## FEATURES

Dual independent digitally controlled VGAs
$\mathbf{- 1 1 . 5}$ to $\mathbf{2 0 ~ d B ~ G a i n ~ R a n g e ~}$
0.5 dB step size $\pm 0.1 \mathrm{~dB}$
$150 \Omega$ differential input and output
6 dB noise figure @ maximum gain
OIP3 of $\mathbf{5 0 ~ d B m}$ at $200 \mathbf{~ M H z}$
-3 dB bandwidth of 700 MHz
Multiple control interface options
Parallel 6-bit control interface
Serial peripheral interface
Gain step up/down interface
Wide input dynamic range
High performance power mode
Power-down control
Single 5 V supply operation
40-Lead LFCSP 6 x 6 mm package

## APPLICATIONS

Differential ADC drivers
High IF sampling receivers
High output power IF amplification
Instrumentation


Figure 1.

## GENERAL DESCRIPTION

The ADL5202 is a digitally controlled, variable gain wide bandwidth amplifier that provides precise gain control, high IP3 and low noise figure. The excellent distortion performance and high signal bandwidth makes the ADL5202 an excellent gain control device for a variety of receiver applications.
For wide input dynamic range applications, the ADL5202 provides a broad 31.5 dB gain range with 0.5 dB resolution. The gain is adjustable through multiple gain control interface options: parallel, serial peripheral interface, or gain step up/down.
Using a high speed SiGe process and incorporating proprietary distortion cancellation techniques, the ADL5202 achieves better than 50 dBm output IP3 at frequencies approaching 200 MHz for all gain settings. The ADL5202 is powered on by applying
the appropriate logic level to the PWUP pin. The quiescent current of the ADL5202 is typically 160 mA . It may be configured for higher quiescent current of 220 mA , in high performance power mode, for more demanding applications. When powered down, the ADL5202 consumes less than 18 mA and offers excellent input to output isolation. The gain setting is preserved when powered down.

Fabricated on an ADI's high speed SiGe process, the ADL5202 provides precise gain adjustment capabilities with good distortion performance. The ADL5202 amplifier comes in a compact, thermally enhanced $6 \times 6 \mathrm{~mm} 40$-lead LFCSP package and operates over the temperature range of $-40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$

## Rev. PrE

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## ADL5202

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## SPECIFICATIONS

$\mathrm{V}_{\mathrm{S}}=5 \mathrm{~V}, \mathrm{~T}=25^{\circ} \mathrm{C}, \mathrm{Z}_{\mathrm{S}}=\mathrm{Z}_{\mathrm{L}}=150 \Omega$ at $100 \mathrm{MHz}, \mathrm{PM}=0 \mathrm{~V}, 2 \mathrm{~V}$ p-p differential output unless otherwise noted.
Table 1.

