## CD Digital Signal Processor

## For the availability of this product, please contact the sales office.

## Description

The CXD2508AQ/AR is a digital signal processor for CD players and is equipped with built-in digital filters, no-sound data detection circuit, and 1-bit DAC.

## Features

DSP block

- Digital PLL
- EFM frame sync protection
- SEC strategy-based error correction
- Subcode demodulation, CRC checking
- Digital spindle servo
- Servo auto sequencer
- Asymmetry compensation circuit
- Digital audio interface output
- 16K RAM
- Double-speed playback capability
- New microcomputer interface circuit

Digital filter, DAC block

- Double-speed playback capability
- Digital de-emphasis
- Digital attenuation
- No-sound data detection circuit
- 4 Fs oversampling filter
- Secondary $\Delta \sum$ noise shaper
- PWM-system pulse conversion output


## Recommended Operating Conditions

- Supply voltage VdD Note) 4.5 to 5.5 V

> (double-speed playback)
3.5 to 5.5 V
(normal-speed playback)
3.4 to 5.5 V
(low power consumption or special playback mode)

- Operating temperature



## Applications

CD players

## Structure

Silicon gate CMOS IC

## Absolute Maximum Ratings

- Supply voltage VDD $\quad-0.3$ to $7.0 \quad \mathrm{~V}$
- Input voltage Vı $\quad-0.3$ to $7.0 \quad \mathrm{~V}$
- Input voltage $\quad \mathrm{V}_{\mathrm{IN}} \mathrm{Vss}-0.3 \mathrm{~V}$ (min.) $\mathrm{V} d \mathrm{~d}+0.3$ (max.) V
- Output voltage Vo $\quad-0.3$ to 7.0 V
- Storage temperature

Tstg $\quad-40$ to $125 \quad{ }^{\circ} \mathrm{C}$

- Supply voltage variation
$\begin{array}{lll}\mathrm{V}_{S S}-\mathrm{AVSS} & -0.3 \mathrm{~V}(\text { min. }) & +0.3 \mathrm{~V}(\text { max. }) \\ \mathrm{VDD}-\mathrm{AVDD} & -0.3 \mathrm{~V}(\text { min. }) & +0.3 \mathrm{~V} \text { (max.) }\end{array}$

Topr $\quad-20$ (min.) 75 (max.) $\quad{ }^{\circ} \mathrm{C}$
Note) VDD (min.) is varied by the playback speed and built-in VCO in the CXD2508AQ/AR. 4.5V is the value using the VCO which generates the slower frequency in doublespeed playback. The table below shows the Vdd (min.) for each condition.

| Playback <br> speed | VDD (min.) [V] |  |  |
| :--- | :---: | :---: | :---: |
|  | VCO <br> high-speed | VCO <br> normal-speed | DAC block |
| $\times 2$ | 3.40 | 4.50 | 3.40 |
| $\times 1$ | 3.40 | 3.50 | 3.40 |
| $\times 1^{*}$ | 3.40 | 3.40 | 3.40 |

* When the internal operation of the LSI is set to doublespeed mode and the crystal oscillation frequency is halved, normal-speed playback results.
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## Pin Configuration



## Block Diagram



Note) The pin numbers are for QFP. Refer to the Pin Description for those of LQFP.

## Pin Description

| Pin |  | No. | Symbol | I/O |
| :---: | :---: | :--- | :---: | :--- |
|  |  |  |  |  |  |  |
| R | Q |  |  |  |
| 79 | 1 | SCOR | O | Outputs a high signal when either subcode sync S0 or S1 is detected. |
| 80 | 2 | SBSO | O | Sub P to W serial output. |
| 1 | 3 | EXCK | I | SBSO readout clock input. |
| 2 | 4 | SQSO | O | Sub Q 80-bit serial output. |
| 3 | 5 | SQCK | I | SQSO readout clock input. |
| 4 | 6 | MUTE | I | High: mute; low: release |
| 5 | 7 | SENS | O | SENS output to CPU. |
| 6 | 8 | XRST | I | System reset. Reset when low. |
| 7 | 9 | DATA | I | Serial data input from CPU. |
| 8 | 10 | XLAT | I | Latch input from CPU. Serial data is latched at the falling edge. |
| 9 | 11 | CLOK | I | Serial data transfer clock input from CPU. |
| 10 | 12 | Vss |  | GND. |
| 11 | 13 | SEIN | I | Sense input from SSP. |
| 12 | 14 | CNIN | I | Track jump count signal input. |
| 13 | 15 | DATO | O | Serial data output to SSP. |
| 14 | 16 | XLTO | O | Serial data latch output to SSP. Latched at the falling edge. |
| 15 | 17 | CLKO | O | Serial data transfer clock output to SSP. |
| 16 | 18 | SPOA | I | Microcomputer extended interface (input A). |
| 17 | 19 | SPOB | I | Microcomputer extended interface (input B). |
| 18 | 20 | SPOC | I | Microcomputer extended interface (input C). |
| 19 | 21 | XTSL | I | Crystal selection input. Low for 16.9344MHz; high for 33.8688MHz |
| 20 | 22 | XLON | O | Microcomputer extended interface (output). |
| 21 | 23 | FOK | I | Focus OK input. Used for SENS output and the servo auto sequencer. |
| 22 | 24 | MON | O | Spindle motor on/off control output. |
| 23 | 25 | MDP | O | Spindle motor servo control. |
| 24 | 26 | MDS | O | Spindle motor servo control. |
| 25 | 27 | LOCK | O | GFS is sampled at 460Hz; when GFS is high, this pin outputs a high signal. <br> If GFS is low eight consecutive samples, this pin outputs low. |
| 26 | 28 | TEST | I | TEST pin. Normally GND. |
| 27 | 29 | FILO | O | Master PLL (slave = digital PLL) filter output. |
| 28 | 30 | FILI | I | Master PLL filter input. |
| 29 | 31 | PCO | O | Master PLL charge pump output. |
| 30 | 32 | VDD |  | Digital power supply for DSP. |
| 31 | 33 | AVss1 |  | Analog GND for DSP. |
| 34 | CLTV | I | Master PLL VCO control voltage input. |  |


| Pin No. |  | Symbol | I/O | Description |
| :---: | :---: | :---: | :---: | :---: |
| R | Q |  |  |  |
| 33 | 35 | AVDD1 |  | Analog power supply for DSP. |
| 34 | 36 | RF | 1 | EFM signal input. |
| 35 | 37 | BIAS | 1 | Constant current input of asymmetry compensation circuit. |
| 36 | 38 | ASYI | 1 | Comparator voltage input of asymmetry compensation circuit. |
| 37 | 39 | ASYO | 0 | EFM full-swing output (low = Vss, high = VDD). |
| 38 | 40 | ASYE | 1 | Low: asymmetry compensation off; high: asymmetry compensation on. |
| 39 | 41 | WDCK | 0 | D/A interface for 48-bit slot. Word clock (2Fs). |
| 40 | 42 | LRCK | 0 | D/A interface for 48-bit slot. LR clock (Fs). |
| 41 | 43 | LRCKI | 1 | LR clock input for DAC. (48-bit slot) |
| 42 | 44 | PCMD | 0 | D/A interface. Serial data (two's complement, MSB first). |
| 43 | 45 | PCMDI | 1 | Audio data input for DAC. (48-bit slot) |
| 44 | 46 | BCK | 0 | D/A interface. Bit clock. |
| 45 | 47 | BCKI | 1 | Bit clock input for DAC. (48-bit slot) |
| 46 | 48 | GTOP | 0 | GTOP output. |
| 47 | 49 | XUGF | 0 | XUGF output. |
| 48 | 50 | XPCK | 0 | XPLCK output. |
| 49 | 51 | GFS | 0 | GFS output. |
| 50 | 52 | RFCK | O | RFCK output. |
| 51 | 53 | Vss |  | GND. |
| 52 | 54 | C2PO | 0 | C2PO output. |
| 53 | 55 | XROF | O | XRAOF output. |
| 54 | 56 | MNT3 | O | MNT3 output. |
| 55 | 57 | MNT1 | O | MNT1 output. |
| 56 | 58 | MNT0 | O | MNT0 output. |
| 57 | 59 | FSTT | 0 | 2/3 frequency-divider output for Pins 73 and 74. |
| 58 | 60 | C4M | O | 4.2336 MHz output. |
| 59 | 61 | DOUT | O | Digital Out output. |
| 60 | 62 | EMPH | O | Outputs high signal when the playback disc has emphasis, low signal when no emphasis. |
| 61 | 63 | EMPHI | 1 | DAC de-emphasis on/off. High: on; low: off. |
| 62 | 64 | WFCK | $\bigcirc$ | WFCK (write frame clock) output. |
| 63 | 65 | ZEROL | O | No-sound data detection output; high when no sound data is detected. (Left channel) |
| 64 | 66 | ZEROR | O | No-sound data detection output; high when no sound data is detected. (Right channel) |
| 65 | 67 | DTS1 | 1 | Test pin 1 for DAC; normally low. |
| 66 | 68 | Vdo |  | Digital power supply for DAC. |


