

MASSISIA

Single chip AMPS/ETACS/NAMPS audio/data processor



- Single chip solution for all audio and data processing
- Low power consumption with several power down modes
- SAT decoding and transponding circuitry
- Serial interface

DESCRIPTION

The MAS9191A is a high integration BeCMOS IC for implementing the audio and data signal processing in AMPS, ETACS or NAMPS cellular phones. The power consumption of the device is very low due to several automatic and software controlled power down modes as well as the low power characteristics of the BeCMOS process.

DTMF receiver is also included to enable answering machine functions for the cellular phone. Only a minimal number of external components are needed to meet typical baseband requirements.

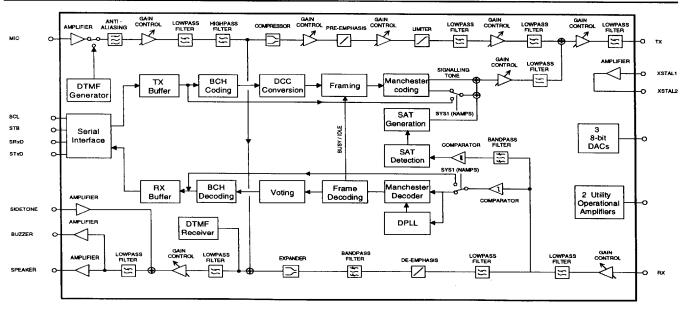
FEATURE

- Voice signal processing including compander, expander and digital gain adjustments
- DTMF and ST generators and DTMF receiver
- Busy/Idle extraction and arbitration with TX block, voting, BCH, data buffering and framing, DCC coding with hardware
- SAT detection and regeneration
 Compatible with AMPS/TACS or NAMPS systems
- Three 8-bit DACs and two operational amplifiers
- On-chip oscillator with clock output for μP
- · Simple 4-wire serial interface
- · Power down modes for all blocks
- 3.3V or 5V operation with low power consumption (RX block at 2mA/3.3V)
- 64-pin TQFP package with -40...85°C operation range

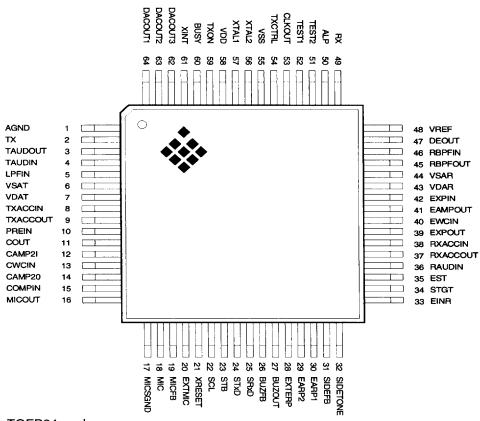
APPLICATION

- AMPS/ETACS Cellular phone
- NAMPS Cellular phone

BLOCK DIAGRAM



PIN CONFIGURATION



TQFP64 package

PIN DESCRIPTION

Pin name	Pin	Туре	Function
AGND	1	AO	Signal ground. The signal ground is generated internally and is equal to V _{DD} /2. The analog ground needs an external capacitor connected to system ground.
TX	2	AO	Transmitted data signal output. Connect this output through a 22nF capacitor to the transmitter.
TAUDOUT	3	AO	TX audio output from the TX audio block
TAUDIN	4	Al	TX audio input. The input for the TX audio signal, normally connected through a 22nF capacitor to TAUDOUT
LPFIN	5	Al	Input for TX limiter, lowpass filter or GC6 depending on the position of switches S15 and S16. The pin is normally left unconnected.
VSAT	6	G	Ground for TX. Connect to system ground.
VDAT	7	Р	Power supply for TX block. Use a bypass capacitor between pins VSAT and VDAT.
TXACCIN	8	AI	TX block extra Op Amp input. See application note in the APPLICATIONS section.
TXACCOUT	9	AO	TX block Op Amp output.



PIN DESCRIPTION

Pin name	Pin	Туре	Function
PREIN	10	Al	Pre-emphasis filter input. Filter has a +6dB/octave (±1dB) frequency response in the range 300Hz3kHz.
COUT	11	AO	Compressor output signal. The compression ratio is 2:1
CAMP2I	12	Al	Compressor 2nd amplifier input as well as GC4 input. Use an external 22nF capacitor between COUT and this pin.
CWCIN	13	AI	Compressor window comparator input. Use an external 22nF capacitor between CAMP2O and this pin.
CAMP2O	14	AO	Compressor 2nd amplifier output.
COMPIN	15	AI	Compressor input. The input is connected through a 22nF capacitor to MICOUT.
MICOUT	16	AO	Microphone amplifier output. See COMPIN. This output is used as a source for the side tone and for detection of the TX audio level.
MICSGND	17	AO	Microphone signal ground. This is the internal signal ground V _{DD} /2. If noise appears on the microphone signal an external capacitor may be needed between this pin and system ground.
MIC	18	Al	Microphone amplifier input. Using this pin and the MICFB output the microphone amplifier frequency response can be adjusted according to the microphone used. The level at this input should be in the range 510mVrms. The maximum gain of the microphone amplifier is 30 dB.
MICFB	19	AO	Microphone amplifier feedback output.
EXTMIC	20	Al	External microphone input. The level should be 100mVrms at 1kHz.
XRESET	21	ı	Master reset. Active low.
SCL	22	l	Serial interface clock input. The data is transferred in both directions at the rising edge of this signal.
STB	23	I	Serial interface strobe signal. With strobe signal the MAS9191 stores the given address from the serial interface buffer and enters the data mode. The serial interface stays in the data mode until eight SCL pulses are received after the strobe signal.
STxD	24	0	Serial interface transmit data output.
SRxD	25	1	Serial interface receive data input
BUZFB	26	Al	Buzzer feedback is the input for the buzzer driver.
BUZOUT	27	AO	Buzzer output.
EXTERP	28	AO	Output for external accessories.
EARP2	29	AO	Earpiece differential outputs of earpiece amplifier. The outputs are capable of driving a
EARP1	30	AO	ceramic earpiece directly.
SIDEFB	31	AO	Side tone feedback output
SIDETONE	32	AI	Side tone input. The level of the side tone is controlled with external components.
EINR	33	AI	External RX input.
STGT	34	Al	Steering control input for DTMF receiver. When the level at this input changes from below $V_{DD}/2$ to above $V_{DD}/2$ the pin is pulled up internally. When this occurs the DTMF tone is stored and an interrupt is generated.
EST	35	AO	Enable Steering output. This pin is high when the DTMF receiver has detected a valid DTMF tone.
RAUDIN	36	Al	Input for filter 6. Connect through a 22nF capacitor to the expander output (EXPOUT).



PIN DESCRIPTION

Pin name	Pin	Туре	Function
RXACCOUT	37	AO	Output of uncommitted Op Amp in MAS9191. The Op Amp is normally used for RX audio level detection. The application circuit for this function is in the APPLICATIONS section. Connect the level detected by the circuit to the A/D converter of the general purpose micro controller.
RXACCIN	38	Al	RX block extra Op Amp input
EXPOUT	39	AO	Expander output. The expander ratio is 1:2.
EWCIN	40	Al	Expander window comparator input. Connect a 22nF capacitor between EWCIN and EAMPOUT.
EAMPOUT	41	AO	Expander amplifier output.
EXPIN	42	Al	Expander input. Connect a 22nF capacitor between EXPIN and RBPFOUT.
VDAR	43	Р	Power supply for RX audio block. Use a bypass capacitor between VDAR and VSAR.
VSAR	44	G	Ground for RX block. Connect to system ground.
RBPFOUT	45	AO	RX bandpass filter output.
RBPFIN	46	AI	RX bandpass filter input. Connect a 22nF capacitor between this pin and DEOUT
DEOUT	47	AO	RX de-emphasis filter output. The filter has a -6dB/octave (±1dB) frequency response in the range 300Hz3kHz.
VREF	48	AO	Reference voltage. Connect a capacitor between this pin and system ground.
RX	49	Al	RX input from RF. This level is 100mVrms at 1kHz.
ALP	50	Al	Audio loop input. Connect through a 22nF capacitor to the TX pin.
TEST2	51	I	Test input. Connect to ground during normal operation.
TEST1	52	I	Test input. Connect to ground during normal operation. If connected to V _{DD} and TEST2 is connected to ground, the external clock can then be connected to XTAL1.
CLKOUT	53	0	4.8 MHz clock output from oscillator circuit.
TXCTRL	54	AO	Transmission control output. If a TX collision occurs this open-collector output is set to low. The TXCTRL will remain low until the TX block is reset with the TXRST bit or with XRESET.
Vss	55	G	Digital ground. Connect a bypass capacitor between Vss and Vpp.
XTAL2	56	0	Crystal oscillator output.
XTAL1	57	l	Crystal oscillator input or external clock input if TEST1 is high and TEST2 is low.
VDD	58	Р	Power supply input for digital block.
TXON	59	0	Transmission detection for debugging. This output indicates when a transmission is occurring.
BUSY	60	0	Busy/Idle output. Indicates the state of the busy/idle bit.
XINT	61	0	Active low interrupt output to micro controller. The interrupt is active until status register 10_{HEX} is read.
DACOUT3	62	AO	Output of DAC 3. The DAC output is connected to ground if the DAC is in power down mode. The output of the DAC is controlled by register 18 _{HEX} . Enter the values in two's complement form into the DAC register.
DACOUT2	63	AO	Output of the DAC 2. The control register is located at 17 _{HEX} .
DACOUT1	64	AO	Output of the DAC 1. The control register is located at 16 _{HEX} .