| Parameter | Conditions | Min | Typ | Max | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: |
| DYNAMIC PERFORMANCE <br> -3 dB Bandwidth <br> Slew Rate | $\mathrm{V}_{\text {OUT }}<2 \mathrm{~V}$ p-p (5.2dBm) |  | $\begin{aligned} & 700 \\ & \text { TBD } \end{aligned}$ |  | MHz <br> V/nsec |
| INPUT STAGE <br> Maximum Input Swing <br> Differential Input Resistance <br> Common-Mode Input Voltage CMRR | Pins VIN+ and VIN- <br> Gain Code $=111111$ <br> Differential <br> Gain Code $=000000$ |  | $\begin{aligned} & 8 \\ & 150 \\ & 1.5 \\ & \text { TBD } \end{aligned}$ |  | $\begin{aligned} & \text { V p-p } \\ & \Omega \\ & V \\ & d B \end{aligned}$ |
| GAIN <br> Maximum Voltage Gain <br> Minimum Voltage Gain <br> Gain Step Size <br> Gain Flatness <br> Gain Temperature Sensitivity <br> Gain Step Response <br> Gain Conformance Error <br> Phase Conformance Error | Gain Code $=000000$ <br> Gain Code $=111111$ <br> $30 \mathrm{MHz}<\mathrm{f}_{\mathrm{c}}<200 \mathrm{MHz}$ <br> Gain Code $=000000$ <br> For $\mathrm{V}_{\mathrm{IN}}=0.2 \mathrm{~V}$, Gain Code 111111to 000000 <br> Normalized to 10 dB gain step <br> Normalized to 10dB gain step |  | $\begin{aligned} & 20 \\ & -11.5 \\ & 0.5 \\ & \text { TBD } \\ & \text { TBD } \\ & 15 \\ & \pm 0.03 \\ & 1.0 \end{aligned}$ |  | dB <br> dB <br> dB <br> dB <br> $\mathrm{mdB} /{ }^{\circ} \mathrm{C}$ <br> ns <br> dB <br> deg |
| OUTPUT STAGE <br> Output Voltage Swing <br> Differential Output Resistance | Pins OUT+ and OUT- <br> At P1dB, Gain Code $=000000$ <br> Differential |  | $\begin{aligned} & 10 \\ & 150 \end{aligned}$ |  | $\begin{aligned} & \text { Vp-p } \\ & \Omega \end{aligned}$ |
| NOISE/HARMONIC PERFORMANCE <br> 46 MHz [High Performance Power Mode] <br> Noise Figure <br> Second Harmonic <br> Third Harmonic <br> Output IP3 <br> Output 1 dB Compression Point | $\text { Gain Code }=000000, \text { LP }=\text { Low }$ $\begin{aligned} & V_{\text {out }}=2 \mathrm{~V} \text { p-p } \\ & \mathrm{V}_{\text {out }}=2 \mathrm{Vp}-\mathrm{p} \end{aligned}$ |  | $\begin{aligned} & 6 \\ & -90 \\ & -100 \\ & \text { TBD } \\ & 18.6 \end{aligned}$ |  | dB <br> dBc <br> dBc <br> dBm <br> dBm |
| 46 MHz [Nominal Power Mode] <br> Noise Figure <br> Second Harmonic <br> Third Harmonic <br> Output IP3 <br> Output 1 dB Compression Point | $\text { Gain Code }=000000, \text { PM }=\text { High }$ $\begin{aligned} & \text { Vout }=2 \mathrm{~V} \text { p-p } \\ & \text { Vout }=2 \mathrm{Vp}-\mathrm{p} \end{aligned}$ |  | $\begin{aligned} & \text { TBD } \\ & -90 \\ & -100 \\ & \text { TBD } \\ & \text { TBD } \end{aligned}$ |  | dB <br> dBc <br> dBc <br> dBm <br> dBm |
| NOISE/HARMONIC PERFORMANCE <br> 70 MHz [High Performance Power Mode] <br> Noise Figure <br> Second Harmonic <br> Third Harmonic <br> Output IP3 <br> Output 1 dB Compression Point | Gain Code $=000000$, LP $=$ Low $\begin{aligned} & V_{\text {out }}=2 \mathrm{~V} \text { p-p } \\ & \mathrm{V}_{\text {out }}=2 \mathrm{Vp}-\mathrm{p} \end{aligned}$ |  | $\begin{aligned} & 6 \\ & -88 \\ & -100 \\ & 46.4 \\ & 19.7 \end{aligned}$ |  | dB <br> dBc <br> dBc <br> dBm <br> dBm |