| Pin No. |  | Symbol | I/O |  |
| :---: | :---: | :--- | :---: | :--- |
| R | Q |  |  |  |
| 67 | 69 | NLPWM | O | Left channel PWM output. (Reverse phase) |
| 68 | 70 | LPWM | O | Left channel PWM output. (Forward phase) |
| 69 | 71 | AVDD2 |  | Power supply for PWM driver. |
| 70 | 72 | AVDD3 |  | Power supply for crystal. |
| 71 | 73 | XTAI | I | 33.8688MHz crystal oscillation circuit input. |
| 72 | 74 | XTAO | O | 33.8688 MHz crystal oscillation circuit output. |
| 73 | 75 | AVss3 |  | GND for crystal. |
| 74 | 76 | AVss2 |  | GND for PWM driver. |
| 75 | 77 | NRPWM | O | Right channel PWM output. (Reverse phase) |
| 76 | 78 | RPWM | O | Right channel PWM output. (Forward phase) |
| 77 | 79 | DTS2 | I | DAC test pin 2; normally low. |
| 78 | 80 | DTS3 | I | DAC test pin 3; normally low. |

## Note)

- PCMD is an MSB first, two's complement output.
- GTOP is used to monitor the frame sync protection status. (High: sync protection window released.)
- XUGF is the negative pulse for the frame sync derived from the EFM signal. It is the signal before sync protection.
- XPLCK is the inverse of the EFM PLL clock. The PLL is designed so that the falling edge of XPLCK and the EFM signal transition point coincide.
- GFS goes high when the frame sync and the insertion protection timing match.
- RFCK is derived with the crystal accuracy. This signal has a cycle of $136 \mu$.
- C2PO represents the data error status.
- XRAOF is generated when the 16 K RAM exceeds the $\pm 4 \mathrm{~F}$ jitter margin.

Electrical Characteristics
DC Characteristics $\quad\left(\mathrm{VDD}=\mathrm{AVDD}=5.0 \mathrm{~V} \pm 5 \%, \mathrm{Vss}=\mathrm{AVss}=0 \mathrm{~V}\right.$, Topr $=-20$ to $\left.+75^{\circ} \mathrm{C}\right)$ NOTE)

| Item |  |  | Conditions | Min. | Typ. | Max. | Unit | Applicable pins |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | High level input voltage | VIH (1) |  | 0.7Vdd |  |  | V | *1 |
|  | Low level input voltage | VIL (1) |  |  |  | 0.3VDD | V |  |
|  | High level input voltage <br> Low level input voltage | $\mathrm{V}_{\mathrm{IH}}(2)$ <br> VIL (2) | Schmitt input | 0.8Vdd |  |  | V | *2 |
|  |  |  |  |  |  | 0.2 VdD | V |  |
|  | Input voltage | Vin (3) | Analog input | Vss |  | Vdo | V | *3 |
|  | High level output voltage | Vor (1) | $\mathrm{IOH}=-4 \mathrm{~mA}$ | VdD-0.8 |  | VDD | V | *4 |
|  | Low level output voltage | Vol (1) | $\mathrm{IoL}=4 \mathrm{~mA}$ | 0 |  | 0.4 | V |  |
|  | High level output voltage | Vot (2) | $\mathrm{IOH}=-2 \mathrm{~mA}$ | VDD-0.8 |  | Vdd | V | *5 |
|  | Low level output voltage | VoL (2) | $\mathrm{loL}=4 \mathrm{~mA}$ | 0 |  | 0.4 | V |  |
|  | High level output voltage | Vor (3) | $\mathrm{IOH}=-0.28 \mathrm{~mA}$ | Vdd-0.5 |  | Vod | V | *6 |
|  | Low level output voltage | VoL (3) | $\mathrm{IoL}=0.36 \mathrm{~mA}$ | 0 |  | 0.4 | V |  |
|  | High level output voltage | Vor (4) | $\mathrm{IOH}=-10 \mathrm{~mA}$ | VdD-0.4 |  | Vdo | V | *7 |
|  | Low level output voltage | Vol (4) | $\mathrm{loL}=10 \mathrm{~mA}$ | 0 |  | 0.4 | V |  |
| Input leak current |  | lL | $\mathrm{V}_{\mathrm{I}}=0$ to 5.25 V |  |  | $\pm 5$ | $\mu \mathrm{A}$ | *1, *2, *3 |
| Tri-state pin output leak current |  | ILo | $\mathrm{Vo}=0$ to 5.25 V |  |  | $\pm 5$ | $\mu \mathrm{A}$ | *8 |

## Applicable pins

*1 XTSL, DATA, XLAT, PCMDI, EMPHI, DTS1, DTS2, DTS3, SPOA, SPOB, SPOC
*2 CLOK, XRST, EXCK, SQCK, MUTE, FOK, SEIN, CNIN, ASYE, LRCKI, BCKI
*3 CLTV, FILI, RF, BIAS, ASYI
*4 MDP, PCO
*5 ASYO, DOUT, FSTT, C4M, SBSO, SQSO, SCOR, EMPH, MON, LOCK, WDCK, DATO, CLKO, XLTO, SENS, MDS, LRCK, WFCK, PCMD, BCK, GTOP, XUGF, XPCK, GFS, RFCK, XROF, MNT0, MNT1, MNT3, ZEROL, ZEROR
FILO
*7 LPWM, NLPWM, RPWM, NRPWM
*8 SENS, MDS, MDP
Note) "AVDD" refers to $A V D D 1, A V d d 2$, and $A V d D 3$. In addition, "AVss" refers to $A V s s 1, A V s s 2$, and $A V s s 3$.

