

**384 OUTPUT TFT-LCD SOURCE DRIVER  
(COMPATIBLE WITH 64 GRAY SCALES)**

The  $\mu$ PD16715 is a source driver for TFT-LCDs capable of dealing with 64 gray scales. Data input is based on digital input configured as 6 bits by 6 dots (2 pixels), which can realize a full-color display of 260,000 colors by output of 64 values  $\gamma$  - corrected by an internal D/A converter and 5-by-2 external power modules. Because the output dynamic range is as large as 13.3 V<sub>P-P</sub>, level inversion operation of the LCD's common electrode is rendered unnecessary. Also, to be able to deal with dot-line inversion when mounted on a single side, this source driver is equipped with a built-in 6-bit D/A converter circuit whose odd output pins and even output pins respectively output gray scale voltages of differing polarity. Assuring a maximum clock frequency of 55 MHz when driving at 3.0 V, this driver is applicable to XGA/SXGA-standard TFT-LCD panels.

**FEATURES**

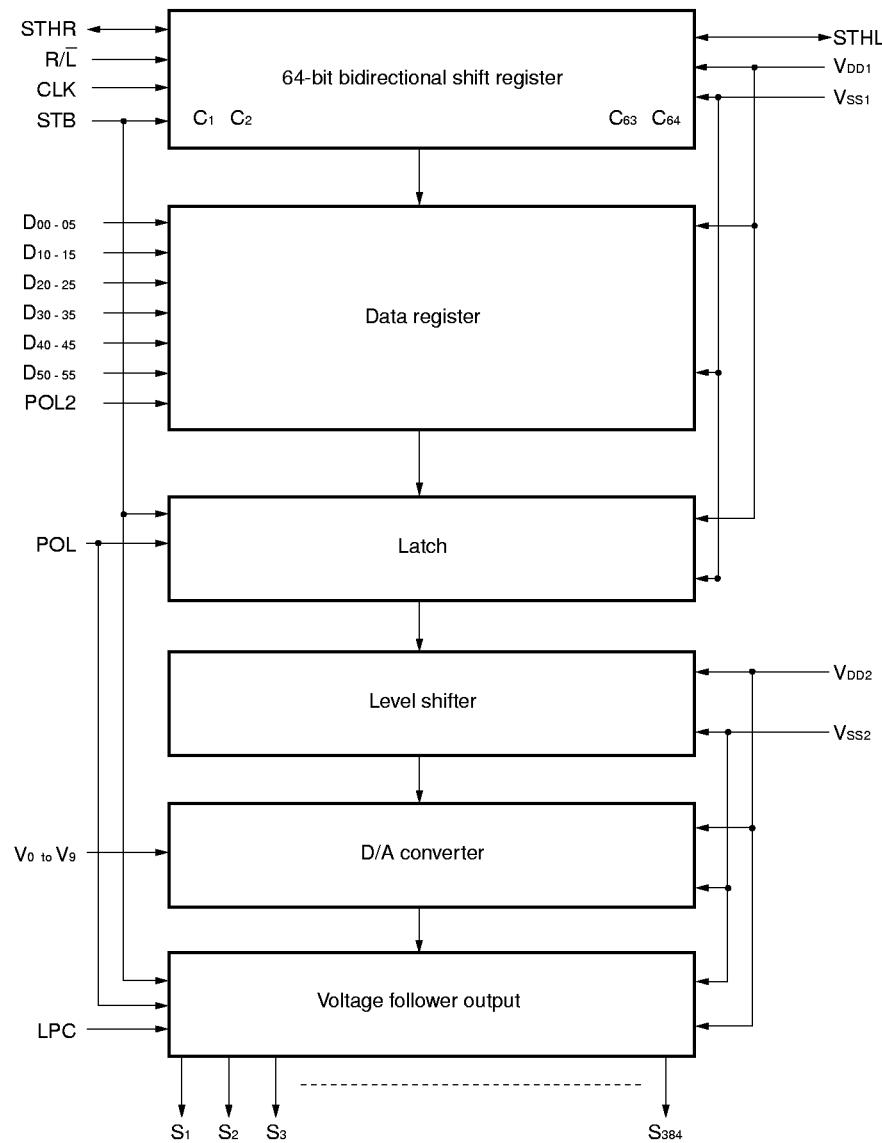
- Capable of outputting 64 values by means of 5-by-2 external power modules (10 units) and a D/A converter
- Output dynamic range 13.3 V<sub>P-P</sub> min. (@  $V_{DD2} = 13.5$  V)
- CMOS level input
- Input of 6 bits (gradation data) by 6 dots
- High-speed data transfer: fmax. = 55 MHz (internal data transfer speed when operating at 3.0 V)
- 384 outputs
- Apply for only dot-line inversion
- Display data inversion function (POL2)
- Single bank arrangement is possible (loaded with slim TCP)