ABSOLUTE MAXIMUM RATINGS

(Gnd = 0V)

Parameter	Symbol	Conditions	Min	Max	Unit
Supply Voltage	V _{oo}			6.0	٧
Storage Temperature	Ts		-55	+125	°C

RECOMMENDED OPERATION CONDITIONS

Parameter	Symbol	Conditions	Min	Тур	Max	Unit
Supply Voltage	V _{DD}	Ta=-4085°C	3.0	3.3	5.25	٧
Supply current	l _{od}	Ta=-4085°C, V _{DD} =3.3V±5%	1.0	2.5	25	mA
Operating temperature	Та		-40		+85	°C

ELECTRICAL CHARACTERISTICS

◆ Digital inputs

(Ta=-40...85°C)

Parameter	Symbol	Conditions	Min	Тур	Max	Unit
Input high voltage	V _{IH}	V _{DO} =5V±5%	2			٧
		V ₀₀ =3.3V±5%	1.4			
Input low voltage	V _{IL}	V ₀₀ =5V±5%			0.8	V
		V _{DD} =3.3V±5%			0.4	
Input leakage current	I _{IL}				±10	μA
Input capacitance load	C _i				1	рF

◆ Digital outputs

(Ta=-40...85°C)

Parameter	Symbol	Conditions	Min	Тур	Max	Unit
Output low voltage	V _{oL}	V _{DD} =5V±5%, +1.8mA	2			V
		V _{DD} =3.3V±5%, -4.0mA	1.4			٧
Output high voltage	V _{OH}	V _{DD} =5V±5%, +0.6mA			0.8	V
		V _{DD} =3.3V±5%, -2,0mA			0.4	V



Analog inputs

(Ta=-40...85°C,f=1kHz)

Parameter	Symbol	Conditions	Min	Тур	Max	Unit
External microphone level	V _{EXTMIC}			100		mV _{rms}
Microphone level	V _{MIC}			10		mV _{m∗}
RXIN input level	V _{RXIN}			100		mV _{rms}

◆ Analog outputs

(Ta=-40...85°C,f=1kHz, default gains)

Parameter	Symbol	Conditions	Min	Тур	Max	Unit
Signal ground	AGND		V _{DD} /2- 0.1V	!	V _{DD} /2+0. 1V	٧
Reference voltage	V _{ref}			AGND+ 1.2V		٧
Earpiece output impedance	z _o				0.5	kΩ
Earpiece load resistance	R _L		1			kΩ
Earpiece series load capacitance	C _L				120	nF
External earpiece load resistance	R _L		30			kΩ
External earpiece load capacitance	CL				1	nF
Earpiece amplifier gain	A _{vol}			-4.26		dВ
External earpiece output	Vo	RX level 100mV _{rms}		200		
Earpiece level non-differential	V _o	RX level 100mV _{rms}		70		
Earpiece level differential	V _o	RX level 100mV _{rms}		155		mV _{rms}
TX level	V _o	EXTMIC level 100 mV _{rms}		200		
DTMF signatures at TV	.,	f = 697 941 Hz	83		176	
DTMF signal levels at TX	v _o	f = 1209 1633Hz	180		385	mV _{rms}
		ETACS		148		
SAT signal level at TX	v _o	AMPS		138		
DATA signal level at TX	V _o			556		
ST signal level at TX	v _o			556		
		NAMPS, DTX mode off		93		
DATA, DST, DSAT level at TX	V _o	NAMPS, DTX mode on		373		
DACs, Output level.	V _o		0.3V		VDD-0.3	٧
DACs, differential nonlinearity	DNL			±0.5		LSB
DACs, integral nonlinearity	INL			±2		LSB
DACs, Settling time		Vdac ±1%			10	ms
DACs, Load resistance	R _L		30			kΩ
DACs, Load capacitance	C _L				80	pF
OPAMPs, Load capacitance	C _L	2.00			1	nF



◆ Expander

(Ta=-40...85°C)

Parameter	Conditions	Min	Тур	Max	Unit
Expander ratio			1:2		
Operation range input		-24		+10	dB
Operation range output		-48		+20	dB
Gain step			1.333		dB
Integral nonlinearity		-1		+1	dB
Attack time		7.4	9.2	14.3	ms
Decay time		9.5	11.9	14.3	ms

♦ Compressor

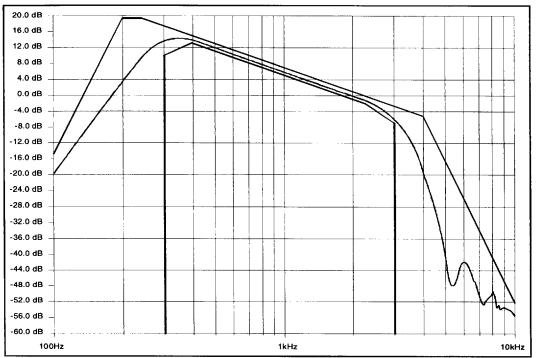
(Ta=-40...85°C)

Parameter	Conditions	Min	Тур	Max	Unit
Compressing ratio			2:1		
Operation range input		-39.4		+20	dB
Operation range output		-19.7		+10	dB
Gain step			1.333		dB
Integral nonlinearity		-0.5		+0.5	dB
Attack time		2.9	3.9	4.6	ms
Decay time		13	16.9	20	ms

♦ AC-characteristics

(Ta=~40...85°C)

Parameter	Min	Тур	Max	Unit
RX S/N ratio		50		dB
TX S/N ratio		50		dB
RX THD		0.1		%
TX THD		0.3		%
Crosstalk RX to TX		50		dB
Crosstalk TX to RX		50		dB
Mute attenuation		50		dB

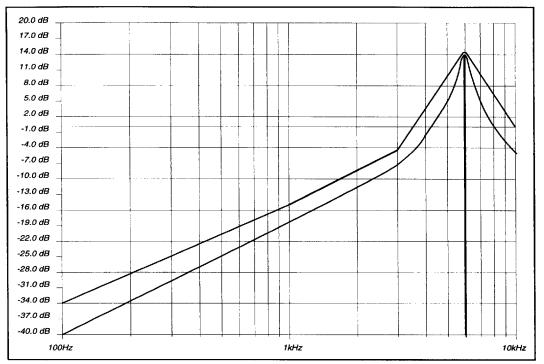


RX Total frequency response



TX Total frequency response





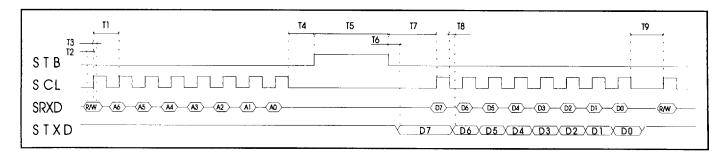
Filter F2 frequency response



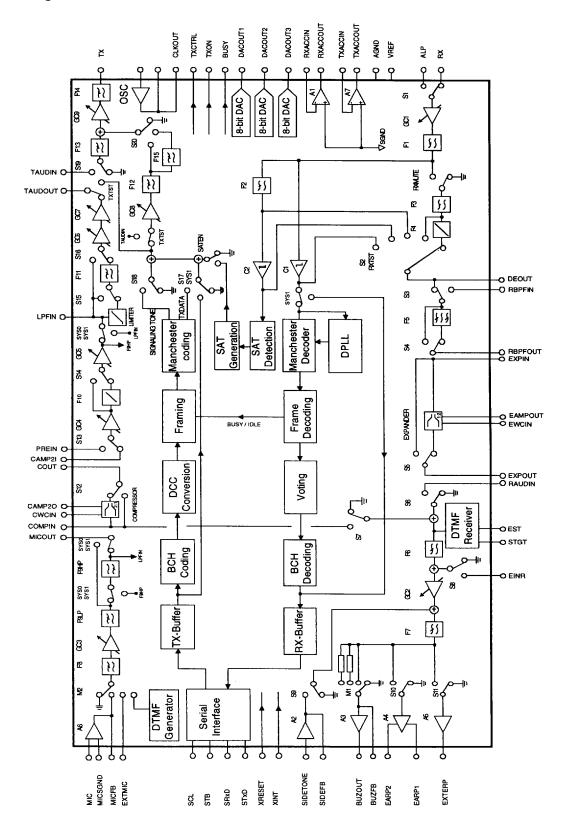
◆ Timing

(Ta=-40...85°C)

