FEMTOCLOCKSTM CRYSTAL-TO-3.3V LVPECL FREQUENCY SYNTHESIZER

GENERAL DESCRIPTION



The ICS843002 is a 2 output LVPECL synthesizer optimized to generate Fibre Channel reference clock frequencies and is a member of the HiPerClocks $^{\text{TM}}$ family of high performance clock solutions from ICS. Using a 26.5625MHz, 18pF

parallel resonant crystal, the following frequencies can be generated based on the 2 frequency select pins (F_SEL[1:0]): 212.5MHz, 187.5MHz, 159.375MHz, 106.25MHz, and 53.125MHz. The ICS843002 uses ICS' 3rd generation low phase noise VCO technology and can achieve 1ps or lower typical rms phase jitter, easily meeting Fibre Channel jitter requirements. The ICS843002 is packaged in a small 20-pin TSSOP package.

FEATURES

- Two 3.3V LVPECL outputs
- Selectable crystal oscillator interface or LVCMOS/LVTTL single-ended input
- Supports the following output frequencies: 212.5MHz, 187.5MHz, 159.375MHz, 106.25MHz and 53.125MHz
- VCO range: 560MHz 680MHz
- RMS phase jitter (637kHz 10MHz): 0.72ps (typical)
- Typical phase noise at 212.5MHz

Phase noise:

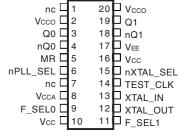
Offset	Noise Power
100Hz	87.7 dBc/Hz
1KHz	111.6 dBc/Hz
10KHz	124.3 dBc/Hz
100KHz	124.3 dBc/Hz

- Full 3.3V supply mode
- · Lead-Free package RoHS compliant
- -30°C to 85°C ambient operating temperature

FREQUENCY SELECT FUNCTION TABLE

Inputs						Output
Input Frequency (MHz)	F_SEL1	F_SEL0	M Divider Value	N Divider Value	M/N Divider Value	Frequency (MHz)
26.5625	0	0	24	3	8	212.5
26.5625	0	1	24	4	6	159.375
26.5625	1	0	24	6	4	106.25
26.5625	1	1	24	12	2	53.125
23.4375	0	0	24	3	8	187.5

PIN ASSIGNMENT

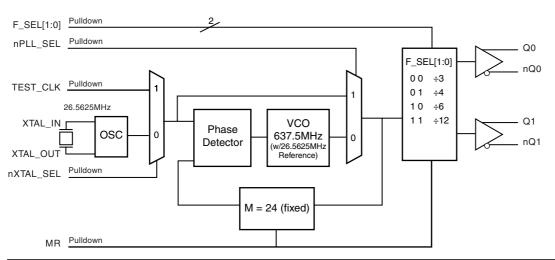


ICS843002 20-Lead TSSOP

6.5mm x 4.4mm x 0.92mm package body

G Package Top View

BLOCK DIAGRAM





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TABLE 1. PIN DESCRIPTIONS

Number	Name	Туре		Description
1, 7	nc	Unused		No connect.
2, 20	V _{cco}	Power		Output supply pins.
3, 4	Q0, nQ0	Ouput		Differential output pair. LVPECL interface levels.
5	MR	Input	Pulldown	Active HIGH Master Reset. When logic HIGH, the internal dividers are reset causing the true outputs Qx to go low and the inverted outputs nQx to go high. When logic LOW, the internal dividers and the outputs are enabled. LVCMOS/LVTTL interface levels.
6	nPLL_SEL	Input	Pulldown	Selects between the PLL and TEST_CLK as input to the dividers. When LOW, selects PLL (PLL Enable). When HIGH, deselects the reference clock (PLL Bypass). LVCMOS/LVTTL interface levels.
8	V_{CCA}	Power		Analog supply pin.
9, 11	F_SEL0, F_SEL1	Input	Pulldown	Frequency select pins. LVCMOS/LVTTL interface levels.
10, 16	V_{cc}	Power		Core supply pin.
12, 13	XTAL_OUT, XTAL_IN	Input		Parallel resonant crystal interface. XTAL_OUT is the output, XTAL_IN is the input.
14	TEST_CLK	Input	Pulldown	LVCMOS/LVTTL clock input.
15	nXTAL_SEL	Input	Pulldown	Selects between crystal or TEST_CLK inputs as the the PLL Reference source. Selects XTAL inputs when LOW. Selects TEST_CLK when HIGH. LVCMOS/LVTTL interface levels.
17	$V_{\sf EE}$	Power		Negative supply pins.
18, 19	nQ1, Q1	Output		Differential output pair. LVPECL interface levels.

