

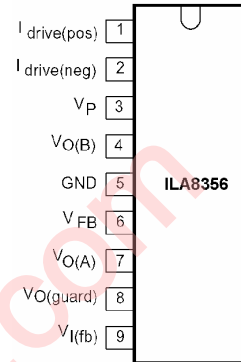
### DC-coupled vertical deflection circuit

The ILA8356 is a power circuit for use in 90° and 110° colour deflection systems for field frequencies of 50 to 120 Hz. The circuit provides a DC driven vertical deflection output circuit, operating as a highly efficient class G system.

#### FEATURES

- Few external components
- Highly efficient fully DC-coupled vertical output bridge circuit
- Vertical flyback switch
- Guard circuit
- Protection against:
  - short-circuit of the output pins (7 and 4)
  - short-circuit of the output pins to  $V_P$
- Temperature (thermal) protection
- High EMC immunity because of common mode inputs

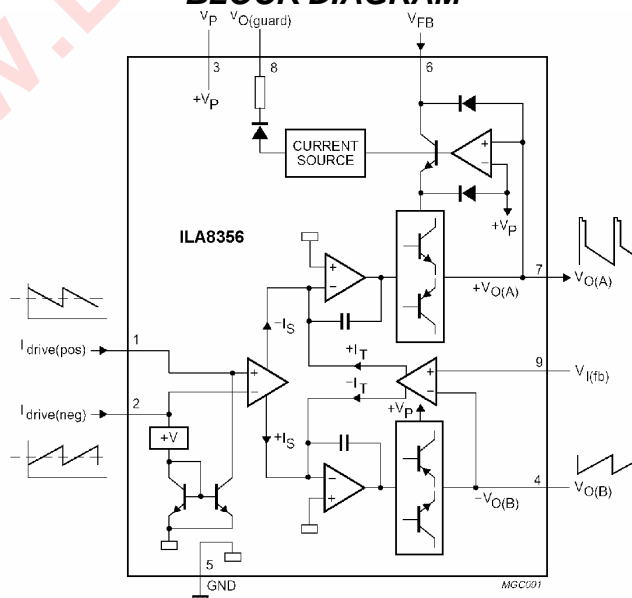
A guard signal in zoom mode



#### PINNING

SYMBOL	PIN	DESCRIPTION
$I_{drive(pos)}$	1	input power-stage (positive); includes $I_{I(s)}$ signal bias
$I_{drive(neg)}$	2	input power-stage (negative); includes $I_{I(s)}$ signal bias
$V_P$	3	operating supply voltage
$V_{O(B)}$	4	output voltage B
GND	5	ground
$V_{FB}$	6	input flyback supply voltage
$V_{O(A)}$	7	output voltage A
$V_{O(guard)}$	8	guard output voltage
$V_{I(fb)}$	9	input feedback voltage

#### BLOCK DIAGRAM



**QUICK REFERENCE DATA**

SYMBOL	PARAMETER	MIN.	TYP.	MAX.	UNIT
<b>DC supply</b>					
$V_P$	supply voltage	9	4.5	25	V
$I_q$	quiescent supply current		30		mA
<b>Vertical circuit</b>					
$I_O(p-p)$	output current (peak-to-peak value)			2	A
$I_{diff}(p-p)$	differential input current (peak-to-peak value)		600		$\mu$ A
$V_{diff}(p-p)$	differential input voltage (peak-to-peak value)		1.5	1.8	V
<b>Flyback switch</b>					
$I_M$	peak output current			+1	A
$V_{FB}$	flyback supply voltage			50	V
<b>Thermal data</b>					
$T_{stg}$	storage temperature	55		+150	$^{\circ}$ C
$T_{amb}$	operating ambient temperature	25		+75	$^{\circ}$ C
$T_{vj}$	virtual junction temperature			150	$^{\circ}$ C

