

## Thermally Enhanced, Fully Integrated, Hall Effect-Based Linear Current Sensor IC with 100 $\mu\Omega$ Current Conductor

### Features and Benefits

- Industry-leading noise performance through proprietary amplifier and filter design techniques
- Integrated shield greatly reduces capacitive coupling from current conductor to die due to high dV/dt signals, and prevents offset drift in high-side, high voltage applications
- Total output error improvement through gain and offset trim over temperature
- Small package size, with easy mounting capability
- Monolithic Hall IC for high reliability
- Ultra-low power loss: 100  $\mu\Omega$  internal conductor resistance
- Galvanic isolation allows use in economical, high-side current sensing in high voltage systems
- 3.0 to 5.5 V, single supply operation

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### Package: 5-pin package



Additional leadforms available for qualifying volumes

### Description

The Allegro® ACS758 family of current sensor ICs provides economical and precise solutions for AC or DC current sensing. Typical applications include motor control, load detection and management, power supply and DC-to-DC converter control, inverter control, and overcurrent fault detection.

The device consists of a precision, low-offset linear Hall circuit with a copper conduction path located near the die. Applied current flowing through this copper conduction path generates a magnetic field which the Hall IC converts into a proportional voltage. Device accuracy is optimized through the close proximity of the magnetic signal to the Hall transducer. A precise, proportional output voltage is provided by the low-offset, chopper-stabilized BiCMOS Hall IC, which is programmed for accuracy at the factory.

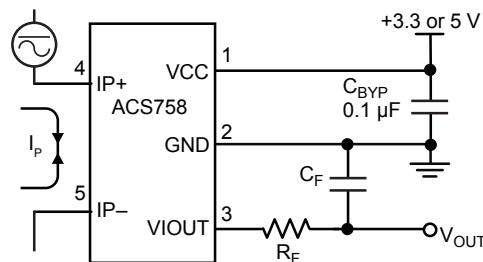
High level immunity to current conductor dV/dt and stray electric fields, offered by Allegro proprietary integrated shield technology, guarantees low output voltage ripple and low offset drift in high-side, high voltage applications.

The output of the device has a positive slope ( $>V_{CC}/2$ ) when an increasing current flows through the primary copper conduction path (from terminal 4 to terminal 5), which is the path used for current sampling. The internal resistance of this conductive path is 100  $\mu\Omega$  typical, providing low power loss.

The thickness of the copper conductor allows survival of the device at high overcurrent conditions. The terminals of the

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### Typical Application



Application 1. The ACS758 outputs an analog signal,  $V_{OUT}$ , that varies linearly with the uni- or bi-directional AC or DC primary sampled current,  $I_p$ , within the range specified.  $C_F$  is for optimal noise management, with values that depend on the application.

# ACS758xCB

## Thermally Enhanced, Fully Integrated, Hall Effect-Based Linear Current Sensor IC with 100 $\mu\Omega$ Current Conductor

### Features and Benefits (continued)

- 120 kHz typical bandwidth
- 3  $\mu\text{s}$  output rise time in response to step input current
- Output voltage proportional to AC or DC currents
- Factory-trimmed for accuracy
- Extremely stable output offset voltage
- Nearly zero magnetic hysteresis

### Description (continued)

conductive path are electrically isolated from the signal leads (pins 1 through 3). This allows the ACS758 family of sensor ICs to be used in applications requiring electrical isolation without the use of opto-isolators or other costly isolation techniques.

The device is fully calibrated prior to shipment from the factory. The ACS758 family is lead (Pb) free. All leads are plated with 100% matte tin, and there is no Pb inside the package. The heavy gauge leadframe is made of oxygen-free copper.

### Selection Guide

Part Number <sup>1</sup>	Package		Primary Sampled Current , $I_P$ (A)	Sensitivity Sens (Typ.) (mV/A)	$T_{OP}$ (°C)	Packing <sup>2</sup>
	Terminals	Signal Pins				
ACS758LCB-050B-PFF-T	Formed	Formed	$\pm 50$	40	-40 to 150	34 pieces per tube
ACS758LCB-100B-PFF-T	Formed	Formed	$\pm 100$	20		
ACS758KCB-150B-PFF-T	Formed	Formed	$\pm 150$	13.3	-40 to 125	
ACS758KCB-150B-PSS-T	Straight	Straight				
ACS758ECB-200B-PFF-T	Formed	Formed	$\pm 200$	10	-40 to 85	
ACS758ECB-200B-PSS-T	Straight	Straight				

<sup>1</sup>Additional leadform options available for qualified volumes

<sup>2</sup>Contact Allegro for additional packing options.



### Absolute Maximum Ratings

Characteristic	Symbol	Notes	Rating	Units
Forward Supply Voltage	$V_{CC}$		8	V
Reverse Supply Voltage	$V_{RCC}$		-0.5	V
Working Voltage for Reinforced Isolation	$V_{WORKING}$	Voltage applied between pins 1-3 and 4-5; tested at 3000 VAC for 1 minute according to UL standard 60950-1	353	VAC
Forward Output Voltage	$V_{IOUT}$		28	V
Reverse Output Voltage	$V_{RIOUT}$		-0.5	V
Output Source Current	$I_{OUT(SOURCE)}$	VIOUT to GND	3	mA
Output Sink Current	$I_{OUT(SINK)}$	VCC to VIOUT	1	mA
Nominal Operating Ambient Temperature	$T_{OP}$	Range E	-40 to 85	°C
		Range K	-40 to 125	°C
		Range L	-40 to 150	°C
Maximum Junction	$T_J(max)$		165	°C
Storage Temperature	$T_{stg}$		-65 to 165	°C

### Typical Overcurrent Capabilities<sup>1,2</sup>

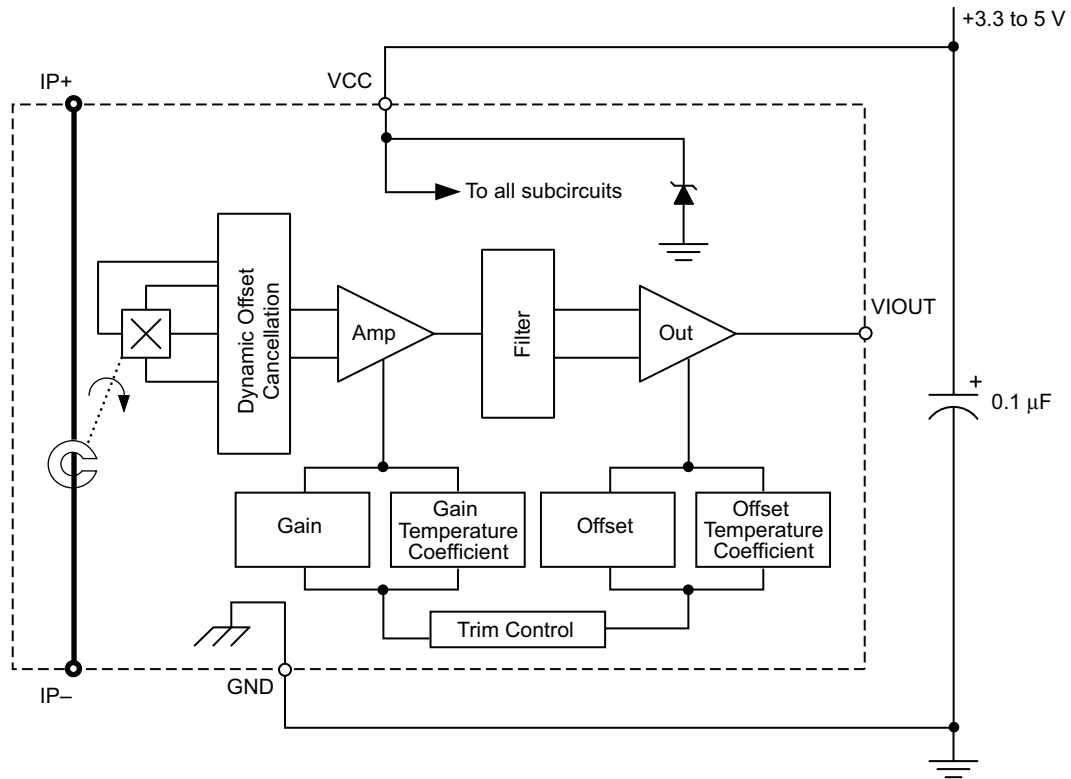
Characteristic	Symbol	Notes	Rating	Units
Overcurrent	$I_{POC}$	$T_A = 25^\circ\text{C}$ , 1s duration, 1% duty cycle	1200	A
		$T_A = 85^\circ\text{C}$ , 1s duration, 1% duty cycle	900	A
		$T_A = 150^\circ\text{C}$ , 1s duration, 1% duty cycle	600	A

<sup>1</sup>Test was done with Allegro evaluation board. The maximum allowed current is limited by  $T_J(max)$  only.