## AC Characteristics

1) XTAI pin
(1) When using self-oscillation (Topr $=-20$ to $+75^{\circ} \mathrm{C}, \mathrm{VDD}=\mathrm{AVDD}=5.0 \mathrm{~V} \pm 5 \%$ )

| Item | Symbol | Min. | Typ. | Max. | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: |
| Oscillation frequency | fmax | 15 |  | 34 | MHz |

(2) When inputting pulses to XTAI

| (Topr $=-20$ to $+75^{\circ} \mathrm{C}, \mathrm{VDD}=\mathrm{AVDD}=5.0 \mathrm{~V} \pm 5 \%$ ) |  |  |  |  |  |
| :--- | :--- | :---: | :---: | :---: | :---: |
| Item | Symbol | Min. | Typ. | Max. | Unit |
| High level <br> pulse width | twhx | 13 |  | 500 | ns |
| Low level <br> pulse width | twLx | 13 |  | 500 | ns |
| Pulse cycle | tck | 26 |  | 1,000 | ns |
| Input high level | VIHX | VDD-1.0 |  |  | V |
| Input low level | VILX |  |  | 0.8 | V |
| Rise time, <br> fall time | $\mathrm{t}_{\mathrm{R}, \mathrm{tF}}$ |  |  | 10 | ns |


(3) When inputting sine waves to XTAI via a capacitor

$$
\text { (Topr }=-20 \text { to }+75^{\circ} \mathrm{C}, \mathrm{VDD}=\mathrm{AVDD}=5.0 \mathrm{~V} \pm 5 \% \text { ) }
$$

| Item | Symbol | Min. | Typ. | Max. | Unit |
| :---: | :--- | :---: | :---: | :---: | :---: |
| Input amplitude | $\mathrm{V}_{1}$ | 2.0 |  | $\mathrm{VDD}_{\mathrm{D}}+0.3$ | $\mathrm{Vp}-\mathrm{p}$ |

2) CLOK, DATA, XLAT, CNIN, SQCK EXCK pins

| (VDD = AVDD $=5.0 \mathrm{~V} \pm 5 \%$, Vss $=$ AVss $=0 \mathrm{~V}$, Topr $=-20$ to $+75^{\circ} \mathrm{C}$ ) |  |  |  |  |  |
| :--- | :--- | :---: | :---: | :---: | :---: |
| Symbol | Min. | Typ. | Max. | Unit |  |
| Clock frequency | fck |  |  | 0.65 | MHz |
| Clock pulse width | twck | 750 |  |  | ns |
| Setup time | tsu | 300 |  |  | ns |
| Hold time | th | 300 |  |  | ns |
| Delay time | to | 300 |  |  | ns |
| Latch pulse width | twL | 750 |  |  | ns |
| EXCK SQCK frequency | ft |  |  | $0.65^{*}$ | MHz |
| EXCK SQCK pulse width | fwt | $750^{*}$ |  |  | ns |



* In pseudo double-speed playback mode, when SLO = SL1 = 1, the maximum operating frequency for SQCK is 300 kHz and the minimum pulse width is $1.5 \mu \mathrm{~s}$.

3) BCKI, LRCKI, and PCMDI pins (VdD $=A V d D=5.0 \mathrm{~V} \pm 5 \%$, $\mathrm{Vss}=\mathrm{AVss}=0 \mathrm{~V}$, $\mathrm{Topr}=-20$ to $+75^{\circ} \mathrm{C}$ )

| Item | Symbol | Conditions | Min. | Typ. | Max. | Unit |
| :--- | :--- | :--- | :---: | :---: | :---: | :---: |
| BCK pulse width | tw |  | 94 | 118 | 141 | nsec |
| DATAL, R setup time | tsu |  | 18 |  |  | nsec |
| DATAL, R hold time | $\mathrm{t}_{\mathrm{H}}$ |  | 18 |  |  | nsec |
| LRCK setup time | tsu |  | 18 |  |  | nsec |

BCKI


1-bit DAC Block Analog Characteristics
$\left(\mathrm{VdD}=\mathrm{AVdD}=5.0 \mathrm{~V}, \mathrm{Vss}=\mathrm{AVss}=0 \mathrm{~V}, \mathrm{Ta}=25^{\circ} \mathrm{C}\right)$

| Item | Symbol | Conditions | Playback mode | Min. | Typ. | Max. | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Total harmonic distortion | THD | $1 \mathrm{kHz},$ <br> OdB data | Normal speed |  |  | 0.015 | \% |
|  |  |  | Pseudo double-speed playback |  |  | 0.025 |  |
| S/N ratio | S/N | 1kHz, <br> OdB data <br> (using filter A) | Normal speed | 87 |  |  | dB |
|  |  |  | Pseudo double-speed playback | 83 |  |  |  |

For both items, Fs=44.1kHz
The circuits for measuring the total harmonic distortion and $\mathrm{S} / \mathrm{N}$ ratio are shown below.


Analog LPF Circuit


Block Diagram for Measuring Analog Characteristics

## Description of Functions

## 1. CPU Interface and Instructions

- CPU interface

This interface uses DATA, CLOK, and XLAT to set the modes. The interface timing chart is shown below.


- Information on each address and the data is provided in Table 1-1.
- The internal registers are initialized by a reset when XRST $=0$; the initialization data is shown in Table 1-2. Note) When XLAT is low, EXCK and SQCK must be set high.
Command Table

|  | Command | Address |  |  |  | Data 1 |  |  |  | Data 2 |  |  |  | Data 3 |  |  |  | Data 4 |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | D3 | D2 | D1 | D0 | D3 | D2 | D1 | Do | D3 | D2 | D1 | D0 | D3 | D2 | D1 | D0 | D3 | D2 | D1 | D0 |
| 4 | Auto sequence | 0 | 1 | 0 | 0 | AS3 | AS2 | AS1 | ASO | - | - | - | - | - | - | - | - | - | - | - | - |
| 5 | Blind (A, E), Overilow (C) <br> Brake (B) | 0 | 1 | 0 | 1 | $\begin{array}{\|l\|} \hline 0.18 \mathrm{~ms} \\ \hline 0.36 \mathrm{~ms} \\ \hline \end{array}$ | 0.00 ms | s 0.05 ms | $0.02 \mathrm{~ms}$ | - | - | - | - | - | - | - | - | - | - | - | - |
| 6 | KICK (D) | 0 | 1 | 1 | 0 | 11.6ms | 5.8 ms | 2.9 ms | 1.45 ms | - | - | - | - | - | - | - | - | - | - | - | - |
| 7 | Auto sequence ( N ) track jump count setting | 0 | 1 | 1 | 1 | 32768 | 16384 | 8192 | 4096 | 2048 | 1024 | 512 | 256 | 128 | 64 | 32 | 16 | 8 | 4 | 2 | 1 |
| 8 | MODE specification | 1 | 0 | 0 | 0 | $\begin{aligned} & \mathrm{CD}-{ }^{\mathrm{RO}} \end{aligned}$ | $\begin{array}{\|l\|l\|} \hline \text { DOUT } \\ \text { MUTE } \end{array}$ | $\begin{array}{r\|l\|} \hline \text { DOUT } \\ \text { DNOFF } \\ \hline \end{array}$ | WSEL | - | - | - | $\begin{aligned} & \text { VCO } \\ & \text { CEL } \end{aligned}$ | - | - | - | - | - | - | - | - |
| 9 | Function specification | 1 | 0 | 0 | 1 | 0 | $\begin{array}{\|l\|} \hline \text { DSPB } \\ \text { ONOFF } \end{array}$ | 0 | 0 | 0 | 0 | 0 | ${ }_{\text {SSLT }}$ | - | - | - | - | - | - | - | - |
| A | Audio CTRL | 1 | 0 | 1 | 0 | 0 | 0 | Mute | ATT | - | - | DADS | - | - | AD6 | AD5 | AD4 | AD3 | AD2 | AD1 | ADO |
| B | Serial bus CTRL | 1 | 0 | 1 | 1 | SL 1 | SLo | CPUSR | 0 | - | - | - | - | - | - | - | - | - | - | - | - |
| C | Servo coefficient setting | 1 | 1 | 0 | 0 | $\begin{array}{\|c\|} \hline \text { Gaia } \\ \text { MDP } \end{array}$ | $\begin{array}{\|l\|l\|} \hline \text { Gain } \\ \text { MDPO } \end{array}$ | Gain | $\begin{aligned} & \text { Gain } \\ & \text { MDSS } \end{aligned}$ | - | - | - | - | - | - | - | - | - | - | - | - |
| D | CLV CTRL | 1 | 1 | 0 | 1 | $\begin{array}{\|l\|l\|l\|} \hline \text { PWMmod } \end{array}$ | TB | TP | $\begin{array}{\|l} \text { CLVS } \\ \text { Gain } \end{array}$ | - | - | - | - | - | - | - | - | - | - | - | - |
| E | CLV mode | 1 | 1 | 1 | 0 | CM3 | CM2 | CM1 | смо | - | - | - | - | - | - | - | - | - | - | - | - |
| F | TEST mode | 1 | 1 | 1 | 1 |  | Don' | 't Use | $\square$ | - | - | - | - | - | - | - | - | - | - | - | - |