**ORDERING INFORMATION**

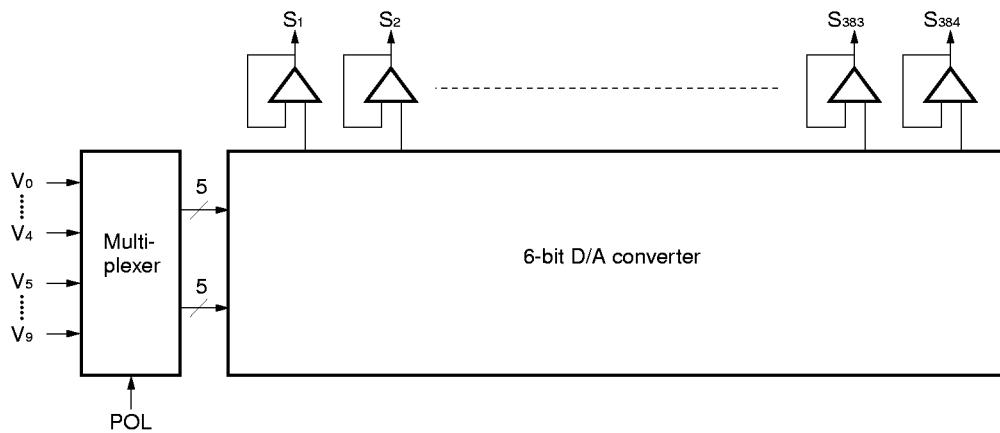
Part Number	Package
$\mu$ PD16715N- x x x	TCP (TAB package)

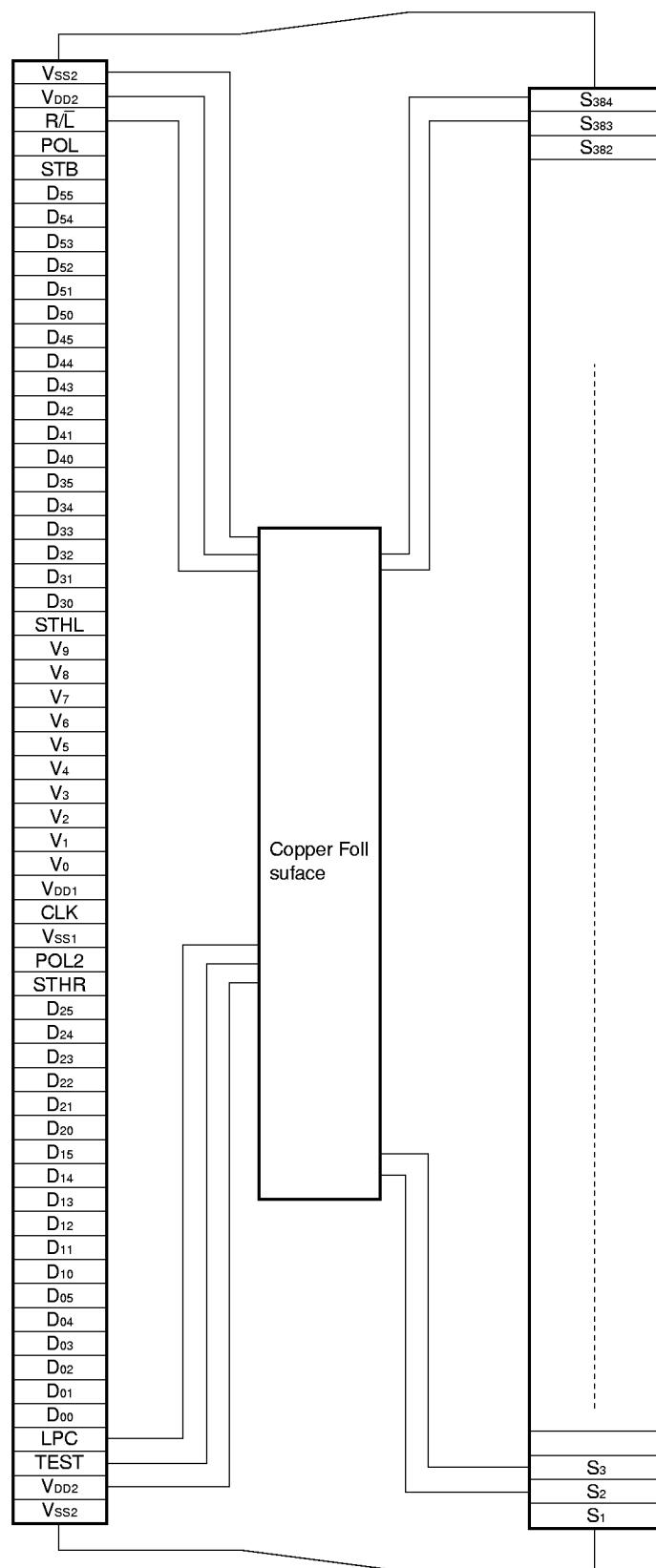
The TCP's external shape is customized. To order your TCP's external shape, please contact a NEC salesperson.