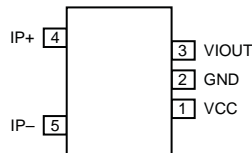
<sup>2</sup>For more overcurrent profiles, please see FAQ on the Allegro website, [www.allegromicro.com](http://www.allegromicro.com).



### Functional Block Diagram



### Pin-out Diagram



### Terminal List Table

Number	Name	Description
1	VCC	Device power supply terminal
2	GND	Signal ground terminal
3	VIOUT	Analog output signal
4	IP+	Terminal for current being sampled
5	IP-	Terminal for current being sampled

### COMMON OPERATING CHARACTERISTICS<sup>1</sup> valid at $T_{OP} = -40^{\circ}\text{C}$ to $150^{\circ}\text{C}$ and $V_{CC} = 5\text{ V}$ , unless otherwise specified

Characteristic	Symbol	Test Conditions	Min.	Typ.	Max.	Units
Supply Voltage	$V_{CC}$		3	5.0	5.5	V
Supply Current	$I_{CC}$	Output open	–	10	13.5	mA
Power-On Delay	$t_{POD}$	$T_A = 25^{\circ}\text{C}$	–	10	–	$\mu\text{s}$
Rise Time <sup>2</sup>	$t_r$	$I_P$ step = 60% of $I_{P+}$ , 10% to 90% rise time, $T_A = 25^{\circ}\text{C}$ , $C_{OUT} = 0.47\text{ nF}$	–	3	–	$\mu\text{s}$
Propagation Delay Time <sup>2</sup>	$t_{PROP}$	$T_A = 25^{\circ}\text{C}$ , $C_{OUT} = 0.47\text{ nF}$	–	1	–	$\mu\text{s}$
Response Time	$t_{RESPONSE}$	Measured as sum of $t_{PROP}$ and $t_r$	–	4	–	$\mu\text{s}$
Internal Bandwidth <sup>3</sup>	$BW_i$	–3 dB; $T_A = 25^{\circ}\text{C}$ , $C_{OUT} = 0.47\text{ nF}$	–	120	–	kHz
Output Load Resistance	$R_{LOAD(MIN)}$	VIOUT to GND	4.7	–	–	k $\Omega$
Output Load Capacitance	$C_{LOAD(MAX)}$	VIOUT to GND	–	–	10	nF
Primary Conductor Resistance	$R_{PRIMARY}$	$T_A = 25^{\circ}\text{C}$	–	100	–	$\mu\Omega$
Symmetry <sup>2</sup>	$E_{SYM}$	Over half-scale of $I_P$	99	100	101	%
Quiescent Output Voltage <sup>4</sup>	$V_{IOUT(Q)}$	$I_P = 0\text{ A}$ , $T_A = 25^{\circ}\text{C}$	–	$V_{CC}/2$	–	V
Ratiometry <sup>2</sup>	$V_{RAT}$	$V_{CC} = 4.5$ to $5.5\text{ V}$	–	100	–	%

<sup>1</sup>Device is factory-trimmed at 5 V, for optimal accuracy.

<sup>2</sup>See Characteristic Definitions section of this datasheet.

<sup>3</sup>Calculated using the formula  $BW_i = 0.35 / t_r$ .

<sup>4</sup> $V_{IOUT(Q)}$  may drift over the lifetime of the device by as much as  $\pm 25\text{ mV}$ .

### X050 PERFORMANCE CHARACTERISTICS<sup>1</sup>: $T_{OP} = -40^{\circ}\text{C}$ to $150^{\circ}\text{C}$ , $V_{CC} = 5\text{ V}$ , unless otherwise specified

Characteristic	Symbol	Test Conditions	Min.	Typ.	Max.	Units
Primary Sampled Current	$I_P$		-50	-	50	A
Sensitivity	$Sens_{TA}$	Full scale of $I_P$ applied for 5 ms, $T_A = 25^{\circ}\text{C}$	-	40	-	mV/A
	$Sens_{(TOP)HT}$	Full scale of $I_P$ applied for 5 ms, $T_{OP} = 25^{\circ}\text{C}$ to $150^{\circ}\text{C}$	-	39.4	-	mV/A
	$Sens_{(TOP)LT}$	Full scale of $I_P$ applied for 5 ms, $T_{OP} = -40^{\circ}\text{C}$ to $25^{\circ}\text{C}$	-	41	-	mV/A
Noise <sup>2</sup>	$V_{NOISE}$	$T_A = 25^{\circ}\text{C}$ , 10 nF on VIOUT pin to GND	-	10	-	mV
Nonlinearity	$E_{LIN}$	Up to full scale of $I_P$ , $I_P$ applied for 5 ms	-1	-	1	%
Electrical Offset Voltage <sup>3</sup>	$V_{OE(TA)}$	$I_P = 0\text{ A}$ , $T_A = 25^{\circ}\text{C}$	-	$\pm 5$	-	mV
	$V_{OE(TOP)HT}$	$I_P = 0\text{ A}$ , $T_{OP} = 25^{\circ}\text{C}$ to $150^{\circ}\text{C}$	-	$\pm 15$	-	mV
	$V_{OE(TOP)LT}$	$I_P = 0\text{ A}$ , $T_{OP} = -40^{\circ}\text{C}$ to $25^{\circ}\text{C}$	-	$\pm 35$	-	mV
Magnetic Offset Error	$I_{ERROM}$	$I_P = 0\text{ A}$ , $T_A = 25^{\circ}\text{C}$ , after excursion of 50 A	-	100	-	mA
Total Output Error <sup>4</sup>	$E_{TOT(HT)}$	Over full scale of $I_P$ , $I_P$ applied for 5 ms, $T_{OP} = 25^{\circ}\text{C}$ to $150^{\circ}\text{C}$	-	-1.2	-	%
	$E_{TOT(LT)}$	Over full scale of $I_P$ , $I_P$ applied for 5 ms, $T_{OP} = -40^{\circ}\text{C}$ to $25^{\circ}\text{C}$	-	2	-	%

<sup>1</sup>See Characteristic Performance Data page for parameter distributions over temperature range.

<sup>2</sup> $\pm 3$  sigma noise voltage.

<sup>3</sup> $V_{OE(TOP)}$  drift is referred to ideal  $V_{IOUT(Q)} = 2.5\text{ V}$ .

<sup>4</sup>Percentage of  $I_P$ . Output filtered.

### X100 PERFORMANCE CHARACTERISTICS<sup>1</sup>: $T_{OP} = -40^{\circ}\text{C}$ to $150^{\circ}\text{C}$ , $V_{CC} = 5\text{ V}$ , unless otherwise specified

Characteristic	Symbol	Test Conditions	Min.	Typ.	Max.	Units
Primary Sampled Current	$I_P$		-100	-	100	A
Sensitivity	$Sens_{TA}$	Full scale of $I_P$ applied for 5 ms, $T_A = 25^{\circ}\text{C}$	-	20	-	mV/A
	$Sens_{(TOP)HT}$	Full scale of $I_P$ applied for 5 ms, $T_{OP} = 25^{\circ}\text{C}$ to $150^{\circ}\text{C}$	-	19.75	-	mV/A
	$Sens_{(TOP)LT}$	Full scale of $I_P$ applied for 5 ms, $T_{OP} = -40^{\circ}\text{C}$ to $25^{\circ}\text{C}$	-	20.5	-	mV/A
Noise <sup>2</sup>	$V_{NOISE}$	$T_A = 25^{\circ}\text{C}$ , 10 nF on VIOUT pin to GND	-	6	-	mV
Nonlinearity	$E_{LIN}$	Up to full scale of $I_P$ , $I_P$ applied for 5 ms	-1.25	-	1.25	%
Electrical Offset Voltage <sup>3</sup>	$V_{OE(TA)}$	$I_P = 0\text{ A}$ , $T_A = 25^{\circ}\text{C}$	-	$\pm 5$	-	mV
	$V_{OE(TOP)HT}$	$I_P = 0\text{ A}$ , $T_{OP} = 25^{\circ}\text{C}$ to $150^{\circ}\text{C}$	-	$\pm 20$	-	mV
	$V_{OE(TOP)LT}$	$I_P = 0\text{ A}$ , $T_{OP} = -40^{\circ}\text{C}$ to $25^{\circ}\text{C}$	-	$\pm 20$	-	mV
Magnetic Offset Error	$I_{ERROM}$	$I_P = 0\text{ A}$ , $T_A = 25^{\circ}\text{C}$ , after excursion of 100 A	-	150	-	mA
Total Output Error <sup>4</sup>	$E_{TOT(HT)}$	Over full scale of $I_P$ , $I_P$ applied for 5 ms, $T_{OP} = 25^{\circ}\text{C}$ to $150^{\circ}\text{C}$	-	-1.3	-	%
	$E_{TOT(LT)}$	Over full scale of $I_P$ , $I_P$ applied for 5 ms, $T_{OP} = -40^{\circ}\text{C}$ to $25^{\circ}\text{C}$	-	2.4	-	%

<sup>1</sup>See Characteristic Performance Data page for parameter distributions over temperature range.