Reset Initialization

| $\begin{aligned} & \stackrel{V}{\pi} \\ & \text { N} \\ & \hline 0 \end{aligned}$ | 응 | 1 |  | \| | \| | $\bigcirc$ | 1 | - | - | \\| | 1 | - | 1 | 1 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | $\bar{\square}$ | 1 |  | \| | I | $\bigcirc$ | 1 | I | - | I | 1 | 1 | I | 1 |
|  | ั | 1 |  | \| | I | - | I | 1 | - | 1 | 1 | \\| | 1 | 1 |
|  | ® | 1 |  | \| | I | $\bigcirc$ | I | I | - | \\| | 1 | 1 | 1 | 1 |
|  | 옹 | I |  | \| | I | $\bigcirc$ | I | 1 | - | \\| | 1 | I | 1 | 1 |
|  | $\bar{\square}$ | 1 |  | \| | I | $\bigcirc$ | I | I | - | I | 1 | I | 1 | 1 |
|  | ก | 1 |  | \\| | I | $\bigcirc$ | I | 1 | - | I | 1 | I | 1 | 1 |
|  | ® | 1 |  | \| | I | $\bigcirc$ | I | 1 | \\| | I | 1 | 1 | 1 | 1 |
| $\begin{aligned} & \text { N } \\ & \underset{\sim}{\tilde{N}} \end{aligned}$ | 앙 | 1 |  | \| | I | - | $\bigcirc$ | $\bigcirc$ | 1 | I | 1 | I | 1 | 1 |
|  | $\bar{\square}$ | I |  | \| | I | - | I | $\bigcirc$ | $\bigcirc$ | I | 1 | I | 1 | 1 |
|  | ก | 1 |  | \\| | I | $\bigcirc$ | 1 | $\bigcirc$ | I | I | 1 | I | 1 | 1 |
|  | ® | 1 |  | 1 | I | - | 1 | $\bigcirc$ | I | I | 1 | I | । | 1 |
|  | 응 | $\bigcirc$ |  | - | - | $\bigcirc$ | $\bigcirc$ | $\bigcirc$ | - | $\bigcirc$ | $\bigcirc$ | - | $\bigcirc$ | $\square$ |
|  | $\bar{\square}$ | $\bigcirc$ |  | $\bigcirc$ | - | $\bigcirc$ | - | $\bigcirc$ | - | - | - | $\bigcirc$ | $\bigcirc$ | $\stackrel{\text { ¢ }}{\sim}$ |
|  | ก | $\bigcirc$ |  | - | - | $\bigcirc$ | $\bigcirc$ | $\bigcirc$ | $\bigcirc$ | $\bigcirc$ | - | $\bigcirc$ | $\bigcirc$ | $\stackrel{\text { ¢ }}{ }$ |
|  | ® | $\bigcirc$ |  | $\bigcirc$ | $\bigcirc$ | $\bigcirc$ | $\bigcirc$ | $\bigcirc$ | $\bigcirc$ | $\bigcirc$ | $\bigcirc$ | - | $\bigcirc$ | $\square$ |
|  | 앙 | $\bigcirc$ |  | - | $\bigcirc$ | - | $\bigcirc$ | - | $\bigcirc$ | - | $\bigcirc$ | - | $\bigcirc$ | - |
|  | $\bar{\square}$ | $\bigcirc$ |  | $\bigcirc$ | - | - | $\bigcirc$ | $\bigcirc$ | - | - | $\bigcirc$ | $\bigcirc$ | - | - |
|  | ก | - |  | - | - | - | $\bigcirc$ | $\bigcirc$ | $\bigcirc$ | - | - | - | - | - |
|  | ® | $\bigcirc$ | $\bigcirc$ |  | $\bigcirc$ | $\bigcirc$ | - | - | - | - | - | - | - | - |
|  |  |  |  |  | $\begin{aligned} & \overline{\mathrm{O}} \\ & \frac{\mathrm{U}}{\mathrm{U}} \end{aligned}$ |  |  |  |  |  |  | $\begin{aligned} & \frac{1}{\mathrm{x}} \\ & \vdots \\ & \frac{1}{0} \end{aligned}$ |  |  |
| $\begin{array}{r} \text { әu } \\ \text { derss } \end{array}$ | $\begin{aligned} & \text { ueu } \\ & \hline \text { bəey } \\ & \hline \end{aligned}$ | - |  | $\bigcirc$ | $\bigcirc$ | $\wedge$ | $\infty$ | の | < | ๓ | 0 | $\bigcirc$ | ш | แ |

1-1. The meaning of the data for each address is explained below.

## \$4X commands

| Command | AS3 | AS2 | AS1 | AS0 |
| :--- | :---: | :---: | :---: | :---: |
| CANCEL | 0 | 0 | 0 | 0 |
| FOCUS-ON | 0 | 1 | 1 | 1 |
| 1 TRACK JUMP | 1 | 0 | 0 | RXF |
| 10 TRACK JUMP | 1 | 0 | 1 | RXF |
| 2N TRACK JUMP | 1 | 1 | 0 | RXF |
| N TRACK MOVE | 1 | 1 | 1 | RXF |
| RXF$=0$ FORWARD |  |  |  |  |
|  | RXF $=0$ | REVERSE |  |  |

- When the FOCUS-ON command ( $\$ 47$ ) is canceled, $\$ 02$ is sent and the auto sequence is interrupted.
- When the TRACK JUMP/MOVE commands ( $\$ 48$ to $\$ 4 \mathrm{~F}$ ) are canceled, $\$ 25$ is sent and the auto sequence is interrupted.


## \$5X commands

Auto sequence timer setting
Setting timers: A, E, C, B

| Command | D3 | D2 | D1 | D0 |
| :--- | :---: | :---: | :---: | :---: |
| Blind (A, E), Over flow (C) | 0.18 ms | 0.09 ms | 0.05 ms | 0.02 ms |
| Brake (B) | 0.36 ms | 0.18 ms | 0.09 ms | 0.05 ms |

Ex.) D2 = D0 = 1, D3 = D1 = 0 (Initial Reset)
$\mathrm{A}=\mathrm{E}=\mathrm{C}=0.11 \mathrm{~ms}$
$B=0.23 \mathrm{~ms}$

## \$6X commands

Auto sequence timer setting
Setting timer: D

| Command | D3 | D2 | D1 | D0 |
| :--- | :---: | :---: | :---: | :---: |
| KICK (D) | 11.6 ms | 5.8 ms | 2.9 ms | 1.45 ms |

Ex.) D3 = 0, D2 = D1 = D0 = 1 (Initial Reset)
$D=10.15 \mathrm{~ms}$

## \$7X commands

Auto sequence TRACK JUMP/MOVE count setting (N)

| Command | Data 1 |  |  |  | Data 2 |  |  |  | Data 3 |  |  |  |  | Data 4 |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | D3 | D2 | D1 | D0 | D3 | D2 | D1 | D0 | D3 | D2 | D1 | D0 | D3 | D2 | D1 | D0 |  |
| Auto sequence track jump <br> number setting | $2^{15}$ | $2^{14}$ | $2^{13}$ | $2^{12}$ | $2^{11}$ | $2^{10}$ | $2^{9}$ | $2^{8}$ | $2^{7}$ | $2^{6}$ | $2^{5}$ | $2^{4}$ | $2^{3}$ | $2^{2}$ | $2^{1}$ | $2^{0}$ |  |

This command is used to set N when a 2 N TRACK JUMP and an N TRACK MOVE are executed for auto sequence.

