

**OBSOLETE PRODUCT  
POSSIBLE SUBSTITUTE PRODUCT  
HGTG40N60C3**

**75A, 600V, Rugged, UFS Series N-Channel IGBT**

This family of IGBTs was designed for optimum performance in the demanding world of motor control operation as well as other high voltage switching applications. These devices demonstrate rugged performance capability when subjected to harsh Short Circuit Withstand Time (SCWT) conditions. The parts have Ultrafast (UFS) switching speed while the on-state conduction losses have been kept at a low level.

The electrical specifications include typical Turn-On and Turn-Off dv/dt ratings. These ratings and the Turn-On ratings include the effect of the diode in the test circuit (Figure 15). The data was obtained with the diode at the same T<sub>J</sub> as the IGBT under test.

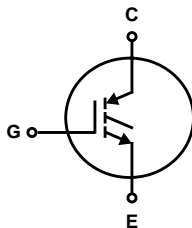
Formerly development type TA49049.

**Ordering Information**

PART NUMBER	PACKAGE	BRAND
HGTG40N60C3R	TO-247	40N60C3R

NOTE: When ordering, use the entire part number.

**Symbol**

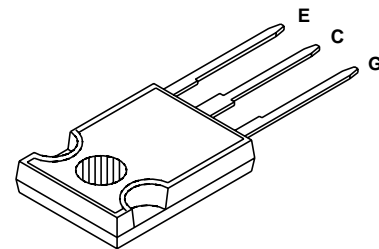


**Features**

- 75A, 600V at T<sub>C</sub> = 25°C
- 600V Switching SOA Capability
- Typical Fall Time at T<sub>J</sub> = 150°C . . . . . 170ns
- Short Circuit Rating at T<sub>J</sub> = 150°C . . . . . 10μs
- Low Conduction Loss

**Packaging**

**JEDEC STYLE TO-247**



**INTERSIL CORPORATION IGBT PRODUCT IS COVERED BY ONE OR MORE OF THE FOLLOWING U.S. PATENTS:**

4,364,073	4,417,385	4,430,792	4,443,931	4,466,176	4,516,143	4,532,534	4,587,713
4,598,461	4,605,948	4,620,211	4,631,564	4,639,754	4,639,762	4,641,162	4,644,637
4,682,195	4,684,413	4,694,313	4,717,679	4,743,952	4,783,690	4,794,432	4,801,986
4,803,533	4,809,045	4,809,047	4,810,665	4,823,176	4,837,606	4,860,080	4,883,767
4,888,627	4,890,143	4,901,127	4,904,609	4,933,740	4,963,951	4,969,027	

# HGTG40N60C3R

## Absolute Maximum Ratings $T_C = 25^\circ\text{C}$ , Unless Otherwise Specified

	HGTG40N60C3R	UNITS
Collector to Emitter Voltage	600	V
Collector Current Continuous		
At $T_C = 25^\circ\text{C}$	75	A
At $T_C = 110^\circ\text{C}$	40	A
Collector Current Pulsed (Note 1)	200	A
Gate to Emitter Voltage Continuous	$\pm 20$	V
Gate to Emitter Voltage Pulsed	$\pm 30$	V
Switching Safe Operating Area at $T_C = 150^\circ\text{C}$	200A at 600V	
Power Dissipation Total at $T_C = 25^\circ\text{C}$	291	W
Power Dissipation Derating $T_C > 25^\circ\text{C}$	2.33	W/ $^\circ\text{C}$
Operating and Storage Junction Temperature Range	-55 to 150	$^\circ\text{C}$
Maximum Lead Temperature for Soldering	260	$^\circ\text{C}$
Reverse Voltage Avalanche Energy	100	mJ
Short Circuit Withstand Time (Note 2) at $V_{GE} = 15\text{V}$	10	$\mu\text{s}$

CAUTION: Stresses above those listed in "Absolute Maximum Ratings" may cause permanent damage to the device. This is a stress only rating and operation of the device at these or any other conditions above those indicated in the operational sections of this specification is not implied.

### NOTES:

1. Repetitive Rating: Pulse width limited by maximum junction temperature.
2.  $V_{CE(PK)} = 440\text{V}$ ,  $T_J = 150^\circ\text{C}$ ,  $R_G = 3\Omega$ .