| Parameter | Conditions | Min | Typ | Max | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: |
| 70 MHz [Nominal Power Mode] <br> Noise Figure <br> Second Harmonic <br> Third Harmonic <br> Output IP3 <br> Output 1 dB Compression Point | $\text { Gain Code }=000000, \text { PM }=\text { High }$ $\begin{aligned} & \mathrm{V}_{\text {out }}=2 \mathrm{~V} \mathrm{p}-\mathrm{p} \\ & \mathrm{~V}_{\text {out }}=2 \mathrm{~V} \text { p-p } \end{aligned}$ |  | $\begin{aligned} & \text { TBD } \\ & -88 \\ & -100 \\ & 40 \\ & \text { TBD } \end{aligned}$ |  | dB <br> dBc <br> dBc <br> dBm <br> dBm |
| NOISE/HARMONIC PERFORMANCE <br> 140 MHz [High Performance Power Mode] <br> Noise Figure <br> Second Harmonic <br> Third Harmonic <br> Output IP3 <br> Output 1 dB Compression Point | $\text { Gain Code }=000000, \text { LP }=\text { Low }$ $\begin{aligned} & \mathrm{V}_{\text {out }}=2 \mathrm{~V} \text { p-p } \\ & \mathrm{V}_{\text {out }}=2 \mathrm{Vp}-\mathrm{p} \end{aligned}$ |  | $\begin{aligned} & 6.4 \\ & -88 \\ & -97 \\ & \text { TBD } \\ & 19.7 \end{aligned}$ |  | dB <br> dBc <br> dBc <br> dBm <br> dBm |
| 140 MHz [Nominal Power Mode] <br> Noise Figure <br> Second Harmonic <br> Third Harmonic <br> Output IP3 <br> Output 1 dB Compression Point | $\text { Gain Code }=000000, \text { PM }=\text { High }$ $\begin{aligned} & \text { Vout }=2 \mathrm{~V} \text { p-p } \\ & \text { Vout }=2 \mathrm{Vp}-\mathrm{p} \end{aligned}$ |  | $\begin{aligned} & \text { TBD } \\ & -88 \\ & -97 \\ & \text { TBD } \\ & \text { TBD } \end{aligned}$ |  | dB <br> dBc <br> dBc <br> dBm <br> dBm |
| NOISE/HARMONIC PERFORMANCE <br> 170 MHz [High Performance Power <br> Mode] <br> Noise Figure <br> Second Harmonic <br> Third Harmonic <br> Output IP3 <br> Output 1 dB Compression Point | $\text { Gain Code }=000000, \text { LP }=\text { Low }$ $\begin{aligned} & \mathrm{V}_{\text {out }}=2 \mathrm{~V} \text { p-p } \\ & \mathrm{V}_{\text {out }}=2 \mathrm{Vp}-\mathrm{p} \end{aligned}$ |  | $\begin{aligned} & 6.5 \\ & -82 \\ & -97 \\ & 46.7 \\ & 19.7 \end{aligned}$ |  | dB <br> dBc <br> dBc <br> dBm <br> dBm |
| 170 MHz [Nominal Power Mode] <br> Noise Figure <br> Second Harmonic <br> Third Harmonic <br> Output IP3 <br> Output 1 dB Compression Point | $\text { Gain Code }=000000, \text { PM }=\text { High }$ $\begin{aligned} & V_{\text {out }}=2 \mathrm{~V} \text { p-p } \\ & \mathrm{V}_{\text {out }}=2 \mathrm{~V} \text { p-p } \end{aligned}$ |  | $\begin{aligned} & \text { TBD } \\ & -77 \\ & -95 \\ & 39.7 \\ & \text { TBD } \end{aligned}$ |  | dB <br> dBc <br> dBc <br> dBm <br> dBm |
| NOISE/HARMONIC PERFORMANCE <br> 240 MHz [High Performance Power <br> Mode] <br> Noise Figure <br> Second Harmonic <br> Third Harmonic <br> Output IP3 <br> Output 1 dB Compression Point | $\text { Gain Code }=000000, \text { LP = Low }$ $\begin{aligned} & \mathrm{V}_{\text {out }}=2 \mathrm{~V} \mathrm{p}-\mathrm{p} \\ & \mathrm{~V}_{\text {out }}=2 \mathrm{Vp}-\mathrm{p} \end{aligned}$ |  | $\begin{aligned} & 6.9 \\ & -78 \\ & -93 \\ & \text { TBD } \\ & 19.7 \end{aligned}$ |  | dB <br> dBc <br> dBc <br> dBm <br> dBm |
| 240 MHz [Nominal Power Mode] <br> Noise Figure <br> Second Harmonic <br> Third Harmonic <br> Output IP3 <br> Output 1 dB Compression Point | $\text { Gain Code }=000000, \text { PM }=\text { High }$ $\begin{aligned} & V_{\text {out }}=2 \mathrm{~V} \text { p-p } \\ & \mathrm{V}_{\text {out }}=2 \mathrm{~V} \text { p-p } \end{aligned}$ |  | $\begin{aligned} & \text { TBD } \\ & -73 \\ & -93 \\ & \text { TBD } \\ & \text { TBD } \end{aligned}$ |  | dB <br> dBc <br> dBc <br> dBm <br> dBm |