NOTE: Pulldown refer to internal input resistors. See Table 2, Pin Characteristics, for typical values.

TABLE 2. PIN CHARACTERISTICS

Symbol	Parameter	Test Conditions	Minimum	Typical	Maximum	Units
C _{IN}	Input Capacitance			4		pF
R _{PULLDOWN}	Input Pulldown Resistor			51		kΩ



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ABSOLUTE MAXIMUM RATINGS

Supply Voltage, V_{CC} 4.6V

Inputs, V_1 -0.5V to V_{cc} + 0.5V

Outputs, I_{\odot}

Continuous Current 50mA Surge Current 100mA

 $\label{eq:package} \begin{array}{ll} \mbox{Package Thermal Impedance, $\theta_{\rm JA}$} & 73.2^{\circ}\mbox{C/W (0 lfpm)} \\ \mbox{Storage Temperature, $T_{\rm STG}$} & -65^{\circ}\mbox{C to 150}^{\circ}\mbox{C} \\ \end{array}$

NOTE: Stresses beyond those listed under Absolute Maximum Ratings may cause permanent damage to the device. These ratings are stress specifications only. Functional operation of product at these conditions or any conditions beyond those listed in the *DC Characteristics* or *AC Characteristics* is not implied. Exposure to absolute maximum rating conditions for extended periods may affect product reliability.

Table 3A. Power Supply DC Characteristics, $V_{CC} = V_{CCA} = V_{CCO} = 3.3V \pm 10\%$, TA = -30°C to 85°C

Symbol	Parameter	Test Conditions	Minimum	Typical	Maximum	Units
V _{cc}	Core Supply Voltage		2.97	3.3	3.63	V
V _{CCA}	Analog Supply Voltage		2.97	3.3	3.63	V
V _{cco}	Output Supply Voltage		2.97	3.3	3.63	V
I _{EE}	Power Supply Current				135	mA
I _{cc}	Core Supply Current				100	mA
I _{CCA}	Analog Supply Current				15	mA
I _{cco}	Output Supply Current				31	mA

Table 3B. LVCMOS / LVTTL DC Characteristics, $V_{CC} = V_{CCA} = V_{CCO} = 3.3V \pm 10\%$, TA = -30°C to 85°C

Symbol	Parameter		Test Conditions	Minimum	Typical	Maximum	Units
V _{IH}	Input High Vol	tage		2		V _{cc} + 0.3	V
V _{IL}	Input	nPLL_SEL, nXTAL_SEL, F_SEL0, F_SEL1, MR		-0.3		0.8	٧
	Low Voltage	TEST_CLK		-0.3		1.0	V
I _{IH}	Input High Current	TEST_CLK, MR, F_SEL0, F_SEL1, nPLL_SEL, nXTAL_SEL,	$V_{CC} = V_{IN} = 3.63V$			150	μΑ
I	Input Low Current	TEST_CLK, MR, F_SEL0, F_SEL1, nPLL_SEL, nXTAL_SEL,	V _{CC} = 3.63V, V _{IN} = 0V	-150			μΑ

Table 3C. LVPECL DC Characteristics, $V_{CC} = V_{CCA} = V_{CCO} = 3.3V \pm 10\%$, TA = -30°C to 85°C

Symbol	Parameter	Test Conditions	Minimum	Typical	Maximum	Units
V _{OH}	Output High Voltage; NOTE 1		V _{cco} - 1.4		V _{cco} - 0.9	٧
V _{OL}	Output Low Voltage; NOTE 1		V _{cco} - 2.0		V _{cco} - 1.7	V
V _{SWING}	Peak-to-Peak Output Voltage Swing		0.6		1.0	V

NOTE 1: Outputs terminated with 50 Ω to V_{cco} - 2V.