## Electrical Specifications $T_C = 25^\circ\text{C}$ , Unless Otherwise Specified

PARAMETER	SYMBOL	TEST CONDITIONS	MIN	TYP	MAX	UNITS	
Collector to Emitter Breakdown Voltage	$BV_{CES}$	$I_C = 250\mu\text{A}$ , $V_{GE} = 0\text{V}$	600	-	-	V	
Collector to Emitter Leakage Current	$I_{CES}$	$V_{CE} = BV_{CES}$	$T_C = 25^\circ\text{C}$	-	-	250	$\mu\text{A}$
			$T_C = 150^\circ\text{C}$	-	-	4.0	mA
Collector to Emitter Saturation Voltage	$V_{CE(SAT)}$	$I_C = I_{C110}$ , $V_{GE} = 15\text{V}$	$T_C = 25^\circ\text{C}$	-	1.8	2.2	V
			$T_C = 150^\circ\text{C}$	-	2.0	2.5	V
Gate to Emitter Threshold Voltage	$V_{GE(TH)}$	$I_C = 250\mu\text{A}$ , $V_{CE} = V_{GE}$	4.5	6.2	7.5	V	
Gate to Emitter Leakage Current	$I_{GES}$	$V_{GE} = \pm 20\text{V}$	-	-	$\pm 100$	nA	
Switching SOA (See Figure 2)	SSOA	$T_J = 150^\circ\text{C}$ , $R_G = 3\Omega$ , $V_{GE} = 15\text{V}$ , $V_{CE(PK)} = 600\text{V}$ , $L = 100\mu\text{H}$	200	-	-	A	
Gate to Emitter Plateau Voltage	$V_{GEP}$	$I_C = I_{C110}$ , $V_{CE} = 0.5 BV_{CES}$	-	9.8	-	V	
On-State Gate Charge	$Q_{g(ON)}$	$I_C = I_{C110}$ , $V_{CE} = 0.5 BV_{ES}$	$V_{GE} = 15\text{V}$	-	230	330	nC
			$V_{GE} = 20\text{V}$	-	330	430	nC
Current Turn-On Delay Time	$t_{d(ON)I}$	$T_J = 150^\circ\text{C}$	-	56	-	ns	
Current Rise Time	$t_{rI}$	$I_{CE} = I_{C110}$	-	75	-	ns	
Current Turn-Off Delay Time	$t_{d(OFF)I}$	$V_{CE(PK)} = 0.8 BV_{CES}$	-	265	500	ns	
Current Fall Time	$t_{fI}$	$V_{GE} = 15\text{V}$	-	170	400	ns	
Turn-Off Voltage $dv/dt$ (Note 3)	$dV_{CE}/dt$	$R_G = 3\Omega$	-	1.9	-	V/ns	
Turn-On Voltage $dv/dt$ (Note 3)	$dV_{CE}/dt$	$L = 500\mu\text{H}$	-	6.8	-	V/ns	
Turn-On Energy (Note 4)	$E_{ON}$	Diode used in test circuit	-	3.5	-	mJ	
Turn-Off Energy (Note 5)	$E_{OFF}$	RHRP30120 at $150^\circ\text{C}$	-	2.5	-	mJ	
Thermal Resistance Junction to Case	$R_{\theta JC}$		-	-	0.43	$^\circ\text{C}/\text{W}$	

### NOTES:

3.  $dV_{CE}/dt$  depends on the diode used and the temperature of the diode.
4. Turn-On Energy Loss ( $E_{ON}$ ) includes losses due to the diode recovery and is defined as the integral of the instantaneous power loss starting at the leading edge of the input pulse and ending at the point where the collector voltage equals  $V_{CE(ON)}$ . This value of  $E_{ON}$  was obtained with a RHRP30120 diode at  $T_J = 150^\circ\text{C}$ . A different diode or temperature will result in a different  $E_{ON}$ . For example with diode at  $T_J = 25^\circ\text{C}$ ,  $E_{ON}$  is about one half the value of  $E_{ON}$  with diode at  $T_J = 150^\circ\text{C}$ .
5. Turn-Off Energy Loss ( $E_{OFF}$ ) is defined as the integral of the instantaneous power loss starting at the trailing edge of the input pulse and ending at the point where the collector current equals zero ( $I_{CE} = 0\text{A}$ ). All devices were tested per JEDEC standard No. 24-1 Method for Measurement of Power Device Turn-Off Switching Loss. This test method produces the true total Turn-Off Energy Loss.