| Parameter | Conditions | Min | Typ | Max | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: |
| NOISE/HARMONIC PERFORMANCE <br> 300 MHz [High Performance Power Mode] <br> Noise Figure <br> Second Harmonic <br> Third Harmonic <br> Output IP3 <br> Output 1 dB Compression Point | $\text { Gain Code }=000000, \text { LP }=\text { Low }$ $\begin{aligned} & \text { Vout }=2 \mathrm{~V} \text { p-p } \\ & \text { Vout }^{2}=2 \mathrm{Vp}-\mathrm{p} \end{aligned}$ |  | $\begin{aligned} & 7.3 \\ & -70 \\ & -88 \\ & \text { TBD } \\ & 19.5 \end{aligned}$ |  | dB <br> dBc <br> dBc <br> dBm <br> dBm |
| 300 MHz [Nominal Power Mode] <br> Noise Figure <br> Second Harmonic <br> Third Harmonic <br> Output IP3 <br> Output 1 dB Compression Point | $\text { Gain Code }=000000, \text { PM }=\text { High }$ $\begin{aligned} & V_{\text {out }}=2 \mathrm{~V} \text { p-p } \\ & V_{\text {out }}=2 \mathrm{Vp}-\mathrm{p} \end{aligned}$ |  | $\begin{aligned} & \text { TBD } \\ & -68 \\ & -88 \\ & \text { TBD } \\ & \text { TBD } \end{aligned}$ |  | dB <br> dBC <br> dBc <br> dBm <br> dBm |
| NOISE/HARMONIC PERFORMANCE <br> 380 MHz [High Performance Power <br> Mode] <br> Noise Figure <br> Second Harmonic <br> Third Harmonic <br> Output IP3 <br> Output 1 dB Compression Point | $\text { Gain Code }=000000, \text { LP }=\text { Low }$ $\begin{aligned} & V_{\text {out }}=2 \mathrm{~V} \text { p-p } \\ & V_{\text {out }}=2 \mathrm{~V} \text { p-p } \end{aligned}$ |  | $\begin{aligned} & 7.8 \\ & -67 \\ & -80 \\ & \text { TBD } \\ & 18.4 \end{aligned}$ |  | dB <br> dBc <br> dBC <br> dBm <br> dBm |
| 380 MHz [Nominal Power Mode] <br> Noise Figure <br> Second Harmonic <br> Third Harmonic <br> Output IP3 <br> Output 1 dB Compression Point | $\text { Gain Code }=000000, \text { PM }=\mathrm{High}$ $\begin{aligned} & V_{\text {out }}=2 \mathrm{~V} \text { p-p } \\ & V_{\text {out }}=2 \mathrm{~V} \text { p-p } \end{aligned}$ |  | $\begin{aligned} & \text { TBD } \\ & -65 \\ & -80 \\ & \text { TBD } \\ & \text { TBD } \end{aligned}$ |  | dB <br> dBc <br> dBc <br> dBm <br> dBm |
| ENABLE INTERFACE <br> Enable Threshold PWUP Input Bias Current | Pin PWUP <br> Minimum voltage to enable the device |  | TBD | 1.4 | $\begin{aligned} & \text { V } \\ & \text { nA } \end{aligned}$ |
| GAIN CONTROL INTERFACE <br> $\mathrm{V}_{\mathrm{H}}$ <br> VIL <br> Maximum Input Bias Current | Digital pins <br> Minimum voltage for a logic high Maximum voltage for a logic low | 1.4 | TBD | 0.8 | V <br> nA |
| POWER-INTERFACE <br> Supply Voltage Quiescent Current <br> Power Down Current | PM = Low (High Performance Power Mode) <br> PM = High (Nominal Power Mode) <br> PWUP Low | 4.5 | $\begin{aligned} & 220 \\ & 160 \\ & 18 \end{aligned}$ | 5.5 | V <br> mA <br> mA <br> mA |

## ABSOLUTE MAXIMUM RATINGS

Table Summary
Table 2.