<sup>2</sup> $\pm 3$  sigma noise voltage.

<sup>3</sup> $V_{OE(TOP)}$  drift is referred to ideal  $V_{IOUT(Q)} = 2.5\text{ V}$ .

<sup>4</sup>Percentage of  $I_P$ . Output filtered.

### X150 PERFORMANCE CHARACTERISTICS<sup>1</sup>: $T_{OP} = -40^{\circ}\text{C}$ to $125^{\circ}\text{C}$ , $V_{CC} = 5\text{ V}$ , unless otherwise specified

Characteristic	Symbol	Test Conditions	Min.	Typ.	Max.	Units
Primary Sampled Current	$I_P$		-150	-	150	A
Sensitivity	$Sens_{TA}$	Full scale of $I_P$ applied for 5 ms, $T_A = 25^{\circ}\text{C}$	-	13.3	-	mV/A
	$Sens_{(TOP)HT}$	Full scale of $I_P$ applied for 5 ms, $T_{OP} = 25^{\circ}\text{C}$ to $125^{\circ}\text{C}$	-	13.1	-	mV/A
	$Sens_{(TOP)LT}$	Full scale of $I_P$ applied for 5 ms, $T_{OP} = -40^{\circ}\text{C}$ to $25^{\circ}\text{C}$	-	13.5	-	mV/A
Noise <sup>2</sup>	$V_{NOISE}$	$T_A = 25^{\circ}\text{C}$ , 10 nF on VIOOUT pin to GND	-	4	-	mV
Nonlinearity	$E_{LIN}$	Up to full scale of $I_P$ , $I_P$ applied for 5 ms	-1	-	1	%
Electrical Offset Voltage <sup>3</sup>	$V_{OE(TA)}$	$I_P = 0\text{ A}$ , $T_A = 25^{\circ}\text{C}$	-	$\pm 5$	-	mV
	$V_{OE(TOP)HT}$	$I_P = 0\text{ A}$ , $T_{OP} = 25^{\circ}\text{C}$ to $125^{\circ}\text{C}$	-	$\pm 14$	-	mV
	$V_{OE(TOP)LT}$	$I_P = 0\text{ A}$ , $T_{OP} = -40^{\circ}\text{C}$ to $25^{\circ}\text{C}$	-	$\pm 24$	-	mV
Magnetic Offset Error	$I_{ERROM}$	$I_P = 0\text{ A}$ , $T_A = 25^{\circ}\text{C}$ , after excursion of 150 A	-	205	-	mA
Total Output Error <sup>4</sup>	$E_{TOT(HT)}$	Over full scale of $I_P$ , $I_P$ applied for 5 ms, $T_{OP} = 25^{\circ}\text{C}$ to $125^{\circ}\text{C}$	-	-1.8	-	%
	$E_{TOT(LT)}$	Over full scale of $I_P$ , $I_P$ applied for 5 ms, $T_{OP} = -40^{\circ}\text{C}$ to $25^{\circ}\text{C}$	-	1.6	-	%

<sup>1</sup>See Characteristic Performance Data page for parameter distributions over temperature range.

<sup>2</sup> $\pm 3$  sigma noise voltage.

<sup>3</sup> $V_{OE(TOP)}$  drift is referred to ideal  $V_{IOOUT(Q)} = 2.5\text{ V}$ .

<sup>4</sup>Percentage of  $I_P$ . Output filtered.

### X200 PERFORMANCE CHARACTERISTICS<sup>1</sup>: $T_{OP} = -40^{\circ}\text{C}$ to $85^{\circ}\text{C}$ , $V_{CC} = 5\text{ V}$ , unless otherwise specified

Characteristic	Symbol	Test Conditions	Min.	Typ.	Max.	Units
Primary Sampled Current	$I_P$		-200	-	200	A
Sensitivity	$Sens_{TA}$	Full scale of $I_P$ applied for 5 ms, $T_A = 25^{\circ}\text{C}$	-	10	-	mV/A
	$Sens_{(TOP)HT}$	Full scale of $I_P$ applied for 5 ms, $T_{OP} = 25^{\circ}\text{C}$ to $85^{\circ}\text{C}$	-	9.88	-	mV/A
	$Sens_{(TOP)LT}$	Full scale of $I_P$ applied for 5 ms, $T_{OP} = -40^{\circ}\text{C}$ to $25^{\circ}\text{C}$	-	10.13	-	mV/A
Noise <sup>2</sup>	$V_{NOISE}$	$T_A = 25^{\circ}\text{C}$ , 10 nF on VIOOUT pin to GND	-	3	-	mV
Nonlinearity	$E_{LIN}$	Up to full scale of $I_P$ , $I_P$ applied for 5 ms	-1	-	1	%
Electrical Offset Voltage <sup>3</sup>	$V_{OE(TA)}$	$I_P = 0\text{ A}$ , $T_A = 25^{\circ}\text{C}$	-	$\pm 5$	-	mV
	$V_{OE(TOP)HT}$	$I_P = 0\text{ A}$ , $T_{OP} = 25^{\circ}\text{C}$ to $85^{\circ}\text{C}$	-	$\pm 15$	-	mV
	$V_{OE(TOP)LT}$	$I_P = 0\text{ A}$ , $T_{OP} = -40^{\circ}\text{C}$ to $25^{\circ}\text{C}$	-	$\pm 25$	-	mV
Magnetic Offset Error	$I_{ERROM}$	$I_P = 0\text{ A}$ , $T_A = 25^{\circ}\text{C}$ , after excursion of 200 A	-	230	-	mA
Total Output Error <sup>4</sup>	$E_{TOT(HT)}$	Over full scale of $I_P$ , $I_P$ applied for 5 ms, $T_{OP} = 25^{\circ}\text{C}$ to $85^{\circ}\text{C}$	-	-1.2	-	%
	$E_{TOT(LT)}$	Over full scale of $I_P$ , $I_P$ applied for 5 ms, $T_{OP} = -40^{\circ}\text{C}$ to $25^{\circ}\text{C}$	-	1.2	-	%

<sup>1</sup>See Characteristic Performance Data page for parameter distributions over temperature range.

<sup>2</sup> $\pm 3$  sigma noise voltage.

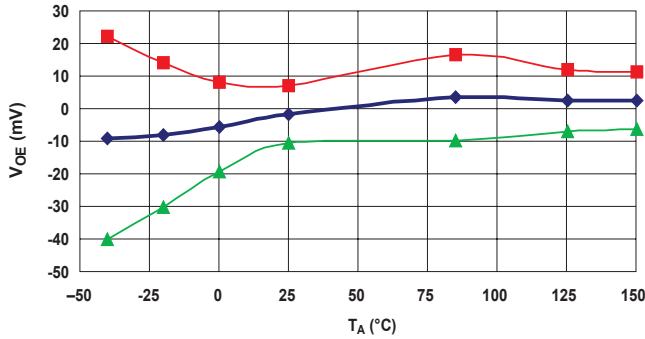
<sup>3</sup> $V_{OE(TOP)}$  drift is referred to ideal  $V_{IOOUT(Q)} = 2.5\text{ V}$ .

<sup>4</sup>Percentage of  $I_P$ . Output filtered.