Typical Performance Curves

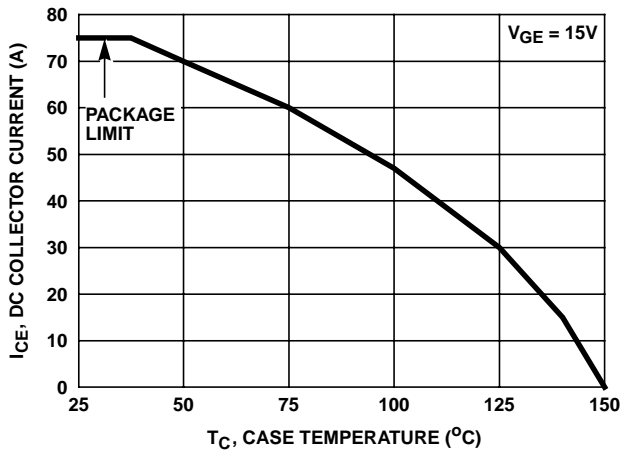


FIGURE 1. DC COLLECTOR CURRENT vs CASE TEMPERATURE

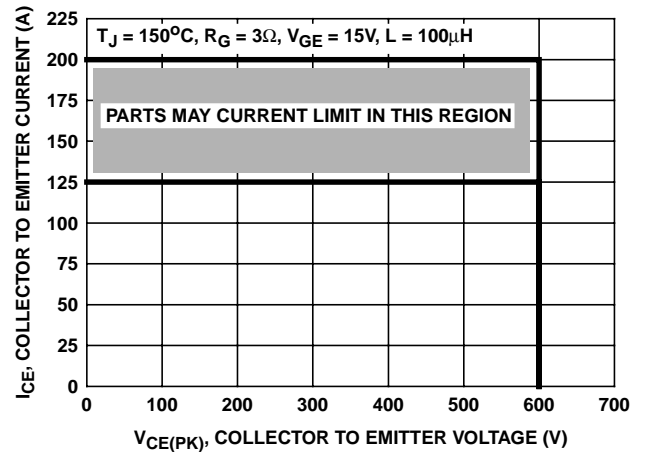


FIGURE 2. SWITCHING SAFE OPERATING AREAS

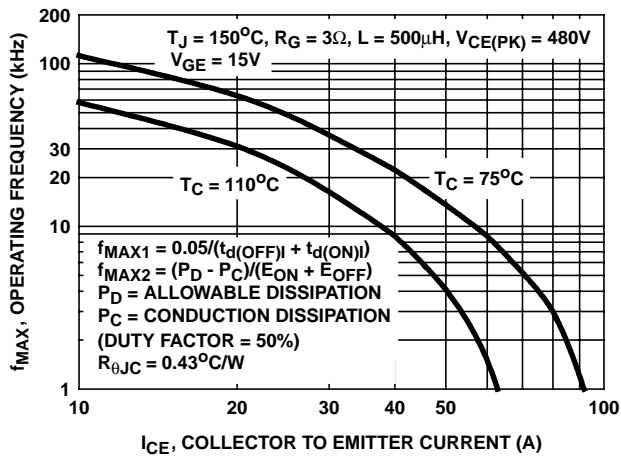


FIGURE 3. OPERATING FREQUENCY vs COLLECTOR TO EMITTER CURRENT