Parameter	Symbol	Condition	Min	Тур	Max	Unit
SCL cycle	Т1		0.5 (2 MHz)			μs
Data setup time	T2		60			ns
Data hold time	ТЗ		20			ns
STB rising edge after SCL falling edge	Т4		0			ns
STB width	Т5		5			μs
MSB data bit valid after STB falling edge	Т6	Register read			5	μs
SCL rising edge after STB falling	77	Register read	5			
edge		Register write	0			μs
Next data bit valid after SCL falling edge	Т8		30			ns
Ready for next address	Т9		7			μs



◆ Functional Diagram



Data Reception

The data reception block is in the power down state after a reset. The power down mode of the block is controlled by bit RXSIP (RX section in power down) in register 07_{HEX}The Manchester encoded data is received through the RX pin. The data is amplified with GC1 and filtered with filter F1. The comparator C1 is used to convert data to a digital signal. The digital PLL circuit recovers the bit clock from the Manchester encoded data. The bit clock is 8kHz in the ETACS mode and 10kHz in the AMPS mode. The mode is set with the SYS0 bit of register 12_{HEX}. The recovered bit clock is used in the Manchester decoder and the data is then transmitted to the frame decoding block. The frame decoding block finds dotting sequences, busy/idle bits and

word syncs from the data. To avoid data being generated by random noise, the frame decoding block enters the data reception mode only after it has received two consecutive word syncs (11100010010) separated by 463 bits in the forward control channel and 77 bits in the forward voice channel. In this case the voting and BCH block are activated. When the frame decoding block loses five consecutive synchronization patterns it rejects the data reception mode and sets the voting and BCH blocks in power down state. The voice and control channel modes are selected with the CTCV bit of register 12_{HEX}.

	Bit Sync	Word sync	Busy/Idle	Data	Busy/Idle	Data	Busy/Idle	Data	Busy/Idle	Data
	10	11	1	10	1	10	1	10	1	10
			1. Repeat o	f word A						
Busy/Idle	Data	Busy/Idle	Data	Busy/Idle	Data	Busy/Idle	Data	Busy/Idle	Data	-
1	10	1	10	1	10	1	10	1	10	L
1. Repeat	of word B							2.Repeat	of Word A	
	Busy/Idle	Data	Busy/Idle	Data	Busy/Idle	Data	Busy/Idle	Bit Sync		•
	. 1	10	1	10	1	10	1	10		
	5. Repeat	of Mord D								

bit sync=1010....01

The busy/idle bits are extracted from the data. Busy/idle bits are used to indicate the current status of the forward control channel. The forward control channel is busy if the busy/idle bit is low and idle if the busy/idle bit is high. The state determination is made with 2-out-of-3

voting. The TX block uses the busy/idle indication for arbitration. The STR bit of register 12_{HEX} selects the words from stream A or stream B.

Forward voice channel data format:

		1	1		1			
BIT SYNC	WORD SYNC	1.REPEAT OF WORD	BIT SYNC	WORD SYNC	2. REPEAT OF WORD	SYNC	WORD	3. REPEAT OF WORD
101	11	40	37	11	40	37	11	40
BIT SYNC	WORD SYNC	4.REPEAT OF WORD	BIT SYNC	WORD SYNC		BIT SYNC	WORD SYNC	11. REPEAT OF WORD
37	11	40	37	11		37	11	40

On the forward voice channel after bit synchronisation and word synchronisation are received and the AUMUT bit in register 13_{HEX} is set to high, the RX audio block will be muted until the 920 bits are received. The repeated words are transferred to the voting block. The voting is done bit by bit, 3-out-of-5. After the voting block the data is transmitted to the BCH (Bose-Chaudhuri-Hocquenghem Code) block. This block performs decoding of the received BCH coded data. The following polynomial is used: $P(x) = x^{12} + x^{10} + x^8 + x^5 + x^4 + x^3 + x^0$

$$P(x) = x^{12} + x^{10} + x^{8} + x^{5} + x^{4} + x^{3} + x^{0}$$

If only one error occurs in data the BCH block can correct it. If more

than one error occurs the BCH block cannot correct them and the BCHER bit of register 11_{HEX} is set to high. The BCH block transfers the data to the RX buffer. When the buffer is full the RXWRD bit of register 10_{HEX} is set to high and this causes interrupt line XINT to go active. When register 10_{HEX} is read the interrupt is cleared. The data buffer can be read by reading register 15_{HEX} four times. If the next word is coming and the previous word has not been read from the buffer, the word is missed.

Data Transmission

After a reset the TX block is in power down. The power down mode is controlled by bit TXSIP (TX section in power down) in register 07_{HEX} The TXRST bit located in register 12 HEX is used every time a TX collision occurs or for any other TX block reset causes. When the device is ready to receive data, the TXWRD bit of register 10_{HEX} is high. If five bytes are written into register 19_{HEX} the data transmission begins. The data is transferred from the serial interface to the TX buffer and the TXWRD bit is set high again, which causes XINT to become active. When the block comes out of power down mode the XINT is active

because the device is ready to receive data (TXWRD goes high). The lower nibble of the fifth byte is ignored. The 36 bits of data are coded by the BCH coder with following polynomial: $P(x) = x^{12} + x^{10} + x^{8} + x^{5} + x^{4} + x^{3} + x^{0}$

The BCH coder adds 12 parity bits to the data and the data is transferred to the DCC coding block. The DCC coder adds a digital color code on the forward control channel according following table:

DCC(1:0)	Coded DCC
00	0000000
01	0011111
10	1100011
11	1111100

The framing block adds bit sync (101010...101) or word sync (11100010010) sequences to the frames and performs needed repeats depending on the mode.

Reverse control channel format: (The numbers below the boxes show the number of bits in each section.)

BIT SYNC	WORD SYNC	CODED DCC	FIRST WORD REPEATED FIVE TIMES	SECOND WORD REPEATED FIVE TIMES	
30	11	7	240	240	

37

voice char	nei format:			····		_	
BIT SYNC	WORD SYNC	1.REPEAT OF WORD 1	BIT SYNC	WORD SYNC	2. REPEAT OF WORD 1	BIT SYNC	WORD SYNC
101	11	48	37	11	48	37	11
BIT SYNC	WORD SYNC	5.REPEAT OF WORD 1	BIT SYNC	WORD SYNC	1. REPEAT OF WORD 2	BIT SYNC	WORD SYNC
37	11	48	37	11	48	37	11
BIT SYNC	WORD SYNC	5.REPEAT OF WORD 2					
37	11	48					

The Manchester coder block encodes the data into a Manchester coded format with bit clock. The bit clock is 8kHz in the ETACS mode and 10kHz in the AMPS mode. The mode can be controlled by bit SYS0 of register $12_{\rm HEX}$. The data polarity can be inverted with bit INVTX of register $13_{\rm HEX}$.If the BUSY bit does not go active between 56 and 104 bits of the transmitted message a transmission collision occurs. In this case the data which is in the TX block and the data that the user is writing to the device will not be transmitted. The TXCOL bit of register

11

10_{HEX} will go high in this case and cause an interrupt. The TXCOL will remain active until the TX block is reset with the TXRST bit of register 12_{HEX}. If the TXCTREN bit is active in register 12_{HEX} the TXCTRL output turns the transmitter off when a TX collision occurs. If the AUMUT bit in register 13_{HEX} is set to high the TX audio block is muted with switch S19 on the voice channel while data transmission is occurring.

◆ Data reception in narrow band mode

Forward voice channel data format for narrow band:

			<u>-</u> 1
DSAT	SYNC WORD	DATA	DSAT
	20	40	

SYNC WORD= 011001010110101001100110100110

DSAT = Digital Supervisory Audio Tone is one of seven 24-bit digital sequences added to the voice transmission. DSAT is transmitted at 200 NRZ bits/second. The following is a list of the seven DSAT sequences:

	T
	DSAT sequence
0	2556CB _{HEX}
1	255B2B _{HEX}
2	256A9B _{HEX}
3	25AD4D _{HEX}
4	26AB2B _{HEX}
5	26B2AD _{HEX}
6	2669AB _{HEX}

The 40-bit long data sequence is generated at a 100 Manchester bits/second rate. The data sequence contains 28 bits of data and 12 parity bits.

