

TPS7201Q, TPS7233Q, TPS7248Q, TPS7250Q
 TPS7201Y, TPS7233Y, TPS7248Y, TPS7250Y
MICROPOWER LOW-DROPOUT (LDO) VOLTAGE REGULATORS
 SLVS102E – MARCH 1995 – REVISED JANUARY 1997

- Available in 5-V, 4.85-V, and 3.3-V Fixed-Output and Adjustable Versions
- Dropout Voltage <85 mV Max at $I_O = 100$ mA (TPS7250)
- Low Quiescent Current, Independent of Load, 180 μ A Typ
- 8-Pin SOIC and 8-Pin TSSOP Package
- Output Regulated to $\pm 2\%$ Over Full Operating Range for Fixed-Output Versions
- Extremely Low Sleep-State Current, 0.5 μ A Max
- Power-Good (PG) Status Output

description

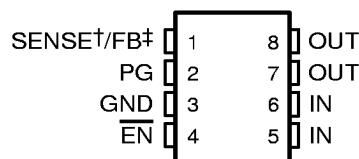
The TPS72xx family of low-dropout (LDO) voltage regulators offers the benefits of low-dropout voltage, micropower operation and miniaturized packaging. These regulators feature extremely low dropout voltages and quiescent currents compared to conventional LDO regulators. Offered in small-outline integrated-circuit (SOIC) packages and 8-terminal thin shrink small-outline (TSSOP), the TPS72xx series devices are ideal for cost-sensitive designs and where board space is at a premium.

A combination of new circuit design and process innovation has enabled the usual pnp pass transistor to be replaced by a PMOS device. Because the PMOS pass element behaves as a low-value resistor, the dropout voltage is very low – maximum of 85 mV at 100 mA of load current (TPS7250) – and is directly proportional to the load current (see Figure 1). Since the PMOS pass element is a voltage-driven device, the quiescent current is very low (300 μ A maximum) and is stable over the entire range of output load current (0 mA to 250 mA). Intended for use in portable systems such as laptops and cellular phones, the low-dropout voltage feature and micropower operation result in a significant increase in system battery operating life.

The TPS72xx also features a logic-enabled sleep mode to shut down the regulator, reducing quiescent current to 0.5 μ A maximum at $T_J = 25^\circ\text{C}$. Other features include a power-good function that reports low output voltage and may be used to implement a power-on reset or a low-battery indicator.

The TPS72xx is offered in 3.3-V, 4.85-V, and 5-V fixed-voltage versions and in an adjustable version (programmable over the range of 1.2 V to 9.75 V). Output voltage tolerance is specified as a maximum of 2% over line, load, and temperature ranges (3% for adjustable version).

**D, P, OR PW PACKAGE
(TOP VIEW)**



† SENSE – Fixed voltage options only
 (TPS7233, TPS7248, and TPS7250)
 ‡ FB – Adjustable version only (TPS7201)

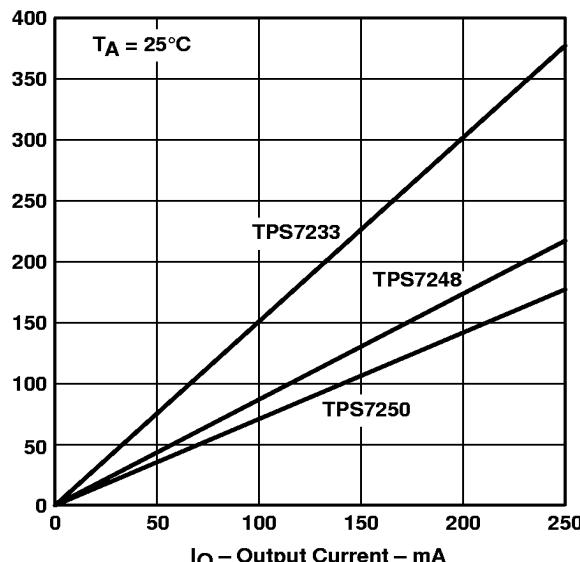


Figure 1. Typical Dropout Voltage Versus Output Current



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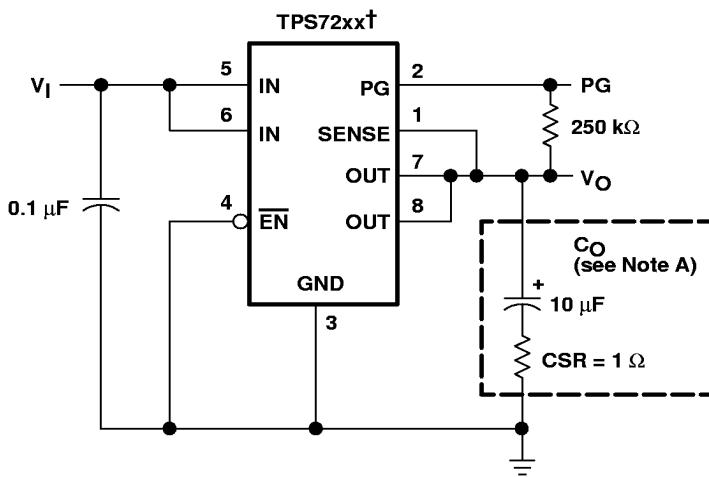
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SLVS102E – MARCH 1995 – REVISED JANUARY 1997

AVAILABLE OPTIONS

T _J	OUTPUT VOLTAGE (V)			PACKAGED DEVICES			CHIP FORM (Y)
	MIN	TYP	MAX	SMALL OUTLINE (D)	PDIP (P)	TSSOP (PW)	
–55°C to 150°C	4.9	5	5.1	TPS7250QD	TPS7250QP	TPS7250QPWLE	TPS7250Y
	4.75	4.85	4.95	TPS7248QD	TPS7248QP	TPS7248QPWLE	TPS7248Y
	3.23	3.3	3.37	TPS7233QD	TPS7233QP	TPS7233QPWLE	TPS7233Y
	Adjustable 1.2 V to 9.75 V		TPS7201QD	TPS7201QP	TPS7201QPWLE	TPS7201Y	

The D package is available taped and reeled. Add R suffix to device type (e.g., TPS7250QDR). The PW package is only available left-end taped and reeled. The TPS7201Q is programmable using an external resistor divider (see application information). The chip form is tested at 25°C.



† TPS7233, TPS7248, TPS7250 (fixed-voltage options)

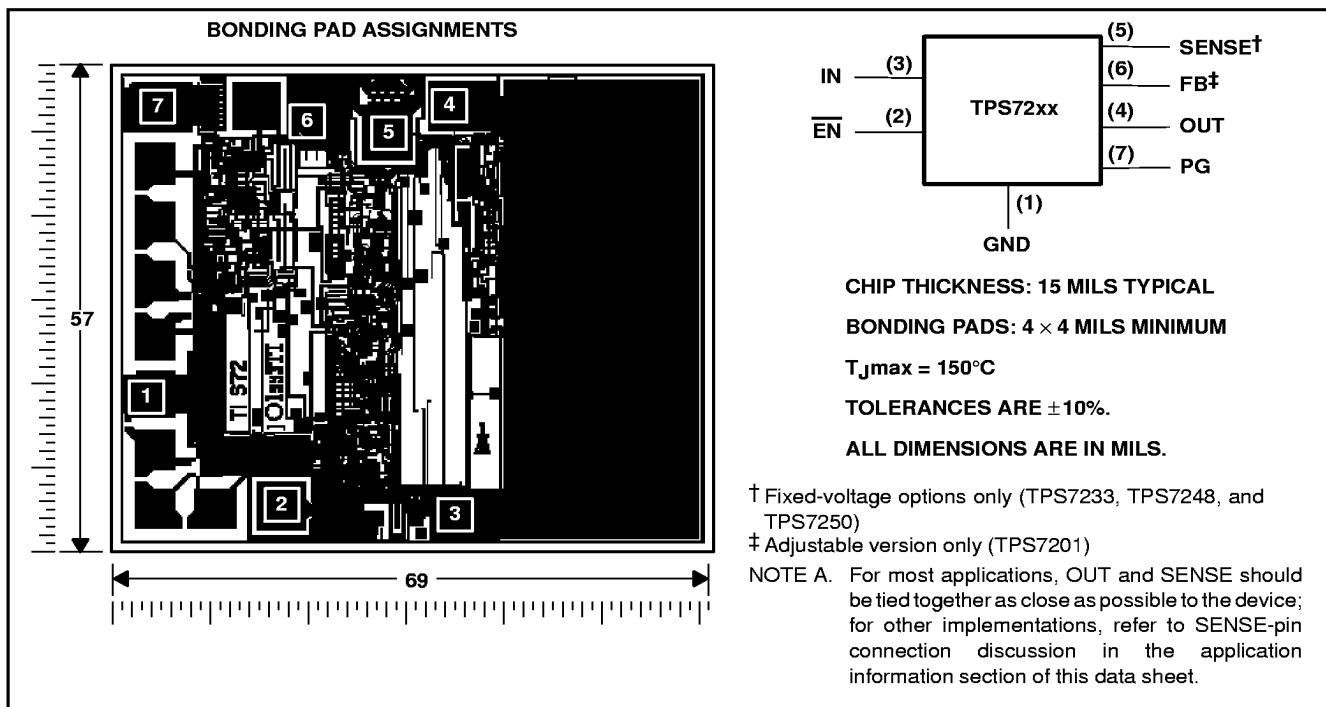
NOTE A: Capacitor selection is nontrivial. See application information section for details.

Figure 2. Typical Application Configuration

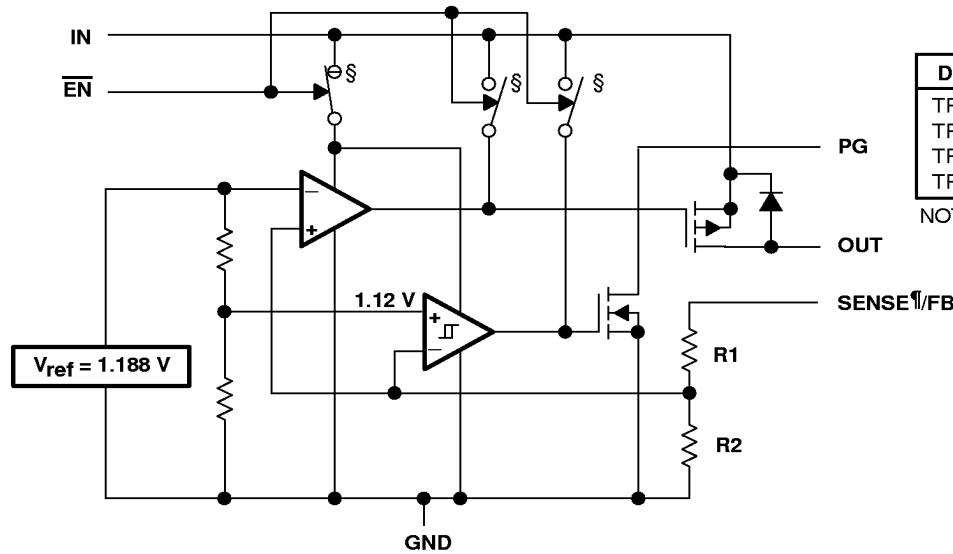
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MICROPOWER LOW-DROPOUT (LDO) VOLTAGE REGULATORS
 SLVS102E – MARCH 1995 – REVISED JANUARY 1997

TPS72xx chip information

These chips, when properly assembled, display characteristics similar to the TPS72xxQ. Thermal compression or ultrasonic bonding may be used on the doped aluminum bonding pads. The chips may be mounted with conductive epoxy or a gold-silicon preform.



functional block diagram



§ Switch positions are shown with EN low (active).

¶ For most applications, SENSE should be externally connected to OUT as close as possible to the device.
 For other implementations, refer to the SENSE-pin connection discussion in application information section.



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**TPS7201Q, TPS7233Q, TPS7248Q, TPS7250Q
TPS7201Y, TPS7233Y, TPS7248Y, TPS7250Y
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SLVS102E – MARCH 1995 – REVISED JANUARY 1997

absolute maximum ratings over operating free-air temperature range (unless otherwise noted)[†]

Input voltage range [‡] , V_I , PG, SENSE, \overline{EN}	–0.3 V to 11 V
Output current, I_O	1.5 A
Continuous total power dissipation	See Dissipation Rating Tables 1 and 2
Operating virtual junction temperature range, T_J	–55°C to 150°C
Storage temperature range, T_{STG}	–65°C to 150°C
Lead temperature 1.6 mm (1/16 inch) from case for 10 seconds	260°C

[†] Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

[‡] All voltage values are with respect to network ground terminal.

DISSIPATION RATING TABLE 1 – FREE-AIR TEMPERATURE (see Note 1 and Figure 3)

PACKAGE	$T_A \leq 25^\circ C$ POWER RATING	DERATING FACTOR ABOVE $T_A = 25^\circ C$	$T_A = 70^\circ C$ POWER RATING	$T_A = 85^\circ C$ POWER RATING	$T_A = 125^\circ C$ POWER RATING
D	725 mW	5.8 mW/°C	464 mW	377 mW	145 mW
P	1175 mW	9.4 mW/°C	752 mW	611 mW	235 mW
PW	525 mW	4.2 mW/°C	336 mW	273 mW	105 mW

DISSIPATION RATING TABLE 2 – CASE TEMPERATURE (see Note 1 and Figure 4)

PACKAGE	$T_C \leq 25^\circ C$ POWER RATING	DERATING FACTOR ABOVE $T_C = 25^\circ C$	$T_C = 70^\circ C$ POWER RATING	$T_C = 85^\circ C$ POWER RATING	$T_C = 125^\circ C$ POWER RATING
D	2063 mW	16.5 mW/°C	1320 mW	1073 mW	413 mW
P	2738 mW	21.9 mW/°C	1752 mW	1424 mW	548 mW
PW	2900 mW	23.2 mW/°C	1856 mW	1508 mW	580 mW

NOTE 1: Dissipation rating tables and figures are provided for maintenance of junction temperature at or below absolute maximum of 150°C. For guidelines on maintaining junction temperature within the recommended operating range, see application information section.