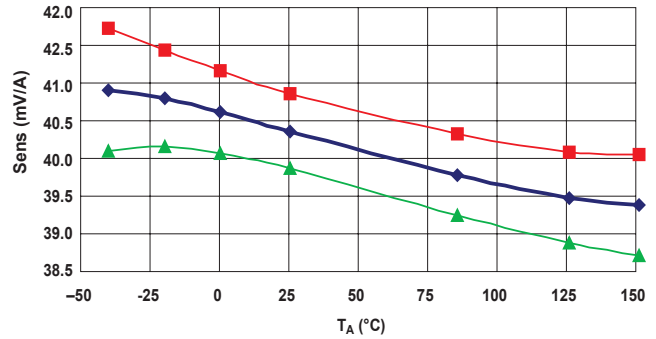
### Characteristic Performance Data Data taken using the ACS758LCB-50B

#### Accuracy Data

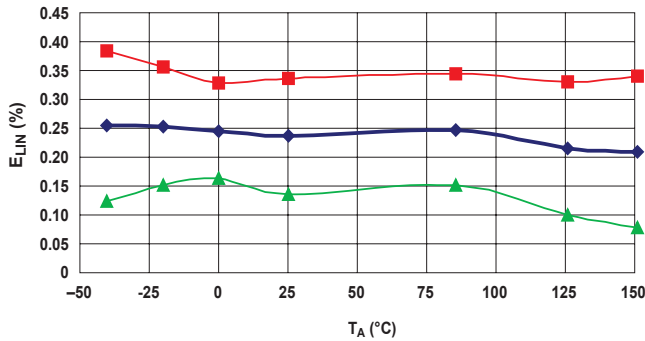
Electrical Offset Voltage versus Ambient Temperature



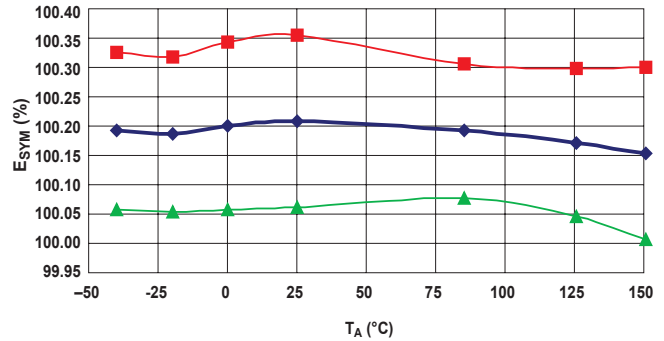
Sensitivity versus Ambient Temperature



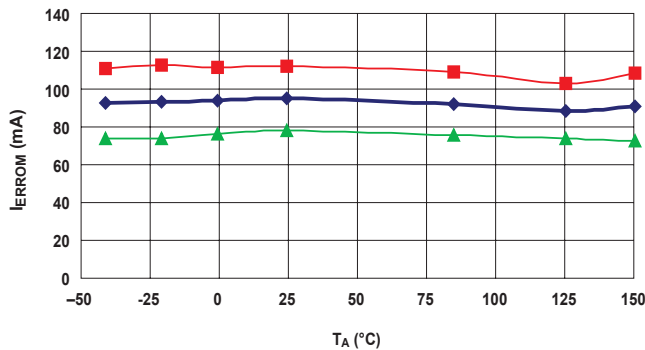
Nonlinearity versus Ambient Temperature



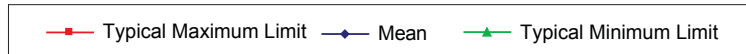
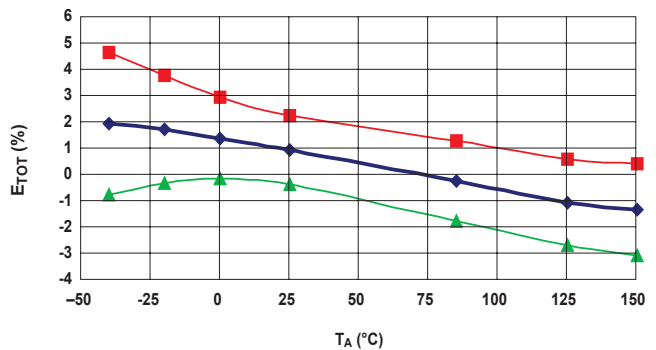
Symmetry versus Ambient Temperature



Magnetic Offset Error versus Ambient Temperature



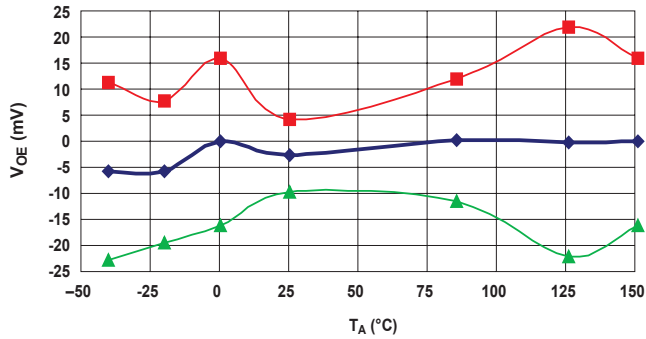
Total Output Error versus Ambient Temperature



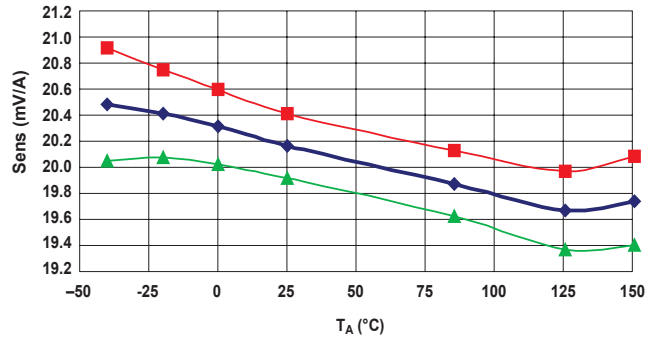
### Characteristic Performance Data Data taken using the ACS758LCB-100B

#### Accuracy Data

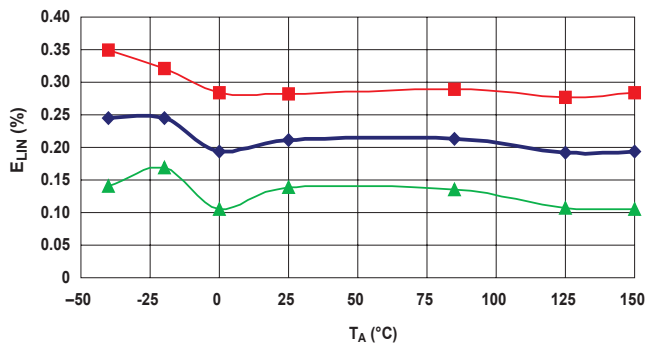
Electrical Offset Voltage versus Ambient Temperature



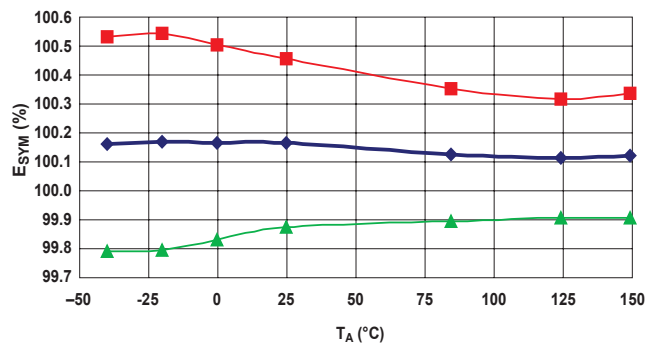
Sensitivity versus Ambient Temperature



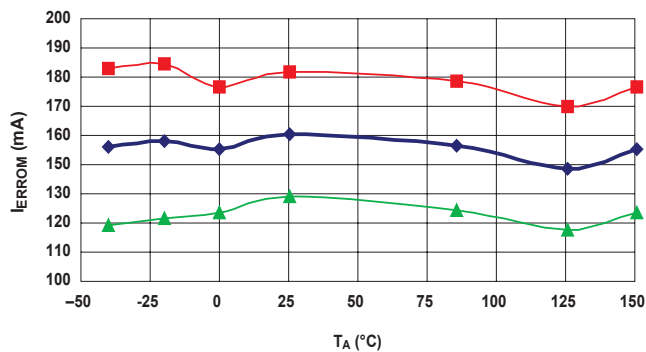
Nonlinearity versus Ambient Temperature



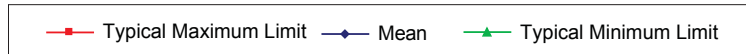
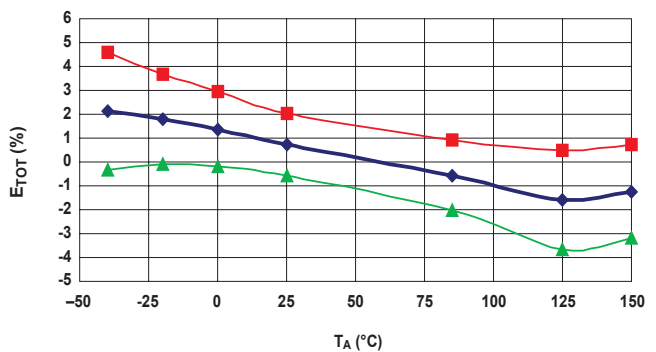
Symmetry versus Ambient Temperature



