



DUAL 5V MULTIFUNCTION VOLTAGE REGULATOR

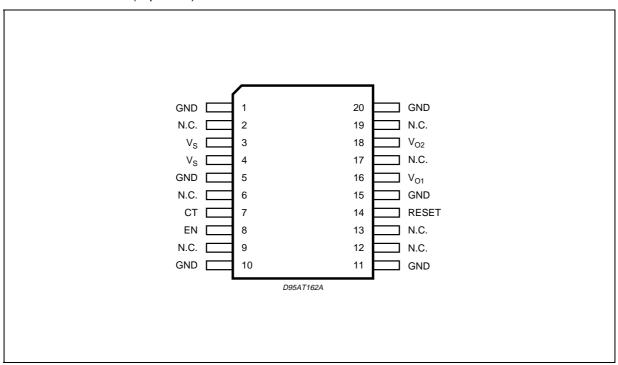
- STANDBY OUTPUT VOLTAGE PRECISION 5V ± 2%
- OUTPUT 2 TRACKED TO THE STANDBY OUT-PUT
- OUTPUT 2 DISABLE FUNCTION FOR STANDBY MODE
- VERY LOW QUIESCENT CURRENT, LESS THAN 260µA, IN STANDBY MODE
- OUTPUT CURRENTS: $I_{01} = 50 \text{mA}$, $I_{02} = 500 \text{mA}$
- VERY LOW DROPOUT (max 0.4V/0.6V)
- OPERATING TRANSIENT SUPPLY VOLTAGE UP TO 40V
- POWER-ON RESET CIRCUIT SENSING THE STANDBY OUTPUT VOLTAGE
- POWER-ON RESET DELAY PULSE DEFINED BY THE EXTERNAL CAPACITOR
- THERMAL SHUTDOWN AND SHORT CIRCUIT PROTECTIONS



DESCRIPTION

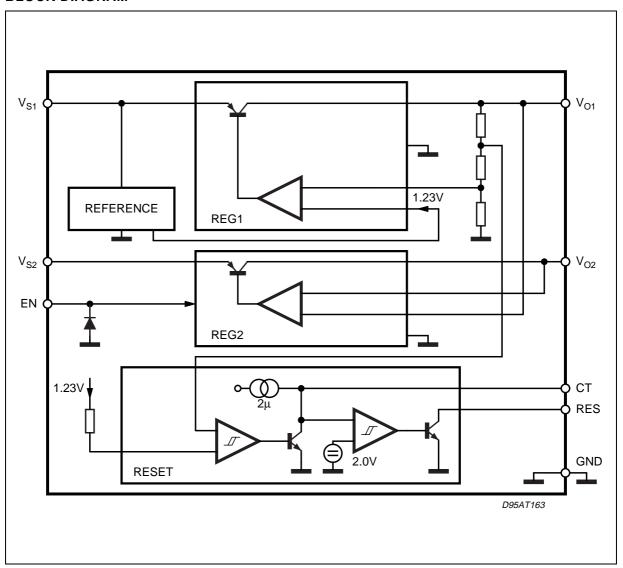
The L4937NPD is a monolithic integrated dual voltage regulators with two very low dropout outputs and additional functions such as power-on reset and input voltage sense. They are designed for supplying microcomputer controlled systems specially in automotive applications.

PIN CONNECTION (top view)



June 2000 1/9

BLOCK DIAGRAM



THERMAL DATA

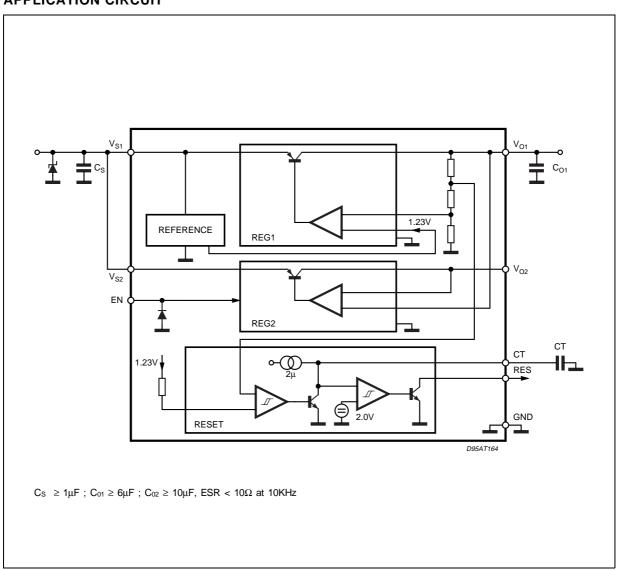
Symbol	Parameter	Value	Unit
R _{thj-case}	Thermal Resistance Junction-Case Max.	1.5	°C/W