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TABLE 4. CRYSTAL CHARACTERISTICS

Parameter	Test Conditions	Minimum	Typical	Maximum	Units
Mode of Oscillation		Fundamental			
Frequency		23.33	26.5625	28.33	MHz
Equivalent Series Resistance (ESR)				50	Ω
Shunt Capacitance				7	pF

NOTE: Characterized using an 18pF parallel resonant crystal.

Table 5. AC Characteristics, $V_{CC} = V_{CCA} = V_{CCO} = 3.3V \pm 10\%$, TA = -30°C to 85°C

Symbol	Parameter	Test Conditions	Minimum	Typical	Maximum	Units
		F_SEL[1:0] = 00	186.67		226.67	MHz
f	Output Frequency	F_SEL[1:0] = 01	140		170	MHz
f _{out}	Output Frequency	F_SEL[1:0] = 10	93.33		113.33	MHz
		F_SEL[1:0] =11	46.67		56.67	MHz
tsk(o)	Output Skew; NOTE 1, 2				20	ps
	RMS Phase Jitter (Random); NOTE 3	212.5MHz, (637KHz - 10MHz)		0.72		ps
#::+(<i>O</i> ()		159.375MHz, (637KHz - 10MHz)		0.76		ps
<i>t</i> jit(Ø)		106.25MHz, (637KHz - 10MHz)		0.84		ps
		53.125MHz, (637KHz - 10MHz)		0.97		ps
t _R / t _F	Output Rise/Fall Time	20% to 80%	300		600	ps
odc	Output Duty Cycle	F_SEL[1:0] =00	46		54	%
ouc	Output Duty Cycle	F_SEL[1:0] ≠ 00	49		51	%

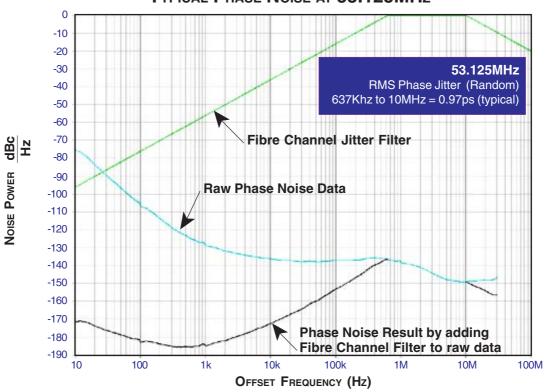
NOTE 1: Defined as skew between outputs at the same supply voltages and with equal load conditions.

Measured at $V_{cco}/2$.

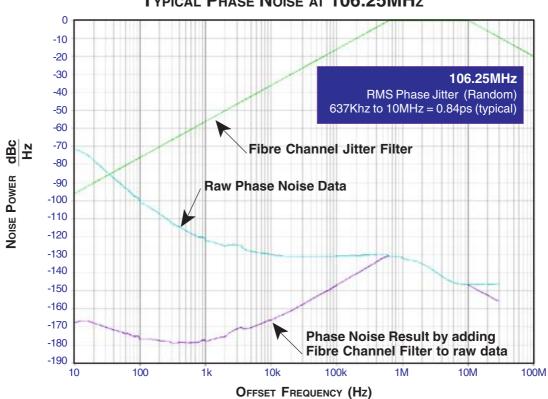
NOTE 2: This parameter is defined in accordance with JEDEC Standard 65. NOTE 3: See Phase Noise plot.