The incoming data is captured by two 8-bit shift registers. When one shift register is full the RXWRDP flag is set and an interrupt is generated. The captured data must be read by the micro controller within 20ms after the interrupt. Meanwhile, the other shift register is being filled and when it is full a new interrupt is generated. The shift registers are clocked in at 400 Hz. Two samples of each state of both the 200 NRZ bits/second data and the 100 Manchester bits/second data are loaded into the registers. Note that there are as many

transitions in the 200 NRZ bits/second data as in the 100 Manchester bits/second data. The micro controller will then be used to filter the digital bit sequence and detect the DSAT, SYNC WORD and DATA out of the bitstream. The DATA must be checked with following algorithm:

$$P(x) = x^{12} + x^{10} + x^8 + x^5 + x^4 + x^3 + x^0$$

◆ Data transmission reverse narrow analog voice channel

Reverse voice channel data format for narrow band:

DSAT/DST	SYNC WORD	DATA	DSAT/DST
	30	48	

SYNC WORD= 01100101011010011001100110

The data contains 36 data bits and 12 parity bits. The transmitted data is 100 bits/sec Manchester code. DSAT, DST and SYNC WORD are transmitted as 200 bits/sec NRZ code. The DSAT on the TX side is similar to the DSAT on RX side. However, under certain conditions the inverted DSAT, or DST (Digital Signalling Tone), must be transmitted. The DST is one of seven 24-bit digital sequences consisting of the logical inverse of the seven DSAT sequences. The conversions

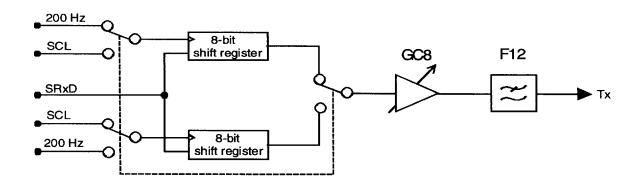
DSAT/ DST and DST/DSAT must be made without disturbing the phase of the DSAT. There is also a special 24-bit digital mask for each of the seven sequences. The mask defines the first bit to be inverted when converting from DSAT to DST or vice versa. Only when the bit in the mask is one can the polarity be changed.

	DSAT	DST	MASK
0	2556CB _{HEX}	DAA934 _{HEX}	FF003E _{HEX}
1	255B2B _{HEX}	DAA4D4 _{HEX}	0BBF82 _{HEX}
2	256A9B _{HEX}	DA9564 _{HEX}	BD780F _{HEX}
3	25AD4D _{HEX}	DA52B2 _{HEX}	3FF118 _{HEX}
4	26AB2B _{HEX}	D954D4 _{HEX}	0AE6F6 _{HEX}
5	26B2AD _{HEX}	D94D52 _{HEX}	8001FF _{HEX}
6	2669AB _{HEX}	D69654 _{HEX}	1C0FCD _{HEX}

MAS9191A does not include frame coding logic for narrow band operation. The DSAT, DST, SYNC WORD and DATA must be generated by micro controller. The BCH function must also be performed by the micro controller using the following algorithm:

$$P(x) = x^{12} + x^{10} + x^8 + x^5 + x^4 + x^3 + x^0$$

The generated bit sequence is written into shift register $19_{\rm HEX}$ 8 bits at a time. While the next byte is written to one of the 8-bit shift registers the other is clocked out with a 200Hz clock. Each time the contents of a shift register transmitted, the TWRDP flag is set and an interrupt is generated. Note that 200 NRZ bits/second data has as many transitions as 100 Manchester bits/second data.



◆ SAT detection & regeneration

SAT detection is active on voice channel in AMPS or TACS mode (SYS1=0). The supervisory audio tone is detected with a digital PLL. The detector compares the received SAT to the given SAT color code (SCC). When the given SAT is detected the SATDET bit of register $11_{\rm HEX}$ is set to high. If SATINTEN bit of register $13_{\rm HEX}$ is on the rising and falling edge of SATDET will cause the interrupt SATINT. On the voice channel the SAT regeneration can be enabled with bit

SATEN of register . By setting bit NOMSAT of register $14_{\rm HEX}$ nominal SAT frequency is generated. Otherwise the SAT output frequency will follow received SAT frequency.

The block is in power down mode when the TXSIP (transmit section in power down) bit is active in register $07_{\rm HEX}$. After a reset the block is in power down mode.

SCC1	SCC0	SAT frequency
0	0	5970 Hz
0	1	6000 Hz
1	0	6030 Hz
1	1	no valid

♦ SAT, ST or Data transmission

The SAT and signalling tone (ST) are sent only on the voice channel in AMPS or TACS mode (SYS1=0). The SATEN and STON can be used to control the SAT and ST transmission. However, the devlce will automatically stop transmitting SAT and ST signals whenever data is being transmitted, even though SATEN or STON is high. The signalling tone is 8kHz in ETACS and 10kHz in AMPS. Also, switch S17 for TX data and switch S18 for ST can be used to disable the transmission. The switches must be on during transmission. The control bits for these switches are located in register 0D_{HEX}. The SAT, TXDATA and the ST are summed and amplified with GC8. The gain of the amplifier is -3.75dB...+3.75dB with 16 steps. The adjustment is

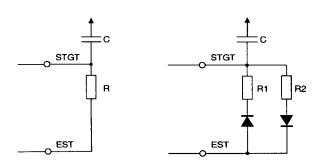
made with bits 0..3 of register 0D $_{\rm HEX}$. After being amplified the signal is filtered with 4th order SC filter F12. The cutoff frequency of the filter is 19kHz in AMPS, 15kHz in ETACS and 200Hz in NAMPS (SYS1=1). The gain of the filter at 1kHz is 0dB. In the NAMPS mode, the signal is then fed to lowpass filter F15, which is a 2nd order RC type filter. After filtering the signal is summed to the TX audio signal depending on the state of switch S20. The switch is controlled by bit 6 of register 01 $_{\rm HEX}$. Register 07 $_{\rm HEX}$ bit AUDIOP is used to set these blocks into power down mode.

◆ DTMF Receiver

For enabling answering machine functions, the chip has an internal DTMF receiver. The receiver is in power down mode after a reset. The DTMFRP bit of register 07_{HEX} controls the receiver power down mode. The receiver has two separate filters for separation of the low and high frequencies. The comparator and logic section measures the low and high frequency periods with an averaging algorithm. When the valid DTMF tone is detected the external steering logic output pin EST is set to high. With an external RC time constant the tone detect time and tone dropout times can be adjusted. The STGT input/output pin has an internal comparator and pull-up and pull-down transistors. When EST is active and STGT goes from below to above the Vref level (VDD/2) the STGT is pulled up with an internal transistor. This causes the STD signal to go high, which causes the XINT line to become active. At the same time the detected DTMF tone is stored in register 04_{HEX}. When

the DTMF tone is not present the EST will go low, which causes the STGT to fall. When the STGT falls below the Vref level (VDD/2) the internal logic pulls the input down. The external RC circuit will filter out very short gaps in the received DTMF tone. The interrupt caused by the DTMF detector is cleared by reading status register $10_{\rm HEX}$. If the interrupt is cleared but register $04_{\rm HEX}$ is not read until the next DTMF tone is received, then the previous tone will be lost.

The formula below can be used to calculate tone present and tone absent times. By adding diodes to the external circuit the tone present and tone dropout times can be altered. If one of the diodes is removed, the absent and present times are calculated using the parallel combination of R1 and R2.



$$Time = R^*C ln(\frac{V_{DD}}{V_{REF}})$$

◆ RX Audio

The RX audio block starts with switch S1. The switch is controlled by bit S1 of register $0E_{\text{HEX}}$. The input ALP is used for enabling the audio loopback mode. In normal operation RX is used. Behind the switch is amplifier GC1 with adjustable gain from -3dB to +3dB. The gain is controlled with bits 0..3 in register $0E_{\text{HEX}}$. The amplifier output is connected to the second order lowpass filter F1. The cutoff frequency of the filter is 50 kHz and the gain at 1kHz is 0dB. The filter output is connected to data comparator C1 and to the SAT bandpass filter F2 and to the rest of the TX audio block. The polarity of the received data can be inverted with bit INVRX, which is located in register 13_{HEX} . The data comparator output is connected to the DPLL and Manchester decoder blocks. The filter F2 is a 6kHz bandpass filter for supervisory audio tone. The filter is a second order SC filter. The filter output is connected to SAT comparator C2 and the SAT detection block.

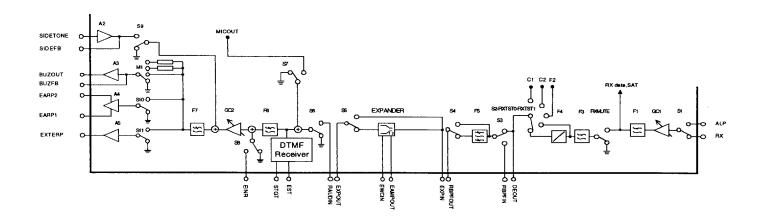
The signal from F1 is connected to filter F3 through a switch which is used by the internal logic when the AUMUT bit of register $13_{\rm HEX}$ is active. In this case if the data is received on the voice channel the lowpass filter F3 input is grounded automatically with this switch. RX de-emphasis filter F4 has a -6dB/octave (±1dB) frequency response in the range 300Hz...3kHz. The filter can be bypassed with S2-RXTST0-RXTST1. Switch S2-RXTST0-RXTST1 is connected to the DEOUT pin and to switch S3. The performance of C1, C2 and F2 can be monitored with switch S2-RXTST0-RXTST1. Bits 5 and 6 in register 02 $_{\rm HEX}$ and bit 5 in register 0E $_{\rm HEX}$ control this switch. An external capacitor is needed between DEOUT and RBPFIN. The RBPFIN input is connected to filter F5, which is a 6th order bandpass filter. The gain of this filter is 0dB at 1kHz. The filter can be bypassed with switch S4. After S4 the signal is connected to the RBPFOUT pin. An external capacitor is used to connect the signal to the EXPIN input.