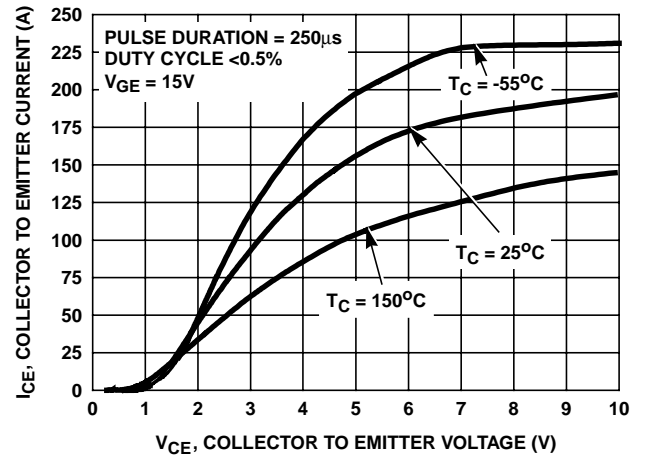


FIGURE 4. COLLECTOR TO EMITTER ON-STATE VOLTAGE

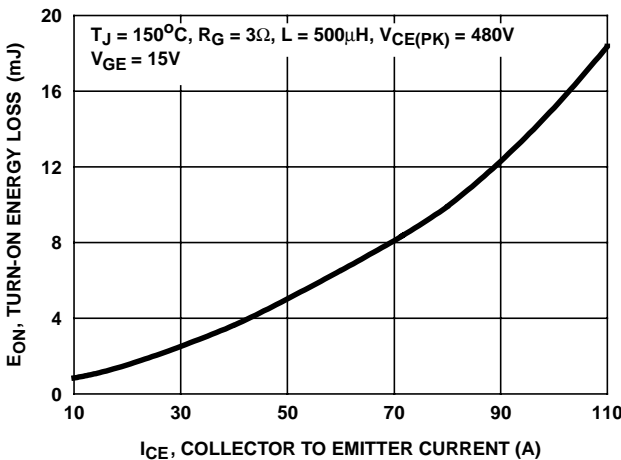


FIGURE 5. TURN-ON ENERGY LOSS vs COLLECTOR TO EMITTER CURRENT

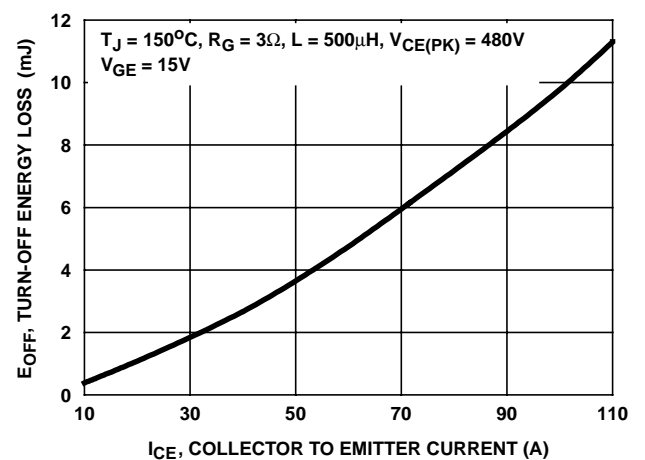


FIGURE 6. TURN-OFF ENERGY LOSS vs COLLECTOR TO EMITTER CURRENT

Typical Performance Curves (Continued)