## 1. BLOCK DIAGRAM



## 2. RELATIONSHIP BETWEEN OUTPUT CIRCUIT AND D/A CONVERTER



3. PIN CONFIGURATION ( $\mu$ PD16715N- $\times \times \times$ )

This figure does not specify the TCP package.

#### 4. PIN FUNCTIONS

Pin Symbol	Pin Name	Description
S <sub>1</sub> to S <sub>384</sub>	Driver output	The D/A converted 64-gray-scale analog voltage is output.
D <sub>00</sub> to D <sub>05</sub>	Display data input	The display data is input with a width of 36 bits, viz.; the gray scale data (6 bits) by 6 dots (2 pixels).
D <sub>10</sub> to D <sub>15</sub>		
D <sub>20</sub> to D <sub>25</sub>		DX0: LSB, DX5: MSB
D <sub>30</sub> to D <sub>35</sub>		
D <sub>40</sub> to D <sub>45</sub>		
D <sub>50</sub> to D <sub>55</sub>		
R/L	Shift direction control input	These refer to the start pulse input/output pins when driver ICs are connected in cascade. The shift directions of the shift registers are as follows. R/L = H : STHR input, S <sub>1</sub> → S <sub>384</sub> , STHL output R/L = L : STHL input, S <sub>384</sub> → S <sub>1</sub> , STHR output
STHR	Right shift start pulse input/output	R/L = H : Becomes the start pulse input pin. R/L = L : Becomes the start pulse output pin.
STHL	Left shift start pulse input/output	R/L = H : Becomes the start pulse output pin. R/L = L : Becomes the start pulse input pin.
CLK	Shift clock input	Refers to the shift register's shift clock input. The display data is incorporated into the data register at the rising edge. At the rising edge of the 64th clock after the start pulse input, the start pulse output reaches the high level, thus becoming the start pulse of the next-level driver. The initial-level driver's 64th clock becomes valid as the next-level driver's start pulse is input. If 66 clock pulses are input after input of the start pulse, input of display data is halted automatically. The contents of the shift register are cleared at the STB's rising edge.
STB	Latch input	The contents of the data register are transferred to the latch circuit at the rising edge. And, at the falling edge, the gray scale voltage is supplied to the driver. It is necessary to ensure input of one pulse per horizontal period.
POL	Polarity input	POL = L; The S <sub>2n-1</sub> output uses V <sub>0</sub> to V <sub>4</sub> as the reference supply; The S <sub>2n</sub> output uses V <sub>5</sub> to V <sub>9</sub> as the reference supply. POL = H; The S <sub>2n-1</sub> output uses V <sub>5</sub> to V <sub>9</sub> as the reference supply; The S <sub>2n</sub> output uses V <sub>0</sub> to V <sub>4</sub> as the reference supply.
POL2	Data inversion	POL2 = H : Display data is inverted. POL2 = L : Display data is not inverted.
LPC	Low power control input	The output buffer constant current source is blocked, reducing current consumption. In lower power mode (LPC = 'H': DC-level input possible), the ordinary static current consumption can be reduced by approx. 20%. The condition that low power mode can be used is that the load constant is at least 5 kW + 100 pF.
V <sub>0</sub> to V <sub>9</sub>	$\gamma$ -corrected power supplies	Input the $\gamma$ -corrected power supplies from outside by using operational amplifier. Make sure to maintain the following relationships. During the gray scale voltage output, be sure to keep the gray scale level power supply at a constant level. $V_{DD2} > V_0 > V_1 > V_2 > V_3 > V_4 > V_5 > V_6 > V_7 > V_8 > V_9 > V_{SS2}$
TEST	Test pin	TEST = H or Open : Standard mode. TEST = L : Test mode. Please input "H" level.
V <sub>DD1</sub>	Logic power supply	3.3 V ±0.3 V
V <sub>DD2</sub>	Driver power supply	9.0 V to 13.5 V
V <sub>SS1</sub>	Logic ground	Grounding
V <sub>SS2</sub>	Driver ground	Grounding