The signal is fed to the SC-type audio expander with expansion ratio 1:2. The expander can also be bypassed with switch S5. The switch is controlled with bit S5 in register $03_{\rm HEX}$. After switch S5 the signal goes to the EXPOUT output pin. An external capacitor is used to connect the signal to the RAUDIN input. The input is connected to switch S6. With

switches S6 and S7 the received audio and transmitted audio can be summed. The control bits of the switches are located in register $03_{\rm HEX}$.

The summed signal goes through filter F6 and to the DTMF receiver. The function of the DTMF receiver is described in the next section. Filter F6 is a 4th order SC-type lowpass filter. The gain at 1kHz is 0dB. After filtering the signal and the external accessory input EINR can be summed with switch S8. The control bit for the switch is in register $03_{\rm HEX}$. The summed signal is amplified with GC2, which has a $-15{\rm dB...}+15{\rm dB}$ gain with 16 steps. The gain is controlled with bits 0..3 in register $03_{\rm HEX}$. The amplified signal can be summed with SIDETONE. The side tone can be switched on with S9, which has a control bit in register $02_{\rm HEX}$. The side tone input has an internal Op Amp with feedback signal SIDEFB. The gain of the Op Amp is controlled by external components. COMPIN is normally used as an input for the side tone amplifier circuit.

Filter F7 is a second order low pass filter. The cutoff frequency is 20kHz and the gain at 1kHz is 0dB. After filtering the signal can be connected to three amplifiers with switches M1, S10 and S11. The control bits are located in registers 02_{HEX} and 03_{HEX}. A4 is the earphone amplifier, which is a single input differential output amplifier. Amplifier A5 is for external accessories and is capable of driving a capacitive load. The load capacitance is 1nF and the block has a 4.82dB gain. The third amplifier A3 is a buzzer driver. It drives the signal to the power transistor, which drives the buzzer. The buzzer represents a high inductive load of 1.2mH/25ohm. The block stabilizes the current flow through the external buzzer which depends on the current gain factor of the external bipolar transistor. The emitter resistor of the transistor must be 6.8 ohms. Three current values (peak values) can be chosen with switch M1:10mA, 66mA and 160mA. The tolerance of the external resistor will directly affect the buzzer current. The RX audio block can be set into power down mode together with the TX audio block. The AUDIOP bit in Register 07_{HEX} is used for this purpose. The blocks are in power down mode after a reset.



◆ TX Audio

The TX audio block is in the power down mode after a reset. The block is set to the operation mode with the AUDIOP bit of register $07_{\rm HEX}$. The same bit also controls the RX audio block. The TX audio block starts with switch M2, which is controlled by bits 4 and 5 in register $00_{\rm HEX}$. Switch M2 connects one of the following blocks as an input source: 1) After a reset the input source is connected to signal ground. 2) In the second position the input is connected to the microphone amplifier A6. The gain of amplifier A6 is determined with external components. The maximum gain is 30 dB. 3) In the third position the input is connected to EXTMIC, which is for external accessories. 4) The fourth position selects the DTMF generator. The DTMF generator is controlled with registers $05_{\rm HEX}$ and $06_{\rm HEX}$. Register $05_{\rm HEX}$ controls the low frequencies and register $06_{\rm HEX}$ controls the high frequencies. A detailed description of how to use these registers is in the REGISTERS section. The DTMF generator is in the power down mode after a reset and can be set up with the DTMFTP bit of register $07_{\rm HEX}$.

The signal is then filtered with anti-aliasing filter F8. The gain at 1kHz is 0dB and the cutoff frequency is 15kHz. After the filter the signal is amplified with GC3. The gain is -3dB...+3dB with 16 steps according to bits 0..3 in register 00_{HEX} . The amplified signal is then fed to lowpass filter F9LP and highpass filter F9HP, which are 4th order SC filters. During normal operation the output of filter F9HP is connected to output MICOUT. With switch SYS0-SYS1 the highpass filter can be transferred to the output of GC5 after the signal has been compressed and pre-emphasized. Bit 6 in register 12 $_{\rm HEX}$ and bit 6 in register 14 $_{\rm HEX}$ control switch SYS0-SYS1. Connect MICOUT through a 22nF capacitor to the compressor input COMPIN. The MICOUT can also be used as a source for the side tone amplifier and for detecting the audio level with the uncommitted Op Amp (See APPLICATIONS section). The compressor is an SC audio type with a 2:1 compressing ratio. The detailed values are found in the ELECTRICAL CHARACTERISTICS section under Compressor. The compressor can be bypassed internally with switch S12 or externally with S13. The bypass gain is 0dB (100 mV_{rms}). The compressor requires one external 22nF capacitor between pins CAMP20 and CWCIN. Another 22nF capacitor is needed between pins CAMP2IN and COUT. This capacitor also serves as an external DC blocking capacitor between the compressor output and gain control GC4 input.

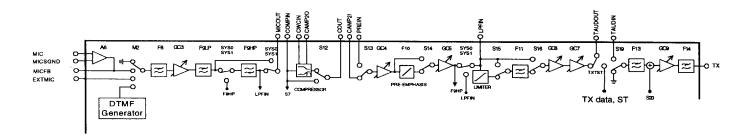
The gain of GC4 can be adjusted in the range -0.67..+5.33dB with bits 0..3 in register 0A_{HEX}. After GC4 the signal is filtered with pre-emphasis

filter F10, which has a +6dB/octave (\pm 1dB) frequency response at the range 300Hz..3kHz. At 1kHz the gain is 0dB. The pre-emphasis filter can be bypassed with switch S14.The signal is then amplified with GC5. Gain control GC5 is an SC-type with a programmable gain adjustment. The range is 0dB..-20 dB with 16 steps according to bits 0..3 in register 0A_{HEX}.

At the output of GC5 is output pin LPFIN (or highpass filter F9HP) and an SC-type limiter. The limiting level is 439mV_{pp} and the tolerance is $\pm5\%$. The 0dB level is 370mV_{pp} . The limiter can be bypassed with switch S15. The lowpass filter F11 with a 6kHz notch follows the limiter. The gain of the filter is 2.94dB at 1kHz. The filter and the limiter can be bypassed with switch S16. GC6 and GC7 are continuous time programmable gain adjustment blocks. The gain of GC6 is - 3dB...+3dB with 16 step according to bits 0..3 in register 0B_{HEX}. The gain of GC7 is adjustable within - 3dB...+1dB with bits 0..3 in register 0B_{HEX} in 16 steps.

GC7 output is connected to pin TAUDOUT through switch TXTST. The data transmission signal can be examined at TAUDOUT with this switch. Also with switch TXTST, the performance of GC8, -3.75..3.75dB., F12 and F15 can be observed with an input at TAUDIN. An external capacitor between pins TADOUT and TAUDIN is required. The TAUDIN input is connected to switch S19. The switch can be used to mute the TX audio block. S19 is controlled with bit 5 in register 01_{HEX}. Switch S19 is also controlled by the TX framing block. During data transmission on the voice channel and with the AUMUT bit in register 13_{HEX} set to high the TX audio block is muted with S19. Following S19 is a second order lowpass filter F13 with 15kHz cutoff frequency. After filtering the signal is summed with the data signal and amplified with GC9. The gain of the amplifier GC9 is adjustable with bits 0..3 located in register 0C_{HEX}. The adjustment is done in 16 steps in the range 3.0.. +3.0dB .The output of the amplifier GC9 is then filtered with F14. The filter is a third order lowpass filter with 54kHz cutoff frequency. The gain at 1kHz is 0dB.

DTX bit of the regiter 0D $_{\rm HEX}$ is used to set DATA,DSAT and DST deviation. When the device is not in DTX (Discontinous Transmission) the DTX = 0 and there is nominal level at TX, 93 mV $_{\rm rms}$. If DTX is on, the level is increased to 373 mV $_{\rm rms}$.



◆ DACs

The device has three 8-bit DACs. The DACs can be set to the power down mode by using bits XPDDAC1, XPDDAC2 and XPDDAC3. The DACs are in power down mode after a reset. The Vref for the DACs is VDD/2. The output values of the DACs are entered in 8-bit two's complement form into registers $16_{\rm HEX}$, $17_{\rm HEX}$ and $18_{\rm HEX}$. The typical step size is 13 mV and the DC output level is in the range 0.3V..VDD-0.3V.