Magnetic Offset Error versus Ambient Temperature



Total Output Error versus Ambient Temperature

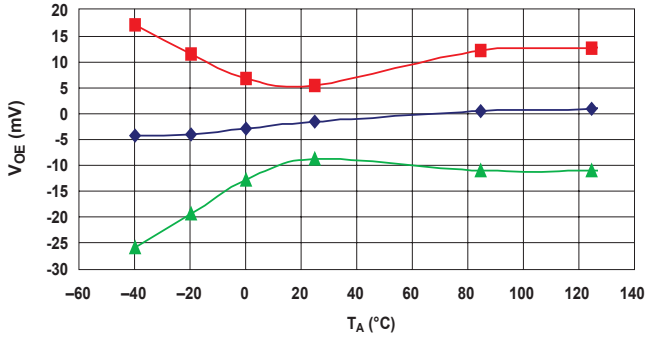




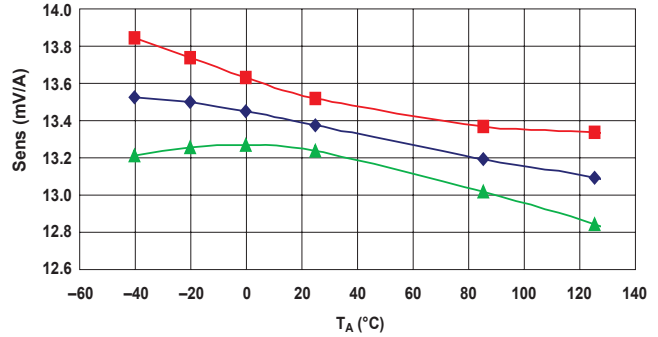
### Characteristic Performance Data Data taken using the ACS758KCB-150B

#### Accuracy Data

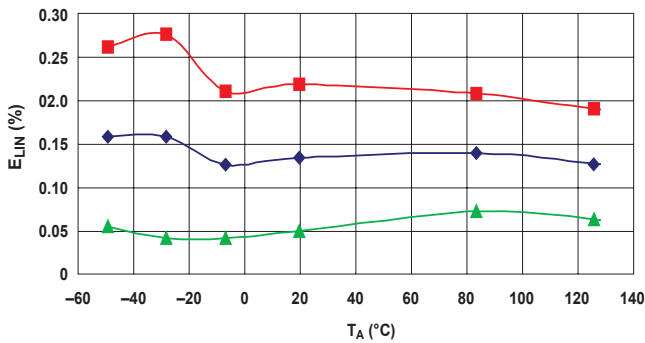
Electrical Offset Voltage versus Ambient Temperature



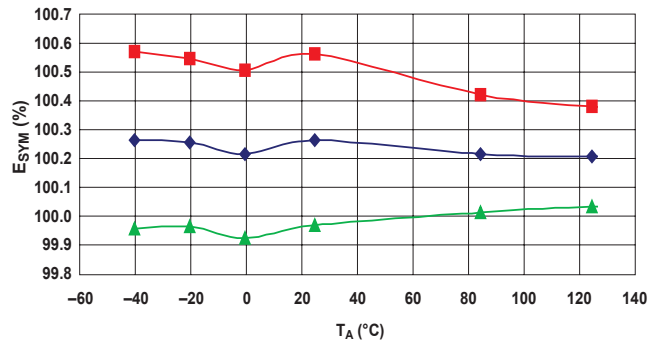
Sensitivity versus Ambient Temperature



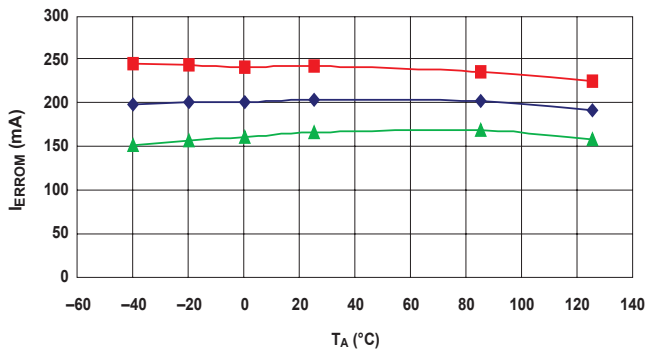
Nonlinearity versus Ambient Temperature



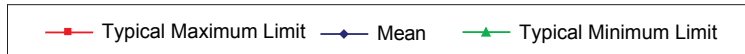
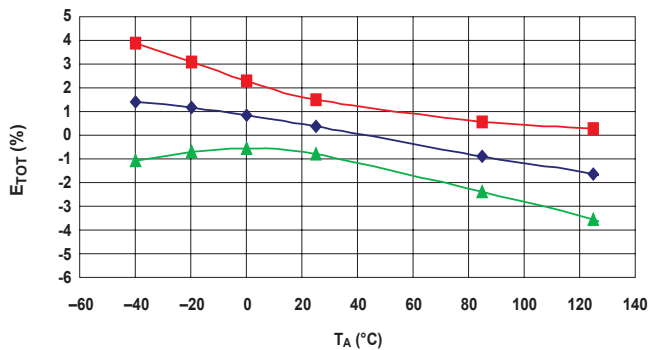
Symmetry versus Ambient Temperature



Magnetic Offset Error versus Ambient Temperature



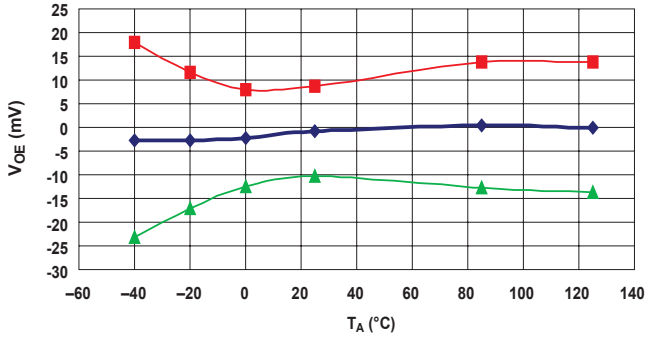
Total Output Error versus Ambient Temperature



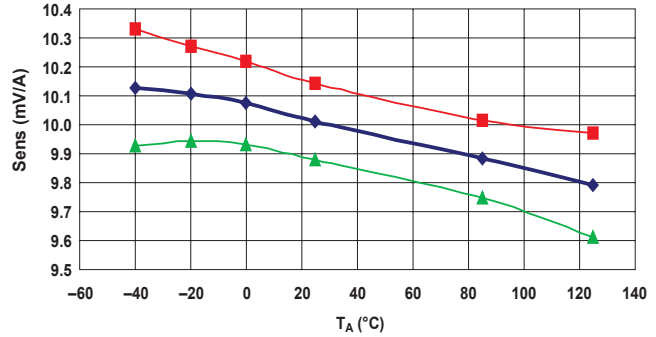
### Characteristic Performance Data Data taken using the ACS758ECB-200B

#### Accuracy Data

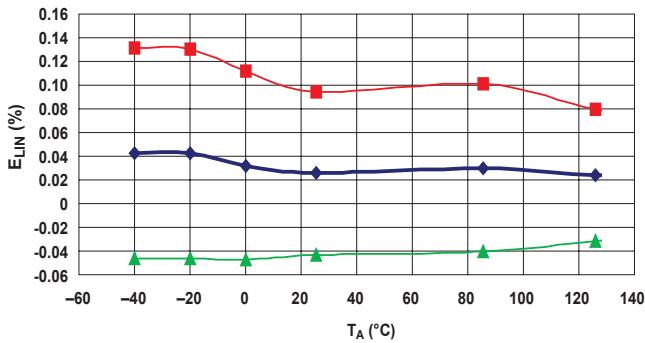
Electrical Offset Voltage versus Ambient Temperature



