

HEX INVERTING SCHMITT TRIGGER

Each circuit of the HEF40106B functions as an inverter with Schmitt-trigger action. The Schmitt-trigger switches at different points for the positive and negative-going input signals. The difference between the positive-going voltage (V_P) and the negative-going voltage (V_N) is defined as hysteresis voltage (V_H).

This device may be used for enhanced noise immunity or to "square up" slowly changing waveforms.

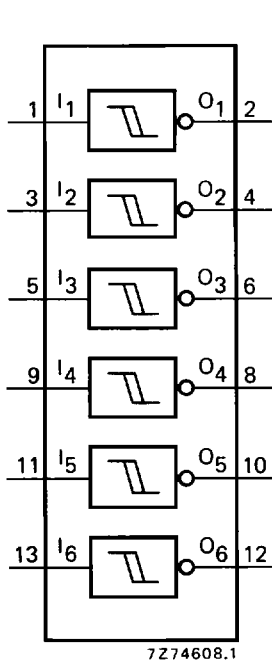


Fig. 1 Functional diagram.

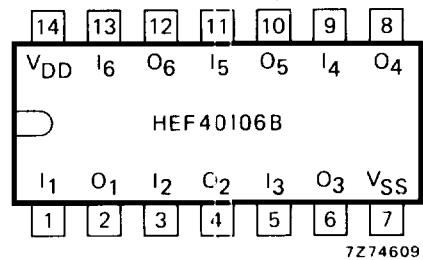


Fig. 2 Pinning diagram.

- HEF40106BP(N): 14-lead DIL; plastic (SOT27-1)
- HEF40106BD(F): 14-lead DIL; ceramic (cerdip) (SOT73)
- HEF40106BT(D): 14-lead SO; plastic (SOT108-1)
- (): Package Designator North America

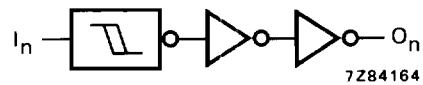


Fig. 3 Logic diagram (one inverter).

FAMILY DATA

I_{DD} LIMITS category GATES

} see Family Specifications

D.C. CHARACTERISTICS

 $V_{SS} = 0\text{ V}$; $T_{amb} = 25\text{ }^{\circ}\text{C}$

	V_{DD} V	symbol	min.	typ.	max.	
Hysteresis voltage	5	V_H	0,5	0,8		V
	10		0,7	1,3		V
	15		0,9	1,8		V
Switching levels positive-going input voltage	5	V_P	2	3,0	3,5	V
	10		3,7	5,8	7	V
	15		4,9	8,3	11	V
negative-going input voltage	5	V_N	1,5	2,2	3	V
	10		3	4,5	6,3	V
	15		4	6,5	10,1	V

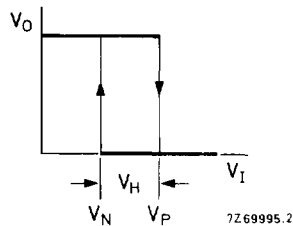
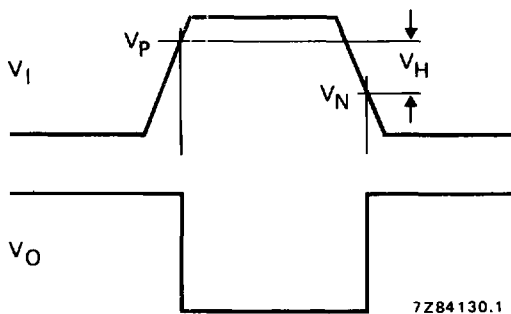


Fig. 4 Transfer characteristic.

Fig. 5 Waveforms showing definition of V_P , V_N and V_H , where V_N and V_P are between limits of 30% and 70%.