Typical Phase Noise at 53.125MHz

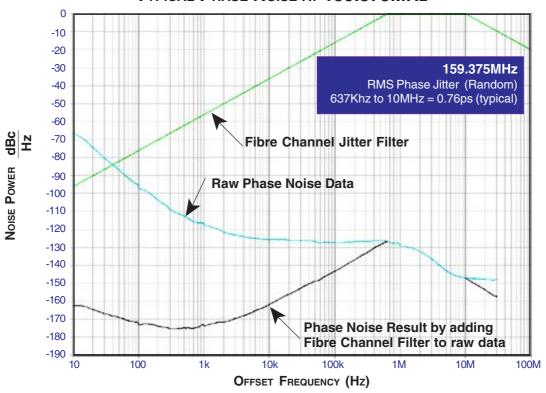


Typical Phase Noise at 106.25MHz

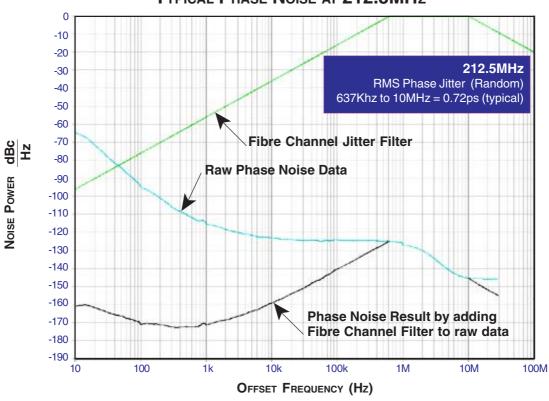




Typical Phase Noise at 159.375MHz

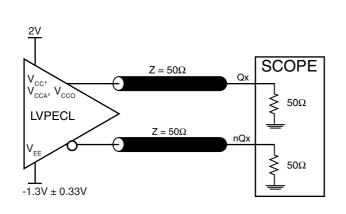


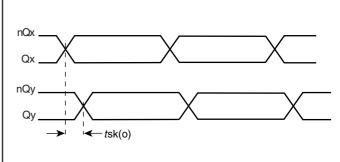
Typical Phase Noise at 212.5MHz



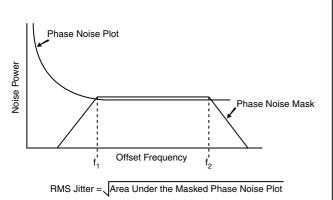
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PARAMETER MEASUREMENT INFORMATION

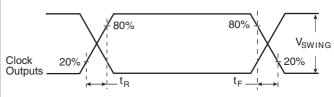




3.3V CORE/3.3V OUTPUT LOAD AC TEST CIRCUIT

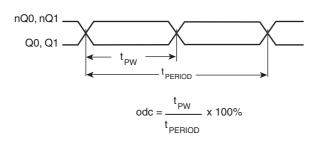


OUTPUT SKEW



RMS PHASE JITTER

OUTPUT RISE/FALL TIME



OUTPUT DUTY CYCLE/PULSE WIDTH/PERIOD

APPLICATION INFORMATION

Power Supply Filtering Techniques

As in any high speed analog circuitry, the power supply pins are vulnerable to random noise. The ICS843002 provides separate power supplies to isolate any high switching noise from the outputs to the internal PLL. $V_{\rm CC}, V_{\rm CCA},$ and $V_{\rm CCO}$ should be individually connected to the power supply plane through vias, and bypass capacitors should be used for each pin. To achieve optimum jitter performance, power supply isolation is required. Figure 1 illustrates how a 10Ω resistor along with a $10\mu F$ and a $.01\mu F$ bypass capacitor should be connected to each $V_{\rm CCA}$.

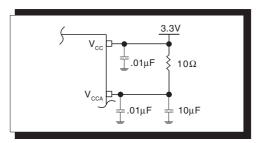


FIGURE 1. POWER SUPPLY FILTERING

TERMINATION FOR 3.3V LVPECL OUTPUT

The clock layout topology shown below is a typical termination for LVPECL outputs. The two different layouts mentioned are recommended only as guidelines.