**MAXIMUM CONTINUOUS DISSIPATION
vs
FREE-AIR TEMPERATURE**

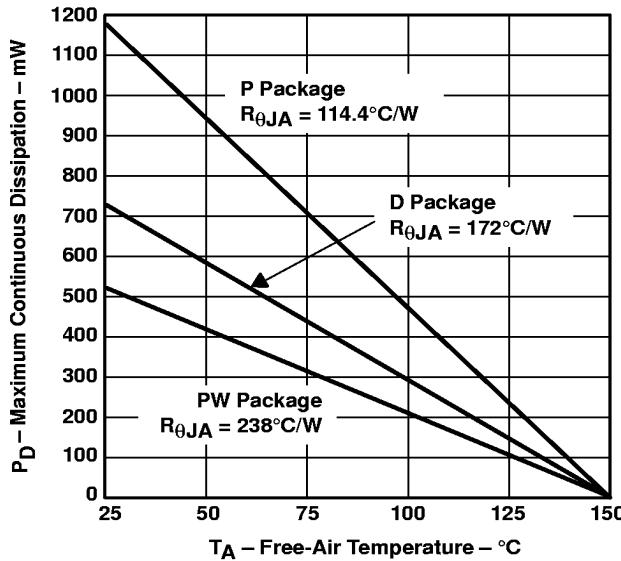


Figure 3

**MAXIMUM CONTINUOUS DISSIPATION
vs
CASE TEMPERATURE**

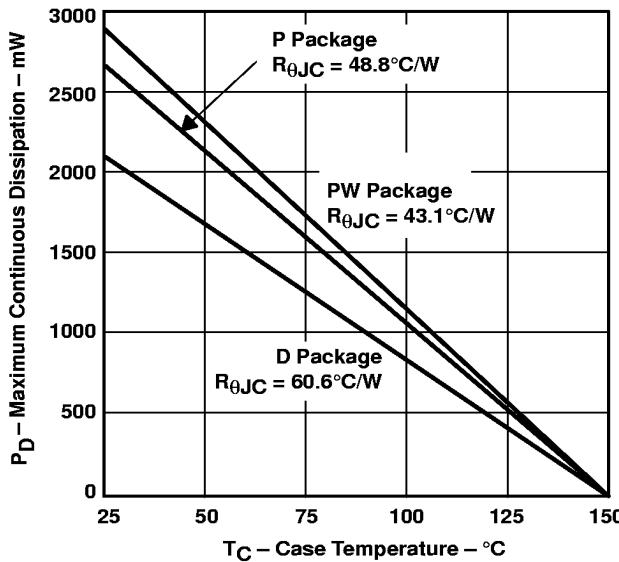


Figure 4



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TPS7201Y, TPS7233Y, TPS7248Y, TPS7250Y
MICROPOWER LOW-DROPOUT (LDO) VOLTAGE REGULATORS**
SLVS102E – MARCH 1995 – REVISED JANUARY 1997

recommended operating conditions

		MIN	MAX	UNIT
Input voltage, V_I [†]	TPS7201Q	2.5	10	V
	TPS7233Q	3.98	10	
	TPS7248Q	5.24	10	
	TPS7250Q	5.41	10	
High-level input voltage at \overline{EN} , V_{IH}		2		V
Low-level input voltage at \overline{EN} , V_{IL}			0.5	V
Output current, I_O		0	250	mA
Operating virtual junction temperature, T_J		-40	125	°C

[†] Minimum input voltage defined in the recommended operating conditions is the maximum specified output voltage plus dropout voltage at the maximum specified load range. Since dropout voltage is a function of output current, the usable range can be extended for lighter loads. To calculate the minimum input voltage for the maximum load current used in a given application, use the following equation:

$$V_{I(min)} = V_{O(max)} + V_{DO(max\ load)}$$

Because the TPS7201 is programmable, $r_{DS(on)}$ should be used to calculate V_{DO} before applying the above equation. The equation for calculating V_{DO} from $r_{DS(on)}$ is given in Note 3 under the TPS7201 electrical characteristics table. The minimum value of 2.5 V is the absolute lower limit for the recommended input-voltage range for the TPS7201.



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**TPS7201Q, TPS7233Q, TPS7248Q, TPS7250Q
 TPS7201Y, TPS7233Y, TPS7248Y, TPS7250Y
 MICROPOWER LOW-DROPOUT (LDO) VOLTAGE REGULATORS**

SLVS102E – MARCH 1995 – REVISED JANUARY 1997

electrical characteristics, $I_O = 10 \text{ mA}$, $\overline{\text{EN}} = 0 \text{ V}$, $C_O = 4.7 \mu\text{F}$ ($\text{CSR}^\dagger = 1 \Omega$), SENSE/FB shorted to OUT (unless otherwise noted)

PARAMETER	TEST CONDITIONS‡	T_J	TPS7201Q, TPS7233Q TPS7248Q, TPS7250Q			UNIT
			MIN	TYP	MAX	
Ground current (active mode)	$\overline{\text{EN}} \leq 0.5 \text{ V}$, $V_I = V_O + 1 \text{ V}$, $0 \text{ mA} \leq I_O \leq 250 \text{ mA}$	25°C	180	225	μA	
		-40°C to 125°C	325			
Input current (standby mode)	$\overline{\text{EN}} = V_I$, $2.7 \text{ V} \leq V_I \leq 10 \text{ V}$	25°C	0.5	μA		
		-40°C to 125°C	1			
Output current limit threshold	$V_O = 0 \text{ V}$ $V_I = 10 \text{ V}$	25°C	0.6	1	A	
		-40°C to 125°C	1.5			
Pass-element leakage current in standby mode	$\overline{\text{EN}} = V_I$, $2.7 \text{ V} \leq V_I \leq 10 \text{ V}$	25°C	0.5	μA		
		-40°C to 125°C	1			
PG leakage current	$V_{PG} = 10 \text{ V}$, Normal operation	25°C	0.5	μA		
		-40°C to 125°C	0.5			
Output voltage temperature coefficient		-40°C to 125°C	31	75	ppm/°C	
Thermal shutdown junction temperature			165		°C	
EN logic high (standby mode)	2.5 V ≤ $V_I \leq 6 \text{ V}$ $6 \text{ V} \leq V_I \leq 10 \text{ V}$	-40°C to 125°C	2	V		
			2.7			
EN logic low (active mode)	2.7 V ≤ $V_I \leq 10 \text{ V}$	25°C	0.5	V		
		-40°C to 125°C	0.5			
EN hysteresis voltage		25°C	50	mV		
EN input current	0 V ≤ $V_I \leq 10 \text{ V}$	25°C	-0.5	μA		
		-40°C to 125°C	-0.5			
Minimum V_I for active pass element		25°C	1.9	2.5	V	
		-40°C to 125°C	2.5			
Minimum V_I for valid PG	$I_{PG} = 300 \mu\text{A}$	25°C	0.95	1.5	V	
		-40°C to 125°C	1.9			

† CSR(compensation series resistance) refers to the total series resistance, including the equivalent series resistance (ESR) of the capacitor, any series resistance added externally, and PWB trace resistance to C_O .

‡ Pulse-testing techniques are used to maintain virtual junction temperature as close as possible to ambient temperature; thermal effects must be taken into account separately.



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**TPS7201Q, TPS7233Q, TPS7248Q, TPS7250Q
TPS7201Y, TPS7233Y, TPS7248Y, TPS7250Y
MICROPOWER LOW-DROPOUT (LDO) VOLTAGE REGULATORS**
SLVS102E – MARCH 1995 – REVISED JANUARY 1997

TPS7201Q electrical characteristics, $I_O = 10 \text{ mA}$, $V_I = 3.5 \text{ V}$, $\overline{EN} = 0 \text{ V}$, $C_O = 4.7 \mu\text{F}$ ($CSR^\dagger = 1 \Omega$), FB shorted to OUT at device leads (unless otherwise noted)

PARAMETER	TEST CONDITIONS‡	T_J	TPS7201Q			UNIT
			MIN	TYP	MAX	
Reference voltage (measured at FB with OUT connected to FB)	$V_I = 3.5 \text{ V}$, $I_O = 10 \text{ mA}$	25°C	1.188			V
	$2.5 \text{ V} \leq V_I \leq 10 \text{ V}$, $5 \text{ mA} \leq I_O \leq 250 \text{ mA}$, See Note 2	-40°C to 125°C	1.152	1.224		V
Reference voltage temperature coefficient			-40°C to 125°C	31	75	ppm/°C
Pass-element series resistance (see Note 3)	$V_I = 2.4 \text{ V}$, $50 \mu\text{A} \leq I_O \leq 100 \text{ mA}$	25°C	2.1	4.2		Ω
		-40°C to 125°C		4.8		
	$V_I = 2.4 \text{ V}$, $100 \text{ mA} \leq I_O \leq 200 \text{ mA}$	25°C	2.9	4.4		
		-40°C to 125°C		4.6		
	$V_I = 2.9 \text{ V}$, $50 \mu\text{A} \leq I_O \leq 250 \text{ mA}$	25°C	1.6	2.7		
		-40°C to 125°C		4.5		
Input regulation	$V_I = 2.5 \text{ V to } 10 \text{ V}$, $50 \mu\text{A} \leq I_O \leq 250 \text{ mA}$, See Note 2	25°C		23		mV
		-40°C to 125°C		36		
Output regulation	$I_O = 5 \text{ mA to } 250 \text{ mA}$, $2.5 \text{ V} \leq V_I \leq 10 \text{ V}$, See Note 2	25°C	15	25		mV
		-40°C to 125°C		36		
	$I_O = 50 \mu\text{A to } 250 \text{ mA}$, $2.5 \text{ V} \leq V_I \leq 10 \text{ V}$, See Note 2	25°C	17	27		
		-40°C to 125°C		43		
Ripple rejection	$f = 120 \text{ Hz}$	$I_O = 50 \mu\text{A}$	25°C	49	60	dB
			-40°C to 125°C	32		
		$I_O = 250 \text{ mA}$, See Note 2	25°C	45	50	
			-40°C to 125°C	30		
Output noise spectral density	$f = 120 \text{ Hz}$	25°C		2		µV/√Hz
Output noise voltage	$10 \text{ Hz} \leq f \leq 100 \text{ kHz}$, $CSR^\dagger = 1 \Omega$	$C_O = 4.7 \mu\text{F}$	25°C	235		µVRms
		$C_O = 10 \mu\text{F}$	25°C	190		
		$C_O = 100 \mu\text{F}$	25°C	125		
PG trip-threshold voltage§	V_{FB} voltage decreasing from above V_{PG}		-40°C to 125°C	$0.95 \times V_{FB(\text{nom})}$		V
PG hysteresis voltage§	Measured at V_{FB}		25°C	12		mV
PG output low voltage§	$I_{PG} = 400 \mu\text{A}$, $V_I = 2.13 \text{ V}$	25°C		0.1	0.4	V
		-40°C to 125°C			0.4	
FB input current		25°C	-10	0.1	10	nA
		-40°C to 125°C	-20		20	

† CSR refers to the total series resistance, including the ESR of the capacitor, any series resistance added externally, and PWB trace resistance to C_O .

‡ Pulse-testing techniques are used to maintain virtual junction temperature as close as possible to ambient temperature; thermal effects must be taken into account separately.

§ Output voltage programmed to 2.5 V with closed-loop configuration (see application information).

NOTES: 2. When $V_I < 2.9 \text{ V}$ and $I_O > 100 \text{ mA}$ simultaneously, pass element $r_{DS(on)}$ increases (see Figure 10) to a point such that the resulting dropout voltage prevents the regulator from maintaining the specified tolerance range.

3. To calculate dropout voltage, use equation:

$$V_{DO} = I_O \cdot r_{DS(on)}$$

$r_{DS(on)}$ is a function of both output current and input voltage. The parametric table lists $r_{DS(on)}$ for $V_I = 2.4 \text{ V}$, 2.9 V , 3.9 V , and 5.9 V , which corresponds to dropout conditions for programmed output voltages of 2.5 V, 3 V, 4 V, and 6 V, respectively. For other programmed values, refer to Figures 10 and 11.



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**TPS7201Q, TPS7233Q, TPS7248Q, TPS7250Q
 TPS7201Y, TPS7233Y, TPS7248Y, TPS7250Y
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SLVS102E – MARCH 1995 – REVISED JANUARY 1997

TPS7233Q electrical characteristics, $I_O = 10 \text{ mA}$, $V_I = 4.3 \text{ V}$, $\bar{EN} = 0 \text{ V}$, $C_O = 4.7 \mu\text{F}$ ($CSR^\dagger = 1 \Omega$), SENSE shorted to OUT (unless otherwise noted)

PARAMETER	TEST CONDITIONS‡	T_J	TPS7233Q			UNIT
			MIN	TYP	MAX	
Output voltage	$V_I = 4.3 \text{ V}$, $I_O = 10 \text{ mA}$	25°C	3.3			V
	$4.3 \text{ V} \leq V_I \leq 10 \text{ V}$, $5 \text{ mA} \leq I_O \leq 250 \text{ mA}$	-40°C to 125°C	3.23		3.37	
Dropout voltage	$I_O = 10 \text{ mA}$, $V_I = 3.23 \text{ V}$	25°C	14	20		mV
		-40°C to 125°C		30		
	$I_O = 100 \text{ mA}$, $V_I = 3.23 \text{ V}$	25°C	140	180		
		-40°C to 125°C		232		
Pass-element series resistance	$(3.23 \text{ V} - V_O)/I_O$, $I_O = 250 \text{ mA}$	25°C	360	460		Ω
		-40°C to 125°C		610		
	$V_I = 3.23 \text{ V}$, $I_O = 250 \text{ mA}$	25°C	1.5	1.84		
		-40°C to 125°C		2.5		
Input regulation	$V_I = 4.3 \text{ V}$ to 10 V , $50 \mu\text{A} \leq I_O \leq 250 \text{ mA}$	25°C	8	25		mV
		-40°C to 125°C		33		
Output regulation	$I_O = 5 \text{ mA}$ to 250 mA , $4.3 \text{ V} \leq V_I \leq 10 \text{ V}$	25°C	32	42		mV
		-40°C to 125°C		71		
	$I_O = 50 \mu\text{A}$ to 250 mA , $4.3 \text{ V} \leq V_I \leq 10 \text{ V}$	25°C	41	55		
		-40°C to 125°C		98		
Ripple rejection	$f = 120 \text{ Hz}$	$I_O = 50 \mu\text{A}$	25°C	40	52	dB
			-40°C to 125°C	38		
		$I_O = 250 \text{ mA}$	25°C	35	44	
			-40°C to 125°C	33		
Output noise spectral density	$f = 120 \text{ Hz}$		25°C	2		$\mu\text{V}/\sqrt{\text{Hz}}$
Output noise voltage	$10 \text{ Hz} \leq f \leq 100 \text{ kHz}$, $CSR^\dagger = 1 \Omega$	$C_O = 4.7 \mu\text{F}$	25°C	265		μVRms
		$C_O = 10 \mu\text{F}$	25°C	212		
		$C_O = 100 \mu\text{F}$	25°C	135		
PG trip-threshold voltage	V_O voltage decreasing from above V_{PG}		-40°C to 125°C	$0.95 \times V_{O(\text{nom})}$		V
PG hysteresis voltage			25°C	32		mV
PG output low voltage	$I_{PG} = 1 \text{ mA}$, $V_I = 2.8 \text{ V}$		25°C	0.22	0.4	V
			-40°C to 125°C		0.4	

† CSR refers to the total series resistance, including the ESR of the capacitor, any series resistance added externally, and PWB trace resistance to C_O .