ABSOLUTE MAXIMUM RATINGS

Symbol	Parameter	Value	Unit
Vs	DC Supply Voltage	28	V
	Transient Supply Voltage (T < 1s)	40	V
T _j , T _{stg}	Junction and Storage Temperature Range	-55 to 150	°C
I _{EN}	Enable Input Current (V _{EN} ≤0.3V)	±1	mA
V _{EN}	Enable Input Voltage	Vs	
V _{RES}	Reset Output Voltage	20	V
I _{RES}	Reset Output Current	5	mA
P _D	Power Dissipation	875	mW

Note: The circuit is ESD protected according to MIL-STD-883C.

APPLICATION CIRCUIT



ELECTRICAL CHARACTERISTICS ($V_S = 14V$; $-40^{\circ}C \le T_J \le 125^{\circ}C$ unless otherwise specified)

Symbol	Parameter	Test Conditions	Min.	Тур.	Max.	Unit
Vs	Operating Supply Voltage				25	V
V _{O1}	Standby Output Voltage	$ 6V \le V_S \le 25V $		5.00	5.10	V
V ₀₂ - V ₀₁	Output Voltage 2 Tracking Error	$6V \le V_S \le 25V$ -25 $5mA \le I_{O2} \le 500mA$ Enable = LOW			+25	mV
V_{DP1}	Dropout Voltage 1	I _{O1} = 10mA I _{O1} = 50mA	I _{O1} = 10mA		0.25 0.4	V V
V _{IO1}	Input to Output Voltage Difference in Undervoltage Condition	Vs = 4V, I _{O1} = 35mA			0.4	٧
V_{DP2}	Dropout Voltage 2	IO1 = 100mA I _{O1} = 500mA		0.2 0.3	0.3 0.6	V
V _{IO2}	Input to Output Voltage Difference in Undervoltage Condition	Vs = 4.6V, I _{O1} = 350mA			0.6	V
V _{OL 1.2}	Line Regulation	$6V \le V_S \le 25V$ $I_{O1} = 1 \text{mA}; I_{O2} = 5 \text{mA}$			20	mV
V_{OLO1}	Load Regulation 1	$1mA \le I_{O1} \le 50mA$			25	mV
V _{OLO2}	Load Regulation 2	$5mA \le I_{O2} \le 500mA$			50	mV
I _{LIM1}	Current Limit 1	V _{O1} = 4.5V V _{O1} = 0V (note 1)	55 25	100 50	200 100	mA mA
I _{LIM2}	Current Limit 2	$V_{O2} = 0V$	550	1000	1700	mA
I_{QSB}	Quiescent Current Standby Mode (output 2 disabled)	$I_{O1} = 0.3 \text{mA}; T_J < 100^{\circ}\text{C}$ $V_{EN} \ge 2.4 \text{V}$ $V_S = 14 \text{V}$ $V_S = 3.5 \text{V}$		210 340	290 850	μ Α μ Α
IQ	Quiescent Current	I _{O1} = 50mA I _{O1} = 500mA			30	mA

ENABLE

V _{ENL}	Enable Input LOW Voltage (output 2 active)		-0.3		1.5	V
V _{ENH}	Enable Input HIGH Voltage		2.4		7	V
V _{ENhyst}	Enable Hysteresis		30	75	200	mV
I _{EN}	Enable Input Current	0V < V _{EN} < 1.2V 2.5V < V _{EN} < 7V	–10 –1	-1.5 0	-0.5 +1	μA μA

ELECTRICAL CHARACTERISTICS (continued)

RESET

Symbol	Parameter	Test Conditions	Min.	Тур.	Max.	Unit
V_{Rt}	Reset Low Threshold Voltage		V _{o1} -0.4	4.7	V _{o1} -0.1	V
V_{Rth}	Reset Threshold Hysteresis		50	100	200	mV
t _{RD}	Reset Pulse Delay	$C_T = 100 nF; t_R > 100 \mu s$	55	100	180	ms
t _{RR}	Reset Reaction Time	C _T = 100nF	1	10	50	μs
V_{RL}	Reset Output LOW Voltage	$R_{RES} = 10 K\Omega$ to V_{01} $V_S = 1.5 V$			0.4	V
I _{LRES}	Reset Output HIGH Leakage	V _{RES} = 5V			1	μΑ
V _{CTh}	Delay Comparator Threshold			2.0		V
V _{CTh, hyst}	Delay Comparator Threshold Hysteresis			100		mV