- The maximum track count is 65,535 , but note that with 2 N track jumps the maximum track jump count is determined by the mechanical limitations of the optical system.
- The number of track jump is counted according to the signals input from CNIN pin.
\$8X commands

| Command | Data 1 |  |  |  | Data 2 |  |  |  |  |
| :--- | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | D3 | D2 | D1 | D0 | D3 | D2 | D1 | D0 |  |
| MODE <br> specification | CDROM | DOUT <br> MUTE | DOUT <br> ON-OFF | WSEL | 0 | 0 | 0 | VCO <br> SEL |  |


| Command bit | C2PO timing | Processing |
| :---: | :---: | :--- |
| CDROM $=1$ | $1-3$ | CDROM mode; average value interpolation and <br> pre-value hold are not performed. |
| CDROM $=0$ | $1-3$ | Audio mode; average value interpolation and pre-value <br> hold are performed. |


| Command bit | Processing |
| :---: | :--- |
| DOUT MUTE $=1$ | Digital Out output is muted. (DA output is not muted.) |
| DOUT MUTE $=0$ | When no other mute conditions are set, Digital Out output is not muted. |


| Command bit |  |
| :---: | :--- |
| DOUT ON-OFF $=1$ | Digital Out is output from the DOUT pin. |
| DOUT ON-OFF $=0$ | Digital Out is not output from the DOUT pin. |


| Command bit | Sync protection window width | Application |
| :---: | :--- | :--- |
| WSEL $=1$ | $\pm 26$ channel clock* | Anti-rolling is enhanced. |
| WSEL $=0$ | $\pm 6$ channel clock | Sync window protection is enhanced. |

* In normal-speed playback, channel clock $=4.3218 \mathrm{MHz}$.

| Command bit | Processing | Application |
| :---: | :--- | :--- |
| VCOSEL $=1$ | VCO for double-speed <br> playback is selected. | Double-speed playback or low voltage <br> operation is possible. |
| VCOSEL $=0$ | VCO for normal-speed <br> playback is selected. | The selection is made for the normal <br> speed playback. |

## \$9X commands

| Command | Data 1 |  |  |  | Data 2 |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | D3 | D2 | D1 | D0 | D3 | D2 | D1 | D0 |  |
| Function <br> specifications | 0 | DSPB <br> ON-OFF | 0 | 0 | 0 | 0 | 0 | FSTT <br> SEL |  |


| Command bit |  | Processing |
| :---: | :--- | :--- |
| DSPB $=0$ | Normal-speed playback |  |
| DSPB $=1$ | Double-speed playback |  |


| Command bit |  |
| :--- | :--- |
| FSTTSEL $=0$ | The clock with two-thirds frequency of crystal is output to FSTT pin. |
| FSTTSEL $=1$ | The clock with the sixth frequency of crystal is output to FSTT pin. |

## \$AX commands

| Command | Data 1 |  |  |  | Data 2 |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | D3 | D2 | D1 | D0 | D3 | D2 | D1 | D0 |
| Audio CTRL | 0 | 0 | Mute | ATT | - | - | DADS | - |


| Command bit | Meaning |
| :---: | :--- |
| Mute $=0$ | Mute off. |
| Mute $=1$ | Mute on. O data is output <br> from DSP. |


| Command bit | Meaning |
| :---: | :--- |
| ATT $=0$ | Attenuation off. |
| ATT $=1$ | -12 dB |


| Command bit | Processing |
| :---: | :--- |
| DADS $=0$ | Normal-speed playback for DAC block |
| DADS $=1$ | Double-speed playback for DAC block |

In the case of using the crystal of 768 Fs ( $\mathrm{Fs}=44.1 \mathrm{kHz}$ )

## Digital Attenuation

The audio output level from DAC can be attenuated by setting AD6 to AC0 of register A. (with a built-in primary noise shaper)

| Command | Data 3 |  |  | Data 4 |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | D2 | D1 | D0 | D3 | D2 | D1 | D0 |
| Audio CTRL | AD6 | AD5 | AD4 | AD3 | AD2 | AD1 | AD0 |


| Command bit <br> AD6 to AD0 | Audio output |
| :---: | :---: |
| $7 \mathrm{~F}(\mathrm{H})$ | 0 dB |
| $7 \mathrm{HE}(\mathrm{H})$ |  |
| to |  |
| $01(\mathrm{H})$ | -0.13 dB <br> to <br> -42.144 dB |
| $00(\mathrm{H})$ | $-\infty$ |

The attenuation data consists of seven bits (AD6 to AC0), and 127 settings are possible.
Audio output from $01(H)$ to $7 \mathrm{E}(\mathrm{H})$ is determined according to the following formula:
Audio output $=20 \log \left(\frac{\text { attenuation data }}{128}\right) \mathrm{dB}$
Ex.) When the attenuation data is $7 \mathrm{~A}(\mathrm{H})$
Audio output $=20 \log \left(\frac{122}{128}\right) \mathrm{dB}=-0.417 \mathrm{~dB}$

## Soft Mute

With soft mute function, when the attenuation data goes from $7 \mathrm{~F}(\mathrm{H})(\mathrm{OdB})$ to $00(\mathrm{H})(-\infty)$ or vice versa, muting is turned on/off with a muting time of $1024 / \mathrm{Fs}[\mathrm{s}]=23.2[\mathrm{~ms}](\mathrm{Fs}=44.1 \mathrm{kHz})$.

## Attenuation Operation

Assume attenuation data $\mathrm{X} 1, \mathrm{X} 2$, and X 3 , where $\mathrm{X} 1>\mathrm{X} 3>\mathrm{X} 2$, and audio output Y 1 , Y 2 , and Y 3 , where $\mathrm{Y} 1>$ $Y 3>Y 2$. First, assume $X 1$ is transferred and then $X 2$ is transferred. If $X 2$ is transferred before $Y 1$ is reached (state " A " in the diagram), then the value continues approaching Y 2 . Next, if X 3 is transferred before Y 2 is reached (either state " B " or " C " in the diagram), the value begins approaching Y 3 from the current value at that point.


## \$BX commands

| Command | D3 | D2 | D1 | D0 |
| :---: | :---: | :---: | :---: | :---: |
| Serial bus CTRL | SL1 | SL0 | CPUSR | 0 |

This command switches the method of interfacing with the CPU. With the CDL500 series, the number of signal lines between the CPU and the DSP can be reduced in comparison with the CDL40 series. Also, the error rate can be measured with the CPU.

| Command bits |  |  |
| :---: | :---: | :--- |
| SL1 | SL0 |  |
| 0 | 0 | Same interface mode as the CDL40 series. |
| 0 | 1 | SBSO is output from SQSO pin. In other words, subcodes P to W are <br> read out from SQSO. Input the readout clock to SQCK. |
| 1 | 0 | SENS is output from SQSO pin. |
| 1 | 1 | Each output signal is output from SQSO pin. <br> Input the readout clock to SQCK. <br> (See the Timing Chart 1-2.) |


| Command bits |  | Processing |
| :---: | :--- | :--- |
| CPUSR $=1$ | XLON pin is high. |  |
| CPUSR $=0$ | XLON pin is low. |  |