FOUT and nFOUT are low impedance follower outputs that generate ECL/LVPECL compatible outputs. Therefore, terminating resistors (DC current path to ground) or current sources must be used for functionality. These outputs are

designed to drive 50Ω transmission lines. Matched impedance techniques should be used to maximize operating frequency and minimize signal distortion. Figures 2A and 2B show two different layouts which are recommended only as guidelines. Other suitable clock layouts may exist and it would be recommended that the board designers simulate to guarantee compatibility across all printed circuit and clock component process variations.

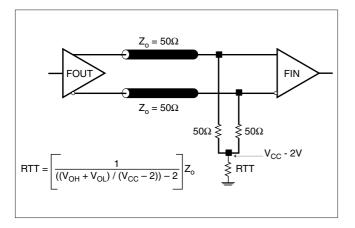


FIGURE 2A. LVPECL OUTPUT TERMINATION

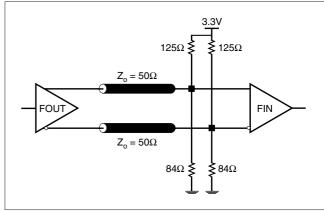
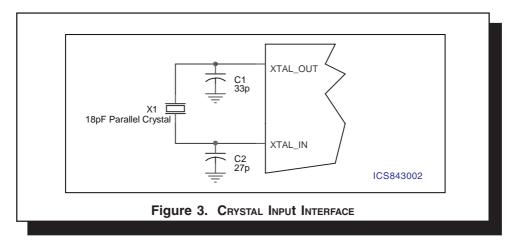


FIGURE 2B. LVPECL OUTPUT TERMINATION

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CRYSTAL INPUT INTERFACE

The ICS843002 has been characterized with 18pF parallel resonant crystals. The capacitor values shown in Figure 3 below were determined using a 26.5625MHz 18pF parallel resonant crystal and were chosen to minimize the ppm error.



3.3V LVPECL Frequency Synthesizer

LAYOUT GUIDELINE

Figure 4A shows a schematic example of the ICS843002. An example of LVEPCL termination is shown in this schematic. Additional LVPECL termination approaches are shown in the LVPECL Termination Application Note. In this example, an 18 pF

parallel resonant 26.5625MHz crystal is used. The C1=27pF and C2=33pF are recommended for frequency accuracy. For different board layout, the C1 and C2 may be slightly adjusted for optimizing frequency accuracy.

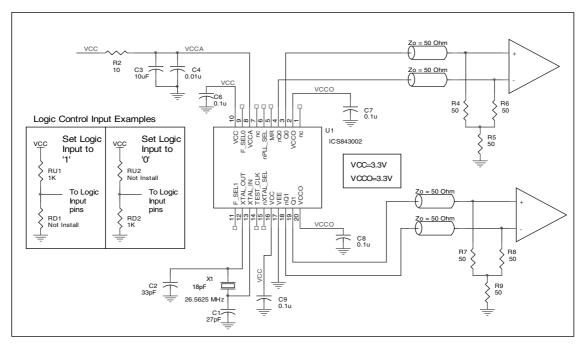


FIGURE 4A. ICS843002 SCHEMATIC EXAMPLE

PC BOARD LAYOUT EXAMPLE

Figure 4B shows an example of ICS843002 P.C. board layout. The crystal X1 footprint shown in this example allows installation of either surface mount HC49S or through-hole HC49 package. The footprints of other components in this example are listed in the Table 6. There should be at least one decoupling capacitor per power pin. The decoupling capacitors should be located as close as possible to the power pins. The layout assumes that the board has clean analog power ground plane.

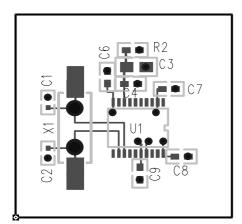


TABLE 6. FOOTPRINT TABLE

Reference	Size
C1, C2	0402
C3	0805
C4, C5, C6, C7, C8	0603
R2	0603

NOTE: Table 6, lists component sizes shown in this layout example.

FIGURE 4B. ICS843002 PC BOARD LAYOUT EXAMPLE

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Power Considerations

This section provides information on power dissipation and junction temperature for the ICS843002. Equations and example calculations are also provided.