Note: 1: Foldback characteristic

FUNCTIONAL DESCRIPTION

The L4937ND is based on the SGS-THOMSON Microelectronics modular voltage regulator approach. Several out-standing features and auxiliary functions are provided to meet the requirements of supplying the microprocessor systems used in automotive applications.

Furthermore the device is suitable also in other applications requiring two stabilized voltages.

The modular approach allows other features and functions to be realized easily when required.

STANDBY REGULATOR

The standby regulator uses an Isolated Collector Vertical PNP transistor as the regulating element. This structure allows a very low dropout voltage at currents up to 50mA. The dropout operation of the standby regulator is maintained down to 2V input supply voltage. The output voltage is regulated up to the transient input supply voltage of 40V. This feature avoids functional interruptions which could be generated by overvoltage pulses.

The typical curve of the standby output voltage as a function of the input supply voltage is shown in fig. 1.

The current consumption of the device (quiescent current) is less than $260\mu A$ when output 2 is disabled (standby mode). The dropout voltage is controlled to reduce the quiescent current peak in the undervoltage region and to improve the transient response in this region.

The quiescent current is shown in fig. 2 as a function of the supply input voltage 2.

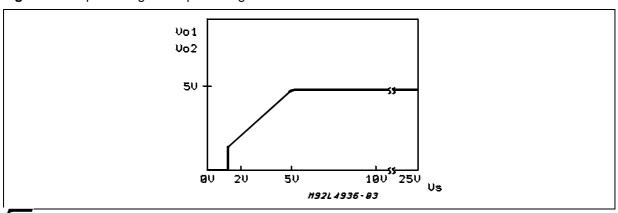
OUTPUT 2 VOLTAGE

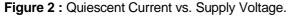
The output 2 regulator uses the same output structure as the standby regulator, but rated for an output current of 500mA.

The output 2 regulator works in tracking mode with the standby output voltage as a reference voltage when the output 2.

The output 2 regulator can be switched off via the Enable input.

Figure 1 : Output Voltage vs. Input Voltage.





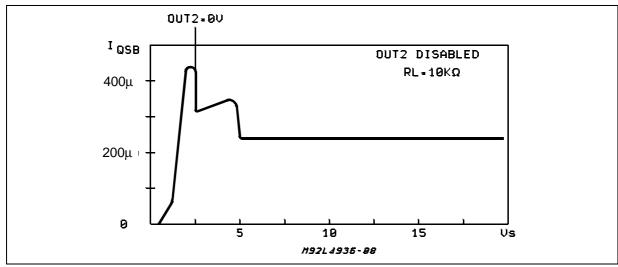
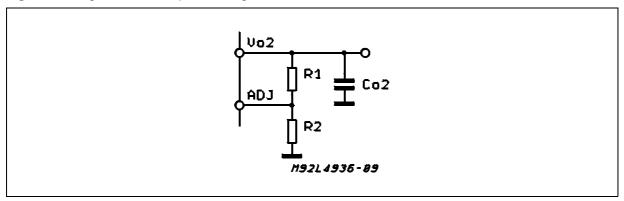


Figure 3 : Programmable Output 2 Voltage with External Resistors.



RESET CIRCUIT

The block circuit diagram of the reset circuit is shown in fig. 4. The reset circuit supervises the standby output voltage. The reset threshold of 4.7V is defined by the internal reference voltage and the standby output divider.

The reset pulse delay time t_{RD} , is defined by the charge time of an external capacitor C_T :

$$t_{RD} = \frac{C_T \times 2V}{2\mu A}$$

The reaction time of the reset circuit depends on the discharge time limitation of the reset capacitor C_T and is proportional to the value of C_T .

The reaction time of the reset circuit increases the noise immunity. In fact, if the standby output voltage drops below the reset threshold for a time shorter than the reaction time t_{RR} , no reset output variation occurs. The nominal reset delay is generated for standby output voltage drops longer than the time necessary for the complete discharging of the capacitor C_T . This time is typically equal to $50\mu s$ if $C_T = 100nF$. The typical reset output waveforms are shown in fig. 5.