The differential nonlinearity is ± 0.5 LSB and the integral ± 2 LSB. The settling time is 10ms (max). The minimum load resistance is 30k and the maximum load capacitance is 80pF.

◆ Op Amps

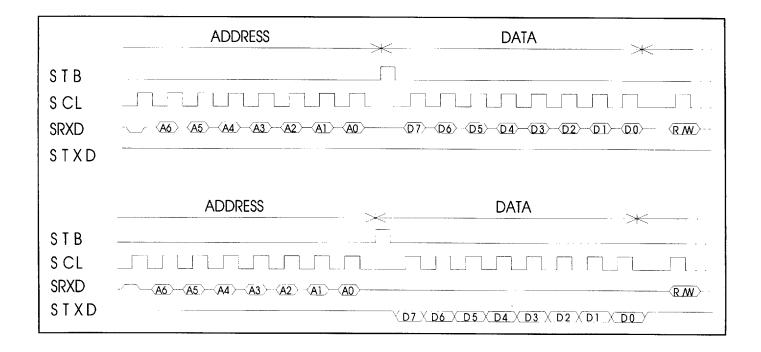
There are two uncommitted inverting operational amplifiers in the device. The input pins are RXACCIN and TXACCIN. The output pins are RXACCOUT and TXACCOUT. The Op Amps are capable of driving

capacitive loads up to 1nF.

Serial interface

The serial interface has three inputs: SCL (serial clock), STB (strobe) and SRxD(received data). Output pin STxD is used for transmitting data. The data is latched with the rising edge of the SCL signal. The MSB is received first and the LSB last. Before the STB signal, eight address bits must first be shifted in. The STB signal sets the device into the data mode. The MSB of the address byte defines the read/write operation. If the MSB is high the data is read from the device. If the MSB is low, the data is written to the device. After the STB the written data is shifted in with the rising edge of the SCL. If the data is to be

read from the device, the STxD output is in the state of MSB data bit after the STB signal. The falling edge of the SCL shifts the next data bit to the STxD output. Eight data bits must be shifted out at which time the device exits the data mode. Because the serial interface transmit buffer is dynamic, data will be valid on the buffer only 200 uS after the STB signal appears. This means that the SCL frequency must be at least 5kHz.





◆ Registers

The Register Map:

Address	1/0	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
00 _{HEX}	Write	0	0	M2[1:0]		GC	3(3:0)	
01 _{HEX}	Write	0	S20	S19	S16	S15	S14	S13	S12
02 _{HEX}	Write	0	RXTS	Τ[1:0]	S9	S11	S10	M1[1:0]
03 _{HEX}	Write	S8	S 7	S6	S5		GC	C2[3:0]	
04 _{HEX}	Write	0		V _{ref} [3:0]		0	0	0	0
(84 _{HEX})	Read	х	х	х	х		DT	MF[3:0]	
05 _{HEX}	Write	0	0			LO	WF[5:0]		
06 _{HEX}	Write			<u> </u>	HIG	HF[7:0]		·	
07 _{HEX}	Write	XPDDAC3	XPDDAC2	XPDDAC1	RXSIP	TXSIP	AUDIOP	DTMFRP	DTMFTP
OA _{HEX}	Write		GC4	[3:0]			G	C5[3:0]	
OB HEX	Write		GC7	[3:0]	· · · · · · · · · · · · · · · · · · ·		G	C6[3:0]	************
OC HEX	Write	0	0	0	TXTST		G	C9[3:0]	
OD HEX	Write	0	S18	S17	DTX	GC8[3:0]			
OE HEX	Write	S4	S3	S2	S1	GC1[3:0]			
10 _{HEX} (90 _{HEX})	Read	×	×	×	SATINT	TXCOL	TXWRD	RXWRD	STD
11 _{HEX} (11 _{HEX})	Read	×	BCHERR	BUSY	TXON	WSYNC	DOT	SATDET	х
12 _{HEX}	Write	SATEN	SYS0	STR	RXRST	TXRST	TXCTREN	стсч	STON
13 _{HEX}	Write	SATINTEN	STD	RXINTE	AUMUT	INVRX	INVTX	DCC	(1:0)
14 _{HEX}	Write	NOMSAT	SYS1	0	0	0	0	sco	(1:0)
15 HEX. SYS1 = 0	Read				187	F FIX[0:7]			
(95 _{HEX})					2ND	RX[8:15]			
					ЗRD	RX[16:23]			
			4TH[24:27]		0	0	0	0
15 _{HEX} , SYS1 = 1 (95 _{HEX})	Read		8-bits of	captured data	after data comp	arator with shif	t register clock	ed with 400Hz.	
16 _{HEX}	Write				Di	AC1[7:0]			
17 _{HEX}	Write				D	AC2[7:0]			
18 _{HEX}	Write				D	AC3[7:0]			
19 _{HEX} , SYS1 = 0	Write				18	T TX[0:7]			
					201	TX[8:15]			
					3RD	TX[16:23]			
1					4TH	TX[24:31]			
			5ТН Т	X[32:35]		0	0	0	0
19 HEX , SYS1 = 1	Write		8-bits	s of bit sequen	cy to be transm	itted with shift	regiter clocked	with 200Hz.	



Registers cont.

Register	O _{HEX} , write only				
ВІТ	NAME	STATE	Function		
3-0	GC3[3:0]	0000-1111	-3.0 dB +3.0 dB 0.4dB/step		
		0000	-3.0 dB		
		0001	-2.6 dB		
		0010	-2.2 dB		
		1111	+3.0 dB		
		1000	+0.2 dB default value		
5-4	M2[1:0]	00	AGND		
	1	01	A6-OUT	Connected to GC3 input	
		10	EXTMIC		
		11	DTMFGEN		
6	-	0	reserved for future use, set to 0.		
7	<u> </u>	0	reserved for future use, set to 0		

віт	NAME	STATE	FUNCTION		
0	S12	0	Compressor output	connected to COUT	
		1	Compressor input		
1	S13	0	COUT	connected to GC4 input	
		1	PREIN		
2	S14	0	F10 output	connected to GC5-IN	
		1	F10 input		
3	S15	o	Limiter output	connected to F11 input	
		1	Limiter input		
4	S16	o	F11 output	connected to GC6 input	
		1	Limiter input		
5	S19	0	AGND		
		1	TAUDIN	connected to F13 input	
6	S20	0 (SYS1 = 0)	AGND		
		1 (SYS1 = 0)	F12 output		
		0 (SYS1 = 1)	AGND	summed to GC9 input	
		1 (SYS1 = 1)	F15 output		
7	reserved	0	reserved for future use, set t	O ZARO	



Register (02 _{HEX} , write only			
віт	NAME	STATE	FUNCTION	
1-0	M2[1:0]	00	AGND	Connected to amplified AD input
		01	G0	Connected to amplifier A3 input
		10	G1	
		11	G2	
2	S10	0	AGND	Connected to amplifier A4 input
		1	Filter F7 output	
3	S11	0	AGND	Connected to amplifier A5 input
		1	Filter F7 output	
4	S9	0	AGND	Summed into F7 input
		1	amplifier A2 output	
6-5	RXTST[1:0]	00	determined by S2	Connected to DEOUT
		01	C1	
		10	F2	
	<u> </u>	11	C2	
7	reserved	0	Operating mode.	

Register 0	3 _{H∈X} , write only			
BIT	NAME	STATE	FUNCTION	
3-0	GC2[3:0]	0000-1111	-15.0 dB +15.0 dB gain of GC2 2.0 dB/step	
		0000	-15.0 dB	
		0001	-13.0 dB	
		0010	-11.0 dB	
		1111	+15.0 dB	
		1000	+1.0 dB default value	
4	S5	0	EXPOUT	Connected to EXPOUT
		1	EXPIN	
5	S6	٥	AGND	Ourseld late FO least
		1	RAUDIN	Summed into F6 input
6	S7	0	AGND	
		1	COMPIN	
7	S8	0	AGND	Summed into GC2 input
		1	EINR	



Register (04 _{HEX} , bits 7-4 write only, bits	3-0 read only	
BIT	NAME	STATE	FUNCTION
3-0	DTMF[3:0]	0001	'1' Tone detected
		0010	'2' Tone detected
		0011	'3' Tone detected
		0100	'4' Tone detected
		0101	'5' Tone detected
		0110	'6' Tone detected
		0111	'7' Tone detected
		1000	'8' Tone detected
		1001	'9' Tone detected
		1010	'0' Tone detected
		1011	*** Tone detected
		1100	'#' Tone detected
		1101	'A' Tone detected
		1110	'B' Tone detected
		1111	'C' Tone detected
		0000	'D' Tone detected
6-4	V _{REF} [2:0]	Internal refer	ence voltage adjustment.
		000 001 010 011 100 101 110 111	+0.75dB +0.50dB +0.25dB 0.00dB -0.25dB -0.50dB -0.75dB -1.00dB 0.00dB default value
7	For testing purposes only.	o	In operating mode set to 0.