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SLVS102E – MARCH 1995 – REVISED JANUARY 1997

TPS7248Q electrical characteristics, $I_O = 10 \text{ mA}$, $V_I = 5.85 \text{ V}$, $\bar{EN} = 0 \text{ V}$, $C_O = 4.7 \mu\text{F}$ ($CSR^\dagger = 1 \Omega$), SENSE shorted to OUT (unless otherwise noted)

PARAMETER	TEST CONDITIONS‡	T_J	TPS7248Q			UNIT
			MIN	TYP	MAX	
Output voltage	$V_I = 5.85 \text{ V}$, $I_O = 10 \text{ mA}$	25°C	4.85			V
	5.85 V $\leq V_I \leq 10 \text{ V}$, 5 mA $\leq I_O \leq 250 \text{ mA}$	-40°C to 125°C	4.75	4.95		
Dropout voltage	$I_O = 10 \text{ mA}$, $V_I = 4.75 \text{ V}$	25°C	10	19		mV
		-40°C to 125°C		30		
	$I_O = 100 \text{ mA}$, $V_I = 4.75 \text{ V}$	25°C	90	100		
		-40°C to 125°C		150		
Pass-element series resistance	$(4.75 \text{ V} - V_O)/I_O$, $V_I = 4.75 \text{ V}$, $I_O = 250 \text{ mA}$	25°C	216	250		Ω
		-40°C to 125°C		285		
	$V_I = 5.85 \text{ V}$ to 10 V, 50 $\mu\text{A} \leq I_O \leq 250 \text{ mA}$	25°C	0.8	1		
		-40°C to 125°C		1.4		
Input regulation	$V_I = 5.85 \text{ V}$ to 10 V, 50 $\mu\text{A} \leq I_O \leq 250 \text{ mA}$	25°C	34			mV
		-40°C to 125°C		50		
Output regulation	$I_O = 5 \text{ mA}$ to 250 mA, 5.85 V $\leq V_I \leq 10 \text{ V}$	25°C	43	55		mV
		-40°C to 125°C		95		
	$I_O = 50 \mu\text{A}$ to 250 mA, 5.85 V $\leq V_I \leq 10 \text{ V}$	25°C	55	75		
		-40°C to 125°C		135		
Ripple rejection	$f = 120 \text{ Hz}$	$I_O = 50 \mu\text{A}$	25°C	42	53	dB
			-40°C to 125°C	36		
		$I_O = 250 \text{ mA}$	25°C	36	46	
			-40°C to 125°C	34		
Output noise spectral density	$f = 120 \text{ Hz}$		25°C	2		$\mu\text{V}/\sqrt{\text{Hz}}$
Output noise voltage	10 Hz $\leq f \leq 100 \text{ kHz}$, $CSR^\dagger = 1 \Omega$	$C_O = 4.7 \mu\text{F}$	25°C	370		μVRms
		$C_O = 10 \mu\text{F}$	25°C	290		
		$C_O = 100 \mu\text{F}$	25°C	168		
PG trip-threshold voltage	V_O voltage decreasing from above V_{PG}		-40°C to 125°C	$0.95 \times V_{O(\text{nom})}$		V
PG hysteresis voltage			25°C	50		mV
PG output low voltage	$ I_{PG} = 1.2 \text{ mA}$, $V_I = 4.12 \text{ V}$	25°C	0.2	0.4		V
		-40°C to 125°C		0.4		

† CSR refers to the total series resistance, including the ESR of the capacitor, any series resistance added externally, and PWB trace resistance to C_O .

‡ Pulse-testing techniques are used to maintain virtual junction temperature as close as possible to ambient temperature; thermal effects must be taken into account separately.



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**TPS7201Q, TPS7233Q, TPS7248Q, TPS7250Q
 TPS7201Y, TPS7233Y, TPS7248Y, TPS7250Y
 MICROPOWER LOW-DROPOUT (LDO) VOLTAGE REGULATORS**

SLVS102E – MARCH 1995 – REVISED JANUARY 1997

TPS7250Q electrical characteristics, $I_O = 10 \text{ mA}$, $V_I = 6 \text{ V}$, $\overline{EN} = 0 \text{ V}$, $C_O = 4.7 \mu\text{F}$ ($CSR^{\dagger} = 1 \Omega$), SENSE shorted to OUT (unless otherwise noted)

PARAMETER	TEST CONDITIONS‡	T_J	TPS7250Q			UNIT
			MIN	TYP	MAX	
Output voltage	$V_I = 6 \text{ V}$, $6 \text{ V} \leq V_I \leq 10 \text{ V}$, $I_O = 10 \text{ mA}$, $5 \text{ mA} \leq I_O \leq 250 \text{ mA}$	25°C –40°C to 125°C	5 4.9		5.1	V
		25°C –40°C to 125°C	8 30			
Dropout voltage	$I_O = 10 \text{ mA}$, $V_I = 4.88 \text{ V}$	25°C –40°C to 125°C	76 136		85	mV
	$I_O = 100 \text{ mA}$, $V_I = 4.88 \text{ V}$	25°C –40°C to 125°C	190 312		206	
	$I_O = 250 \text{ mA}$, $V_I = 4.88 \text{ V}$	25°C –40°C to 125°C				
	$(4.88 \text{ V} - V_O)/I_O$, $I_O = 250 \text{ mA}$	25°C –40°C to 125°C	0.76 1.25		0.825	Ω
	$V_I = 6 \text{ V}$ to 10 V , $50 \mu\text{A} \leq I_O \leq 250 \text{ mA}$	25°C –40°C to 125°C			28 35	mV
	$I_O = 5 \text{ mA}$ to 250 mA , $6 \text{ V} \leq V_I \leq 10 \text{ V}$	25°C –40°C to 125°C	46 100		61	mV
Output regulation	$I_O = 50 \mu\text{A}$ to 250 mA , $6 \text{ V} \leq V_I \leq 10 \text{ V}$	25°C –40°C to 125°C	59 150		79	
		25°C –40°C to 125°C				
	$f = 120 \text{ Hz}$	$I_O = 50 \mu\text{A}$	25°C –40°C to 125°C	41 37	52	dB
		$I_O = 250 \text{ mA}$	25°C –40°C to 125°C	36 32	46	
Output noise spectral density	$f = 120 \text{ Hz}$		25°C		2	μV/√Hz
Output noise voltage	$10 \text{ Hz} \leq f \leq 100 \text{ kHz}$, $CSR^{\dagger} = 1 \Omega$	$C_O = 4.7 \mu\text{F}$	25°C		390	μVrms
		$C_O = 10 \mu\text{F}$	25°C		300	
		$C_O = 100 \mu\text{F}$	25°C		175	
PG trip-threshold voltage	V_O voltage decreasing from above V_{PG}		–40°C to 125°C	$0.95 \times V_{O(\text{nom})}$		V
PG hysteresis voltage			25°C		50	mV
PG output low voltage	$I_{PG} = 1.2 \text{ mA}$, $V_I = 4.25 \text{ V}$		25°C		0.19	V
			–40°C to 125°C		0.4	

† CSR refers to the total series resistance, including the ESR of the capacitor, any series resistance added externally, and PWB trace resistance to C_O .

‡ Pulse-testing techniques are used to maintain virtual junction temperature as close as possible to ambient temperature; thermal effects must be taken into account separately.



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 POST OFFICE BOX 1443 • HOUSTON, TEXAS 77251-1443

**TPS7201Q, TPS7233Q, TPS7248Q, TPS7250Q
TPS7201Y, TPS7233Y, TPS7248Y, TPS7250Y
MICROPOWER LOW-DROPOUT (LDO) VOLTAGE REGULATORS**
SLVS102E – MARCH 1995 – REVISED JANUARY 1997

electrical characteristics, $I_O = 10 \text{ mA}$, $\overline{EN} = 0 \text{ V}$, $C_O = 4.7 \mu\text{F}$ ($CSR^\dagger = 1 \Omega$), $T_J = 25^\circ\text{C}$, SENSE/FB shorted to OUT (unless otherwise noted)

PARAMETER	TEST CONDITIONS‡	TPS7201Y, TPS7233Y TPS7248Y, TPS7250Y			UNIT
		MIN	TYP	MAX	
Ground current (active mode)	$\overline{EN} \leq 0.5 \text{ V}$, $0 \text{ mA} \leq I_O \leq 250 \text{ mA}$		$V_I = V_O + 1 \text{ V}$,	180	μA
Output current limit threshold	$V_O = 0 \text{ V}$,		$V_I = 10 \text{ V}$	0.6	A
Thermal shutdown junction temperature				165	$^\circ\text{C}$
\overline{EN} hysteresis voltage				50	mV
Minimum V_I for active pass element				1.9	V
Minimum V_I for valid PG	$I_{PG} = 300 \mu\text{A}$			0.95	V

electrical characteristics, $I_O = 10 \text{ mA}$, $\overline{EN} = 0 \text{ V}$, $C_O = 4.7 \mu\text{F}$ ($CSR^\dagger = 1 \Omega$), $T_J = 25^\circ\text{C}$, FB shorted to OUT at device leads (unless otherwise noted)

PARAMETER	TEST CONDITIONS‡	TPS7201Y			UNIT
		MIN	TYP	MAX	
Reference voltage (measured at FB with OUT connected to FB)	$V_I = 3.5 \text{ V}$, $I_O = 10 \text{ mA}$			1.188	V
Pass-element series resistance (see Note 3)	$V_I = 2.4 \text{ V}$, $50 \mu\text{A} \leq I_O \leq 100 \text{ mA}$			2.1	Ω
	$V_I = 2.4 \text{ V}$, $100 \text{ mA} \leq I_O \leq 200 \text{ mA}$			2.9	
	$V_I = 2.9 \text{ V}$, $50 \mu\text{A} \leq I_O \leq 250 \text{ mA}$			1.6	
	$V_I = 3.9 \text{ V}$, $50 \mu\text{A} \leq I_O \leq 250 \text{ mA}$			1	
	$V_I = 5.9 \text{ V}$, $50 \mu\text{A} \leq I_O \leq 250 \text{ mA}$			0.8	
Output regulation	$2.5 \text{ V} \leq V_I \leq 10 \text{ V}$, See Note 2	$I_O = 5 \text{ mA to } 250 \text{ mA}$		15	mV
	$2.5 \text{ V} \leq V_I \leq 10 \text{ V}$, See Note 2	$I_O = 50 \mu\text{A to } 250 \text{ mA}$		17	
Ripple rejection	$V_I = 3.5 \text{ V}$, $f = 120 \text{ Hz}$	$I_O = 50 \mu\text{A}$		60	dB
		$I_O = 250 \text{ mA}$, See Note 2		50	
Output noise spectral density	$V_I = 3.5 \text{ V}$, $f = 120 \text{ Hz}$			2	$\mu\text{V}/\sqrt{\text{Hz}}$
Output noise voltage	$V_I = 3.5 \text{ V}$, $10 \text{ Hz} \leq f \leq 100 \text{ kHz}$, $CSR^\dagger = 1 \Omega$	$C_O = 4.7 \mu\text{F}$		235	μVRms
		$C_O = 10 \mu\text{F}$		190	
		$C_O = 100 \mu\text{F}$		125	
PG hysteresis voltage§	$V_I = 3.5 \text{ V}$,	Measured at V_{FB}		12	mV
PG output low voltage§	$V_I = 2.13 \text{ V}$,	$I_{PG} = 400 \mu\text{A}$		0.1	V
FB input current	$V_I = 3.5 \text{ V}$			0.1	nA

† CSR refers to the total series resistance, including the ESR of the capacitor, any series resistance added externally, and PWB trace resistance to C_O .

‡ Pulse-testing techniques are used to maintain virtual junction temperature as close as possible to ambient temperature; thermal effects must be taken into account separately.

§ Output voltage programmed to 2.5 V with closed-loop configuration (see application information).

NOTES: 2 When $V_I < 2.9 \text{ V}$ and $I_O > 100 \text{ mA}$ simultaneously, pass element $r_{DS(on)}$ increases (see Figure 10) to a point such that the resulting dropout voltage prevents the regulator from maintaining the specified tolerance range.

3 To calculate dropout voltage, use equation:

$$V_{DO} = I_O \cdot r_{DS(on)}$$

$r_{DS(on)}$ is a function of both output current and input voltage. The parametric table lists $r_{DS(on)}$ for $V_I = 2.4 \text{ V}, 2.9 \text{ V}, 3.9 \text{ V}$, and 5.9 V , which corresponds to dropout conditions for programmed output voltages of 2.5 V, 3 V, 4 V, and 6 V, respectively. For other programmed values, refer to Figures 10 and 11.



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POST OFFICE BOX 1443 • HOUSTON, TEXAS 77251-1443

**TPS7201Q, TPS7233Q, TPS7248Q, TPS7250Q
 TPS7201Y, TPS7233Y, TPS7248Y, TPS7250Y
 MICROPOWER LOW-DROPOUT (LDO) VOLTAGE REGULATORS**

SLVS102E – MARCH 1995 – REVISED JANUARY 1997

electrical characteristics, $I_O = 10 \text{ mA}$, $\overline{EN} = 0 \text{ V}$, $C_O = 4.7 \mu\text{F}$ ($CSR^\dagger = 1 \Omega$), $T_J = 25^\circ\text{C}$, SENSE shorted to OUT (unless otherwise noted)

PARAMETER	TEST CONDITIONS [‡]	TPS7233Y			UNIT
		MIN	TYP	MAX	
Output voltage	$V_I = 4.3 \text{ V}$, $I_O = 10 \text{ mA}$		3.3		V
Dropout voltage	$V_I = 3.23 \text{ V}$, $I_O = 10 \text{ mA}$		14		mV
	$V_I = 3.23 \text{ V}$, $I_O = 100 \text{ mA}$		140		
	$V_I = 3.23 \text{ V}$, $I_O = 250 \text{ mA}$		360		
Pass-element series resistance	$(3.23 \text{ V} - V_O)/I_O$, $V_I = 3.23 \text{ V}$, $I_O = 250 \text{ mA}$		1.5		Ω
Input regulation	$V_I = 4.3 \text{ V}$ to 10 V , $50 \mu\text{A} \leq I_O \leq 250 \text{ mA}$		8		mV
Output regulation	$4.3 \text{ V} \leq V_I \leq 10 \text{ V}$, $I_O = 5 \text{ mA}$ to 250 mA		32		mV
	$4.3 \text{ V} \leq V_I \leq 10 \text{ V}$, $I_O = 50 \mu\text{A}$ to 250 mA		41		
Ripple rejection	$V_I = 4.3 \text{ V}$, $I_O = 50 \mu\text{A}$		52		dB
	$f = 120 \text{ Hz}$, $I_O = 250 \text{ mA}$		44		
Output noise spectral density	$V_I = 4.3 \text{ V}$, $f = 120 \text{ Hz}$		2		$\mu\text{V}/\sqrt{\text{Hz}}$
Output noise voltage	$V_I = 4.3 \text{ V}$, $C_O = 4.7 \mu\text{F}$		265		μVRms
	$10 \text{ Hz} \leq f \leq 100 \text{ kHz}$, $CSR^\dagger = 1 \Omega$, $C_O = 10 \mu\text{F}$		212		
	$CSR^\dagger = 1 \Omega$, $C_O = 100 \mu\text{F}$		135		
PG hysteresis voltage	$V_I = 4.3 \text{ V}$		32		mV
PG output low voltage	$V_I = 2.8 \text{ V}$, $I_{PG} = 1 \text{ mA}$		0.22		V

[†] CSR refers to the total series resistance, including the ESR of the capacitor, any series resistance added externally, and PWB trace resistance to C_O .