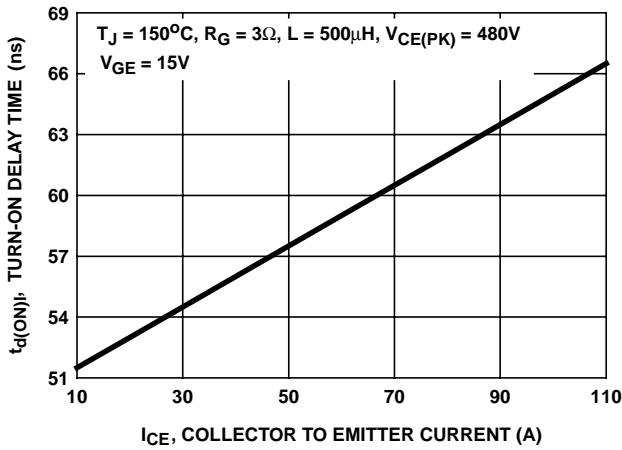


FIGURE 7. TURN-ON DELAY TIME vs COLLECTOR TO EMITTER CURRENT

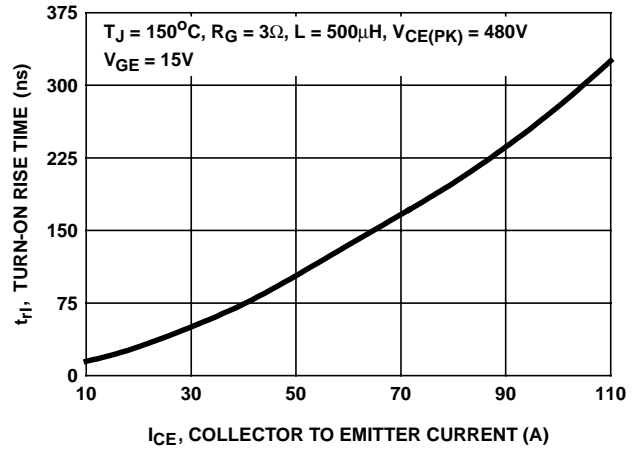


FIGURE 8. TURN-ON RISE TIME vs COLLECTOR TO EMITTER CURRENT

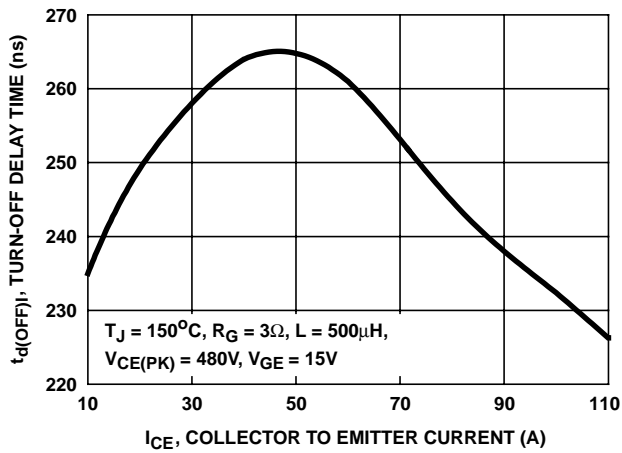


FIGURE 9. TURN-OFF DELAY TIME vs COLLECTOR TO EMITTER CURRENT

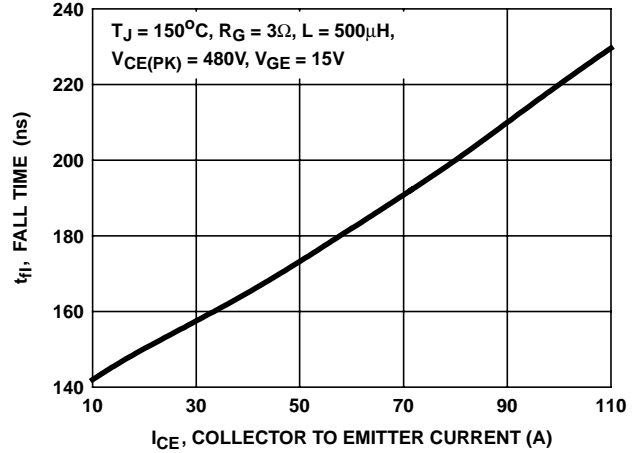


FIGURE 10. TURN-OFF FALL TIME vs COLLECTOR TO EMITTER CURRENT

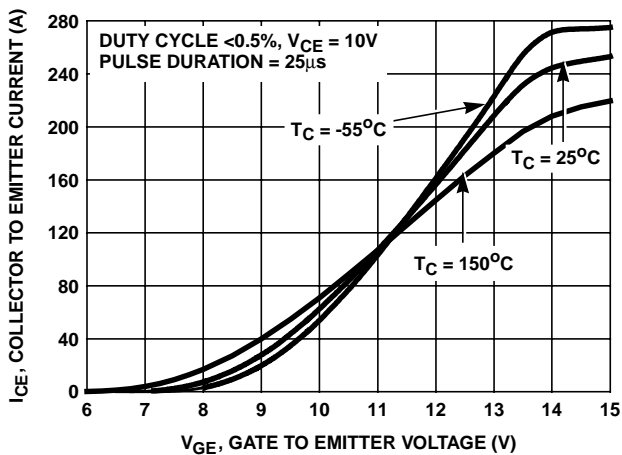


FIGURE 11. TRANSFER CHARACTERISTICS

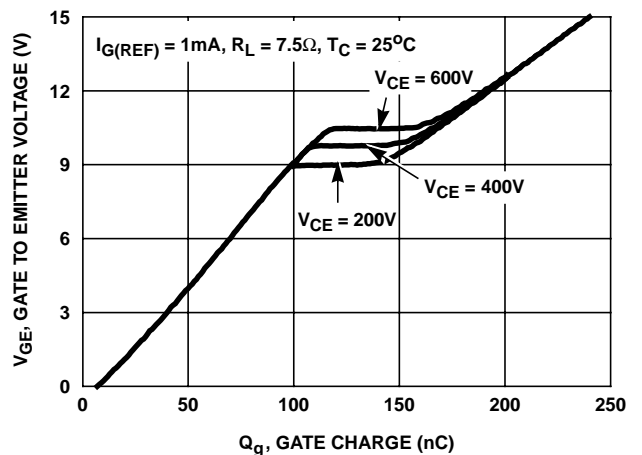


FIGURE 12. GATE CHARGE WAVEFORMS

Typical Performance Curves (Continued)