віт	NAME	STATE	FUNCTION		
5-0	LOWF[5:0]	00H-3FH	Transmitted low frequency DTMF tone		
			Nominal	Real freq.	
		000000	No signal	No signal	
		100110	697	695.8 Hz	
		101010	770	769.0 Hz	
		101111	852	860.6 Hz	
		110011	941	933.8 Hz	
7-6	reserved	00			

The register value for other frequencies can be calculated with the formula:

$$LG[5:0] = \frac{[f_{out} * 2^{17}]}{2.4MHz}$$



Register 0	6 _{HEX} , write only			
ВІТ	NAME	STATE	FUNCTION	
7-0	HIGHF[7:0]	00H-FFH	Transmitted high frequency DTMF tone	
			Nominal	Real freq.
		00000000	No signal	No signal
		01000010	1209 Hz	1208.5 Hz
		01001001	1336 Hz	1336.7 Hz
		01010001	1477 Hz	1483.2 Hz
		01011001	1633 Hz	1629.6 Hz

The register value for other frequencies can be calculated with the formula:

$$HG[7:0] = \frac{[f_{out} * 2^{17}]}{2.4MHz}$$

BIT	NAME	STATE	FUNCTION
		SIAIE	FONCTION
0	DTMFTP	٥	DTMF transmitter in power down
		1	DTMF transmitter active
1	DTMFRP	0	DTMF receiver in power down
		1	DTMF receiver active
2	AUDIOP	0	AUDIO in power down
	:	1	AUDIO active
3	TXSIP	0	Digital TX section in power down
		1	Digital TX section active
4	RXSIP	0	Digital RX section in power down
		1	Digital RX section active
5	XPDDAC1	0	DAC1 in power down
		1	DAC1 active
6	XPDDAC2	0	DAC2 in power down
		1	DAC2 active
7	XPDDAC3	٥	DAC3 in power down
		1	DAC3 active

Register (DA _{HEX} , write only		
BIT	NAME	STATE	FUNCTION
3-0	GC5[3:0]	0000 - 1111	Control the gain of GC5 From 0 to -20 dB. One step \simeq -1.33 dB
		0000	0.00 dB
		0001	-1.33 dB
		0010	-2.67 dB
		0011	-4.00 dB
		1111	-20.00 dB
		0000	0.00 dB default value
7-4	GC4[3:0]	0000 - 1111	Control the gain of GC4. One step is 0.40 dB.
		0000	- 0.67 dB
		0001	- 0.27 dB
		0010	+0.13 dB
		0011	+0.53 dB
		1111	+5.33 dB
		1000	+2.53 dB default value

Register	OB _{HEX} , write only		
BIT	NAME	STATE	FUNCTION
3-0	GC6[3:0]	0000-1111	-3.0 dB +3.0 dB 0.4dB/step
		0000	-3.000 dB
		0001	-2.600 dB
		0010	-2.200 dB
		0011	-1.800 dB
		1111	+3.000 dB
		1000	+0.200 dB default value
7-4	GC7[3:0]	0000-1111	Control the gain of GC5 -3.000+1.000 dB 0.266 dB/step
		0000	-3.000 dB
	,	0001	-2.733 dB
		0010	-2.467 dB
		0011	-2.200 dB
		1111	+1.000 dB
		1011	-0.067 dB default value



Register	OC _{HEX} , write only		
віт	NAME	STATE	FUNCTION
3-0	GC9[3:0]	0000-1111	Control the gain of GC9, -3dB3dB 0.4dB/step
		0000	-3.000 dB
		0001	-2.600 dB
		0010	-2.200 dB
		0011	-1.800 dB
		1111	3.000 dB
		1000	0.200 dB default value
4	тхтѕт	0	GC7 output connected to TAUDOUT ST, SAT and TXDATA connected to GC8 input
		1	ST, SAT and TXDATA connected to TAUDOUT TAUDIN connected to GC8 input
7-5	Reserved for testing	000	Operating mode.

віт	NAME	STATE	FUNCTION	
3-0	GC8[3:0]	0000-1111	Control the gain of GC8 -3.	753.75 dB with 0.5 dB steps
		0000	-3.750 dB	
		0001	-3.250 dB	
		0010	-2.750 dB	
		0011	-2.250 dB	
		1111	3.750 dB	
		1000	0.250 dB default value	
4	DTX	0	Discontinuous transmission disabled	
		1	Discontinuous transmission	enabled
5	S17	0 (SYS1 = 0)	AGND	Summed into GC8 input.
		1 (SYS1 = 0)	TXDATA signal	
		0 (SYS1 = 1)	AGND	
		1 (SYS1 = 1)	TX Buffer output signal	
6	S18	0	AGND	
		1	Signalling Tone	
7	reserved	0	Set to zero	



Register (DE _{HEX} , write only							
BIT	NAME	STATE	FUNCTION					
3-0	GC1[3:0]	0000-1111	Control the gain of GC1 -3.0003.000 dB with 0.4dB step					
		0000	-3.000 dB					
		0001	-2.600 dB					
		0010	-2.200 dB					
		0011	-1.800 dB					
		1111	3.000 dB					
		1000	0.200 dB default value					
4	S1	0	RXIN	Connected to GC1 input				
		1	ALP					
5	S2	o	F4 output	Connected to DEOUT				
		1	F4 input					
6	S3	0	signal from S2-RXTST0- RXTST1	Connected to F5 input				
		1	RBPFIN	<u></u>				
7	S4	0	F5-output	Connected to RBPFOUT				
		1	F5-input					

Register	10 _{HEX} , read only		
віт	NAME	STATE	FUNCTION
0	STD	0	DTMF tone not received
		1	DTMF tone received
1	RXWRD	0	RX-buffer empty
	1	. 1	RX word received, SYS1=0. Next captured byte ready, SYS1=1.
2	TXWRD	0	Transmitting previous byte or word.
		1	TX-buffer empty.
3	TXCOL	0	No TX-collision detected
		1	TX-collision detected
4	SATINT	o	SATDET signal stable
		1	Rising or falling edge of SATDET detected
7-6		xxxx	Ignore

The rising edge of the signals STD, RXWRD, TXWRD, TXCOL and SATINT activates the interrupt line XINT. If the signal(s) changes when register 10_{HEX} or one of the other registers is read, the interrupt line will be activated right after the register read. Reading this register sets the interrupt line inactive. If a new interrupt is generated during register read, the interrupt line will be activated again right after register is read.

Register 1	1 _{HEX} , read only		
ВІТ	NAME	STATE	FUNCTION
0	Reserved for future use	х	
1	SATDET	0	Invalid SAT frequency received
		1	Valid SAT frequency received. See also SCC[1:0] in register 14 _{HEX} .
2	DOT	0	No dotting sequence detected
		1	Dotting sequence detected
3	WSYNC	0	No word sync detected
		1	Word sync detected
4	TXON	0	No transmission
		1	Word transmission on
5	BUSY	0	BUSY bit detected
		1	IDLE bit detected
6	BCHERR	0	No BCH error detected
		1	BCH error detected
7		×	-

ВІТ	NAME	STATE	FUNCTION	
0	STON	0	No Signatting tone transmission	
		1	Signalling tone transmission on	
1	стсу	0	Control channel	
		1	Voice channel	
2	TXCTREN	0	TX control active.	
		1	TX control active. TXCTRL output pin can be used to control transmitter.	
3	TXRST	0	Digital TX section operating	
		1	Digital TX section reset	
4	RXRST	0	Digital RX section operating	
		1	Digital RX section reset	
5	STR	0	Stream A	
		1	Stream B	
6	SYS0	0 (SYS1 = 0)	ETACS mode	
		1 (SYS1 = 0)	AMPS mode	
		0 (SYS1 = 1)	F9HP input connected to F9LP output. Narrow voice channel, NAMPS.	
		1 (SYS1 = 1)	F9HP input connected toGC5 output Narrow voice channel, NAMPS.	
7	SATEN	0	SAT transmission disabled	
		1	SAT transmission enabled	



Register :	Register 13 _{HEX} , write only							
ВІТ	NAME	STATE	FUNCTION					
1-0	DCC[1:0]	00-11	Digital color code					
		00	0000000					
		01	0011111					
		10	1100011					
		11	1111100					
2	INVTX	0	TXDATA not inverted					
		1	Invert TXDATA polarity					
3	INVRX	o	RX input not inverted					
		1	Invert RX input polarity					
4	AUMUT	o	No automatic mute					
		1	TX and RX audio muted automatically on voice channel					
5	RINTE	o	RX interrupts disabled					
		1	RX interrupts enabled					
6	STD	0	STD (DTMF receiver) interrupts disabled					
		1	STD (DTMF receiver) interrupts enabled					
7	SATINTEN	0	SAT Detection intrerrupts disabled					
		1	SAT Detection interrupts enabled					

BIT	NAME	STATE	FUNCTION
1-0	SAT[1:0]	00	5958Hz-5982Hz SAT frequency expected
	-	01	5988Hz-6012Hz SAT frequency expected
		10	6018Hz-6042Hz SAT frequency expected
		11	Invalid combination
5-2	Reserved for testing	0000	Normal operation
6	SYS1	0	Wide band mode (AMPS/ETACS)
		1	Narrow band mode (NAMPS)
7	NOMSAT	0	Transmitted SAT will follow received SAT.
		1	Nominal SAT frequency transmitted.