[‡] Pulse-testing techniques are used to maintain virtual junction temperature as close as possible to ambient temperature; thermal effects must be taken into account separately.

PARAMETER	TEST CONDITIONS [‡]	TPS7248Y			UNIT
		MIN	TYP	MAX	
Output voltage	$V_I = 5.85 \text{ V}$, $I_O = 10 \text{ mA}$		4.85		V
Dropout voltage	$V_I = 4.75 \text{ V}$, $I_O = 10 \text{ mA}$		10		mV
	$V_I = 4.75 \text{ V}$, $I_O = 100 \text{ mA}$		90		
	$V_I = 4.75 \text{ V}$, $I_O = 250 \text{ mA}$		216		
Pass-element series resistance	$(4.75 \text{ V} - V_O)/I_O$, $V_I = 4.75 \text{ V}$, $I_O = 250 \text{ mA}$		0.8		Ω
Output regulation	$5.85 \text{ V} \leq V_I \leq 10 \text{ V}$, $I_O = 5 \text{ mA}$ to 250 mA		43		mV
	$5.85 \text{ V} \leq V_I \leq 10 \text{ V}$, $I_O = 50 \mu\text{A}$ to 250 mA		55		
Ripple rejection	$V_I = 5.85 \text{ V}$, $I_O = 50 \mu\text{A}$		53		dB
	$f = 120 \text{ Hz}$, $I_O = 250 \text{ mA}$		46		
Output noise spectral density	$V_I = 5.85 \text{ V}$, $f = 120 \text{ Hz}$		2		$\mu\text{V}/\sqrt{\text{Hz}}$
Output noise voltage	$V_I = 5.85 \text{ V}$, $C_O = 4.7 \mu\text{F}$		370		μVRms
	$10 \text{ Hz} \leq f \leq 100 \text{ kHz}$, $CSR^\dagger = 1 \Omega$, $C_O = 10 \mu\text{F}$		290		
	$CSR^\dagger = 1 \Omega$, $C_O = 100 \mu\text{F}$		168		
PG hysteresis voltage	$V_I = 5.85 \text{ V}$		50		mV
PG output low voltage	$V_I = 4.12 \text{ V}$, $I_{PG} = 1.2 \text{ mA}$		0.2		V

[†] CSR refers to the total series resistance, including the ESR of the capacitor, any series resistance added externally, and PWB trace resistance to C_O .

[‡] Pulse-testing techniques are used to maintain virtual junction temperature as close as possible to ambient temperature; thermal effects must be taken into account separately.



POST OFFICE BOX 655303 • DALLAS, TEXAS 75265
 POST OFFICE BOX 1443 • HOUSTON, TEXAS 77251-1443

**TPS7201Q, TPS7233Q, TPS7248Q, TPS7250Q
TPS7201Y, TPS7233Y, TPS7248Y, TPS7250Y
MICROPOWER LOW-DROPOUT (LDO) VOLTAGE REGULATORS**
SLVS102E – MARCH 1995 – REVISED JANUARY 1997

electrical characteristics, $I_O = 10 \text{ mA}$, $\overline{EN} = 0 \text{ V}$, $C_O = 4.7 \mu\text{F}$ ($CSR^\dagger = 1 \Omega$), $T_J = 25^\circ\text{C}$, SENSE shorted to OUT (unless otherwise noted) (continued)

PARAMETER	TEST CONDITIONS‡	TPS7250Y			UNIT
		MIN	TYP	MAX	
Output voltage	$V_I = 6 \text{ V}$, $I_O = 10 \text{ mA}$	5			V
Dropout voltage	$V_I = 4.88 \text{ V}$ $I_O = 10 \text{ mA}$	8			mV
	$V_I = 4.88 \text{ V}$ $I_O = 100 \text{ mA}$	76			
	$V_I = 4.88 \text{ V}$, $I_O = 250 \text{ mA}$	190			
Pass-element series resistance	$(4.88 \text{ V} - V_O)/I_O$, $I_O = 250 \text{ mA}$	0.76			Ω
Input regulation	$V_I = 6 \text{ V}$ to 10 V , $50 \mu\text{A} \leq I_O \leq 250 \text{ mA}$				mV
Output regulation	$6 \text{ V} \leq V_I \leq 10 \text{ V}$, $I_O = 5 \text{ mA}$ to 250 mA	46			mV
	$6 \text{ V} \leq V_I \leq 10 \text{ V}$, $I_O = 50 \mu\text{A}$ to 250 mA	59			
Ripple rejection	$V_I = 6 \text{ V}$, $f = 120 \text{ Hz}$	52			dB
		46			
Output noise spectral density	$V_I = 6 \text{ V}$, $f = 120 \text{ Hz}$	2			$\mu\text{V}/\sqrt{\text{Hz}}$
Output noise voltage	$V_I = 6 \text{ V}$, $10 \text{ Hz} \leq f \leq 100 \text{ kHz}$, $CSR^\dagger = 1 \Omega$	$C_O = 4.7 \mu\text{F}$	390		μVRms
		$C_O = 10 \mu\text{F}$	300		
		$C_O = 100 \mu\text{F}$	175		
PG hysteresis voltage	$V_I = 6 \text{ V}$	50			mV
PG output low voltage	$V_I = 4.25 \text{ V}$, $I_{PG} = 1.2 \text{ mA}$	0.19			V

† CSR refers to the total series resistance, including the ESR of the capacitor, any series resistance added externally, and PWB trace resistance to C_O .

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POST OFFICE BOX 655303 • DALLAS, TEXAS 75265
POST OFFICE BOX 1443 • HOUSTON, TEXAS 77251-1443

**TPS7201Q, TPS7233Q, TPS7248Q, TPS7250Q
 TPS7201Y, TPS7233Y, TPS7248Y, TPS7250Y
 MICROPOWER LOW-DROPOUT (LDO) VOLTAGE REGULATORS**

SLVS102E – MARCH 1995 – REVISED JANUARY 1997

TYPICAL CHARACTERISTICS

Table of Graphs

			FIGURE
I_Q	Quiescent current	vs Output current	5
		vs Input voltage	6
ΔI_Q^\dagger	Change in quiescent current	vs Free-air temperature	7
V_{DO}	Dropout voltage	vs Output current	8
ΔV_{DO}	Change in dropout voltage	vs Free-air temperature	9
V_{DO}	Dropout voltage (TPS7201 only)	vs Output current	10
$r_{DS(on)}$	Pass-element series resistance	vs Input voltage	11
ΔV_O	Change in output voltage	vs Free-air temperature	12
V_O	Output voltage	vs Input voltage	13
	Line regulation		14
	Load regulation (TPS7233)		15
	Load regulation (TPS7248)		16
	Load regulation (TPS7250)		17
$V_{O(PG)}$	Power-good (PG) voltage	vs Output voltage	18
$r_{DS(on)PG}$	Power-good (PG) on-resistance	vs Input voltage	19
V_I	Minimum input voltage for valid PG	vs Free-air temperature	20
	Output voltage response from enable (EN)		21
	Load transient response (TPS7201/TPS7233)		22
	Load transient response (TPS7248/TPS7250)		23
	Line transient response (TPS7201)		24
	Line transient response (TPS7233)		25
	Line transient response (TPS7248/TPS7250)		26
Ripple rejection		vs Frequency	27
Output Spectral Noise Density		vs Frequency	28
Compensation series resistance (CSR)		vs Output current ($C_O = 4.7 \mu F$)	29
		vs Added ceramic capacitance ($C_O = 4.7 \mu F$)	30
		vs Output current ($C_O = 10 \mu F$)	31
		vs Added ceramic capacitance ($C_O = 10 \mu F$)	32

† This symbol is not currently listed within EIA or JEDEC standards for semiconductor symbology.

TYPICAL CHARACTERISTICS

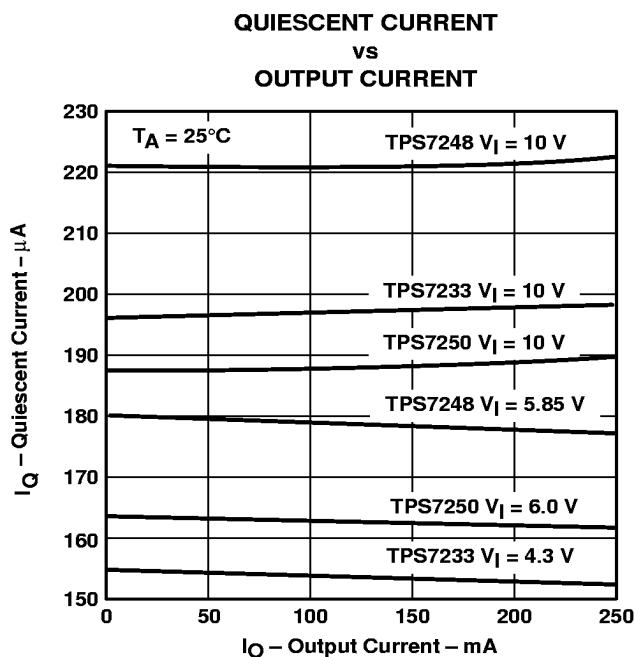


Figure 5

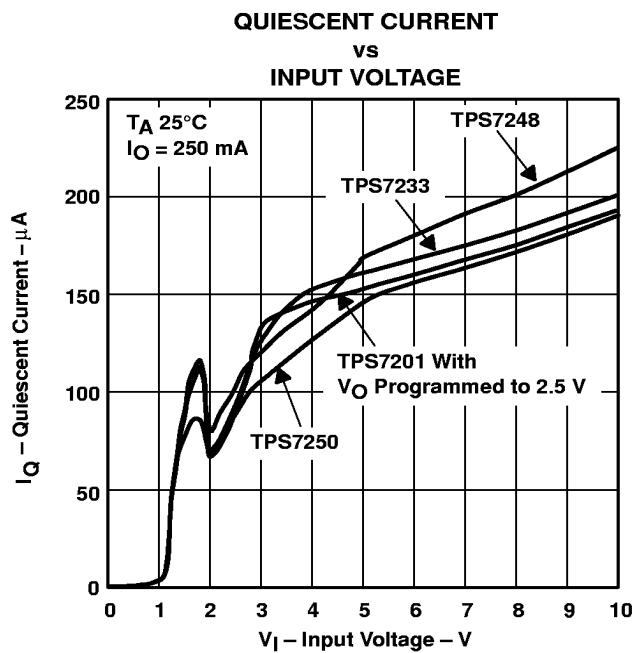


Figure 6

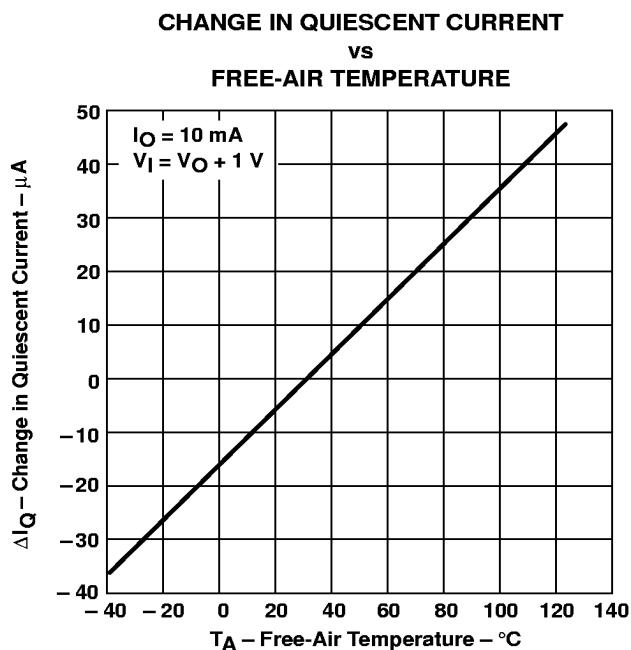


Figure 7

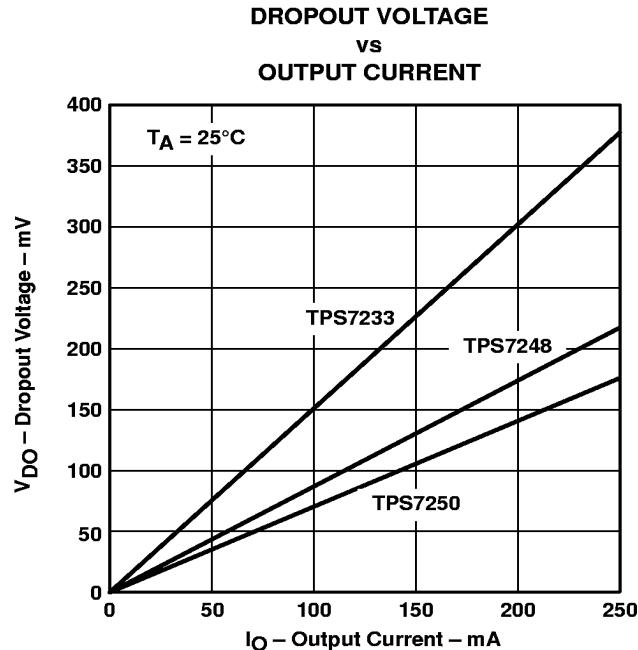


Figure 8



POST OFFICE BOX 655303 • DALLAS, TEXAS 75265
 POST OFFICE BOX 1443 • HOUSTON, TEXAS 77251-1443

**TPS7201Q, TPS7233Q, TPS7248Q, TPS7250Q
 TPS7201Y, TPS7233Y, TPS7248Y, TPS7250Y
 MICROPOWER LOW-DROPOUT (LDO) VOLTAGE REGULATORS**