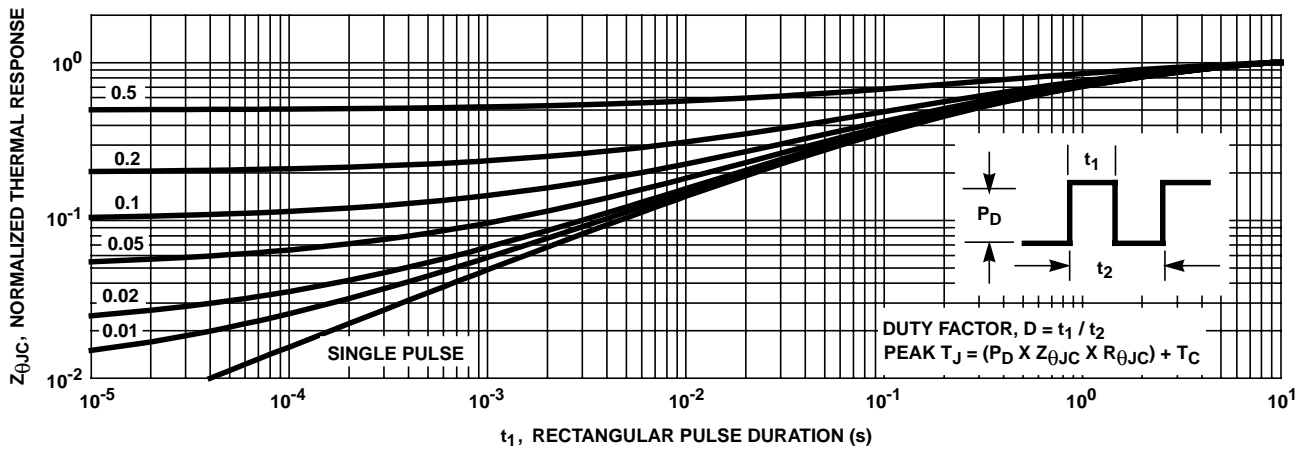


FIGURE 13. IGBT NORMALIZED TRANSIENT THERMAL RESPONSE, JUNCTION TO CASE

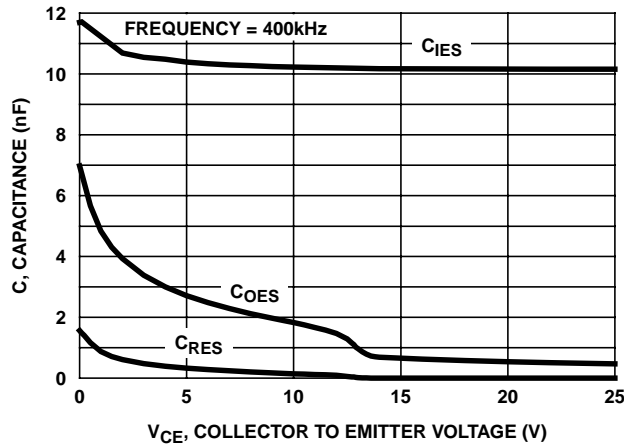


FIGURE 14. CAPACITANCE vs. COLLECTOR TO EMITTER VOLTAGE

Test Circuit and Waveforms

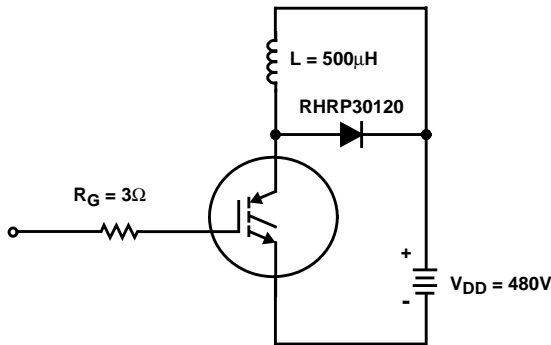


FIGURE 15. INDUCTIVE SWITCHING TEST CIRCUIT

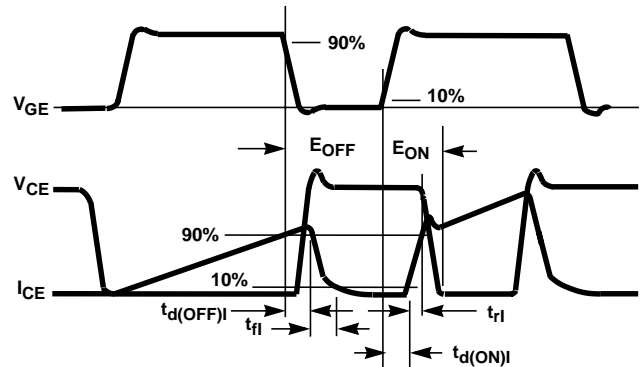


FIGURE 16. SWITCHING TEST WAVEFORMS

## Handling Precautions for IGBTs

Insulated Gate Bipolar Transistors are susceptible to gate-insulation damage by the electrostatic discharge of energy through the devices. When handling these devices, care should be exercised to assure that the static charge built in the handler's body capacitance is not discharged through the device. With proper handling and application procedures, however, IGBTs are currently being extensively used in production by numerous equipment manufacturers in military, industrial and consumer applications, with virtually no damage problems due to electrostatic discharge. IGBTs can be handled safely if the following basic precautions are taken:

1. Prior to assembly into a circuit, all leads should be kept shorted together either by the use of metal shorting springs or by the insertion into conductive material such as "ECCOSORBD LD26™" or equivalent.
2. When devices are removed by hand from their carriers, the hand being used should be grounded by any suitable means - for example, with a metallic wristband.
3. Tips of soldering irons should be grounded.
4. Devices should never be inserted into or removed from circuits with power on.
5. **Gate Voltage Rating** - Never exceed the gate-voltage rating of  $V_{GEM}$ . Exceeding the rated  $V_{GE}$  can result in permanent damage to the oxide layer in the gate region.
6. **Gate Termination** - The gates of these devices are essentially capacitors. Circuits that leave the gate open-circuited or floating should be avoided. These conditions can result in turn-on of the device due to voltage buildup on the input capacitor due to leakage currents or pickup.