#### 5. CAUTIONS

- (1) The power start sequence must be V<sub>DD1</sub>, logic input, and V<sub>DD2</sub> & V<sub>0</sub> to V<sub>9</sub> in that order. Reverse this sequence to shut down. (Simultaneous power application to V<sub>DD2</sub> and V<sub>0</sub> to V<sub>9</sub> is possible.)
- (2) To stabilize the supply voltage, please be sure to insert a 0.47  $\mu$ F bypass capacitor between V<sub>DD1</sub>-V<sub>SS1</sub> and V<sub>DD2</sub>-V<sub>SS2</sub>. Furthermore, for increased precision of the D/A converter, insertion of a bypass capacitor of about 0.01  $\mu$ F is also advised between the  $\gamma$ -corrected power supply terminals (V<sub>0</sub>, V<sub>1</sub>, V<sub>2</sub>, ..., V<sub>9</sub>) and V<sub>SS2</sub>.

## 6. RELATIONSHIP BETWEEN INPUT DATA AND OUTPUT VOLTAGE VALUE

This product incorporates a 6-bit D/A converter whose odd output pins and even output pins output respectively gray scale voltages of differing polarity with respect to the LCD's counter electrode (common electrode) voltage. The D/A converter consists of ladder resistors and switches. The ladder resistors  $r_0$  to  $r_{62}$  are so designed that the ratios between the LCD panel's  $\gamma$ -corrected voltages and  $V_0'$  to  $V_{63}'$  and  $V_0''$  to  $V_{63}''$  are roughly equal; and their respective resistance values are as shown on page 6. Among the 5-by-2  $\gamma$ -corrected voltages, input gray scale voltages of the same polarity with respect to the common voltage, for the respective five  $\gamma$ -corrected voltages of  $V_0$  to  $V_4$  and  $V_5$  to  $V_9$ . If fine gray-scale voltage precision is not necessary, the voltage follower circuit supplied to the  $\gamma$ -corrected power supplies  $V_1$  to  $V_4$  and  $V_{11}$  to  $V_{16}$  can be deleted.

Figure 1 shows the relationship between the driving voltages such as liquid-crystal driving voltages  $V_{DD2}$  and  $V_{SS2}$ , common electrode potential  $V_{COM}$ , and  $\gamma$ -corrected voltages  $V_0$  to  $V_9$  and the input data. Be sure to maintain the voltage relationships of  $V_{DD2} > V_0 > V_1 > V_2 > V_3 > V_4 > V_5 > V_6 > V_7 > V_8 > V_9 > V_{SS2}$ . Figures 2-1 and 2-2 show the relationship between the input data and the output data. Table 1 shows the resistance values of the resistor strings. This driver IC is designed for single-sided mounting. Therefore, please do not use it for  $\gamma$ -corrected power supply level inversion in double-sided mounting. Because the current flowing through ladder resistors  $r_0$  to  $r_{62}$  is small, its use for double-sided mounting impairs the IC's stable operation when the level of the  $\gamma$ -corrected power supply terminal is inverted thus causing display failures.