SLVS102E – MARCH 1995 – REVISED JANUARY 1997

TYPICAL CHARACTERISTICS

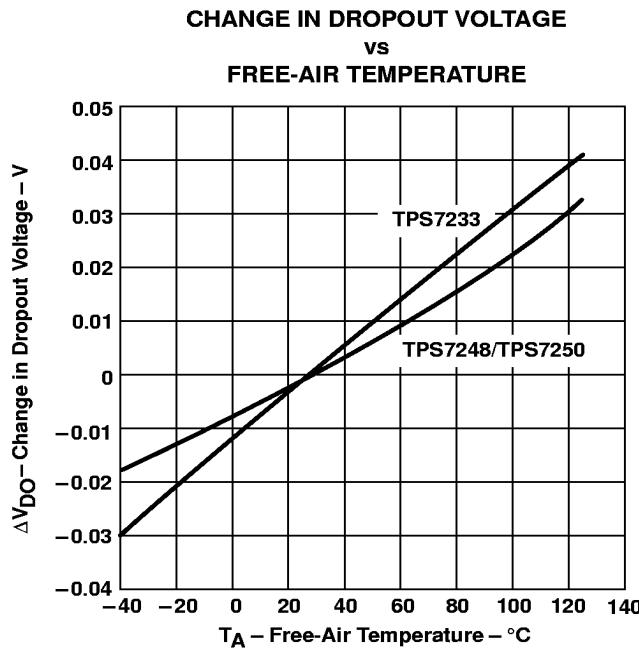


Figure 9

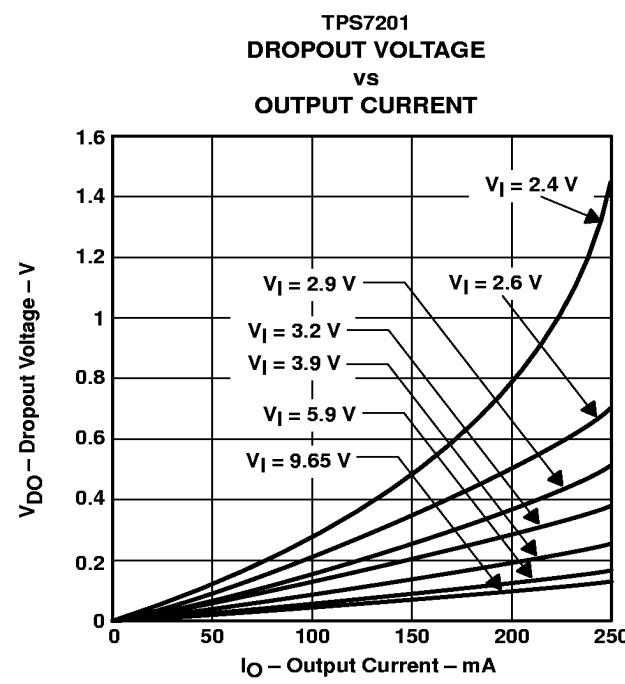


Figure 10

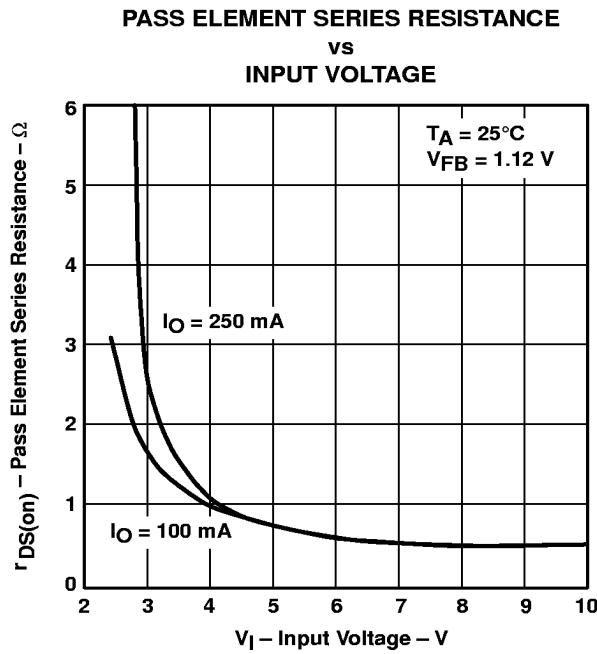


Figure 11

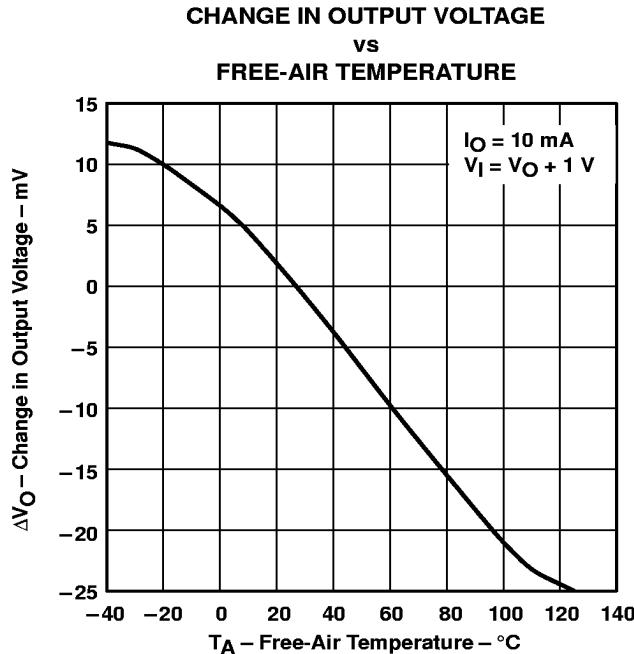


Figure 12

TYPICAL CHARACTERISTICS

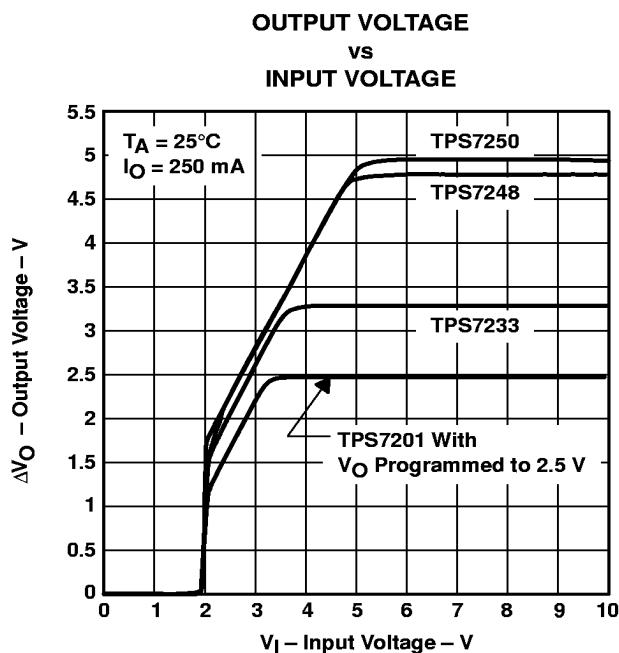


Figure 13

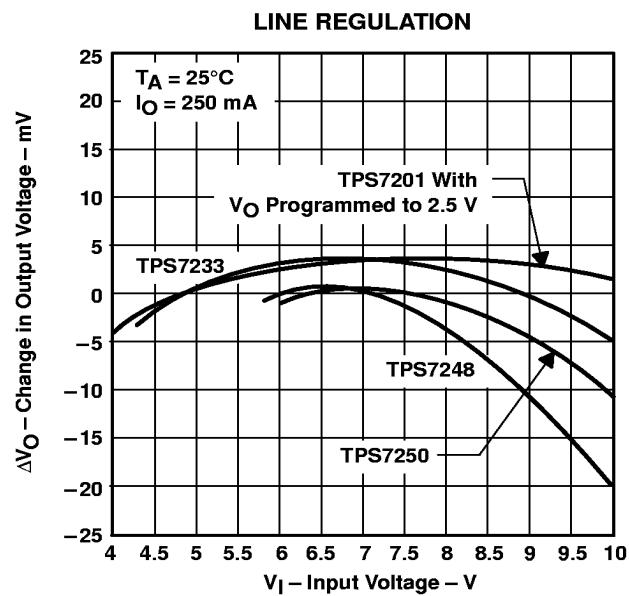


Figure 14

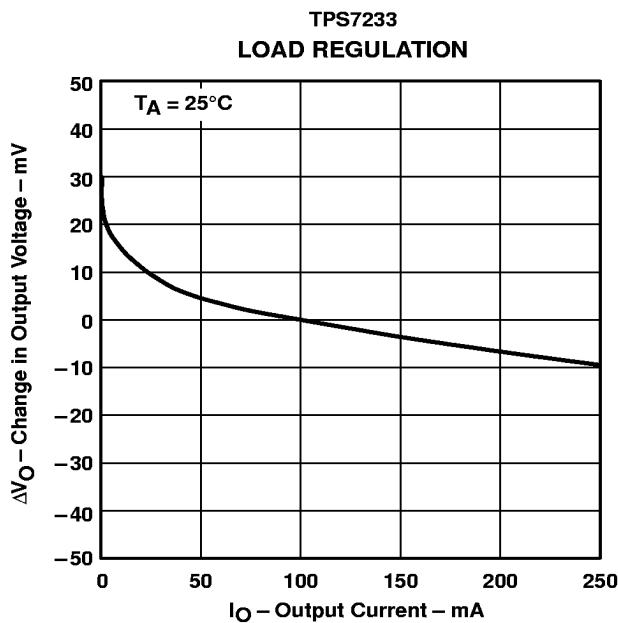


Figure 15

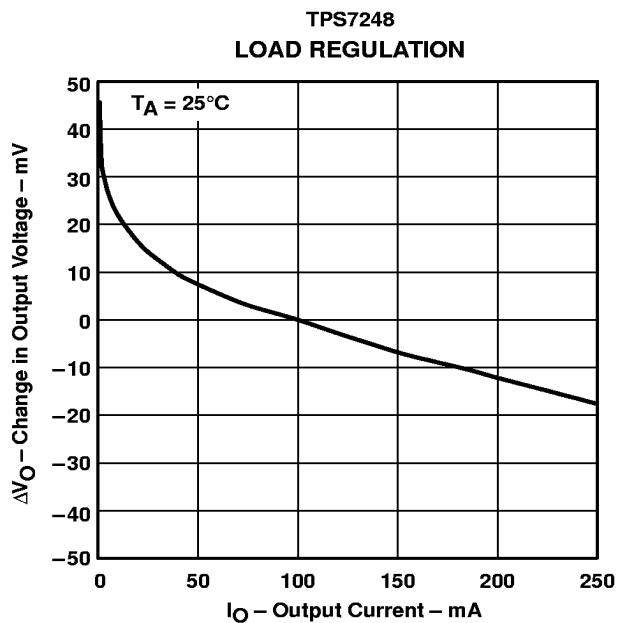


Figure 16

 **TEXAS
INSTRUMENTS**

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 POST OFFICE BOX 1443 • HOUSTON, TEXAS 77251-1443

**TPS7201Q, TPS7233Q, TPS7248Q, TPS7250Q
 TPS7201Y, TPS7233Y, TPS7248Y, TPS7250Y
 MICROPOWER LOW-DROPOUT (LDO) VOLTAGE REGULATORS**

SLVS102E – MARCH 1995 – REVISED JANUARY 1997

TYPICAL CHARACTERISTICS

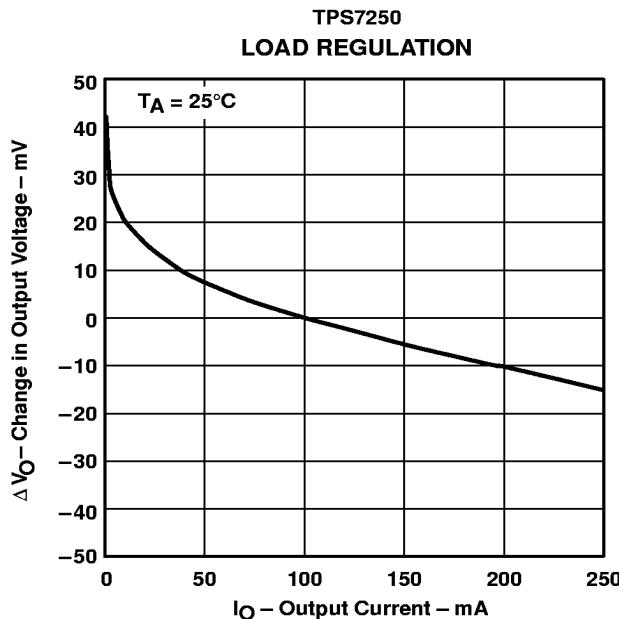
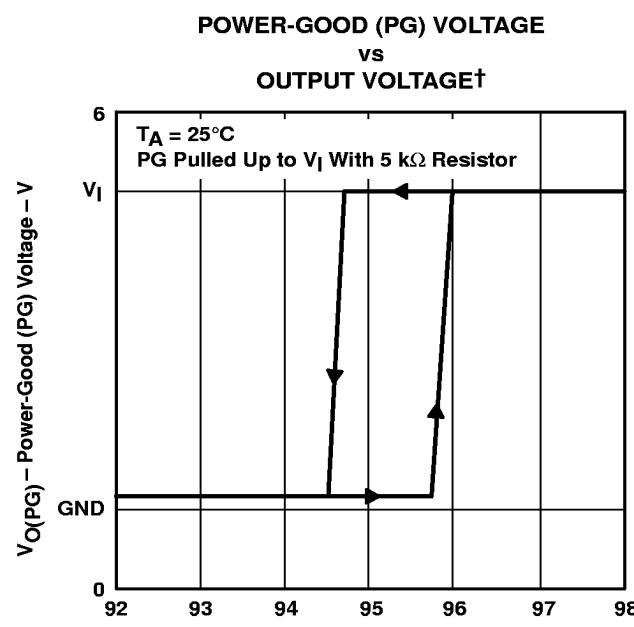


Figure 17



† V_O as a percent of $V_{O(nom)}$.