A.C. CHARACTERISTICS

 $V_{SS} = 0 \text{ V}$; $T_{amb} = 25 \text{ }^\circ\text{C}$; $C_L = 50 \text{ pF}$; input transition times $\leq 20 \text{ ns}$

	V_{DD} V	symbol	typ.	max.		typical extrapolation formula
Propagation delays $I_n \rightarrow O_n$ HIGH to LOW	5	t_{PHL}	90	180	ns	$63 \text{ ns} + (0,55 \text{ ns/pF}) C_L$
	10		35	70	ns	$24 \text{ ns} + (0,23 \text{ ns/pF}) C_L$
	15		30	60	ns	$22 \text{ ns} + (0,16 \text{ ns/pF}) C_L$
LOW to HIGH	5	t_{PLH}	75	150	ns	$48 \text{ ns} + (0,55 \text{ ns/pF}) C_L$
	10		35	70	ns	$24 \text{ ns} + (0,23 \text{ ns/pF}) C_L$
	15		30	60	ns	$22 \text{ ns} + (0,16 \text{ ns/pF}) C_L$
Output transition times HIGH to LOW	5	t_{THL}	60	120	ns	$10 \text{ ns} + (1,0 \text{ ns/pF}) C_L$
	10		30	60	ns	$9 \text{ ns} + (0,42 \text{ ns/pF}) C_L$
	15		20	40	ns	$6 \text{ ns} + (0,28 \text{ ns/pF}) C_L$
LOW to HIGH	5	t_{TLH}	60	120	ns	$10 \text{ ns} + (1,0 \text{ ns/pF}) C_L$
	10		30	60	ns	$9 \text{ ns} + (0,42 \text{ ns/pF}) C_L$
	15		20	40	ns	$6 \text{ ns} + (0,28 \text{ ns/pF}) C_L$

	V_{DD} V	typical formula for P (μW)	where f_i = input freq. (MHz) f_o = output freq. (MHz) C_L = load capacitance (pF) $\Sigma(f_o C_L)$ = sum of outputs V_{DD} = supply voltage (V)
Dynamic power dissipation per package (P)	5	$2\,300 f_i + \Sigma(f_o C_L) \times V_{DD}^2$	
	10	$9\,000 f_i + \Sigma(f_o C_L) \times V_{DD}^2$	
	15	$20\,000 f_i + \Sigma(f_o C_L) \times V_{DD}^2$	

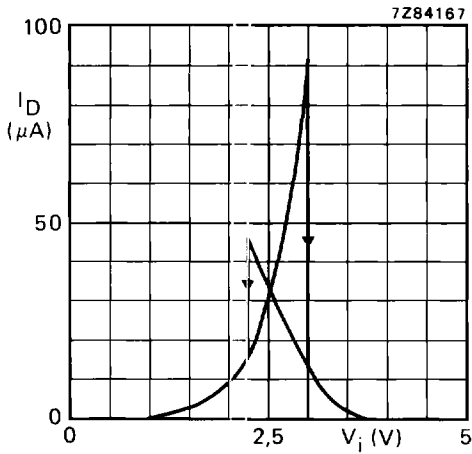


Fig. 6 Typical drain current as a function of input voltage; $V_{DD} = 5\text{ V}$; $T_{amb} = 25\text{ }^\circ\text{C}$.

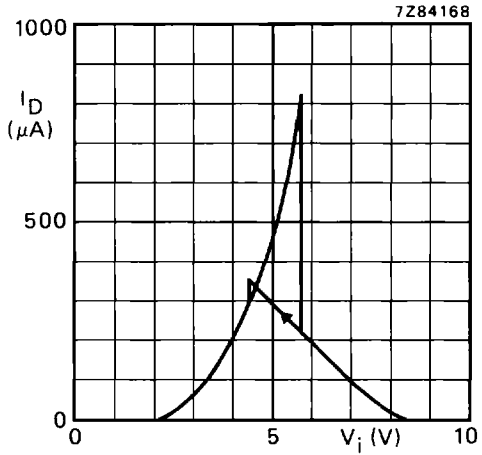


Fig. 7 Typical drain current as a function of input voltage; $V_{DD} = 10\text{ V}$; $T_{amb} = 25\text{ }^\circ\text{C}$.