**Figure 1. Relationship between Input Data and Output Voltage**

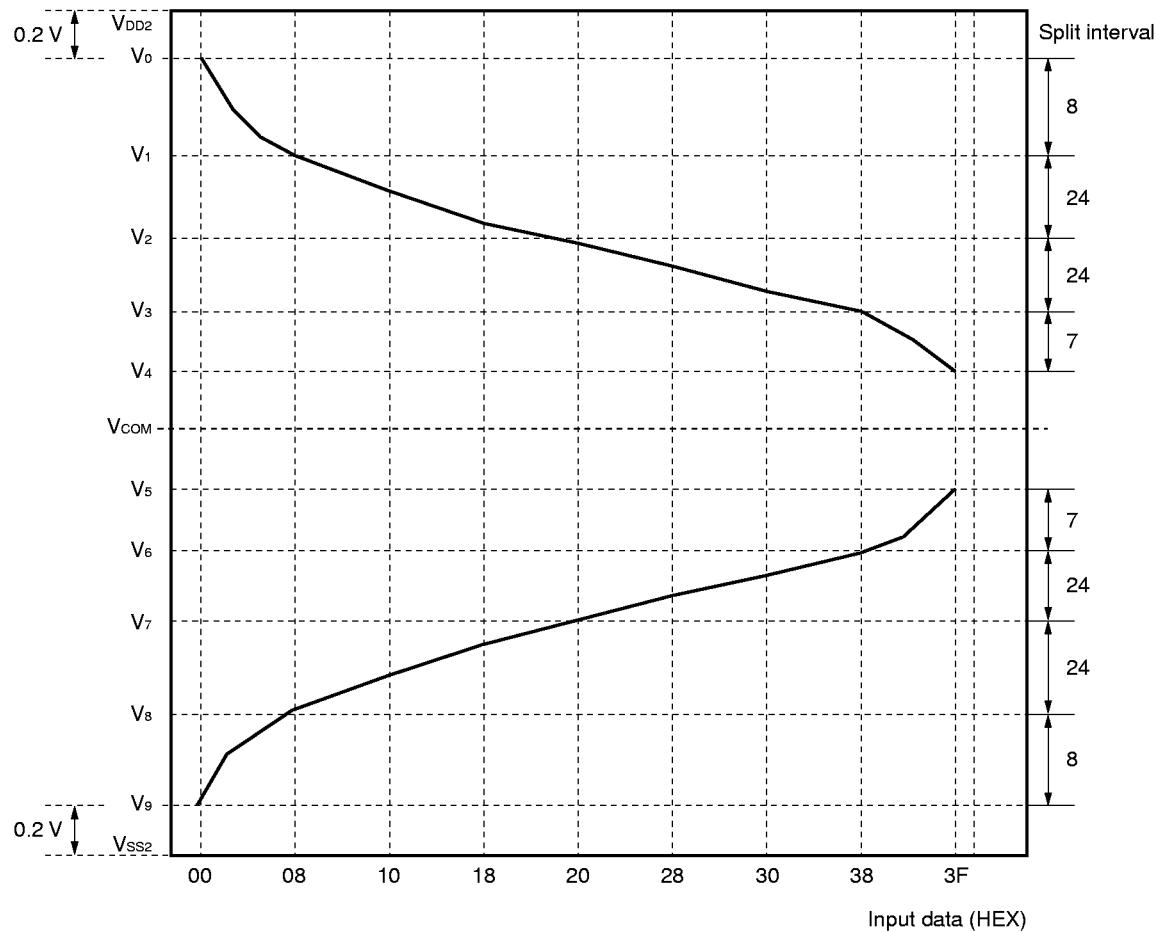
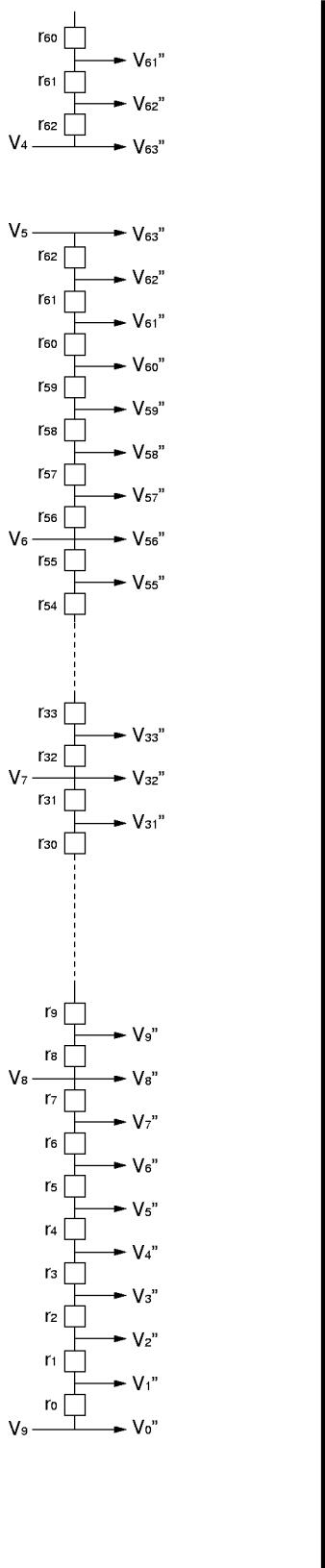


Figure 2-1. Relationship between Input Data and Output Voltage:  $V_{DD2} > V_0 > V_1 > V_2 > V_3 > V_4 > V_5$ 

The table below provides the data mapping for the ladder diagram. The columns represent the data bits D<sub>x5</sub> through D<sub>x0</sub>. The rows represent the output voltages V<sub>0'</sub> through V<sub>63''</sub>.