Figure 18

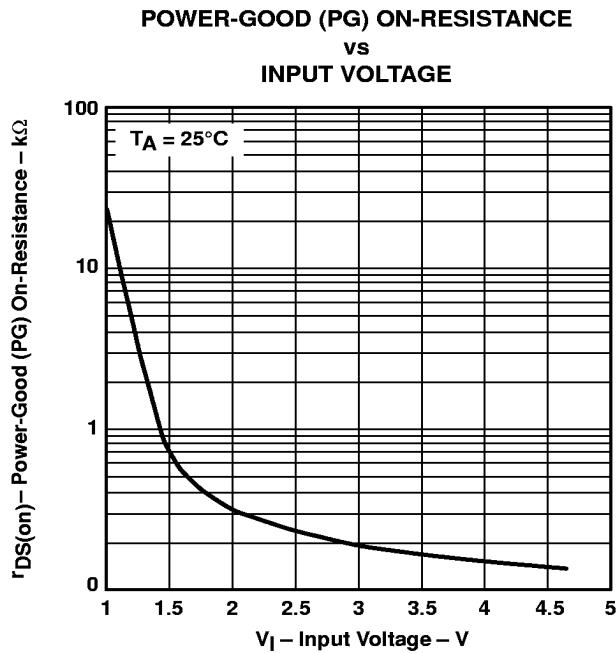


Figure 19

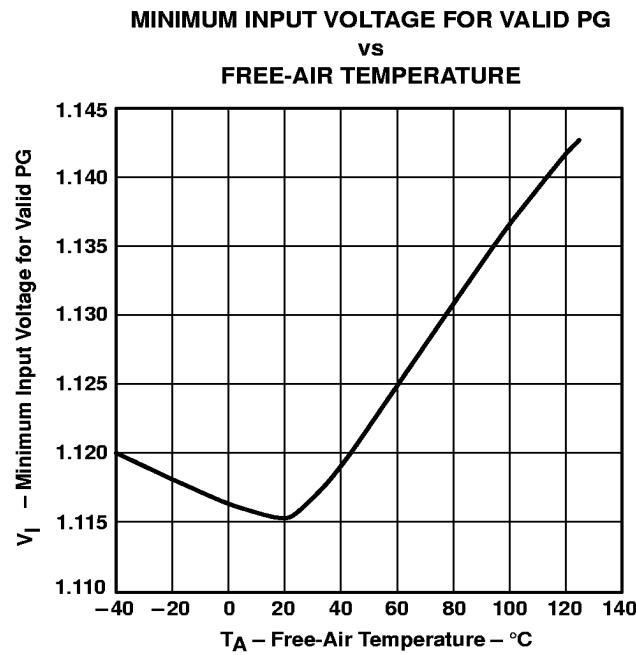


Figure 20



POST OFFICE BOX 655303 • DALLAS, TEXAS 75265
 POST OFFICE BOX 1443 • HOUSTON, TEXAS 77251-1443

**TPS7201Q, TPS7233Q, TPS7248Q, TPS7250Q
TPS7201Y, TPS7233Y, TPS7248Y, TPS7250Y
MICROPOWER LOW-DROPOUT (LDO) VOLTAGE REGULATORS**
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TYPICAL CHARACTERISTICS

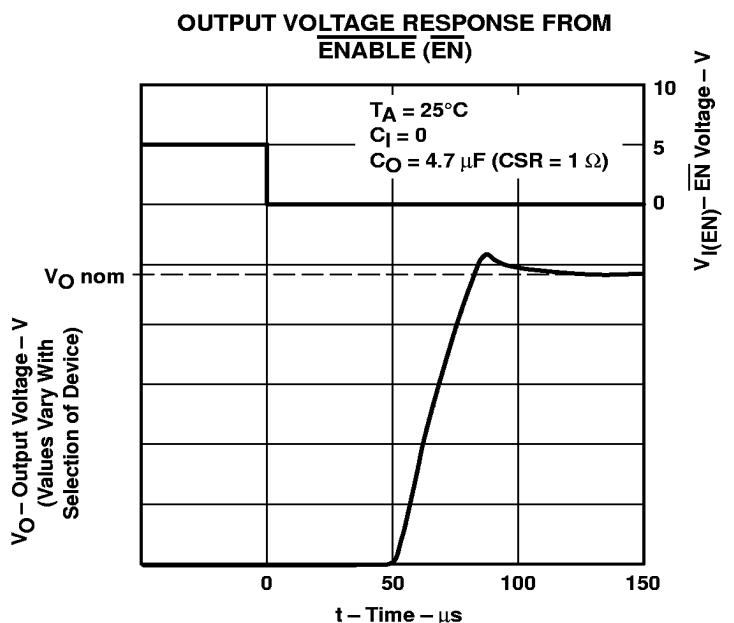


Figure 21

**TPS7201 (WITH V_O PROGRAMMED TO 2.5 V), TPS7233
LOAD TRANSIENT RESPONSE**

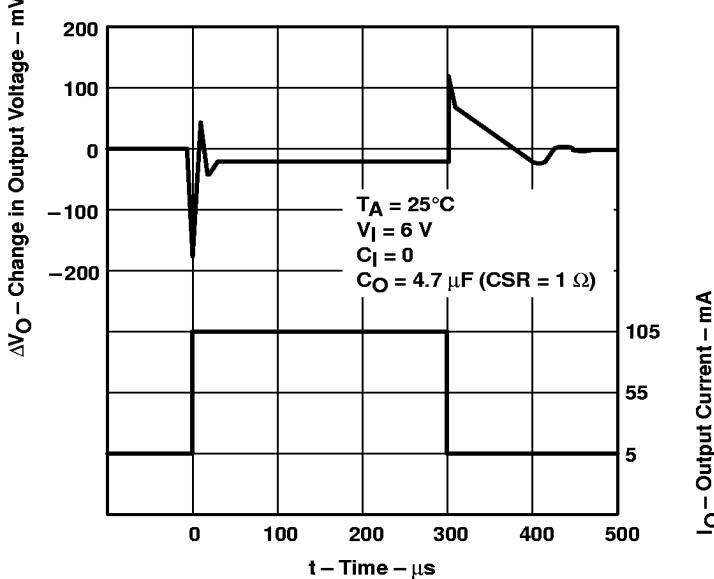


Figure 22



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**TPS7201Q, TPS7233Q, TPS7248Q, TPS7250Q
 TPS7201Y, TPS7233Y, TPS7248Y, TPS7250Y
 MICROPOWER LOW-DROPOUT (LDO) VOLTAGE REGULATORS**

SLVS102E – MARCH 1995 – REVISED JANUARY 1997

TYPICAL CHARACTERISTICS

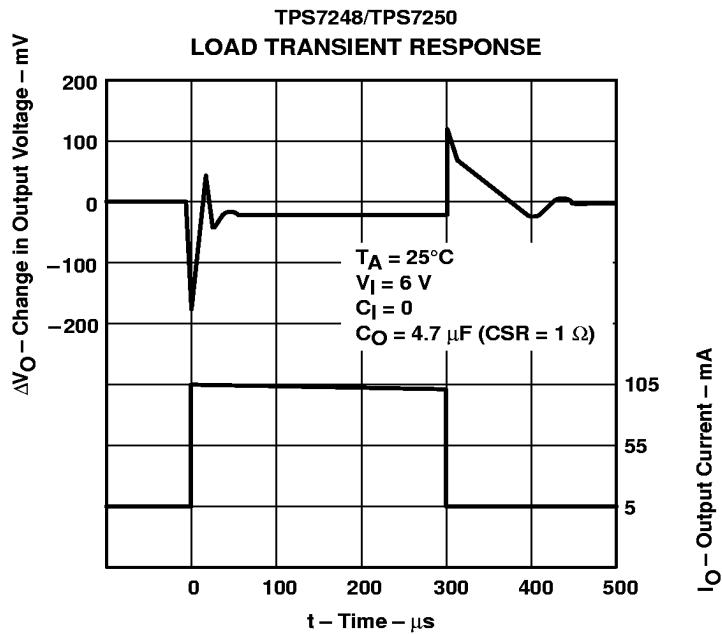


Figure 23

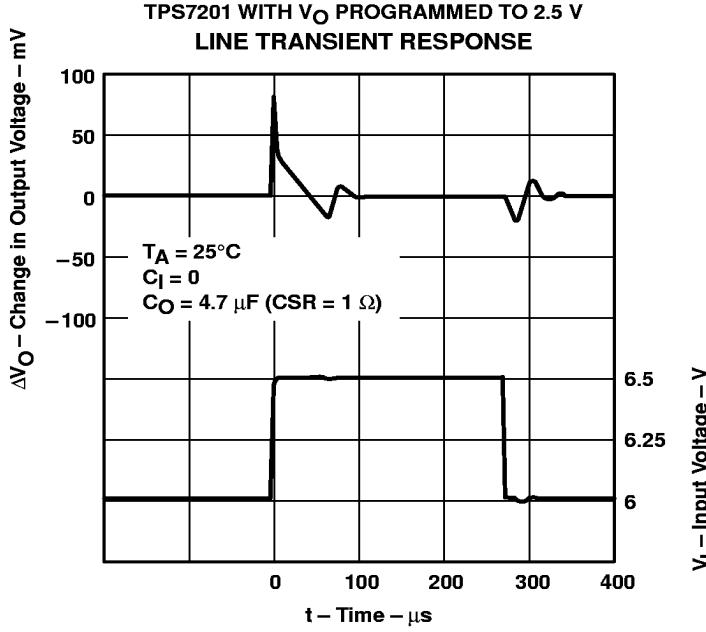


Figure 24

TYPICAL CHARACTERISTICS

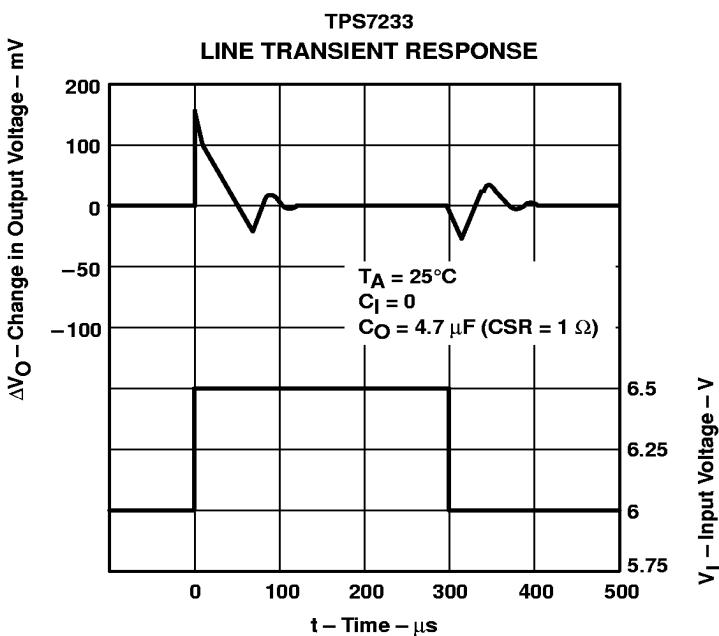


Figure 25

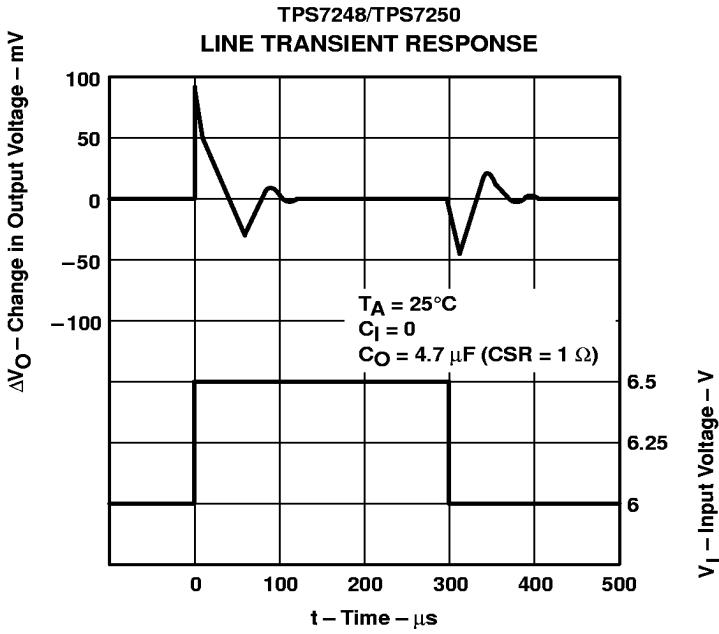


Figure 26



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**TPS7201Q, TPS7233Q, TPS7248Q, TPS7250Q
 TPS7201Y, TPS7233Y, TPS7248Y, TPS7250Y
 MICROPOWER LOW-DROPOUT (LDO) VOLTAGE REGULATORS**

