

# ET-AMS1085M-XX

August 2005

lnv № 428

# LOW DROPOUT POSITIVE VOLTAGE REGULATOR

### DESCRIPTION

The AMS1085M series of positive adjustable and fixed regulators are designed to provide 3A with high efficiency. All internal circuitry is designed to operate down to 1.3V input to output differential. On-chip trimming adjusts the reference voltage to 1%.

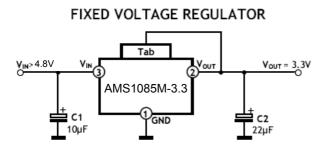
## **APPLICATIONS**

- High Efficiency Linear Regulators
- Post Regulators for Switching Supplies
- Adjustable Power Supply

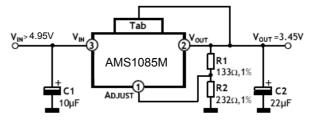
#### FEATURES

- Adjustable or Fixed Output
- Output Current of 3A
- Low Dropout, 1.5V max at 3A Output Current
- 0.04% Line Regulation
- 0.2 % Load Regulation
- 100% Thermal Limit Burn-In
- Fast Transient Response

# **TYPICAL APPLICATION DATA**



# ADJUSTABLE VOLTAGE REGULATOR



NOTES:

- C1 needed if device is far from filter capacitors
- C2 minimum value required for stability

$$\boldsymbol{V}_{\text{out}} = \boldsymbol{V}_{\text{ref}} \times \left(1 + \frac{R2}{R1}\right) + \boldsymbol{I}_{\text{adj}} \times R2$$

		PIN NO.	FIXED VERSION	ADJUSTABLE VERSION	Tab	
T0-220-3	Tab 1 2 3	1	GROUND	ADJUST		-263-3
		2	Vout	νουτ	1 2 3 PPP	-0T
52-2	Tab	3	V <sub>IN</sub>	V <sub>IN</sub>	2	263-2
T0-2		Tab	V <sub>out</sub>	Vout	Tab 1	T0-2

### **PIN DESCRIPTION**



# ET-AMS1085M-XX

Inv № 428

LOW DROPOUT POSITIVE VOLTAGE REGULATOR

## ABSOLUTE MAXIMUM RATINGS

SYMBOL	PARAMETER	VALUE	UNIT			
V <sub>CC</sub>	Power Dissipation	Internally Limited	W			
V <sub>IN</sub>	Input Voltage	15	V			
	Operation Junction Temperature Range					
TJ	Control Section	0 to 125	٥C			
	Power Transistor	0 to 150	C			
T <sub>STG</sub>	Storage Temperature Range	-65 to +150	°C			
T <sub>LEAD</sub>	Lead Temperature (Soldering 10 sec)	300	0 <sup>0</sup> C			

### **ELECTRICAL CHARACTERISTICS**

 $I_{LOAD}$  = 0mA and  $T_{J}$  = +25  $^{0}C$  (unless otherwise noted)

PARAMETER	DEVICE	TEST CONDITIONS		MIN.	TYP.	MAX.	UNIT	
	AMS1085M	V <sub>IN</sub> =5V, I <sub>LOAD</sub> = 10mA		1.238	1.250	1.262	V	
Reference Voltage Note 1		$V_{IN} - V_{OUT}$ = 1.5V to 10V, I <sub>LOAD</sub> = 10mA to 3A	*	1.225	1.250	1.275		
	All fixed versions	V <sub>IN</sub> – V <sub>OUT</sub> = 1.5V, Variator from nominal V <sub>OUT</sub>		-1	-	+1	%	
Output Voltage Note 1		$V_{IN} - V_{OUT} = 1.5V$ to 10V $I_{LOAD} = 0$ mA to 3A, Variator from nominal $V_{OUT}$	*	-2	-	+2		
Line Regulation Note 1	All	I <sub>LOAD</sub> = 10mA, * - V <sub>IN</sub> - V <sub>OUT</sub> = 1.5V to 10V		0.04	0.20	%		
Load Regulation Note 1	All	V <sub>IN</sub> – V <sub>OUT</sub> = 1.5V I <sub>LOAD</sub> = 10mA to 3A	*	-	0.2	0.40	70	
Minimum Load Current	AMS1085M	$V_{IN} = 5V, V_{ADJ} = 0V$	*	-	3	7	mA	
Ground Pin Current	All fixed versions	$V_{IN} - V_{OUT} = 1.5V$ $I_{LOAD} = 10$ mA to 3A	*	-	7	10	mA	
Adjust Pin Current	AMS1085M	$V_{IN} - V_{OUT} = 1.5V$ to 10V I <sub>LOAD</sub> = 10mA	*	-	40	90	μA	
Current Limit	All	$V_{IN} - V_{OUT} = 1.5V$	*	3	4.5	-	А	
Ripple Rejection Note 2	All	$V_{IN} - V_{OUT} = 3V$ $I_{LOAD} = 3A$	*	60	65	-	dB	
Dropout Voltage Note 1,3	All	I <sub>LOAD</sub> = 3A	*	-	1.3	1.5	V	
Temperature Coefficient	All	$V_{IN} - V_{OUT} = 1.5V,$ $I_{LOAD} = 10mA$	*	-	0.005	-	%/ <sup>0</sup> C	

The \* denotes the specifications which apply over the full temperature range (see previous table 1,  $T_{\rm J})$  NOTES:

1: Low duty pulse testing with Kelvin connections required.