Data	D <sub>x5</sub> D <sub>x4</sub> D <sub>x3</sub> D <sub>x2</sub> D <sub>x1</sub> D <sub>x0</sub>	Output Voltage	
00H	0 0 0 0 0 0	V <sub>0'</sub>	V <sub>0</sub>
01H	0 0 0 0 0 1	V <sub>1'</sub>	$V_1 + (V_0 - V_1) \times 4300/5100$
02H	0 0 0 0 1 0	V <sub>2'</sub>	$V_1 + (V_0 - V_1) \times 3550/5100$
03H	0 0 0 0 1 1	V <sub>3'</sub>	$V_1 + (V_0 - V_1) \times 2850/5100$
04H	0 0 0 1 0 0	V <sub>4'</sub>	$V_1 + (V_0 - V_1) \times 2200/5100$
05H	0 0 0 1 0 1	V <sub>5'</sub>	$V_1 + (V_0 - V_1) \times 1600/5100$
06H	0 0 0 1 1 0	V <sub>6'</sub>	$V_1 + (V_0 - V_1) \times 1050/5100$
07H	0 0 0 1 1 1	V <sub>7'</sub>	$V_1 + (V_0 - V_1) \times 500/5100$
08H	0 0 1 0 0 0	V <sub>8'</sub>	V <sub>0</sub>
09H	0 0 1 0 0 1	V <sub>9'</sub>	$V_1 + (V_0 - V_1) \times 5200/5700$
0AH	0 0 1 0 1 0	V <sub>10'</sub>	$V_1 + (V_0 - V_1) \times 4800/5700$
0BH	0 0 1 0 1 1	V <sub>11'</sub>	$V_1 + (V_0 - V_1) \times 4400/5700$
0CH	0 0 1 1 0 0	V <sub>12'</sub>	$V_1 + (V_0 - V_1) \times 4050/5700$
0DH	0 0 1 1 0 1	V <sub>13'</sub>	$V_1 + (V_0 - V_1) \times 3700/5700$
0EH	0 0 1 1 1 0	V <sub>14'</sub>	$V_1 + (V_0 - V_1) \times 3350/5700$
0FH	0 0 1 1 1 1	V <sub>15'</sub>	$V_1 + (V_0 - V_1) \times 3050/5700$
10H	0 1 0 0 0 0	V <sub>16'</sub>	$V_1 + (V_0 - V_1) \times 2750/5700$
11H	0 1 0 0 0 1	V <sub>17'</sub>	$V_1 + (V_0 - V_1) \times 2450/5700$
12H	0 1 0 0 1 0	V <sub>18'</sub>	$V_1 + (V_0 - V_1) \times 2200/5700$
13H	0 1 0 0 1 1	V <sub>19'</sub>	$V_1 + (V_0 - V_1) \times 1950/5700$
14H	0 1 0 1 0 0	V <sub>20'</sub>	$V_1 + (V_0 - V_1) \times 1700/5700$
15H	0 1 0 1 0 1	V <sub>21'</sub>	$V_1 + (V_0 - V_1) \times 1500/5700$
16H	0 1 0 1 1 0	V <sub>22'</sub>	$V_1 + (V_0 - V_1) \times 1300/5700$
17H	0 1 0 1 1 1	V <sub>23'</sub>	$V_1 + (V_0 - V_1) \times 1100/5700$
18H	0 1 1 0 0 0	V <sub>24'</sub>	$V_1 + (V_0 - V_1) \times 950/5700$
19H	0 1 1 0 0 1	V <sub>25'</sub>	$V_1 + (V_0 - V_1) \times 800/5700$
1AH	0 1 1 0 1 0	V <sub>26'</sub>	$V_1 + (V_0 - V_1) \times 650/5700$
1BH	0 1 1 0 1 1	V <sub>27'</sub>	$V_1 + (V_0 - V_1) \times 500/5700$
1CH	0 1 1 1 0 0	V <sub>28'</sub>	$V_1 + (V_0 - V_1) \times 400/5700$
1DH	0 1 1 1 0 1	V <sub>29'</sub>	$V_1 + (V_0 - V_1) \times 300/5700$
1EH	0 1 1 1 1 0	V <sub>30'</sub>	$V_1 + (V_0 - V_1) \times 200/5700$