\$CX commands

| Command | D3 | D2 | D1 | D0 |
| :---: | :---: | :---: | :---: | :---: |
| Servo coefficient setting | Gain <br> MDP1 | Gain <br> MDP0 | Gain <br> MDS1 | Gain <br> MDS0 |
| CLV CTRL (\$DX) |  |  |  |  |

- CLVS mode gain setting: GCLVS

| Gain <br> MDS1 | Gain <br> MDSO | Gain <br> CLVS | GCLVS |
| :---: | :---: | :---: | :---: |
| 0 | 0 | 0 | -12 dB |
| 0 | 0 | 1 | -6 dB |
| 0 | 1 | 0 | -6 dB |
| 0 | 1 | 1 | 0 dB |
| 1 | 0 | 0 | 0 dB |
| 1 | 0 | 1 | +6 dB |

- CLVP mode gain setting: GMDP, GMDS

| Gain <br> MDP1 | Gain <br> MDP0 | GMDP |
| :---: | :---: | :---: |
| 0 | 0 | -6 dB |
| 0 | 1 | 0 dB |
| 1 | 0 | +6 dB |


| Gain <br> MDS1 | Gain <br> MDS0 | GMDS |
| :---: | :---: | :---: |
| 0 | 0 | -6 dB |
| 0 | 1 | 0 dB |
| 1 | 0 | +6 dB |

## \$DX commands

| Command | D3 | D2 | D1 | D0 |
| :---: | :---: | :---: | :---: | :---: |
| CLV CTRL | DCLV <br> PWM MD | TB | TP | CLVS <br> Gain |

See the \$CX command.

| Command bit | Explanation (See the Timing Chart 1-3.) |
| :---: | :--- |
| DCLV PWM MD $=1$ | CLV PWM mode specified. Both MDS and MDP are used. |
| DCLV PWM MD $=0$ | CLV PWM mode specified. Ternary MDP values are output. |


| Command bit | Explanation |
| :---: | :--- |
| $\mathrm{TB}=0$ | Bottom hold in CLVS mode at cycle of RFCK/32 |
| $\mathrm{TB}=1$ | Bottom hold in CLVS mode at cycle of RFCK/16 |
| $\mathrm{TP}=0$ | Peak hold in CLVS mode at cycle of RFCK/4 |
| $\mathrm{TP}=1$ | Peak hold in CLVS mode at cycle of RFCK/2 |

## \$EX commands

| Command | D3 | D2 | D1 | D0 |
| :--- | :---: | :---: | :---: | :---: |
| CLV mode | CM3 | CM2 | CM1 | CM10 |


| CM3 | CM2 | CM1 | CM0 | Mode | Explanation |
| :---: | :---: | :---: | :---: | :---: | :---: |
| 0 | 0 | 0 | 0 | STOP | See the Timing Chart 1-4. |
| 1 | 0 | 0 | 0 | KICK | See the Timing Chart 1-5. |
| 1 | 0 | 1 | 0 | BRAKE | See the Timing Chart 1-6. |
| 1 | 1 | 1 | 0 | CLVS |  |
| 1 | 1 | 1 | 1 | CLVP |  |
| 0 | 1 | 1 | 0 | CLVA |  |

STOP : Spindle motor stop mode
KICK : Spindle motor forward rotation mode
BRAKE : Spindle motor reverse rotation mode
CLVS : Rough servo mode. When RF-PLL circuit lock is disengaged, this mode is used to pull the disc rotations within the RF-PLL capture range.
CLVP : PLL servo mode.
CLVA : Automatic CLVS/CLVP switching mode. This mode is normally used during playback.
Timing Chart 1-1
If C2 Pointer $=1$,
data is NG
Rch 16bit C2 Pointer
Lch C2 Pointer
CDROM $=0$
C2P0
CDROM $=1$
C2PO
LRCK

$\square$

$\square$

WDCK
Rch C2 Pointer
Timing Chart 1-2

| C1F1 | C1F2 | C1 correction status |
| :---: | :---: | :--- |
| 0 | 0 | No Error |
| 1 | 0 | Single error correction |
| 1 | 1 | Irretrievable error |

## Timing Chart 1-3



DCLV PWM MD $=0$


## Timing Chart 1-4



Timing Chart 1-5


DCLV PWM MD = 1


Timing Chart 1-6


## 1-2. Description of SENS Output

The following signals are output from SENS, depending on the microcomputer serial register value (latching not required).

| Microcomputer serial register value (latching not required) | SENS output | Meaning |
| :---: | :---: | :---: |
| \$0X, 1X, 2X, 3X | SEIN | SEIN, a signal input to the this IC from the SSP, is output. |
| \$4X | XBUSY | Low while the auto sequencer is in operation, high when operation terminates. |
| \$5X | FOK | Outputs the signal input to the FOK pin. Normally, FOK (from RF) is input. High for "focus OK". |
| \$6X | SEIN | SEIN, a signal input to this IC from the SSP, is output. |
| \$AX | GFS | High when the played back frame sync is obtained with the correct timing. |
| \$EX | $\overline{\text { OV64 }}$ | Low when the EFM signal, after passing through the sync detection filter, is lengthened by 64 channel clock pulses or more. |
| $\begin{aligned} & \text { \$7X, 8X, 9X, BX, CX, } \\ & \text { DX, FX } \end{aligned}$ | "L" | SENS pin is fixed low. |

Note that the SENS output can be read from SQSO pin when $\operatorname{SL1}=1$ and SLO = 0 . (See the $\$ B X$ commands.)

## 2. Subcode Interface

This section explains the subcode interface.
There are two methods for reading out a subcode externally. The 8-bit subcodes P to W can be read from SBSO by inputting EXCK to the CXD2508AQ/AR.
Sub Q can be read out after the CRC check of the 80 bits data in the subcode frame. This accomplished, after checking SCOR and CRCF, by inputting 80 clock pulses to SQCK and reading data from SQSO pin.

## 2-1. P to W Subcode Read

Data can be read out by inputting EXCK immediately after WFCK falls. (See Fig. 2-1.)
Also, SBSO can be read out from SQSO pin when SL1 = 0 and SL0 = 1. (See the \$BX commands.)

## 2-2. 80-bit Sub Q Read

Fig. 2-2 shows the peripheral block of the 80-bit Sub Q register.

- First, Sub Q, regenerated at one bit per frame, is input to the 80 -bit serial/parallel register and the CRC check circuit.
- 96-bit Sub $Q$ is input, and if the CRC is OK, it is output to SQSO with CRCF = 1 . In addition, the 80 bits are loaded into the parallel/serial register.
When SQSO goes high $400 \mu$ s or more later (monostable multivibrator time constant) after the subcode is read out, the CPU determines that new data (which passed the CRC check) has been loaded.
- In the CXD2508AQ/AR, when 80-bit data is loaded, the order of the MSB and LSB is inverted for each byte. As a result, although the sequence of bytes is the same, the bits within the bytes are now ordered LSB first.
- Once the fact that the 80-bit data has been loaded is confirmed, SQCK is input so that the data can be read. In the CXD2508AQ/AR, the SQCK input is detected, and when it is low the retriggerable monostable multivibrator is reset.
- The retriggerable monostable multivibrator has a time constant from 270 to $400 \mu \mathrm{~s}$. When the duration of SQCK is high is less than this time constant, the monostable multivibrator is kept reset; during this interval, the S/P register is not loaded into the P/S register.
- While the monostable multivibrator is being reset, data can not be loaded in the 80 -bit parallel/serial register. In other words, while reading out with a clock cycle shorter than the monostable multivibrator time constant, the register will not be rewritten by CRCOK and others.
- Fig. 2-3 shows Timing Chart.
- Although a clock is input from SQCK pin to actually perform these operations, the high and low intervals for this clock should be between 750 ns and $120 \mu \mathrm{~s}$.