**FUNCTIONAL DESCRIPTION**

The vertical driver circuit is a bridge configuration. The deflection coil is connected between the output amplifiers, which are driven in phase opposition. An external resistor ( $R_M$ ) connected in series with the deflection coil provides internal feedback information. The differential input circuit is voltage driven. An external resistor ( $R_{CON}$ ) connected between the differential input determines the output current through the deflection coil. The relationship between the differential input current and

the output current is defined by:  $I_{diff} \times R_{CON} = I_{coil} \times R_M$ . The output current is adjustable from 0.5 A (p-p) to 2 A (p-p) by varying  $R_M$ . The maximum input differential voltage is 1.8 V. In the application it is recommended that  $V_{diff} = 1.5$  V (typ). This is recommended because of the spread of input current and the spread in the value of  $R_{CON}$ . The flyback voltage is determined by an additional supply voltage  $V_{FB}$ . The principle of operating with two supply voltages (class G) makes it possible to fix the supply voltage  $V_P$  optimum for the scan voltage and the second supply voltage  $V_{FB}$  optimum for the flyback voltage. Using this method, very high efficiency is achieved.

The supply voltage  $V_{FB}$  is almost totally available as flyback voltage across the coil, this being possible due to the absence of a decoupling capacitor (not necessary, due to the bridge configuration). The output circuit is fully protected against the following:

- thermal protection
- short-circuit protection of the output pins (pins 4 and 7)
- short-circuit of the output pins to  $V_P$ .

A guard circuit  $V_O(\text{guard})$  is provided. The guard circuit is activated at the following conditions:

- during flyback
- during short-circuit of the coil and during short-circuit of the output pins (pins 4 and 7) to  $V_P$  or ground
- during open loop
- when the thermal protection is activated.

This signal can be used for blanking the picture tube screen.

**LIMITING VALUES**

SYMBOL	PARAMETER	CONDITIONS	MIN.	MAX.	UNIT
<b>DC supply</b>					
$V_P$	supply voltage	non-operating		40	V
				25	V
$V_{FB}$	flyback supply voltage			50	V
<b>Vertical circuit</b>					
$I_{O(p-p)}$	output current (peak-to-peak value)	note 1		2	A
$V_{O(A)}$	output voltage (pin 7)			52	V
<b>Flyback switch</b>					
$I_M$	peak output current			+1.5	A
<b>Thermal data</b>					
$T_{stg}$	storage temperature		55	+150	°C
$T_{amb}$	operating ambient temperature		25	+75	°C
$T_{vj}$	virtual junction temperature			150	°C
$R_{th\ vj-c}$	resistance $v_j$ -case			4	K/W
$R_{th\ vj-a}$	resistance $v_j$ -ambient in free air			40	K/W
$t_{sc}$	short-circuiting time	note 2		1	hr

**Notes**

- $I_O$  maximum determined by current protection.
- Up to  $V_P = 18$  V.

**CHARACTERISTICS**

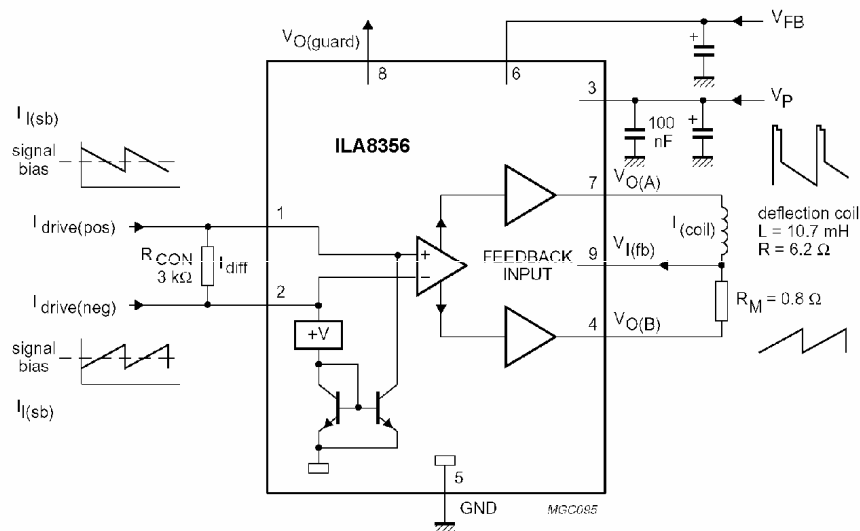
$V_P = 14.5$  V;  $T_{amb} = 25$  °C;  $V_{FB} = 45$  V;  $f_i = 50$  Hz;  $I_{I(sB)} = 400$  A; unless otherwise specified.