| Parameter | Rating |
| :--- | :--- |
| Supply Voltage, $\mathrm{V}_{\text {POS }}$ | 5.5 V |
| PWUP, Digital Pins | -0.6 to $\left(\mathrm{V}_{\text {Pos }}+0.6 \mathrm{~V}\right)$ |
| Input Voltage, $\mathrm{V}_{\mathrm{N}+}, \mathrm{V}_{\mathrm{IN}}$ | -0.6 to +3.1 V |
| Internal Power Dissipation | TBD mW |
| $\theta_{\mathrm{JA}}$ (Exposed paddle soldered down) | $\mathrm{TBD}{ }^{\circ} \mathrm{C} / \mathrm{W}$ |
| $\theta_{\mathrm{JA}}$ (Exposed paddle not soldered down) | $\mathrm{TBD}{ }^{\circ} \mathrm{C} / \mathrm{W}$ |
| $\theta_{\mathrm{JC}}($ At exposed paddle soldered down) | $\mathrm{TBD}{ }^{\circ} \mathrm{C} / \mathrm{W}$ |
| Maximum Junction Temperature | $\mathrm{TBD}{ }^{\circ} \mathrm{C}$ |
| Operating Temperature Range | $-40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$ |
| Storage Temperature Range | $-65^{\circ} \mathrm{C}$ to $+150^{\circ} \mathrm{C}$ |

Stresses above those listed under Absolute Maximum Ratings may cause permanent damage to the device. This is a stress rating only; functional operation of the device at these or any other conditions above those indicated in the operational section of this specification is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

## THERMAL RESISTANCE

$\theta_{\text {IA }}$ is specified for the worst-case conditions, that is, a device soldered in a circuit board for surface-mount packages.

Table 3. Thermal Resistance

| Package Type | $\theta_{\mathrm{JA}}$ | $\theta_{\mathrm{Jc}}$ | Unit |
| :--- | :--- | :--- | :--- |
|  |  |  |  |
|  |  |  |  |

## ESD CAUTION

|  | ESD (electrostatic discharge) sensitive device. <br> Charged devices and circuit boards can discharge <br> without detection. Although this product features <br> patented or proprietary protection circuitry, damage <br> may occur on devices subjected to high energy ESD. <br> Therefore, proper ESD precautions should be taken to <br> avoid performance degradation or loss of functionality. |
| :--- | :--- |

## PIN CONFIGURATION AND FUNCTION DESCRIPTIONS



Table 4. Pin Function Descriptions

| Pin No. | Mnemonic | Description |
| :---: | :---: | :---: |
| 1 | $\overline{\mathrm{CSA}} / \mathrm{A} 3$ | Multi function pin: When Serial mode is enabled, a logic low on this pin selects Channel A. In Parallel mode, this bit 3 for the gain control interface. |
| 2 | A4 | Bit 4 for channel A parallel gain control interface. |
| 3 | A5 | Bit 5, (MSB) for channel A parallel gain control interface. |
| 4 | MODE1 | MSB for the mode control parallel, SPI, up/down interface. |
| 5 | MODE0 | LSB for the mode control parallel, SPI, up/down interface. |
| 6 | PM | A logic low on this pin enables high performance mode. A logic high enables nominal performance mode. |
| $\begin{aligned} & 7,18,33 \\ & E P^{1} \end{aligned}$ | GND | Ground |
| 8 | SDIO/B5 | Multi function pin: When $\overline{C S A}$ or $\overline{C S B}$ is pulled low, SDIO is used for reading and writing to the SPI port. In parallel mode, This bit is $5(\mathrm{MSB})$ for the channel $B$ parallel gain control interface. |
| 9 | SCLK/B4 | Multi function pin: When SPI mode is selected this pin is the serial clock input. In parallel mode this pin is bit 4 for channel $B$ gain interface. |
| 10 | GS1/[CSB/B3 | Multi function pin: When the UP/DOWN mode is enabled, this pin is the MSB for the gain step size control. When serial mode is enabled, a logic low on this pin selects channel B. In parallel mode, this is bit 3 of the gain control interface. |
| 11 | GS0 FA_B/B2 | Multi function pin: When the UP/DOWN mode is enabled, this pin is the LSB for the gain step size control. A logic high enables the channel B SPI port fast attack mode. In parallel mode this pin is bit 2 for channel $B$ gain interface. |
| 12 | UPDN_CLK_B/B1 | Multi function pin: this pin is the clock interface for channel B UPDN function. In Parallel mode this pin is bit1 for channel B gain interface. |
| 13 | UPDN_DAT_B/B0 | Multi function pin: this pin is the data pin for channel B UPDN function. In parallel mode this is bit 0 for channel $B$ gain interface. |
| 14 | LATCHB | Latch, a low input results in gain change. A high input results in no gain change. |
| 15 | VINB- | Channel $B$ negative input. |
| 16 | VINB+ | Channel B positive input. |