**2**: 120Hz input ripple ( $C_{ADJ}$  for ADJ = 25 $\mu$ F,  $C_{OUT}$  = 25 $\mu$ F)

**3**:  $\Delta V_{OUT}$ ,  $\Delta V_{REF} = 1\%$ 

#### **APPLICATION INFORMATION**

The AMS1085M series of adjustable and fixed regulators are easy to use and have all the protection features expected in high performance voltage regulators: short circuit protection and thermal shut-down.

Pin compatible with older three terminal adjustable regulators, these devices offer the advantage of a lower dropout voltage, more precise reference tolerance and improved reference stability with temperature.

#### STABILITY

The circuit design used in the AMS1085M series requires the use of an output capacitor as part of the device frequency compensation.

The addition of 150µF aluminum electrolytic or a 22µF solid tantalum on the output will ensure stability for all operating conditions.

When the adjustment terminal is bypassed with a capacitor to improve the ripple rejection, the requirement for an output capacitor increases. The value of 22µF tantalum or 150µF aluminum covers all cases of bypassing the adjustment terminal. Without bypassing the adjustment terminal smaller capacitors can be used with equally good results.



# ET-AMS1085M-XX

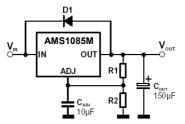
August 2005

Inv № 428

# LOW DROPOUT POSITIVE VOLTAGE REGULATOR

To ensure good transient response with heavy load current changes capacitor values on the order of  $100\mu$ F are used in the output of many regulators. To further improve stability and transient response of these devices larger values of output capacitor can be used.

#### **PROTECTION DIODES**



Unlike older regulators, the AMS1085M family does not need any protection diodes between the adjustment pin and the output and from the output to the input to prevent over-stressing the die.

Internal resistors are limiting the internal current paths on the AMS1085M adjustment pin, therefore even with capacitors on the adjustment pin no protection diode is needed to ensure device safety under short-circuit conditions.

Diodes between the input and output are not usually needed.

Microsecond surge currents of 50A to 100A can be handled by the internal diode between the input and output pins of the device. In normal operations it is difficult to get those values of surge currents even with the use of large output capacitances. If high value output capacitors are used, such as  $1000\mu$ F to  $5000\mu$ F and the input pin is instantaneously shorted to ground, damage can occur. A diode from output to input is recommended, when a crowbar circuit at the input of the AMS1085M is used. Normal power supply cycling or even plugging and unplugging in the system will not generate current large enough to do any damage.

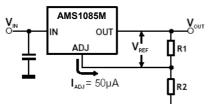
The adjustment pin can be driven on a transient basis  $\pm 25V$ , with respect to the output without any device degradation. As with any IC regulator, none the protection circuitry will be functional and the internal transistors will break down if the maximum input to output voltage differential is exceeded.

#### **RIPPLE REJECTION**

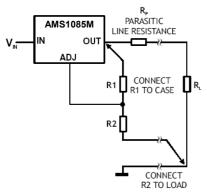
The ripple rejection values are measured with the adjustment pin bypassed. The impedance of the adjust pin capacitor at the ripple frequency should be less than the value of R1 (normally  $100\Omega$  to  $120\Omega$ ) for a proper bypassing and ripple rejection approaching the values shown. The size of the required adjust pin capacitor is a function of the input ripple frequency. If R1= $100\Omega$  at 120Hz the adjust pin capacitor should be 25µF. At 10kHz only  $0.22\mu$ F is needed.

The ripple rejection will be a function of output voltage, in circuits without an adjust pin bypass capacitor. The output ripple will increase directly as a ratio of the output voltage to the reference voltage ( $V_{OUT} / V_{REF}$ ).

#### **OUTPUT VOLTAGE**



#### LOAD REGULATION



The AMS1085M series develops a 1.25V reference voltage between the output and the adjust terminal. Placing a resistor between these two terminals causes a constant current to flow through R1 and down through R2 to set the overall output voltage.

This current is normally the specified minimum load current of 10mA. Because  $I_{ADJ}$  is very small and constant it represents a small error and it can usually be ignored.

True remote load sensing it is not possible to provide, because the AMS1085M is a three terminal device. The resistance of the wire connecting the regulator to the load will limit the load regulation.

The data sheet specification for load regulation is measured at the bottom of the package. Negative side sensing is a true Kelvin connection, with the bottom of the output divider returned to the negative side of the load.

The best load regulation is obtained when the top of the resistor divider R1 is connected directly to the case not to the load. If R1 were connected to the load, the effective resistance between the regulator and the load would be:

$$\frac{\mathbf{R}_{P} \times (\mathbf{R} 2 + \mathbf{R} 1)}{\mathbf{R} 1}, \ \mathbf{R}_{P} = \text{Parasitic Line Resistance}$$

Connected as shown Fig.3,  $R_P$  is not multiplied by the divider ratio. Using 16-gauge wire the parasitic line resistance is about 0.004 $\Omega$  per foot, translating to 4mV/ft at 1A load current. It is important to keep the positive lead between regulator and load as short as possible and use large wire or PC board traces.