1. Power Dissipation.

The total power dissipation for the ICS843002 is the sum of the core power plus the power dissipated in the load(s). The following is the power dissipation for $V_{CC} = 3.3V + 10\% = 3.63V$, which gives worst case results.

NOTE: Please refer to Section 3 for details on calculating power dissipated in the load.

- Power (core)_{MAX} = V_{CC_MAX} * I_{EE_MAX} = 3.63V * 135mA = 490mW
- Power (outputs)_{MAX} = 30mW/Loaded Output pair
 If all outputs are loaded, the total power is 2 * 30mW = 60mW

Total Power (3.63V, with all outputs switching) = 490mW + 60mW = 550mW

2. Junction Temperature.

Junction temperature, Tj, is the temperature at the junction of the bond wire and bond pad and directly affects the reliability of the device. The maximum recommended junction temperature for HiPerClockS $^{\text{TM}}$ devices is 125 $^{\circ}$ C.

The equation for Tj is as follows: Tj = θ_{JA} * Pd_total + T_A

Tj = Junction Temperature

 θ_{JA} = Junction-to-Ambient Thermal Resistance

Pd_total = Total Device Power Dissipation (example calculation is in section 1 above)

 T_{Δ} = Ambient Temperature

In order to calculate junction temperature, the appropriate junction-to-ambient thermal resistance $\theta_{\rm JA}$ must be used. Assuming a moderate air flow of 200 linear feet per minute and a multi-layer board, the appropriate value is 66.6°C/W per Table 7 below.

Therefore, Tj for an ambient temperature of 85°C with all outputs switching is:

 $85^{\circ}\text{C} + 0.550\text{W} * 66.6^{\circ}\text{C/W} = 121.6^{\circ}\text{C}$. This is below the limit of 125°C .

This calculation is only an example. Tj will obviously vary depending on the number of loaded outputs, supply voltage, air flow, and the type of board (single layer or multi-layer).

Table 7. Thermal Resistance θ_{JA} for 20-pin TSSOP, Forced Convection

0200500Single-Layer PCB, JEDEC Standard Test Boards114.5°C/W98.0°C/W88.0°C/WMulti-Layer PCB, JEDEC Standard Test Boards73.2°C/W66.6°C/W63.5°C/W

θ₁ by Velocity (Linear Feet per Minute)

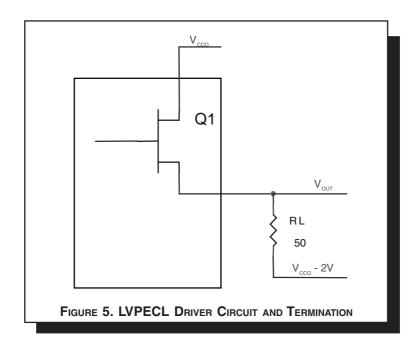
NOTE: Most modern PCB designs use multi-layered boards. The data in the second row pertains to most designs.



3. Calculations and Equations.

The purpose of this section is to derive the power dissipated into the load.

LVPECL output driver circuit and termination are shown in Figure 5.



To calculate worst case power dissipation into the load, use the following equations which assume a 50Ω load, and a termination voltage of V_{CCO} - 2V.

• For logic high,
$$V_{OUT} = V_{OH_MAX} = V_{CCO_MAX} - 0.9V$$

$$(V_{CCO_MAX} - V_{OH_MAX}) = 0.9V$$

• For logic low,
$$V_{OUT} = V_{OL_MAX} = V_{CCO_MAX} - 1.7V$$

$$(V_{CCO_MAX} - V_{OL_MAX}) = 1.7V$$

Pd_H is power dissipation when the output drives high. Pd_L is the power dissipation when the output drives low.