| Pin No. | Mnemonic | Description |
| :--- | :--- | :--- |
| 17 | PWUPB | Channel B power up. A logic high on this pin enables the part. |
| 19,21 | VOUTB+ | Channel B positive output. |
| 20,22 | VOUTB- | Channel B negative output. |
| $23,24,25$, | VPOS | Positive power supply. |
| $26,27,28$, |  |  |
| 29,31 | VOUTA+ | Channel A positive output |
| 30,32 | VOUTA- | Channel A negative output |
| 34 | PWUPA | Channel A power up. A logic high on this pin enables the part. |
| 35 | VINA+ | Channel A positive input. |
| 36 | VINA- | Channel A negative input. |
| 37 | LATCHA | Latch, a low input results in gain change. A high input results in no gain change. <br> 38 |
| 39 | UPDN_DAT_A/AO | Multi function pin: this pin is the data pin for channel A UPDN function. In parallel mode this is bit 0 for <br> channel A gain interface. |
| 40 | UPDN_CLK_A/A1 | Multi function pin: this pin is the clock interface for channel A UPDN function. In Parallel mode this pin is <br> bit1 for channel A gain interface. <br> Multi function pin: When the UP/DOWN mode is enabled, this pin is the LSB for the gain step size control. |
|  |  | A logic high enables the channel A SPI port fast attack mode. In parallel mode this pin is bit 2 for channel <br> A gain interface. |

[^0]
## DIGITAL INTERFACE OVERVIEW

The ADL5202 DVGA has three digital control interface options:

- Parallel Control Interface
- Serial Peripheral Interface
- Gain Step Up/Down Interface

The digital control interface selection is made via 2 digital pins, MODE1 and MODE0, as shown in Table 5. There are two common digital control pins, PM and PWUP. PM selects between two power modes. PWUP is a power up pin. The gain code used is 6 bit binary.
Physical pins are shared between 3 interfaces resulting in as many as 3 different functions per digital pin (see Table 4)

Table 5. Digital control interface selection truth table

| MODE1 | MODE0 | Interface |
| :--- | :--- | :--- |
| 0 | 0 | Parallel |
| 0 | 1 | Serial (SPI) |
| 1 | 0 | Up/Down |
| 1 | 1 | Up/Down |
| Parallel Digital Interface |  |  |

## Parallel Digital Interface

The parallel digital interface uses 6 gain control bits and a latch pin per amplifier. The latch pin controls whether the input data latch is transparent or latched. In transparent mode, gain changes as input gain control bits change. In latched mode, gain is determined by the latched gain setting and does not change with changing input gain control bits.

## Serial Peripheral Interface (SPI)

The SPI uses 3 pins (SDIO, SCLK, and /CSA or /CSB). The SPI data register consists of 2 bytes: 6 gain control bits, 2 attenuation step size address bits, 1 read/write bit, and 7 do not care bits.
The SPI uses a bidirectional pin, SDIO, for writing to the SPI register and for reading from the SPI register. In order to write to the SPI register, $\overline{\mathrm{CSA}}$ or $\overline{\overline{C S B}}$ needs to be pulled low and 16 clock pulses must be applied. Individual channel SPI registers can be selected by pulling low $\overline{C S A}$ or $\overline{C S B}$. By simultaneously pulling low the $\overline{\mathrm{CSA}}$ and $\overline{\mathrm{CSB}}$ pins, the same data can be written to both SPI registers.
In order to read the SPI register value, the R/W bit needs to be set high, $\overline{\mathrm{CSA}}$ or $\overline{\mathrm{CSB}}$ needs to be pulled low, and the part clocked. Once the register has been read out the R/W bit needs to be set low and SPI put in write mode. Note that there is only one SDIO pin. Read back from the registers should be done individually.