## Timing Chart 2-1



SCOR


EXCK


Block Diagram 2-2

Timing Chart 2-3


## 3. Description of Other Functions

## 3-1. Channel Clock Regeneration by Digital PLL Circuit

- The channel clock is necessary for demodulating the EFM signal regenerated by the optical system. Assuming T as the channel clock cycle, the EFM signal is demodulated in an integer multiple of T from 3 T to 11T. In order to read the information in the EFM signal, this integer value must be read correctly. As a result, T , that is channel clock, is required.
In an actual player, the fluctuation in the spindle rotation alters the width of the EFM signal pulses, making a PLL necessary for regenerating channel clock.

The block diagram of this PLL is shown in Fig. 3-1.
The CXD2508AQ/AR has a built-in two-stage PLL as shown in the diagram.

- The first-stage PLL generates a high-frequency clock needed by the second-stage digital PLL.
- The second-stage PLL is a digital PLL that regenerates the actual channel clock, and has a $\pm 250 \mathrm{kHz}$ (normal state) or more capture range.


## Block Diagram 3-1



## 3-2. Frame Sync Protection

- In a CD player operating at normal speed, a frame sync is recorded approximately every $136 \mu \mathrm{~s}(7.35 \mathrm{kHz})$. This signal is used as a reference to know which data is the data within a frame. Conversely, if the frame sync can not be recognized, the data is processed as error data because it can not be recognized what the data is. As a result, recognizing the frame sync properly is extremely important for improving playability.
- In the CXD2508AQ/AR, window protection and forward protection/backward protection have been adopted for frame sync protection. The adoption of these functions achieves very powerful frame sync protection.
There are two window widths: one for cases where a rotational disturbance affects the player and the other for cases where there is no rotational disturbance (WSEL $=0 / 1$ ). In addition, the forward protection counter is fixed to 13 , and the backward protection counter is fixed to 3 . In other words, when the frame sync is being played back normally and then can not be detected due to scratches, a maximum of 13 frames are inserted. If frame sync can not be detected for 13 frames or more, the window is released and the frame sync is resynchronized.
In addition, immediately after the window is released and resynchronization is executed, if a proper frame sync can not be detected within 3 frames, the window is released immediately.


## 3-3. Error Correction

- In the CD format, one 8-bit data contains two error correction codes, C1 and C2. For C1 correction, the code is created with 28 -byte information and 4 -byte C1 parity.
For C 2 correction, the code is created with 24 -byte information and 4 -byte parity.
Both C1 and C2 are Reed-Solomon codes with a minimum distance of 5 .
- The CXD2508AQ/AR SEC strategy provides excellent playability through powerful frame sync protection and C1 and C2 error corrections.
- The correction status can be monitored outside the LSI.

See Table 3-1.

- When the C 2 pointer is high, the data in question was uncorrectable. Either the pre-value was held for that data, or an average value interpolation was made.

| MNT3 | MNT1 | MNT0 | Description |
| :---: | :---: | :---: | :--- |
| 0 | 0 | 0 | No C1 errors |
| 0 | 0 | 1 | One C1 errors corrected |
| 0 | 1 | 1 | C1 correction impossible |
| 1 | 0 | 0 | No C2 errors |
| 1 | 0 | 1 | One C2 errors corrected |
| 1 | 1 | 1 | C2 correction impossible |

Table 3-1.

## Timing Chart 3-2



## 3-4. DA Interface

- The CXD2508AQ/AR DA interface is as described below.

This interface includes 48 cycles of the bit clock within one LRCK cycle, and is MSB first. When LRCK is high, the data is for the left channel.
Timing Chart 3-3
48bit slot Normal-Speed Playback


## 3-5. Digital Out

There are three Digital Out formats: the type 1 format for broadcasting stations, the type 2 form 1 format for home use, and the type 2 form 2 format for the manufacture of software.
The CXD2508AQ/AR supports type 2 form 1.
Sub Q data which are matched twice in succession after a CRC check are input to the first four bits (bit 0 to 3) of channel status.

Digital Out C bit


Table 3-2.

## 3-6. Servo Auto Sequencer

This function performs a series of controls, including auto focus and track jumps. When the auto sequence command is received from the CPU, auto focus, 1 track jump, 2 N track jumps, and N track move are executed automatically.
SSP (servo signal processor LSI) is used in an exclusive manner during the auto sequence execution (when XBUSY = low), so that commands from the CPU are not transferred to the SSP, but they can be sent to the CXD2508AQ/AR.
Connect the CPU, RF and SSP as shown in Fig. 3-4.
When CLOK goes from low to high while XBUSY is low, XBUSY does not become high for a maximum of $100 \mu s$ after that point. This is designed to prevent the transfer of erroneous data to the SSP when XBUSY changes from low to high by the monostable multivibrator, which is reset by CLOK being low (when XBUSY is low).
(a) Auto Focus (\$47)

Focus search up is performed, FOK and FZC are checked, and the focus servo is turned on.
If $\$ 47$ is received from the CPU, the focus servo is turned on according to Fig. 3-5. The auto focus starts with focus search up, and the pickup should be lowered beforehand (focus search down). In addition, blind E of register 5 is used to eliminate FZC chattering. In other words, the focus servo is turned on at the falling edge of FZC after FZC has been continuously high for a longer time than E.

## Connection diagram for using auto sequencer (example)



Fig. 3-4.


Fig. 3-5-(a). Auto Focus Flow Chart


Fig. 3-5-(b). Auto Focus Timing Chart
(b) Track Jump

1, 10, and 2 N -track jumps are performed respectively. Always use this when focus, tracking, and the sled servo are on. Note that tracking gain up and braking on (\$17) should be sent beforehand because they are not performed.

- 1-track jump

When \$48 (\$49 for REV) is received from the CPU, an FWD (REV) 1-track jump is performed in accordance with Fig. 3-6. Set blind A and brake B with register 5 .

- 10-track jump

When \$4A (\$4B for REV) is received from the CPU, an FWD (REV) 10-track jump is performed in accordance with Fig. 3-7. The principal difference between the 10 -track jump and the 1 -track jump is whether to kick the sled or not. In addition, after kicking the actuator, 5 tracks have been counted through CNIN , and the brake is applied to the actuator. Then, the actuator speed is found to have slowed up enough (determined by the CNIN cycle becoming longer than the overflow C set in register 5), and the tracking and sled servos are turned on.

- 2 N -track jump

When \$4C (\$4D for REV) is received from the CPU, an FWD (REV) 2N-track jump is performed in accordance with Fig. $3-8$. The track jump count " N " is set in register 7 . Although N can be set to $2^{16}$ tracks, note that the setting is actually limited by the actuator. CNIN is used for counting the number of jumps.
Although the 2 N -track jump basically follows the same sequence as the 10 -track jump, the one difference is that after the tracking servo is turned on, the sled continues to move only for " D ", set in register 6.

- N-track move

When $\$ 4 \mathrm{E}$ ( $\$ 4 \mathrm{~F}$ for REV) is received from the CPU, an FWD (REV) N-track move is performed in accordance with Fig. 3-9. N can be set to a maximum of $2^{16}$ tracks. CNIN is used for counting the number of jumps. This N -track move uses a method in which only the sled is moved, and is suited for moves over thousands of tracks.


Fig. 3-6-(a). 1-Track Jump Flow Chart


Fig. 3-6-(b). 1-Track Jump Timing Chart


Fig. 3-7-(a). 10-Track Jump Flow Chart


Fig. 3-7-(b). 10-Track Jump Timing Chart


Fig. 3-8-(a). 2N-Track Jump Flow Chart


Fig. 3-8-(b). 2N-Track Jump Timing Chart


Fig. 3-9-(a). N-Track Move Flow Chart


Fig. 3-9-(b). N-Track Move Timing Chart

## 3-7. Digital CLV

Fig. 3-10 shows the Block Diagram. Digital CLV makes PWM output in CLVS and CLVP with the MDS error and MDP error signal sampling frequency increased to 130 kHz during normal-speed operation. In addition, the digital spindle servo can set the gain.