1FH	0 1 1 1 1 1	V <sub>31'</sub>	$V_1 + (V_0 - V_1) \times 100/5700$
20H	1 0 0 0 0 0	V <sub>32'</sub>	V <sub>2</sub>
21H	1 0 0 0 0 1	V <sub>33'</sub>	$V_3 + (V_2 - V_3) \times 2450/2550$
22H	1 0 0 0 1 0	V <sub>34'</sub>	$V_3 + (V_2 - V_3) \times 2350/2550$
23H	1 0 0 0 1 1	V <sub>35'</sub>	$V_3 + (V_2 - V_3) \times 2250/2550$
24H	1 0 0 1 0 0	V <sub>36'</sub>	$V_3 + (V_2 - V_3) \times 2150/2550$
25H	1 0 0 1 0 1	V <sub>37'</sub>	$V_3 + (V_2 - V_3) \times 2050/2550$
26H	1 0 0 1 1 0	V <sub>38'</sub>	$V_3 + (V_2 - V_3) \times 1950/2550$
27H	1 0 0 1 1 1	V <sub>39'</sub>	$V_3 + (V_2 - V_3) \times 1850/2550$
28H	1 0 1 0 0 0	V <sub>40'</sub>	$V_3 + (V_2 - V_3) \times 1750/2550$
29H	1 0 1 0 0 1	V <sub>41'</sub>	$V_3 + (V_2 - V_3) \times 1650/2550$
2AH	1 0 1 0 1 0	V <sub>42'</sub>	$V_3 + (V_2 - V_3) \times 1550/2550$
2BH	1 0 1 0 1 1	V <sub>43'</sub>	$V_3 + (V_2 - V_3) \times 1450/2550$
2CH	1 0 1 1 0 0	V <sub>44'</sub>	$V_3 + (V_2 - V_3) \times 1350/2550$
2DH	1 0 1 1 0 1	V <sub>45'</sub>	$V_3 + (V_2 - V_3) \times 1250/2550$
2EH	1 0 1 1 1 0	V <sub>46'</sub>	$V_3 + (V_2 - V_3) \times 1150/2550$
2FH	1 0 1 1 1 1	V <sub>47'</sub>	$V_3 + (V_2 - V_3) \times 1050/2550$
30H	1 1 0 0 0 0	V <sub>48'</sub>	$V_3 + (V_2 - V_3) \times 950/2550$
31H	1 1 0 0 0 1	V <sub>49'</sub>	$V_3 + (V_2 - V_3) \times 850/2550$
32H	1 1 0 0 1 0	V <sub>50'</sub>	$V_3 + (V_2 - V_3) \times 750/2550$
33H	1 1 0 0 1 1	V <sub>51'</sub>	$V_3 + (V_2 - V_3) \times 650/2550$
34H	1 1 0 1 0 0	V <sub>52'</sub>	$V_3 + (V_2 - V_3) \times 550/2550$
35H	1 1 0 1 0 1	V <sub>53'</sub>	$V_3 + (V_2 - V_3) \times 450/2550$
36H	1 1 0 1 1 0	V <sub>54'</sub>	$V_3 + (V_2 - V_3) \times 300/2550$
37H	1 1 0 1 1 1	V <sub>55'</sub>	$V_3 + (V_2 - V_3) \times 150/2550$
38H	1 1 1 0 0 0	V <sub>56'</sub>	V <sub>3</sub>
39H	1 1 1 0 0 1	V <sub>57'</sub>	$V_4 + (V_3 - V_4) \times 2300/2500$
3AH	1 1 1 0 1 0	V <sub>58'</sub>	$V_4 + (V_3 - V_4) \times 2100/2500$
3BH	1 1 1 0 1 1	V <sub>59'</sub>	$V_4 + (V_3 - V_4) \times 1850/2500$
3CH	1 1 1 1 0 0	V <sub>60'</sub>	$V_4 + (V_3 - V_4) \times 1600/2500$
3DH	1 1 1 1 0 1	V <sub>61'</sub>	$V_4 + (V_3 - V_4) \times 1300/2500$
3EH	1 1 1 1 1 0	V <sub>62'</sub>	$V_4 + (V_3 - V_4) \times 800/2500$
3FH	1 1 1 1 1 1	V <sub>63'</sub>	V <sub>4</sub>