SPI fast attack mode is controlled by FA_A or FA_B. A logic high on the FA pin results in an attenuation selected by FA1 and FA0 bits in the SPI register.

Table 6. SPI 2-bit attenuation step size truth table

| FA1 | FAO | Step Size (dB) |
| :--- | :--- | :--- |
| 0 | 0 | 2 |
| 0 | 1 | 4 |
| 1 | 0 | 8 |
| 1 | 1 | 16 |

## UP/DOWN Interface

The UP/DOWN interface uses two digital pins to control the gain. Gain is increased by a clock pulse on UPDN_CLK (rising and falling edges) when UPDN_DAT is high. Gain is decreased by a clock pulse on UPDN_CLK when UPDN_DAT is low. Reset is detected by a rising edge latching data having one polarity with the falling edge latching the opposite polarity. Reset results in minimum gain code $111111_{\text {bin }}$.


Figure 3. UP/DOWN Timing

The step size is selectable by pins GS1 and GS0. The default step size is 0.5 dB . The gain code count will rail at the top and bottom of the control range.

Table 7. Step size control truth table

| GS1 | GS0 | Step Size (dB) |
| :--- | :--- | :--- |
| 0 | 0 | 0.5 |
| 0 | 1 | 1 |
| 1 | 0 | 2 |
| 1 | 1 | 4 |



Figure 4. 16-bit SPI Register

## TYPICAL PERFORMANCE CHARACTERISTICS




Figure 5. OIP3 vs. Power @ 5 Gains


Figure 6. P1dB Vs. Frequency at Max Gain


Figure 7. OIP3 Vs. Gain


Figure 8. Noise Figure Vs. Frequency at Max Gain


Figure 9. S11, S12 and S22 Vs. Frequency


Figure 10. Harmonic Distortion Vs. Frequency 2Vp-p Out

## EVALUATION BOARD

The ADL5202 evaluation board is available with software control to program the variable gain control. It is a 4-layer board with split ground plane for analog and digital sections. Special care is taken to place the power decoupling capacitors close to the device pins. The board is designed for easy singleended (through a Mini-Circuits TC3-1T+ RF transformer) or differential configuration for each channel.

## EVALUATION BOARD CONTROL SOFTWARE

The ADL5202 evaluation board is configured with a USBfriendly interface to program the gain of the ADL5202. The software GUI (see Figure 11) allows users to select a particular frequency to write to the device and also
to read back data from the SDO pin that shows the currently programmed filter setting. The software setup files can be downloaded from the ADL5202 product page at www.analog.com.


Figure 11. Evaluation Control Software

## SCHEMATICS AND ARTWORK



Figure 12. Evaluation Board Schematic


Figure 13.RF Output Detail

## Preliminary Technical Data

## OUTLINE DIMENSIONS

40-Lead Lead Frame Chip Scale Package [LFCSP_WQ] $6 \times 6 \mathrm{~mm}$ Body, Very Very Thin Quad (CP-40-10)
Dimensions shown in millimeters

*COMPLIANT TO JEDEC STANDARDSMO-220-WJUD-6
Figure 14. 40-Lead Lead Frame Chip Scale Package [LFCSP_WQ]
$6 \mathrm{~mm} \times 6 \mathrm{~mm}$ Body, Very Thin Quad
(CP-40-10)
Dimensions shown in millimeters

## ORDERING GUIDE

| Model | Temperature Range | Package Description | Package Option |
| :--- | :--- | :--- | :--- |
| ADL5202XCPZ-R7 ${ }^{1}$ | $-40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$ | 40 Lead LFCSP_WQ, 7"Reel | CP-40-10 |
| ADL5202XCPZ-WP $^{1}$ | $-40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$ | 40 Lead LFCSP_WQ, Waffle Pack | CP-40-10 |
| ADL5202-EVALZ ${ }^{1}$ | $-40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$ | Evaluation Board |  |

${ }^{1} \mathrm{Z}=$ RoHS Compliant Part


[^0]:    ${ }^{1}$ Exposed Paddle

