRENT-1 | LOW |  | STANDARD |  | MAX |  |
| 10 | NC CTL | MAX |  | STANDARD |  | LOW |  |
| Pin No. | Function | L (to 0.5V) |  |  |  | OPEN OR H (4.0V to ) |  |
| 18 | COMB THROUGH CONTROL | KIL+SP-REC-NO-CORR |  |  |  | KIL+SP-REC |  |
| Pin No. | Function | L (to 0.5V) |  |  |  | H (4.0V to ) |  |
| 8 | REC-I ARRANGEMENT | SP-CURRENT MORE |  |  |  | SP/EP SAME CURRENT |  |
| Pin No. | Function | L (to 0.5 V ) |  | M (2.5V ) |  | H (4.0V to ) |  |
| 11 | REC/PAUSE/PB (Head Amp) | PB |  | REC PAUSE |  | REC |  |

## Test Mode Control

| Pin No. | Function |  |
| :---: | :--- | :--- |
| 22 | Phase EQ Q adjust/Y-test | PULL-UP = Y-test |
| 23 | EQ monitor/R-EQ slope control | PULL-UP = REC EQ slope is changes to slant |
| 29 | Phase EQ F0 adjust/F-test | PULL-UP = F-test |
| 36 | PB EQ Low side band CTL/REC <br> EQ in | REC EQ external signal: input with bias (3.5V) |
| 66 | REC ENV monitor/Head Amp in | Head Amp external signal: input with bias (3.5V) |
| 72 | PB-C monitor/FM Mute \& Child Lock <br> ISync Slice Level | PULL-UP = REC: FM mute at pin66 <br> R/P pin23 changes to TEST mode <br> Child Lock mode |
|  |  | additional resistor to GND: <br> Sync Slice Level changes to Pedestal side. |
| 74 | PB Head Amp monitor/ENV in | External ENV signal: input with bias (3.5V) |

## Test Circuits



- Specifications of any and all SANYO Semiconductor products described or contained herein stipulate the performance, characteristics, and functions of the described products in the independent state, and are not guarantees of the performance, characteristics, and functions of the described products as mounted in the customer's products or equipment. To verify symptoms and states that cannot be evaluated in an independent device, the customer should always evaluate and test devices mounted in the customer's products or equipment.
$\square$ SANYO Semiconductor Co., Ltd. strives to supply high-quality high-reliability products. However, any and all semiconductor products fail with some probability. It is possible that these probabilistic failures could give rise to accidents or events that could endanger human lives, that could give rise to smoke or fire, or that could cause damage to other property. When designing equipment, adopt safety measures so that these kinds of accidents or events cannot occur. Such measures include but are not limited to protective circuits and error prevention circuits for safe design, redundant design, and structural design.
$■$ In the event that any or all SANYO Semiconductor products (including technical data,services) described or contained herein are controlled under any of applicable local export control laws and regulations, such products must not be exported without obtaining the export license from the authorities concerned in accordance with the above law.
$\square$ No part of this publication may be reproduced or transmitted in any form or by any means, electronic or mechanical, including photocopying and recording, or any information storage or retrieval system, or otherwise, without the prior written permission of SANYO Semiconductor Co., Ltd.
- Any and all information described or contained herein are subject to change without notice due to product/technology improvement, etc. When designing equipment, refer to the "Delivery Specification" for the SANYO Semiconductor product that you intend to use.
- Information (including circuit diagrams and circuit parameters) herein is for example only; it is not guaranteed for volume production. SANYO Semiconductor believes information herein is accurate and reliable, but no guarantees are made or implied regarding its use or any infringements of intellectual property rights or other rights of third parties.

This catalog provides information as of February, 2006. Specifications and information herein are subject to change without notice.