SYMBOL	PARAMETER	CONDITIONS	MIN.	TYP.	MAX.	UNIT
<b>DC supply</b>						
$V_P$	operating supply voltage		9.0	4.5	25	V
$V_{FB}$	flyback supply voltage		$V_P$		50	V
$I_P$	supply current	no signal; no load		30	55	mA
<b>Vertical circuit</b>						
$V_O$	output voltage swing (scan)	$I_{diff} = 0.6$ mA (p-p); $V_{diff} = 1.8$ V (p-p); $I_O = 2$ A (p-p)	13.2			V
LE	linearity error	$I_O = 2$ A (p-p); $I_O = 50$ mA (p-p);		1	4	%
$V_O$	output voltage swing (flyback) $V_{O(A)} - V_{O(B)}$	$I_{diff} = 0.3$ mA; $I_O = 1$ A (M)		40		V
$V_{DF}$	forward voltage of the internal efficiency diode ( $V_{O(A)} - V_{FB}$ )	$I_O = 1$ A (M); $I_{diff} = 0.3$ mA			1.5	V
$ I_{os} $	output offset current	$I_{diff} = 0$ ; $I_{I(sB)} = 50$ to $500$ $\mu$ A			40	mA
$ V_{os} $	offset voltage at the input of the feedback amplifier ( $V_{I(fb)} - V_{O(B)}$ )	$I_{diff} = 0$ ; $I_{I(sB)} = 50$ to $500$ $\mu$ A			24	mV
$\Delta V_{osT}$	output offset voltage as a function of temperature	$I_{diff} = 0$			72	$\mu$ V/K
$V_{O(A)}$	DC output voltage	$I_{diff} = 0$ ;		6.5		V
$G_{vo}$	open-loop voltage gain ( $V_{7.4}/V_{1.2}$ )			80		dB

	open loop voltage gain ( $V_{7.4}/V_{9.4}$ ; $V_{1-2} = 0$ )			80			dB
$V_R$	voltage ratio $V_{1-2}/V_{9.4}$			0			dB
$f_{res}$	frequency response (3 dB)	open loop;		40			Hz
$G_I$	current gain ( $I_O/I_{diff}$ )			5000			
$\Delta G_c T$	current gain drift as a function of temperature					$10^{-4}$	K
$I_{I(sb)}$	signal bias current		50	400	500		$\mu A$
$I_{FB}$	flyback supply current	during scan			100		$\mu A$
PSRR	power supply ripple rejection			80			dB
$V_{I(DC)}$	DC input voltage			2.7			V
$V_{I(CM)}$	common mode input voltage	$I_{I(sb)} = 0$	0		1.6		V
$I_{bias}$	input bias current	$I_{I(sb)} = 0$		0.1	0.5		$\mu A$
$I_{O(CM)}$	common mode output current	$\Delta I_{I(sb)} = 300 \mu A$ (p-p); $f_i = 50$ Hz; $I_{diff} = 0$		0.2			mA
<b>Guard circuit</b>							
$I_O$	output current	not active; $V_{O(guard)} = 0$ V				50	$\mu A$
		active; $V_{O(guard)} = 4.5$ V	1			2.5	mA
$V_{O(guard)}$	output voltage on pin 8	$I_O = 100 \mu A$				5.5	V
	allowable voltage on pin 8	maximum leakage current = $10 \mu A$ ;				40	V

$V_P = 13.5$  V;  $I_{O(p-p)} = 1.87$  A;  $I_{I(sb)} = 400 \mu A$ ;  $I_{diff(p-p)} = 500 \mu A$ ;  $V_{FB} = 42$  V;  $t_{FB} = 0.6$  ms.

**APPLICATION DIAGRAM.**



• 9-Pin Plastic Power Single-in-Line (SIL-9MPF, SOT 131-2)