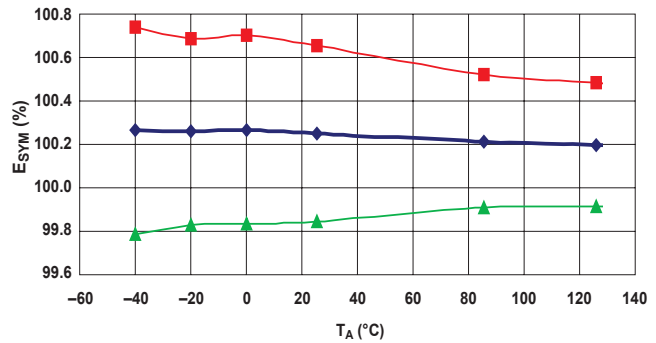
Sensitivity versus Ambient Temperature



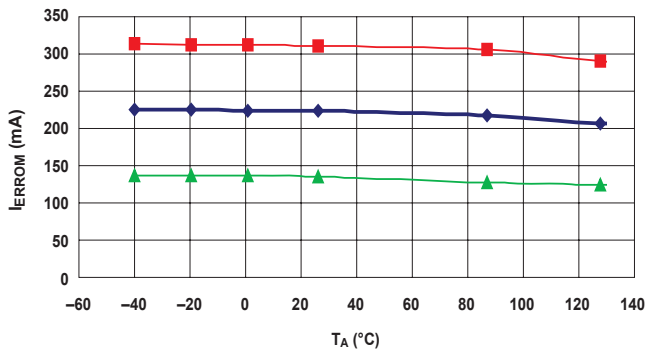
Nonlinearity versus Ambient Temperature



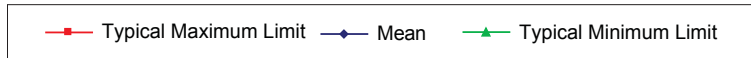
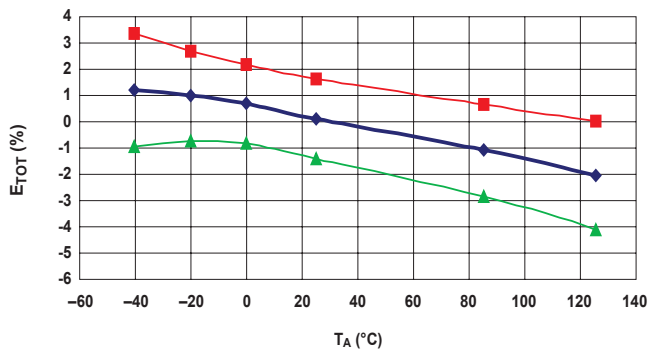
Symmetry versus Ambient Temperature



Magnetic Offset Error versus Ambient Temperature



Total Output Error versus Ambient Temperature

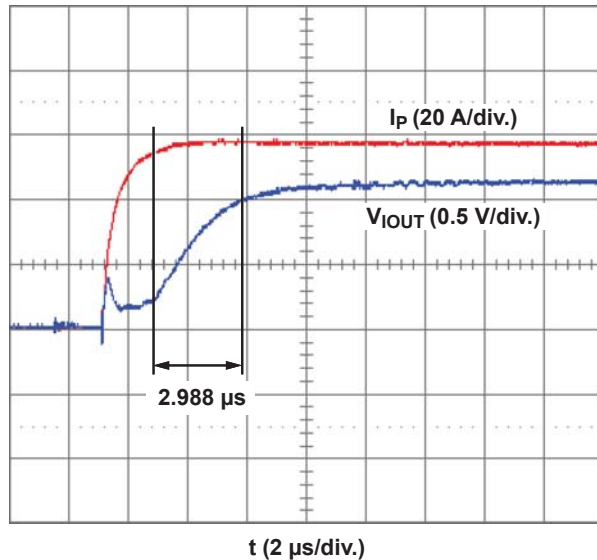


### Characteristic Performance Data

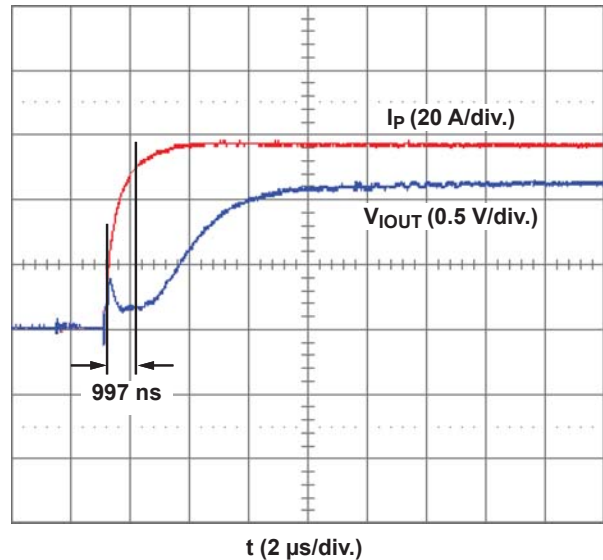
Data taken using the ACS758LCB-100

#### Timing Data

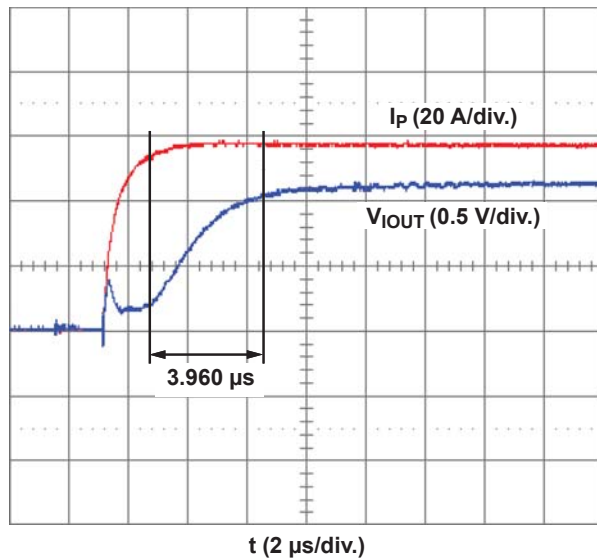
Rise Time



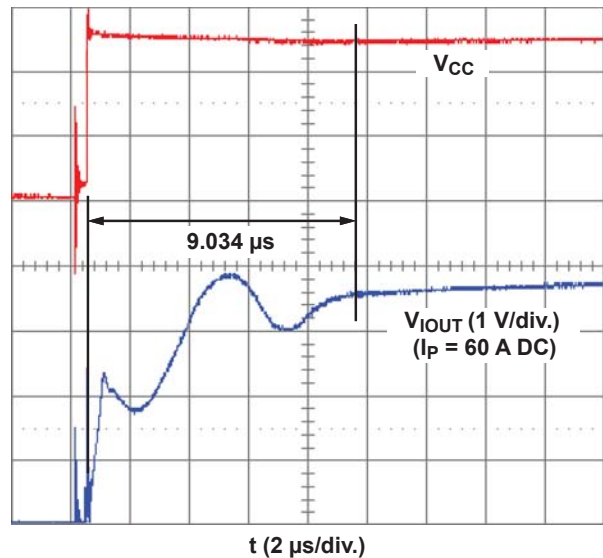
Propagation Delay Time



Response Time



Power-on Delay



### Characteristic Definitions

#### Definitions of Accuracy Characteristics

**Sensitivity (Sens).** The change in device output in response to a 1 A change through the primary conductor. The sensitivity is the product of the magnetic circuit sensitivity (G/A) and the linear IC amplifier gain (mV/G). The linear IC amplifier gain is programmed at the factory to optimize the sensitivity (mV/A) for the half-scale current of the device.

**Noise ( $V_{NOISE}$ ).** The noise floor is derived from the thermal and shot noise observed in Hall elements. Dividing the noise (mV) by the sensitivity (mV/A) provides the smallest current that the device is able to resolve.

**Nonlinearity ( $E_{LIN}$ ).** The degree to which the voltage output from the IC varies in direct proportion to the primary current through its half-scale amplitude. Nonlinearity in the output can be attributed to the saturation of the flux concentrator approaching the half-scale current. The following equation is used to derive the linearity:

$$100 \left\{ 1 - \left[ \frac{\Delta \text{ gain} \times \% \text{ sat} (V_{IOUT\_half\text{-}scale \text{ amperes}} - V_{IOUT(Q)})}{2 (V_{IOUT\_quarter\text{-}scale \text{ amperes}} - V_{IOUT(Q)})} \right] \right\}$$

where

$\Delta$  gain = the gain variation as a function of temperature changes from 25°C,

% sat = the percentage of saturation of the flux concentrator, which becomes significant as the current being sampled approaches half-scale  $\pm I_P$ , and

$V_{IOUT\_half\text{-}scale \text{ amperes}}$  = the output voltage (V) when the sampled current approximates half-scale  $\pm I_P$ .

**Symmetry ( $E_{SYM}$ ).** The degree to which the absolute voltage output from the IC varies in proportion to either a positive or negative half-scale primary current. The following equation is used to derive symmetry:

$$100 \left( \frac{V_{IOUT\_+ \text{ half-scale amperes}} - V_{IOUT(Q)}}{V_{IOUT(Q)} - V_{IOUT\_ - \text{ half-scale amperes}}} \right)$$

**Ratiometry.** The device features a ratiometric output. This means that the quiescent voltage output,  $V_{IOUTQ}$ , and the magnetic sensitivity, Sens, are proportional to the supply voltage,  $V_{CC}$ .