Register 1	h _{ex} , read only										
віт	NAME	STATE	FUNCTION								
7-0	RX BYTE	00H-FFH	When RXW	/RD is	high						
	SYS1=0		1st byte	b o	b 1	v 2	Ьσ	b 4	D 5	b 6	b 7
			2nd byte	b 8	b 9	b 1 0	ь 1	b 1 2	b 1 3	b 1 4	b 1 5
			3rd byte	b 1 6	b 1 7	b 1 8	b 1 9	b 2 0	b 2 1	b 2 2	b 2 3
			4th byte	b 2 4	b 2 5	b 2 6	b 2 7	٥	٥	0	0
7-0	RX BYTE SYS1=1	00H-FFH	Captured byte with 400 Hz after data comparator. MSB = first capture bit								

The received word is read by reading register ${\bf 15}_{\rm HEX}$ four times.

Register	Register 16 _{HEX} , write only							
віт	NAME	STATE	FUNCTION					
7-0	DAC1[7:0]	00H-FFH 80H FFH 00H 7FH	Control port for DAC1 output level OOH x LSB 7FH x LSB 80H x LSB FFH x LSB					

віт	NAME	STATE	FUNCTION	
BH	NAME	STATE	FUNCTION	
7-0	DAC2[7:0]	00H-FFH	Control port for DAC2 output level	
	1 ' '	80H	00H x LSB	

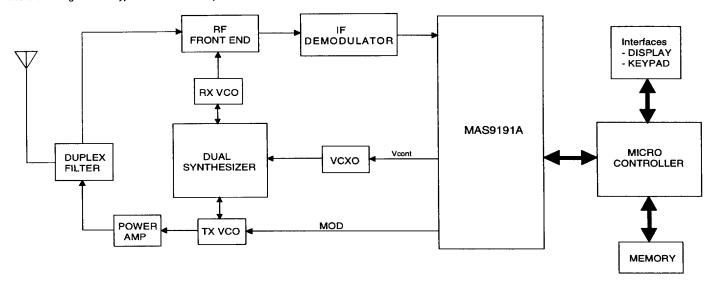
Register	egister 18 _{HEX} , write only							
віт	NAME	STATE	FUNCTION					
7-0	DAC3[7:0]	00H-FFH 80H FFH 00H 7FH	Control port for DAC3 output level OH x LSB 7FH x LSB 80H x LSB FFH x LSB					

Register 1	H _{EX} , write only										
віт	NAME	STATE	FUNCTION								
7-0	TX BYTE SYS1=0	00H-FFH	When TXW	'RD is	high						
	3131=0		1st byte	ьо	Б	v 2	ьз	b 4	b 5	b 6	b 7
			2nd byte	b 8	b 9	ь 1 0	b 1	b 1 2	b 1 3	b 1 4	b 1 5
			3rd byte	b 1 6	b 1 7	b 1 8	b 1 9	b 2 0	b 2 1	b 2 2	ь 2 3
			4th byte	b 2 4	b 2 5	b 2 6	b 2 7	b 2 8	b 2 9	b 3 0	B 3 1
			5th byte	b 3 2	b 3 3	b 3 4	b 3 5	0	0	٥	0
7-0	TX BYTE SYS1=1	00H-FFH	Byte for tra MSB will be				ited o	ut with	200H	z cloc	k.

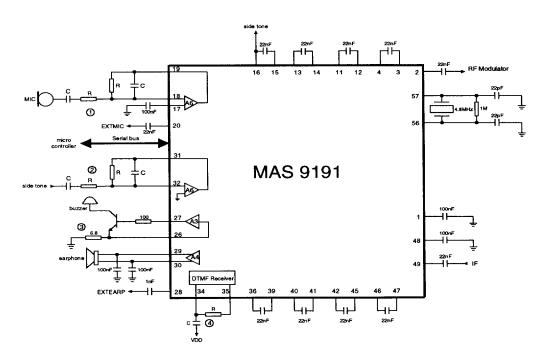
The transmitted word is written by writing to register $\mathbf{19}_{\text{HEX}}$ five times.

APPLICATION INFORMATION

The block diagram of a typical AMPS/TACS phone made with MAS9191A:



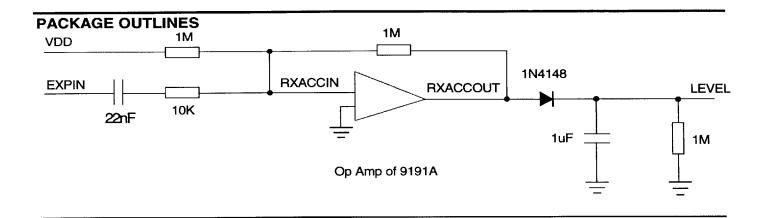
All of the necessary external components are shown in the diagram below.

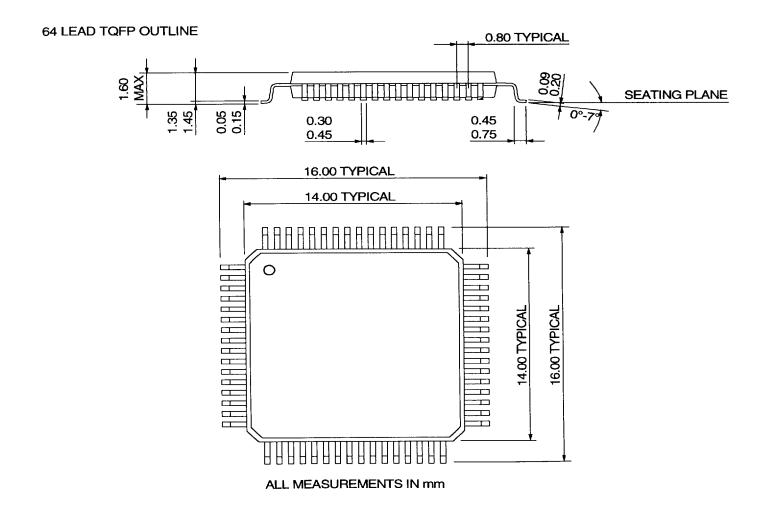


- no components determine the gain of microphone amplifier A6
- @ components determine the gain of SIDETONE amplifier A2
- 3 NPN transistor for the buzzer
- 4 components determine RC time constant if DTMF receiver is used

APPLICATION INFORMATION

The MAS9191 does not have any A/D converters because ordinary general purpose micro controllers have internal A/D converters with a multiplexer. This schematic shows how to use the MAS9191 internal Op Amps for detecting the audio level of the expander input. A similar circuit can be connected to MICOUT for detecting the microphone amplifier level.







ORDERING INFORMATION

Product Code	Product	Package	Comments
MAS9191AJ	AMPS/ETACS single chip audio/data processor	TQFP64	
MAS9191AJ-T	AMPS/ETACS single chip audio/data processor	TQFP64	Tape & Reel

LOCAL DISTRIBUTOR

MICRONAS CONTACTS

◆ SALES OFFICE

Micronas Semiconductor GmbH	Tel. (08165) 9521 0	
Lohweg 29	Tel. Int. + 49 8165 9521 0	
D-85375 NEUFAHRN, GERMANY	Telefax + 49 8165 9521 99	

◆ HEAD OFFICE/FACTORIES

MICRONAS INC. Kamreerintie 2, P.O.Box 51 FIN-02771 ESPOO, FINLAND Tel. (90) 80521 Tel. Int. +358 0 80521

Telefax +358 0 8053213

Micronas Semiconductor SA Ch. Chapons-des-Prés CH-2022 Bevaix Switzerland

Tel. Int. +41 38 460 111 Telefax +41 38 461 930

NOTICE

Micronas reserves the right to make changes to the products contained in this data sheet in order to improve the design or performance and to supply the best possible products. Micronas assumes no responsibility for the use of any circuits shown in this data sheet, conveys no license under any patent or other rights unless otherwise specified in this data sheet, and makes no claim that the circuits are free from patent infringement. Applications for any devices shown in this data sheet are for illustration only and Micronas makes no claim or warranty that such applications will be suitable for the use specified without further testing or modification.