SLVS102E – MARCH 1995 – REVISED JANUARY 1997

TYPICAL CHARACTERISTICS

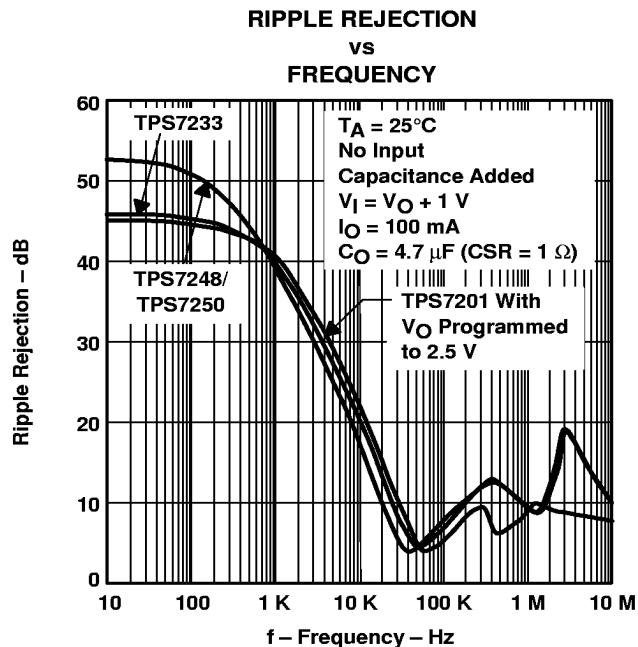


Figure 27

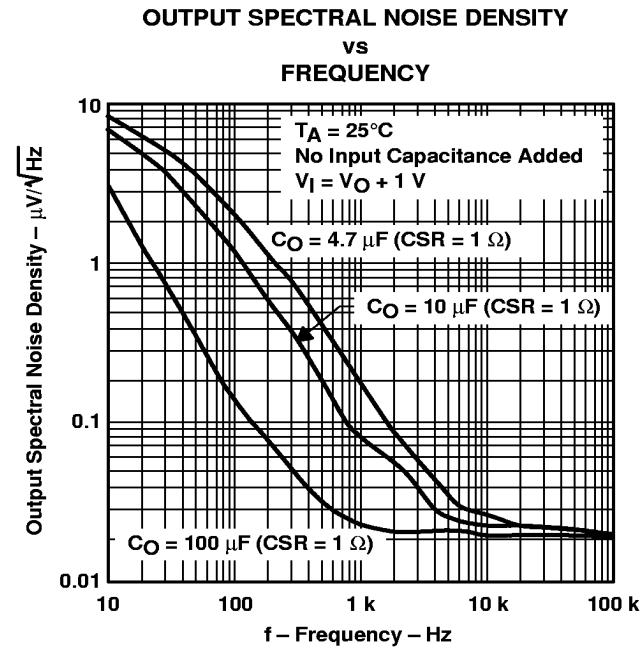


Figure 28

TYPICAL CHARACTERISTICS

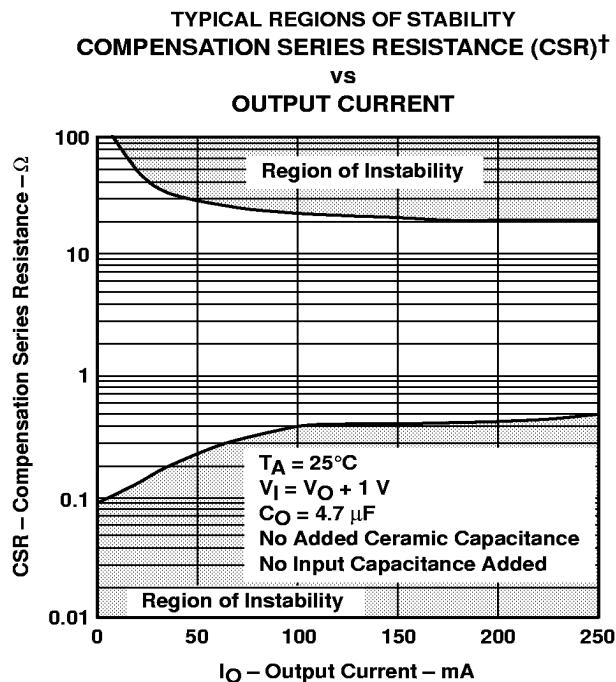


Figure 29

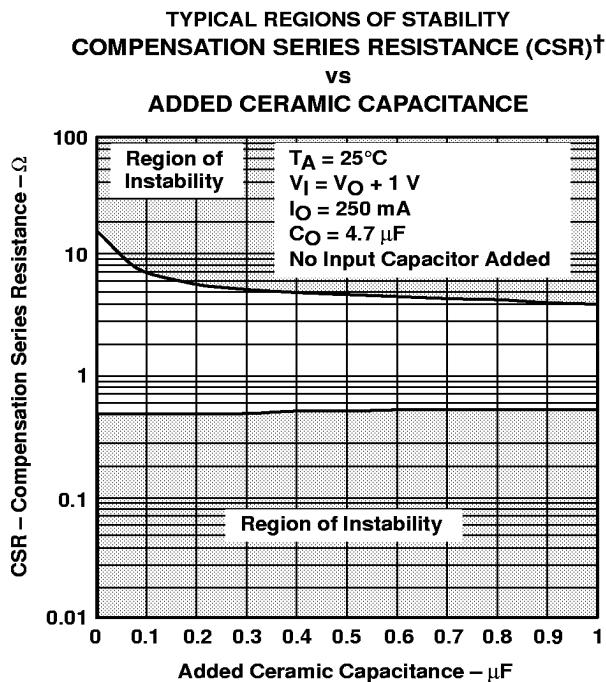


Figure 30

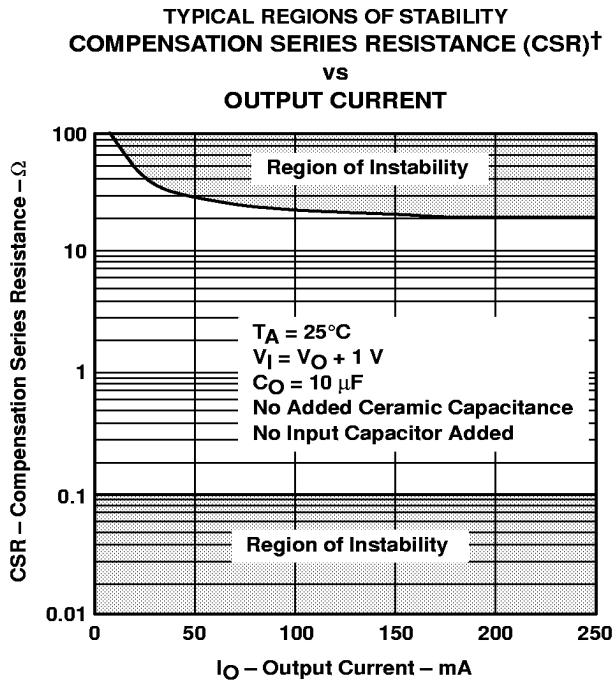


Figure 31

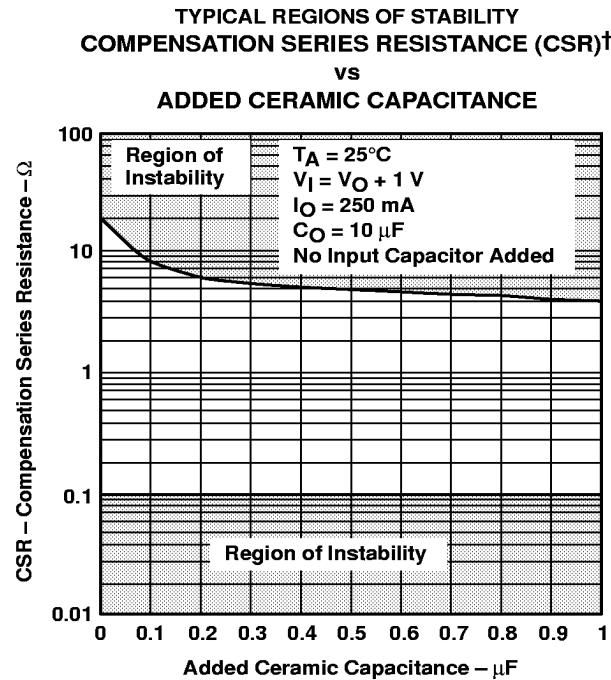


Figure 32

[†] CSR refers to the total series resistance, including the ESR of the capacitor, any series resistance added externally, and PWB trace resistance to C_O .

**TPS7201Q, TPS7233Q, TPS7248Q, TPS7250Q
TPS7201Y, TPS7233Y, TPS7248Y, TPS7250Y
MICROPOWER LOW-DROPOUT (LDO) VOLTAGE REGULATORS**

SLVS102E – MARCH 1995 – REVISED JANUARY 1997

APPLICATION INFORMATION

The design of the TPS72xx family of low-dropout (LDO) regulators is based on the higher-current TPS71xx family. These new families of regulators have been optimized for use in battery-operated equipment and feature extremely low dropout voltages, low supply currents that remain constant over the full-output-current range of the device, and an enable input to reduce supply currents to less than 0.5 μ A when the regulator is turned off.

device operation

The TPS72xx uses a PMOS pass element to dramatically reduce both dropout voltage and supply current over more conventional PNP-pass-element LDO designs. The PMOS transistor is a voltage-controlled device and, unlike a PNP transistor, does not require increased drive current as output current increases. Supply current in the TPS72xx is essentially constant from no-load to maximum.

Current limiting and thermal protection prevent damage by excessive output current and/or power dissipation. The device switches into a constant-current mode at approximately 1 A; further load increases reduce the output voltage instead of increasing the output current. The thermal protection shuts the regulator off if the junction temperature rises above 165°C. Recovery is automatic when the junction temperature drops approximately 5°C below the high temperature trip point. The PMOS pass element includes a back diode that safely conducts reverse current when the input voltage level drops below the output voltage level.

A logic high on the enable input, \overline{EN} , shuts off the output and reduces the supply current to less than 0.5 μ A. \overline{EN} should be grounded in applications where the shutdown feature is not used.

Power good (PG) is an open-drain output signal used to indicate output-voltage status. A comparator circuit continuously monitors the output voltage. When the output drops to approximately 95% of its nominal regulated value, the comparator turns on and pulls PG low.

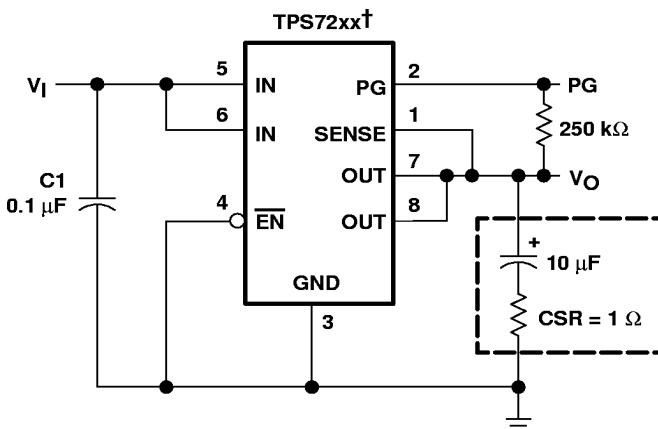
Transient loads or line pulses can also cause activation of PG if proper care is not taken in selecting the input and output capacitors. Load transients that are faster than 5 μ s can cause a signal on PG if high-ESR output capacitors (greater than approximately 7 Ω) are used. A 1- μ s transient causes a PG signal when using an output capacitor with greater than 3.5 Ω of ESR. It is interesting to note that the output-voltage spike during the transient can drop well below the reset threshold and still not trip if the transient duration is short. A 1- μ s transient must drop at least 500 mV below the threshold before tripping the PG circuit. A 2- μ s transient trips PG at just 400 mV below the threshold. Lower-ESR output capacitors help by reducing the drop in output voltage during a transient and should be used when fast transients are expected.

A typical application circuit is shown in Figure 33.



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APPLICATION INFORMATION



† TPS7233, TPS7248, TPS7250 (fixed-voltage options).

Figure 33. Typical Application Circuit

external capacitor requirements

Although not required, a 0.047- μF to 0.1- μF ceramic bypass input capacitor, connected between IN and GND and located close to the TPS72xx, is recommended to improve transient response and noise rejection. A higher-value electrolytic input capacitor may be necessary if large, fast-rise-time load transients are anticipated and the device is located several inches from the power source.

An output capacitor is required to stabilize the internal feedback loop. For most applications, a 10- μF to 15- μF solid-tantalum capacitor with a 0.5- Ω resistor (see capacitor selection table) in series is sufficient. The maximum capacitor ESR should be limited to 1.3 Ω to allow for ESR doubling at cold temperatures. Figure 34 shows the transient response of a 5-mA to 85-mA load using a 10- μF output capacitor with a total ESR of 1.7 Ω .

A 4.7- μF solid-tantalum capacitor in series with a 1- Ω resistor may also be used (see Figures 29 and 30) provided the ESR of the capacitor does not exceed 1 Ω at room temperature and 2 Ω over the full operating temperature range.



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TPS7201Y, TPS7233Y, TPS7248Y, TPS7250Y
MICROPOWER LOW-DROPOUT (LDO) VOLTAGE REGULATORS**

SLVS102E – MARCH 1995 – REVISED JANUARY 1997

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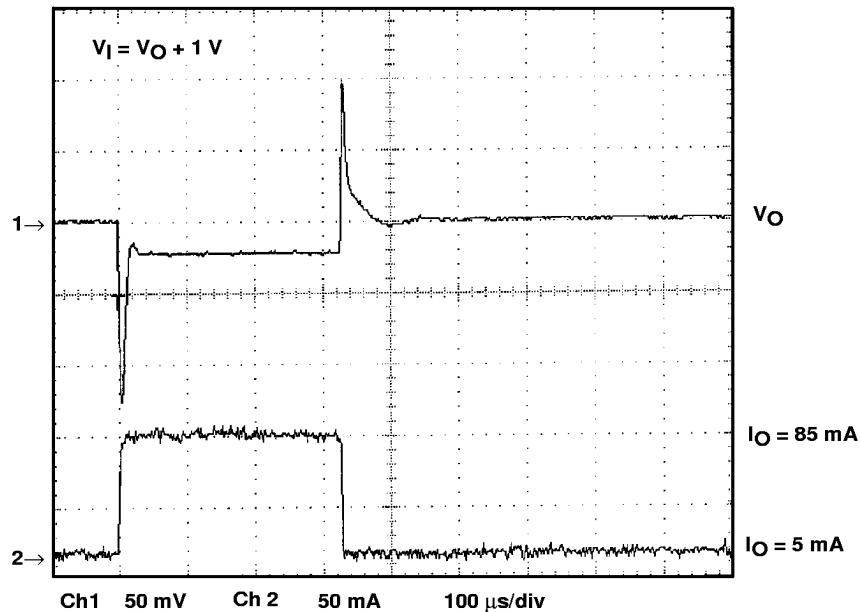


Figure 34. Load Transient Response (CSR total = 1.7 Ω), TPS7248Q

A partial listing of surface-mount capacitors usable with the TPS72xx family is provided below. This information (along with the stability graphs, Figures 29 through 32) is included to assist the designer in selecting suitable capacitors.

CAPACITOR SELECTION

PART NO.	MFR.	VALUE	MAX ESR†	SIZE (H × L × W)†
592D156X0020R2T	Sprague	15 μ F, 20 V	1.1	1.2 × 7.2 × 6
595D156X0025C2T	Sprague	15 μ F, 25 V	1	2.5 × 7.1 × 3.2
595D106X0025C2T	Sprague	10 μ F, 25 V	1.2	2.5 × 7.1 × 3.2
695D106X0035G2T	Sprague	10 μ F, 35 V	1.3	2.5 × 7.6 × 2.5

† Size is in mm. ESR is maximum resistance in ohms at 100 kHz and $T_A = 25^\circ\text{C}$. Listings are sorted by height.

sense-pin connection

SENSE must be connected to OUT for proper operation of the regulator. Normally this connection should be as short as possible; however, remote sense may be implemented in critical applications when proper care of the circuit path is exercised. SENSE internally connects to a high-impedance wide-bandwidth amplifier through a resistor-divider network, and any noise pickup on the PCB trace will feed through to the regulator output. SENSE must be routed to minimize noise pickup. Filtering SENSE using an RC network is not recommended because of the possibility of inducing regulator instability.



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APPLICATION INFORMATION

output voltage programming

The output voltage of the TPS7201 adjustable regulator is programmed using an external resistor divider as shown in Figure 35. The output voltage is calculated using:

$$V_O = V_{\text{ref}} \cdot \left(1 + \frac{R_1}{R_2} \right) \quad (1)$$

where

$V_{\text{ref}} = 1.188 \text{ V typ}$ (the internal reference voltage)

Resistors R1 and R2 should be chosen for approximately 7- μA divider current. Lower value resistors can be used but offer no inherent advantage and waste more power. Higher values should be avoided as leakage currents at FB increase the output voltage error. The recommended design procedure is to choose $R_2 = 169 \text{ k}\Omega$ to set the divider current at 7 μA and then calculate R1 using:

$$R_1 = \left(\frac{V_O}{V_{\text{ref}}} - 1 \right) \cdot R_2 \quad (2)$$

OUTPUT VOLTAGE PROGRAMMING GUIDE		
OUTPUT VOLTAGE (V)	DIVIDER RESISTANCE ($\text{k}\Omega$)†	
	R1	R2
2.5	191	169
3.3	309	169
3.6	348	169
4	402	169
5	549	169
6.4	750	169

† 1% values shown.

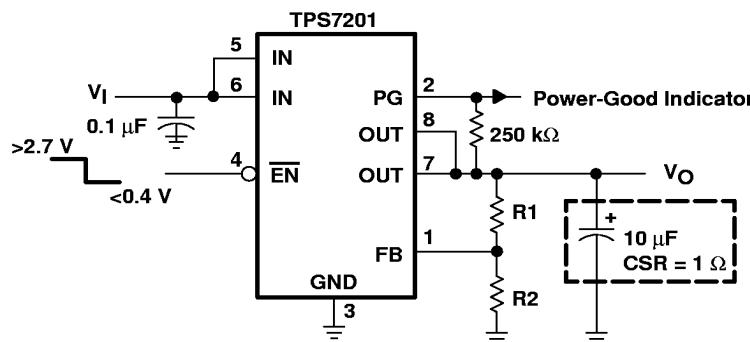


Figure 35. TPS7201 Adjustable LDO Regulator Programming

**TPS7201Q, TPS7233Q, TPS7248Q, TPS7250Q
TPS7201Y, TPS7233Y, TPS7248Y, TPS7250Y
MICROPOWER LOW-DROPOUT (LDO) VOLTAGE REGULATORS**

SLVS102E – MARCH 1995 – REVISED JANUARY 1997

APPLICATION INFORMATION

power dissipation and junction temperature

Specified regulator operation is assured to a junction temperature of 125°C; the maximum junction temperature allowable to avoid damaging the device is 150°C. These restrictions limit the power dissipation that the regulator can handle in any given application. To ensure the junction temperature is within acceptable limits, calculate the maximum allowable dissipation, $P_{D(\max)}$, and the actual dissipation, P_D , which must be less than or equal to $P_{D(\max)}$.