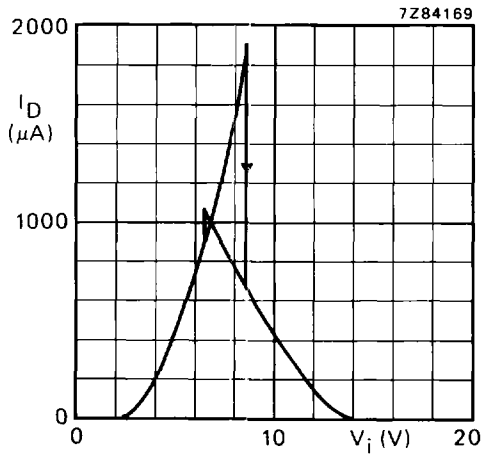


Fig. 8 Typical drain current as a function of input voltage; $V_{DD} = 15\text{ V}$; $T_{amb} = 25\text{ }^\circ\text{C}$.

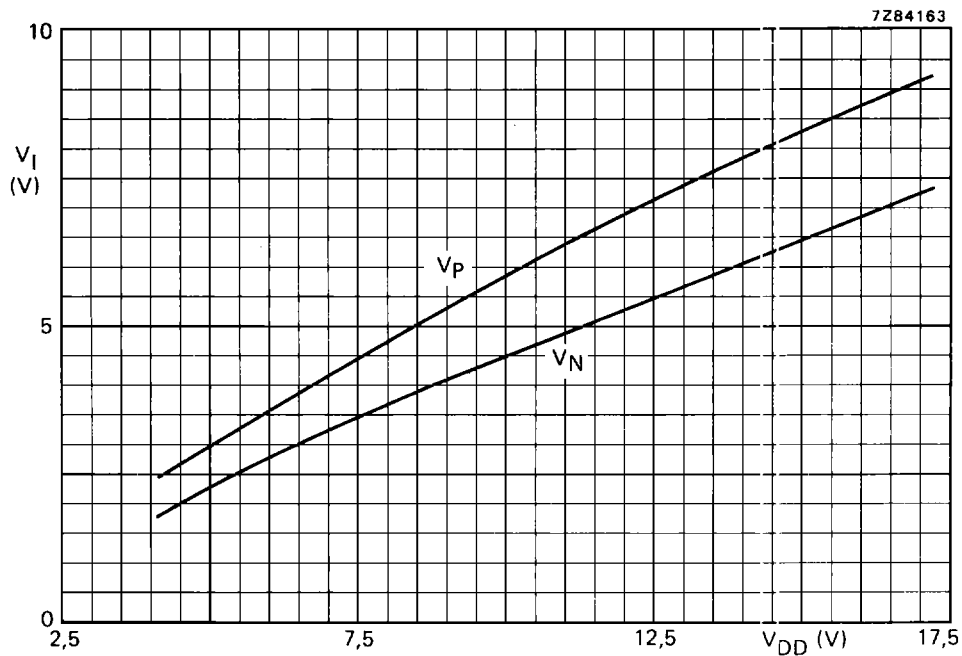


Fig. 9 Typical switching levels as a function of supply voltage V_{DD} ; $T_{amb} = 25\text{ }^{\circ}\text{C}$.

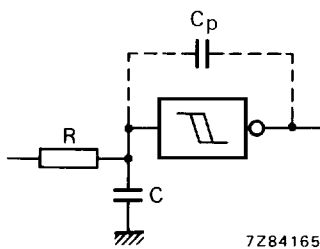


Fig. 10 Schmitt trigger driven via a high impedance ($R > 1\text{ k}\Omega$).

If a Schmitt trigger is driven via a high impedance ($R > 1\text{ k}\Omega$) then it is necessary to incorporate a capacitor C of such value that: $\frac{C}{C_p} > \frac{V_{DD}-V_{SS}}{V_H}$, otherwise oscillation can occur on the edges of a pulse.

C_p is the external parasitic capacitance between input and output; the value depends on the circuit board layout.

APPLICATION INFORMATION

Some examples of applications for the HEF40106B are:

- Wave and pulse shapers
- Astable multivibrators
- Monostable multivibrators.

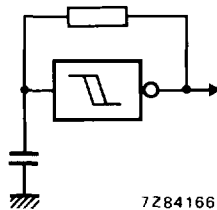


Fig. 11 The HEF40106B used as an astable multivibrator.