7. **Gate Protection** - These devices do not have an internal monolithic Zener diode from gate to emitter. If gate protection is required an external Zener is recommended.

## Operating Frequency Information

Operating frequency information for a typical device (Figure 3) is presented as a guide for estimating device performance for a specific application. Other typical frequency vs collector current ( $I_{CE}$ ) plots are possible using the information shown for a typical unit in Figures 4, 5, 6, 7 and 9. The operating frequency plot (Figure 3) of a typical device shows  $f_{MAX1}$  or  $f_{MAX2}$  whichever is smaller at each point. The information is based on measurements of a typical device and is bounded by the maximum rated junction temperature.

$f_{MAX1}$  is defined by  $f_{MAX1} = 0.05 / (t_{d(OFF)I} + t_{d(ON)I})$ . Deadtime (the denominator) has been arbitrarily held to 10% of the on-state time for a 50% duty factor. Other definitions are possible.  $t_{d(OFF)I}$  and  $t_{d(ON)I}$  are defined in Figure 16. Device turn-off delay can establish an additional frequency limiting condition for an application other than  $T_{JM}$ .  $t_{d(OFF)}$  is important when controlling output ripple under a lightly loaded condition.

$f_{MAX2}$  is defined by  $f_{MAX2} = (P_D - P_C) / (E_{OFF} + E_{ON})$ . The allowable dissipation ( $P_D$ ) is defined by  $P_D = (T_{JM} - T_C) / R_{\theta JC}$ . The sum of device switching and conduction losses must not exceed  $P_D$ . A 50% duty factor was used (Figure 3) and the conduction losses ( $P_C$ ) are approximated by  $P_C = (V_{CE} \times I_{CE}) / 2$ .

$E_{ON}$  and  $E_{OFF}$  are defined in the switching waveforms shown in Figure 16.  $E_{ON}$  is the integral of the instantaneous power loss ( $I_{CE} \times V_{CE}$ ) during turn-on and  $E_{OFF}$  is the integral of the instantaneous power loss ( $I_{CE} \times V_{CE}$ ) during turn-off. All tail losses are included in the calculation for  $E_{OFF}$ ; i.e., the collector current equals zero ( $I_{CE} = 0$ ).

All Intersil semiconductor products are manufactured, assembled and tested under **ISO9000** quality systems certification.

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