The maximum-power-dissipation limit is determined using the following equation:

$$P_{D(\max)} = \frac{T_{J\max} - T_A}{R_{\theta JA}}$$

Where

$T_{J\max}$ is the maximum allowable junction temperature, i.e., 150°C absolute maximum and 125°C recommended operating temperature.

$R_{\theta JA}$ is the thermal resistance junction-to-ambient for the package, i.e., 172°C/W for the 8-terminal SOIC and 238°C/W for the 8-terminal TSSOP.

T_A is the ambient temperature.

The regulator dissipation is calculated using:

$$P_D = (V_I - V_O) \cdot I_O$$

Power dissipation resulting from quiescent current is negligible.

regulator protection

The TPS72xx PMOS-pass transistor has a built-in back diode that safely conducts reverse currents when the input voltage drops below the output voltage (e.g., during power down). Current is conducted from the output to the input and is not internally limited. If extended reverse voltage is anticipated, external limiting might be appropriate.

The TPS72xx also features internal current limiting and thermal protection. During normal operation, the TPS72xx limits output current to approximately 1 A. When current limiting engages, the output voltage scales back linearly until the overcurrent condition ends. While current limiting is designed to prevent gross device failure, care should be taken not to exceed the power dissipation ratings of the package. If the temperature of the device exceeds 165°C, thermal-protection circuitry shuts it down. Once the device has cooled, regulator operation resumes.



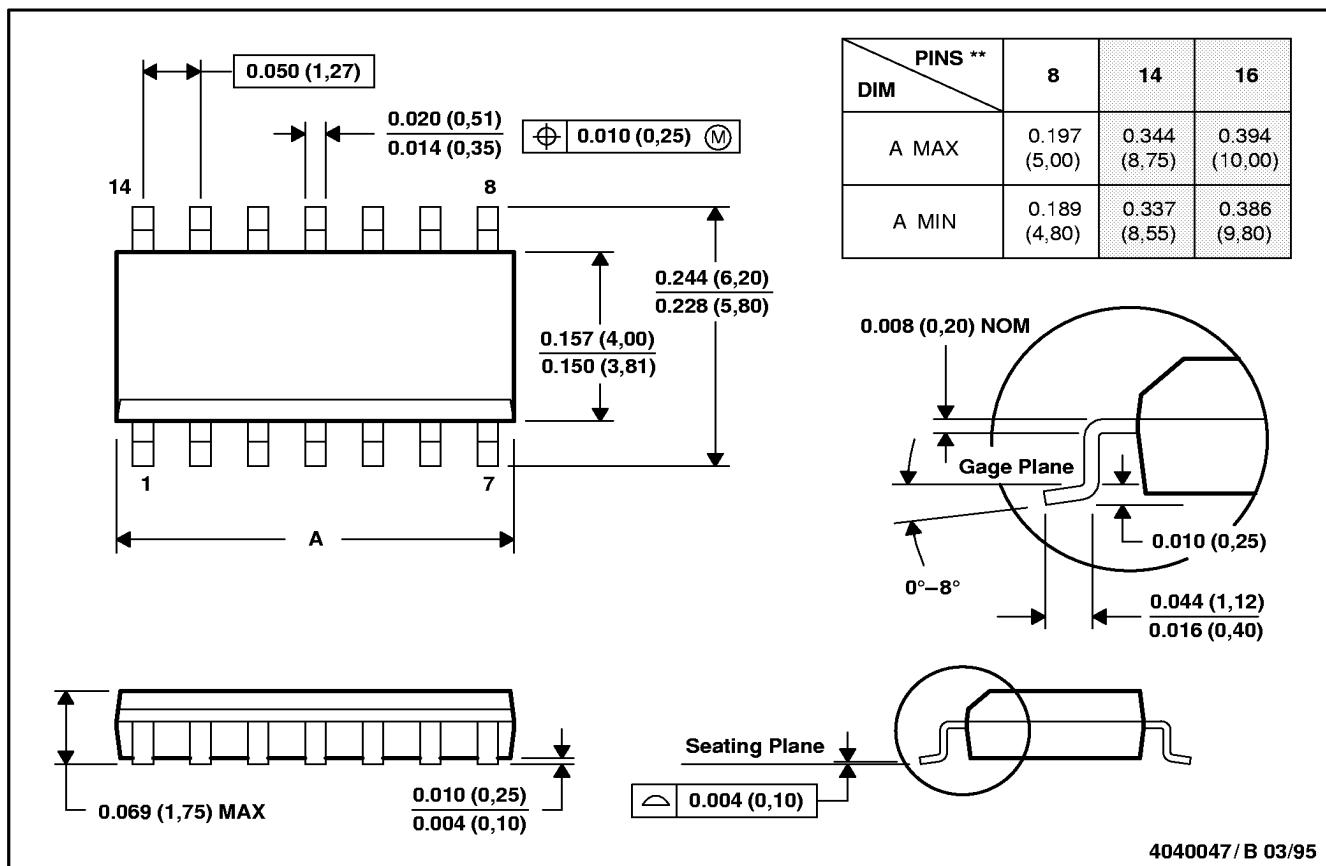
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TPS7201Y, TPS7233Y, TPS7248Y, TPS7250Y
MICROPOWER LOW-DROPOUT (LDO) VOLTAGE REGULATORS**
SLVS102E – MARCH 1995 – REVISED JANUARY 1997

MECHANICAL DATA

D (R-PDSO-G**) PLASTIC SMALL-OUTLINE PACKAGE

14 PIN SHOWN



- NOTES:
- B. All linear dimensions are in inches (millimeters).
 - C. This drawing is subject to change without notice.
 - D. Body dimensions do not include mold flash or protrusion, not to exceed 0.006 (0.15).
 - E. Four center pins are connected to die mount pad.
 - F. Falls within JEDEC MS-012



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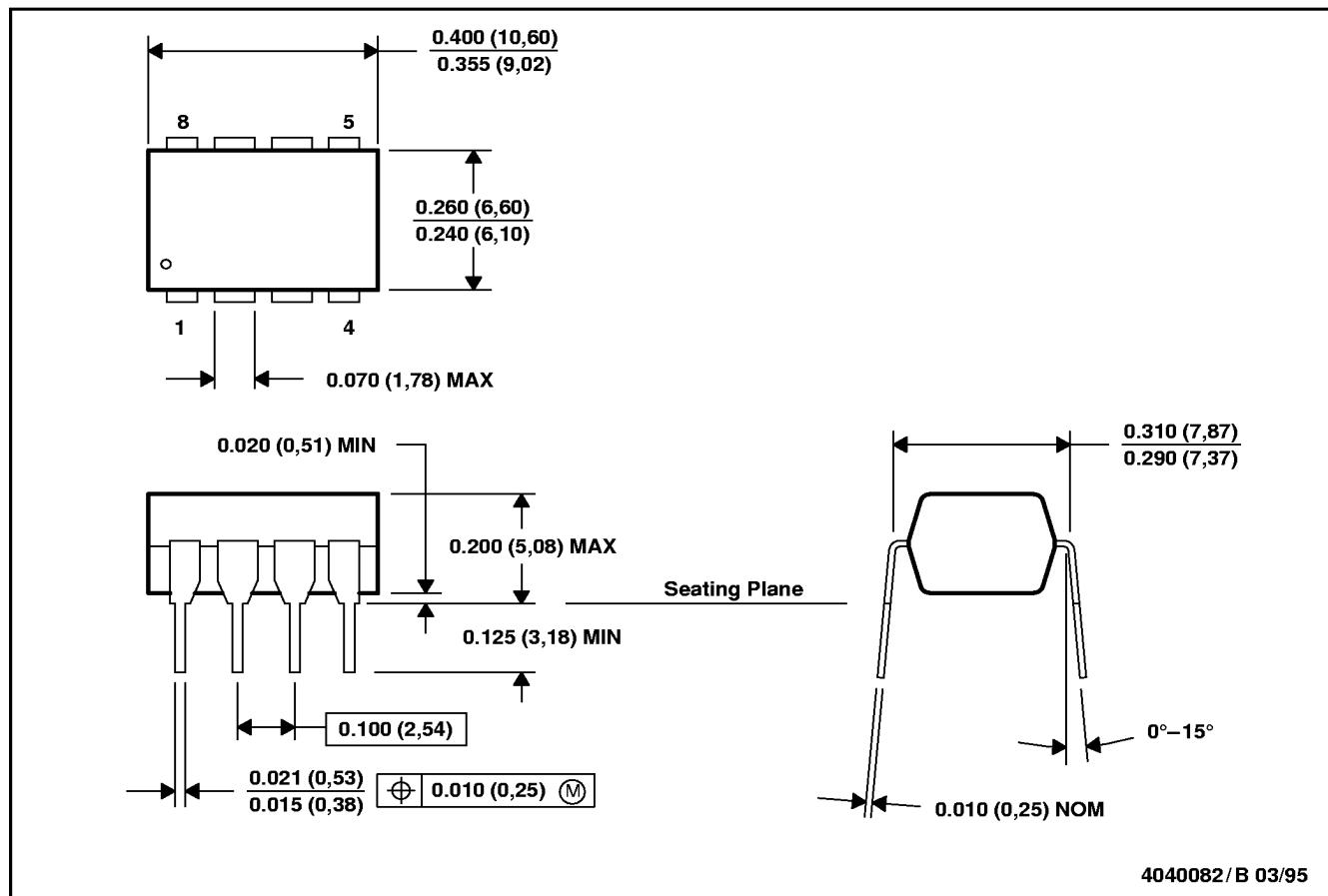
**TPS7201Q, TPS7233Q, TPS7248Q, TPS7250Q
 TPS7201Y, TPS7233Y, TPS7248Y, TPS7250Y
 MICROPOWER LOW-DROPOUT (LDO) VOLTAGE REGULATORS**

SLVS102E – MARCH 1995 – REVISED JANUARY 1997

MECHANICAL DATA

P (R-PDIP-T8)

PLASTIC DUAL-IN-LINE PACKAGE



4040082/B 03/95

- NOTES: A. All linear dimensions are in inches (millimeters).
 B. This drawing is subject to change without notice.
 C. Falls within JEDEC MS-001

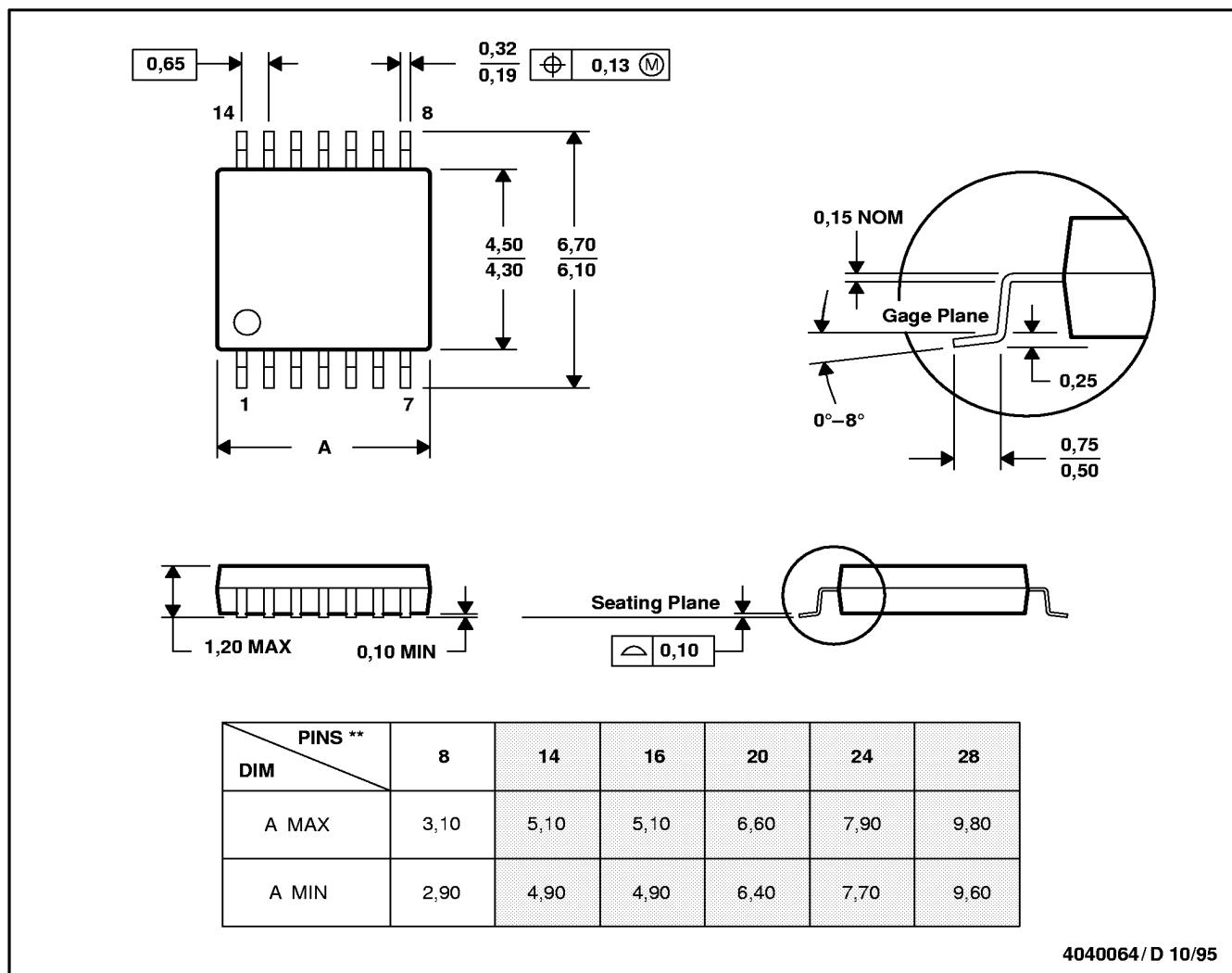
**TPS7201Q, TPS7233Q, TPS7248Q, TPS7250Q
TPS7201Y, TPS7233Y, TPS7248Y, TPS7250Y
MICROPOWER LOW-DROPOUT (LDO) VOLTAGE REGULATORS**
SLVS102E – MARCH 1995 – REVISED JANUARY 1997

MECHANICAL DATA

PW (R-PDSO-G)**

PLASTIC SMALL-OUTLINE PACKAGE

14 PIN SHOWN



- NOTES:
- A. All linear dimensions are in millimeters.
 - B. This drawing is subject to change without notice.
 - C. Body dimensions do not include mold flash or protrusion not to exceed 0.15.
 - D. Falls within JEDEC MO-153



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