The ratiometric change (%) in the quiescent voltage output is defined as:

$$\Delta V_{IOUTQ(\Delta V)} = \frac{V_{IOUTQ(V_{CC})} / V_{IOUTQ(5V)}}{V_{CC} / 5V} \times 100\%$$

and the ratiometric change (%) in sensitivity is defined as:

$$\Delta \text{Sens}_{(\Delta V)} = \frac{\text{Sens}_{(V_{CC})} / \text{Sens}_{(5V)}}{V_{CC} / 5V} \times 100\%$$

**Quiescent output voltage ( $V_{IOUT(Q)}$ ).** The output of the device when the primary current is zero. For a unipolar supply voltage, it nominally remains at  $V_{CC}/2$ . Thus,  $V_{CC} = 5V$  translates into  $V_{IOUT(Q)} = 2.5V$ . Variation in  $V_{IOUT(Q)}$  can be attributed to the resolution of the Allegro linear IC quiescent voltage trim, magnetic hysteresis, and thermal drift.

**Electrical offset voltage ( $V_{OE}$ ).** The deviation of the device output from its ideal quiescent value of  $V_{CC}/2$  due to nonmagnetic causes.

**Magnetic offset error ( $I_{ERROM}$ ).** The magnetic offset is due to the residual magnetism (remnant field) of the core material. The magnetic offset error is highest when the magnetic circuit has been saturated, usually when the device has been subjected to a full-scale or high-current overload condition. The magnetic offset is largely dependent on the material used as a flux concentrator. The larger magnetic offsets are observed at the lower operating temperatures.

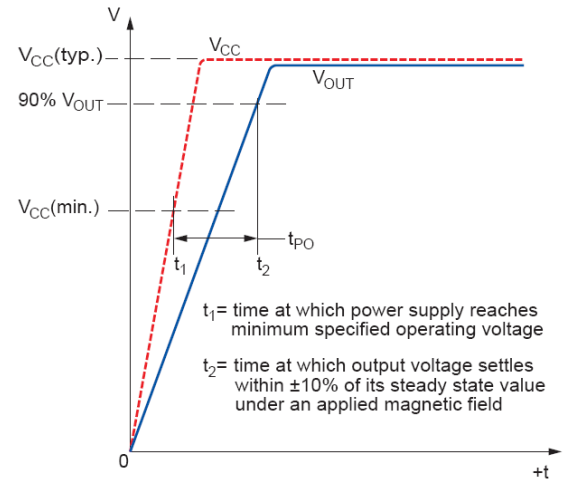
**Total Output Error ( $E_{TOT}$ ).** The maximum deviation of the actual output from its ideal value, also referred to as *accuracy*, illustrated graphically in the output voltage versus current chart on the following page.

$E_{TOT}$  is divided into four areas:

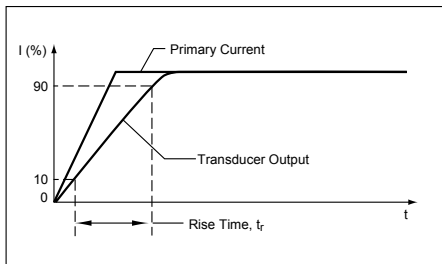
- **0 A at 25°C.** Accuracy at the zero current flow at 25°C, without the effects of temperature.
- **0 A over  $\Delta$  temperature.** Accuracy at the zero current flow including temperature effects.
- **Half-scale current at 25°C.** Accuracy at the the half-scale current at 25°C, without the effects of temperature.
- **Half-scale current over  $\Delta$  temperature.** Accuracy at the half-scale current flow including temperature effects.

### Definitions of Dynamic Response Characteristics

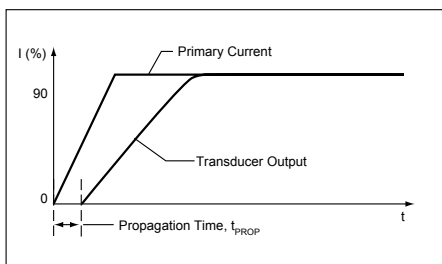
**Power-On Time ( $t_{PO}$ ).** When the supply is ramped to its operating voltage, the device requires a finite time to power its internal components before responding to an input magnetic field. Power-On Time,  $t_{PO}$ , is defined as the time it takes for the output voltage to settle within  $\pm 10\%$  of its steady state value under an applied magnetic field, after the power supply has reached its minimum specified operating voltage,  $V_{CC(min)}$ , as shown in the chart at right.



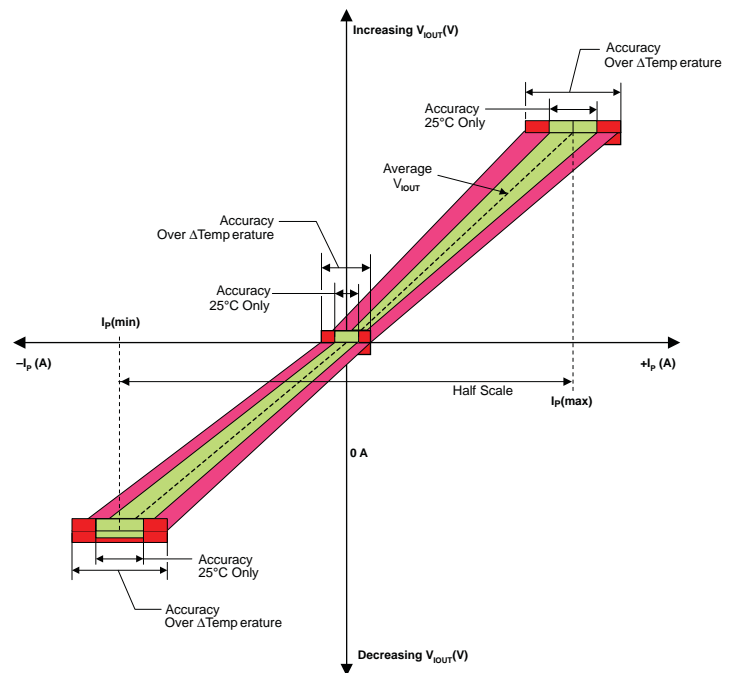
**Rise time ( $t_r$ ).** The time interval between a) when the device reaches 10% of its full scale value, and b) when it reaches 90% of its full scale value. The rise time to a step response is used to derive the bandwidth of the device, in which  $f(-3 \text{ dB}) = 0.35 / t_r$ . Both  $t_r$  and  $t_{RESPONSE}$  are detrimentally affected by eddy current losses observed in the conductive IC ground plane.



**Propagation delay ( $t_{PROP}$ ).** The time required for the device output to reflect a change in the primary current signal. Propagation delay is attributed to inductive loading within the linear IC package, as well as in the inductive loop formed by the primary conductor geometry. Propagation delay can be considered as a fixed time offset and may be compensated.



### Output Voltage versus Sampled Current Total Output Error at 0 A and at Half-Scale Current



### Chopper Stabilization Technique

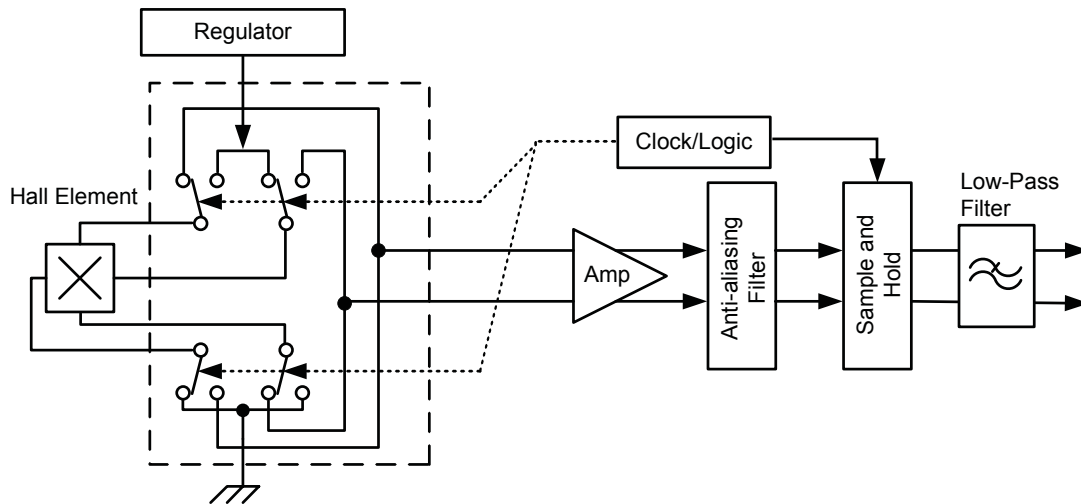
Chopper Stabilization is an innovative circuit technique that is used to minimize the offset voltage of a Hall element and an associated on-chip amplifier. Allegro patented a Chopper Stabilization technique that nearly eliminates Hall IC output drift induced by temperature or package stress effects.