Fig. 3-10. Block Diagram

## 3-8. Asymmetry Compensation

Fig. 3-11 shows the Block Diagram and Circuit Example.


Fig. 3-11. Example of Asymmetry Compensation Application Circuit

## 3-9. Setting Method of the CXD2508AQ/AR Playback Speed

## (a) Signal processing block (DSP block)

The playback mode shown below can be selected by the combination of crystal, XTSL pin and doublespeed command (DSPB) in the CXD2508AQ/AR.

Playback mode at DSP block

| Mode | Crystal | XTSL | DSPB | Speed at DSP block |
| :---: | :---: | :---: | :---: | :---: |
| 1 | $768 F s$ | 1 | 0 | $\times 1$ |
| 2 | $768 F s$ | 1 | 1 | $\times 2$ |
| 3 | $384 F s$ | 0 | 0 | $\times 1$ |
| 4 | $384 F s$ | 0 | 1 | $\times 2$ |
| $5^{* 1}$ | $384 F s$ | 1 | 1 | $\times 1$ |

$\mathrm{Fs}=44.1 \mathrm{kHz}$
*1 Low power consumption mode. The processing speed is halved in the LSI so that the power consumption can be decreased.
(b) DAC block

The operating speed at DAC block is determined by the crystal and the double-speed command DADS in DAC block in spite of the operating conditions of DSP block mentioned above. Then, the playback mode for DAC block and DSP block can be determined independently. (For example, normal-speed playback for DSP block; low power consumption playback for DAC block.) The DAC block supports the normal speed and double speed.

* DADS is controlled by sending the command to DSP block.

> Playback mode at DAC block

| Mode | Crystal | DADS | Speed at DAC block |
| :---: | :---: | :---: | :---: |
| 1 | 768 Fs | 0 | $\times 1$ |
| 2 | 768 Fs | 1 | $\times 2$ |
| $3^{* 2}$ | 384 Fs | 1 | $\times 1$ |

*2 Low power consumption mode. The processing speed is halved in the LSI so that the power consumption can be decreased.

## 4. 1-bit DAC Block

## 4-1. PWM Output Pattern

In the CXD2508AQ/AR, PWM outputs from the DAC include forward phase PWM (RPWM, LPWM) and inverted PWM (NRPWM, NLPWM). By determining the difference between these PWM outputs in the subsequent analog LPF, the noise and others can be canceled in the digital block. In addition, this method also yields improvements in the analog characteristics.
The PWM output waveforms differ for each of the CXD2508AQ/AR three playback modes (normal, doublespeed, and pseudo double-speed). (In the following explanation, $F s=44.1 \mathrm{kHz}$.)
During normal speed playback ( $\mathrm{DSPB}=0$, crystal $=768 \mathrm{Fs}$ ), eleven values (integers from -5 to 5 ) are taken within the 32 Fs cycle. The minimum pulse width is -5 , and the maximum pulse width is +5 . The minimum variation width of change for PWM is the 384 Fs cycle. (See Fig. 4-3.)


Fig. 4-1.
In double-speed playback ( $\mathrm{DSPB}=1$, crystal $=768 \mathrm{Fs}$ ), five values $(-4,-2,0,2,4)$ are taken within the 64 Fs cycle. (See Fig. 4-4.)


Fig. 4-2.

In pseudo double-speed playback ( $\mathrm{DSPB}=1$, crystal $=384 \mathrm{Fs}$ ), five values $(-4,-2,0,2,4)$ are taken within the 32Fs cycle. (See Fig. 4-5.)


Fig. 4-3.

## 4-2. Input Timing for DAC Block

Fig. 4-4 shows the input timing for DAC section.
In the CXD2508AQ/AR, there is no internal transfer of sound data from the CD signal processing block to DAC block. Therefore, data can be transferred to DAC block through an audio DSP and others.
When data is input to DAC block without passing through an audio DSP or similar device, data should be connected externally. In that case, EMPH, LRCK, and PCMD can be connected directly with EMPHI, LRCKI, and PCMDI respectively. (See the Application Circuit.)
Normal-Speed Playback



Fig. 4-4. Input Timing for DAC Block

## 4-4. Description of Functions

## No-Sound Data Detection

The no-sound data detection function detects low-level data on both the left and right channels in audio data from the 1Fs 48 -bit slot and outputs a zero detection signal when that data continues unchanged for a certain period of time.
The audio data is in two's complement format and data in which the upper 12 bits are all " 0 " or all " 1 " is regarded as low level data. When this data continues unchanged for 32,768 samples ( 743 ms when Fs = 44.1 kHz ), the zero detection signal is output. In other words, once a certain period of time during which lowlevel data is detected elapses, the signal is regarded to be in the no-sound state.

The zero detection signal is output from ZEROL (left channel) and ZEROR (right channel) pins. The zero detection output timing is shown in Fig. 4-5.


ZEROR and ZEROL


Fig. 4-5. Zero Detection Output Timing

## Forced Mute

The forced mute can be executed independently for DSP block and DAC block.
DSP can be forcibly muted by setting "1" in MUTE for D1 of register A. This mute can be released by setting " 0 " in MUTE for D1 of register A. Also, the both of left and right channels can be forcibly muted by inputting a high signal to MUTE pin for DAC block (in this event, a soft mute is not performed). In this instance, a fixed pattern is output for the PWM output. To release the mute, input a low signal to MUTE pin.

## Digital De-emphasis

When EMPHI pin (Pin 63) is set high, de-emphasis can be applied by using the IIR filter. However, in normalplayback mode the time constants are as follows:
$\tau 1=50 \mu \mathrm{~s}, \tau 2=15 \mu \mathrm{~s}$.

## Application Circuit



Application circuits shown are typical examples illustrating the operation of the devices. Sony cannot assume responsibility for any problems arising out of the use of these circuits or for any infringement of third party patent and other right due to same.

## Package Outline Unit: mm

## CXD2508AQ

80PIN QFP (PLASTIC)

PACKAGE STRUCTURE

| PACKAGE MATERIAL | EPOXY RESIN |
| :--- | :--- |
| LEAD TREATMENT | SOLDER PLATING |
| LEAD MATERIAL | COPPER / 42 ALLOY |
| PACKAGE WEIGHT | 1.6 g |

CXD2508AQ

80PIN QFP (PLASTIC)

22.6


## CXD2508AQ

QFP 80PIN (PLASTIC)

NOTE: Dimension "*" does not include mold protrusion.
PACKAGE STRUCTURE

|  |  |
| :--- | :--- |
| SONY CODE | QFP-80P-L051 |
| EIAJ CODE | $*$ QFP080-P-1420-AH |
| JEDEC CODE | - |$\quad$| PACKAGE MATERIAL | EPOXY RESIN |  |
| :--- | :--- | :--- |
|  | LEAD TREATMENT | SOLDER PLATING |
| LEAD MATERIAL | 42 ALLOY |  |
| PACKAGE WEIGHT | 1.6 g |  |

## CXD2508AR

80PIN LQFP (PLASTIC)

NOTE: Dimension "*" does not include mold protrusion.
DETAIL A

| SONY CODE | LQFP-80P-L01 |
| :--- | :--- |
| EIAJ CODE | *QFP080-P-1212-A |
| JEDEC CODE |  |

PACKAGE STRUCTURE

| PACKAGE MATERIAL | EPOXY/PHENOL RESIN |
| :--- | :--- |
| LEAD TREATMENT | SOLDER PLATING |
| LEAD MATERIAL | 42 ALLOY |
| PACKAGE WEIGHT | 0.5 g |