Figure 4 :Block Diagram of the Reset Circuit.

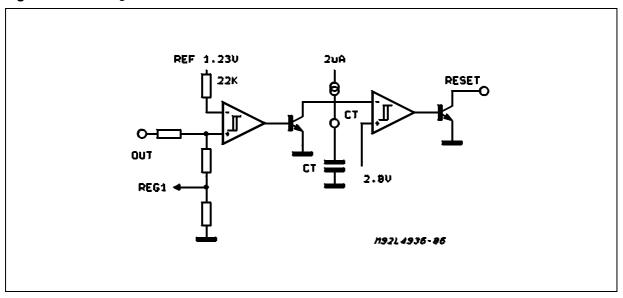
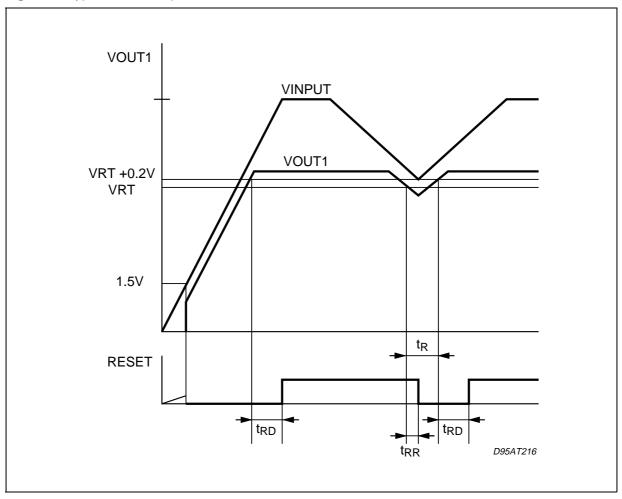


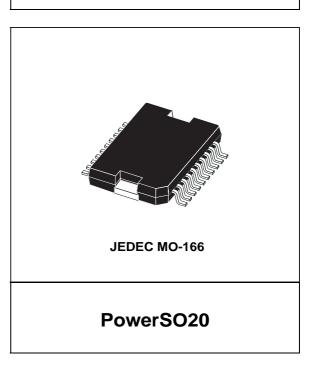
Figure 5 : Typical Reset Output Waveforms.

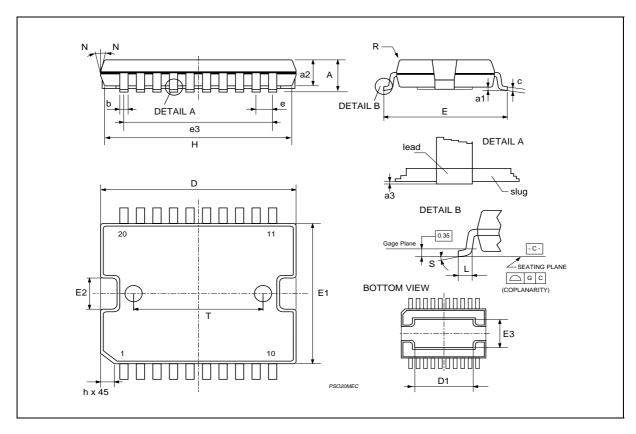


DIM.		mm			inch	
DIIVI.	MIN.	TYP.	MAX.	MIN.	TYP.	MAX.
Α			3.6			0.142
a1	0.1		0.3	0.004		0.012
a2			3.3			0.130
a3	0		0.1	0.000		0.004
b	0.4		0.53	0.016		0.021
С	0.23		0.32	0.009		0.013
D (1)	15.8		16	0.622		0.630
D1	9.4		9.8	0.370		0.386
E	13.9		14.5	0.547		0.570
е		1.27			0.050	
e3		11.43			0.450	
E1 (1)	10.9		11.1	0.429		0.437
E2			2.9			0.114
E3	5.8		6.2	0.228		0.244
G	0		0.1	0.000		0.004
Н	15.5		15.9	0.610		0.626
h			1.1			0.043
L	0.8		1.1	0.031		0.043
N	10° (max.)					
S	8° (max.)					
Т		10			0.394	

(1) "D and F" do not include mold flash or protrusions. - Mold flash or protrusions shall not exceed 0.15 mm (0.006"). - Critical dimensions: "E", "G" and "a3"

OUTLINE AND MECHANICAL DATA





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