This offset reduction technique is based on a signal modulation-demodulation process. Modulation is used to separate the undesired DC offset signal from the magnetically induced signal in the frequency domain. Then, using a low-pass filter, the modulated DC offset is suppressed while the magnetically induced signal passes through the filter. The anti-aliasing filter prevents aliasing from happening in applications with high frequency signal com-

ponents which are beyond the user's frequency range of interest.

As a result of this chopper stabilization approach, the output voltage from the Hall IC is desensitized to the effects of temperature and mechanical stress. This technique produces devices that have an extremely stable Electrical Offset Voltage, are immune to thermal stress, and have precise recoverability after temperature cycling.

This technique is made possible through the use of a BiCMOS process that allows the use of low-offset and low-noise amplifiers in combination with high-density logic integration and sample and hold circuits.

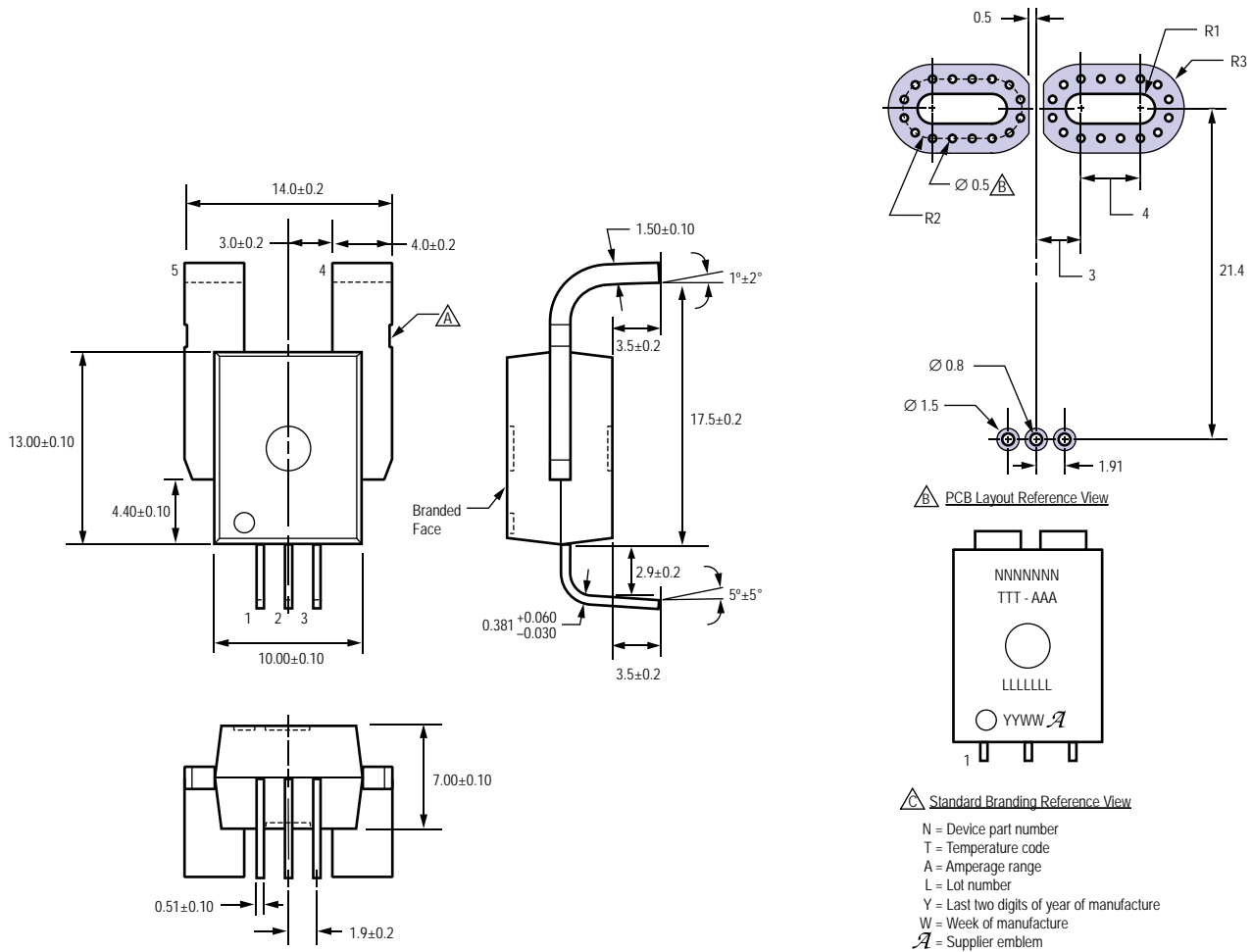


Concept of Chopper Stabilization Technique

# ACS758xCB

## Thermally Enhanced, Fully Integrated, Hall Effect-Based Linear Current Sensor IC with 100 $\mu\Omega$ Current Conductor

### Package CB, 5-pin package, leadform PFF

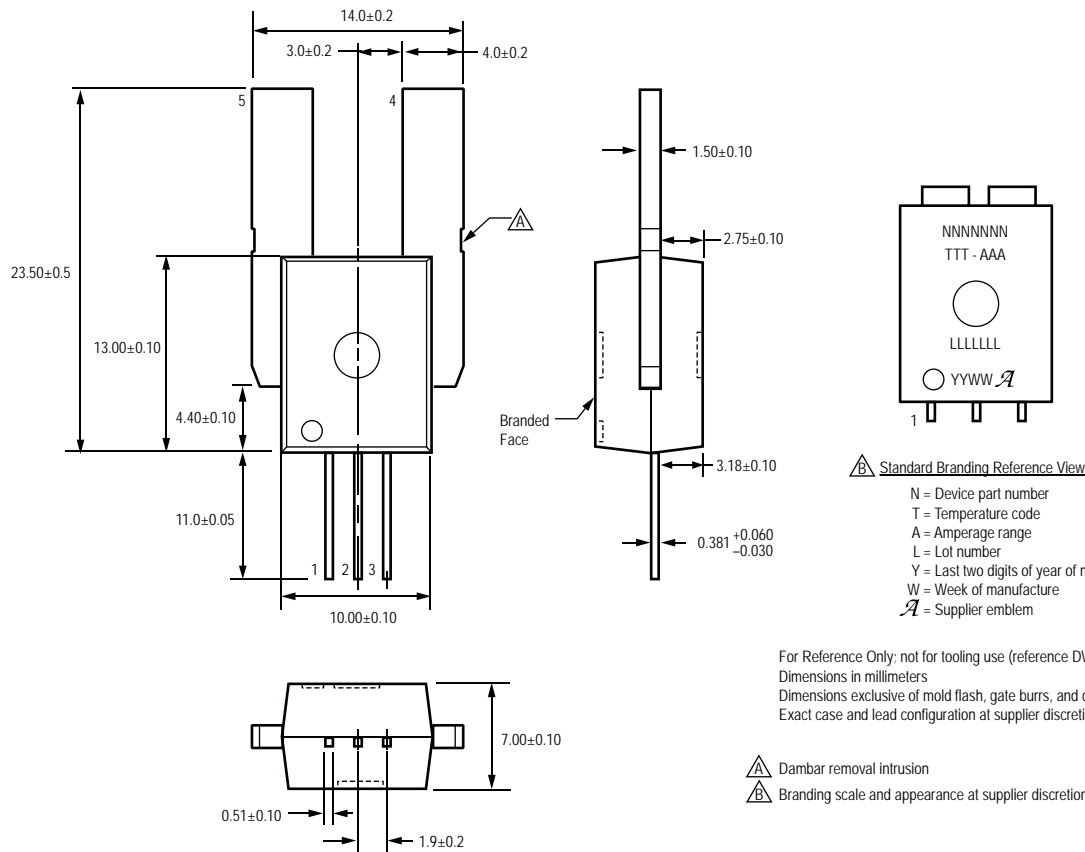


For Reference Only; not for tooling use (reference DWG-9111, DWG-9110)  
 Dimensions in millimeters  
 Dimensions exclusive of mold flash, gate burrs, and dambar protrusions  
 Exact case and lead configuration at supplier discretion within limits shown

- Dambar removal intrusion
- Perimeter through-holes recommended
- Branding scale and appearance at supplier discretion

Creepage distance, current terminals to signal pins: 7.25 mm  
 Clearance distance, current terminals to signal pins: 7.25 mm  
 Package mass: 4.63 g typical

### Package CB, 5-pin package, leadform PSS



Creepage distance, current terminals to signal pins: 7.25 mm  
Clearance distance, current terminals to signal pins: 7.25 mm  
Package mass: 4.63 g typical

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