$$Pd_{-}H = [(V_{OH_MAX} - (V_{CCO_MAX} - 2V))/R_{L}] * (V_{CCO_MAX} - V_{OH_MAX}) = [(2V - (V_{CCO_MAX} - V_{OH_MAX}))/R_{L}] * (V_{CCO_MAX} - V_{OH_MAX}) = [(2V - 0.9V)/50\Omega] * 0.9V = 19.8mW$$

$$Pd_{L} = [(V_{\text{OL_MAX}} - (V_{\text{CCO_MAX}} - 2V))/R_{L}] * (V_{\text{CCO_MAX}} - V_{\text{OL_MAX}}) = [(2V - (V_{\text{CCO_MAX}} - V_{\text{OL_MAX}}))/R_{L}] * (V_{\text{CCO_MAX}} - V_{\text{OL_MAX}}) = [(2V - 1.7V)/50\Omega] * 1.7V = 10.2mW$$

Total Power Dissipation per output pair = Pd_H + Pd_L = **30mW**

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RELIABILITY INFORMATION

Table 8. $\theta_{\text{JA}} \text{vs. Air Flow Table for 20 Lead TSSOP}$

θ_{AA} by Velocity (Linear Feet per Minute)

 0
 200
 500

 Single-Layer PCB, JEDEC Standard Test Boards
 114.5°C/W
 98.0°C/W
 88.0°C/W

 Multi-Layer PCB, JEDEC Standard Test Boards
 73.2°C/W
 66.6°C/W
 63.5°C/W

NOTE: Most modern PCB designs use multi-layered boards. The data in the second row pertains to most designs.

TRANSISTOR COUNT

The transistor count for ICS843002 is: 2578



PACKAGE OUTLINE - G SUFFIX FOR 20 LEAD TSSOP

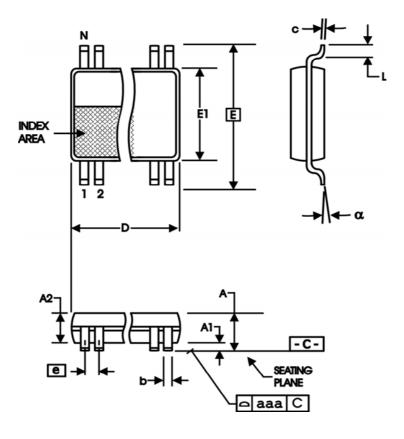


TABLE 9. PACKAGE DIMENSIONS

SYMBOL	Millimeters			
STWIDOL	MIN	MAX		
N	2	0		
А		1.20		
A1	0.05	0.15		
A2	0.80	1.05		
b	0.19	0.30		
С	0.09	0.20		
D	6.40	6.60		
E	6.40 E	BASIC		
E1	4.30	4.50		
е	0.65 E	BASIC		
L	0.45	0.75		
α	0°	8°		
aaa		0.10		

Reference Document: JEDEC Publication 95, MO-153



FEMTOCLOCKSTM CRYSTAL-TO-3.3V LVPECL FREQUENCY SYNTHESIZER

TABLE 10. ORDERING INFORMATION

Part/Order Number	Marking	Package	Shipping Packaging	Temperature
ICS843002AG	ICS843002AG	20 Lead TSSOP	tube	-30°C to 85°C
ICS843002AGT	ICS843002AG	20 Lead TSSOP	2500 tape & reel	-30°C to 85°C
ICS843002AGLF	ICS843002ALF	20 Lead "Lead-Free" TSSOP	tube	-30°C to 85°C
ICS843002AGLFT	ICS843002ALF	20 Lead "Lead-Free" TSSOP	2500 tape & reel	-30°C to 85°C

NOTE: Parts that are ordered with an"LF" to the part number are the Pb-Free configuration and are RoHS compliant.

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REVISION HISTORY SHEET				
Rev	Table	Page	Description of Change	Date
Α		1	Added 187.5MHz to the Frequency Selection Function Table.	8/26/04
Α	T10	15	Ordering Information Table - added Lead Free part number.	9/30/04
Α	T5	4	AC Characteristics Table - corrected typo, f _{out} 180.67 min. to 186.67 min.	12/27/04
А		1	Features section - corrected frequency bullet to read "Supportsoutput frequencies"	2/7/05
	T10	15	Ordering Information Table - updated table.	
В	T5	4	AC Characteristics Table - deleted Propagation Delay.	5/6/05