Figure 2-2. Relationship between Input Data and Output Voltage:  $V_5 > V_6 > V_7 > V_8 > V_9 > V_{ss2}$ 

Data	Dx5 Dx4 Dx3 Dx2 Dx1 Dx0	Output Voltage	
3FH	1 1 1 1 1 1	$V_{63}''$	$V_5$
3EH	1 1 1 1 1 0	$V_{62}''$	$V_6 + (V_5 - V_6) \times 1700/2500$
3DH	1 1 1 1 0 1	$V_{61}''$	$V_6 + (V_5 - V_6) \times 1200/2500$
3CH	1 1 1 1 0 0	$V_{60}''$	$V_6 + (V_5 - V_6) \times 900/2500$
3BH	1 1 1 0 1 1	$V_{59}''$	$V_6 + (V_5 - V_6) \times 650/2500$
3AH	1 1 1 0 1 0	$V_{58}''$	$V_6 + (V_5 - V_6) \times 400/2500$
39H	1 1 1 0 0 1	$V_{57}''$	$V_6 + (V_5 - V_6) \times 200/2500$
38H	1 1 1 0 0 0	$V_{56}''$	$V_6$
37H	1 1 0 1 1 1	$V_{55}''$	$V_7 + (V_6 - V_7) \times 2400/2550$
36H	1 1 0 1 1 0	$V_{54}''$	$V_7 + (V_6 - V_7) \times 2250/2550$
35H	1 1 0 1 0 1	$V_{53}''$	$V_7 + (V_6 - V_7) \times 2100/2550$
34H	1 1 0 1 0 0	$V_{52}''$	$V_7 + (V_6 - V_7) \times 2000/2550$
33H	1 1 0 0 1 1	$V_{51}''$	$V_7 + (V_6 - V_7) \times 1900/2550$
32H	1 1 0 0 1 0	$V_{50}''$	$V_7 + (V_6 - V_7) \times 1800/2550$
31H	1 1 0 0 0 1	$V_{49}''$	$V_7 + (V_6 - V_7) \times 1700/2550$
30H	1 1 0 0 0 0	$V_{48}''$	$V_7 + (V_6 - V_7) \times 1600/2550$
2FH	1 0 1 1 1 1	$V_{47}''$	$V_7 + (V_6 - V_7) \times 1500/2550$
2EH	1 0 1 1 1 0	$V_{46}''$	$V_7 + (V_6 - V_7) \times 1400/2550$
2DH	1 0 1 1 0 1	$V_{45}''$	$V_7 + (V_6 - V_7) \times 1300/2550$
2CH	1 0 1 1 0 0	$V_{44}''$	$V_7 + (V_6 - V_7) \times 1200/2550$
2BH	1 0 1 0 1 1	$V_{43}''$	$V_7 + (V_6 - V_7) \times 1100/2550$
2AH	1 0 1 0 1 0	$V_{42}''$	$V_7 + (V_6 - V_7) \times 1000/2550$
29H	1 0 1 0 0 1	$V_{41}''$	$V_7 + (V_6 - V_7) \times 900/2550$
28H	1 0 1 0 0 0	$V_{40}''$	$V_7 + (V_6 - V_7) \times 800/2550$
27H	1 0 0 1 1 1	$V_{39}''$	$V_7 + (V_6 - V_7) \times 700/2550$
26H	1 0 0 1 1 0	$V_{38}''$	$V_7 + (V_6 - V_7) \times 600/2550$
25H	1 0 0 1 0 1	$V_{37}''$	$V_7 + (V_6 - V_7) \times 500/2550$
24H	1 0 0 1 0 0	$V_{36}''$	$V_7 + (V_6 - V_7) \times 400/2550$
23H	1 0 0 0 1 1	$V_{35}''$	$V_7 + (V_6 - V_7) \times 300/2550$
22H	1 0 0 0 1 0	$V_{34}''$	$V_7 + (V_6 - V_7) \times 200/2550$
21H	1 0 0 0 0 1	$V_{33}''$	$V_7 + (V_6 - V_7) \times 100/2550$
20H	1 0 0 0 0 0	$V_{32}''$	$V_7$
1FH	0 1 1 1 1 1	$V_{31}''$	$V_8 + (V_7 - V_8) \times 5600/5700$
1EH	0 1 1 1 1 0	$V_{30}''$	$V_8 + (V_7 - V_8) \times 5500/5700$
1DH	0 1 1 1 0 1	$V_{29}''$	$V_8 + (V_7 - V_8) \times 5400/5700$
1CH	0 1 1 1 0 0	$V_{28}''$	$V_8 + (V_7 - V_8) \times 5300/5700$
1BH	0 1 1 0 1 1	$V_{27}''$	$V_8 + (V_7 - V_8) \times 5200/5700$
1AH	0 1 1 0 1 0	$V_{26}''$	$V_8 + (V_7 - V_8) \times 5050/5700$
19H	0 1 1 0 0 1	$V_{25}''$	$V_8 + (V_7 - V_8) \times 4900/5700$
18H	0 1 1 0 0 0	$V_{24}''$	$V_8 + (V_7 - V_8) \times 4750/5700$
17H	0 1 0 1 1 1	$V_{23}''$	$V_8 + (V_7 - V_8) \times 4600/5700$
16H	0 1 0 1 1 0	$V_{22}''$	$V_8 + (V_7 - V_8) \times 4400/5700$
15H	0 1 0 1 0 1	$V_{21}''$	$V_8 + (V_7 - V_8) \times 4200/5700$
14H	0 1 0 1 0 0	$V_{20}''$	$V_8 + (V_7 - V_8) \times 4000/5700$
13H	0 1 0 0 1 1	$V_{19}''$	$V_8 + (V_7 - V_8) \times 3750/5700$
12H	0 1 0 0 1 0	$V_{18}''$	$V_8 + (V_7 - V_8) \times 3500/5700$
11H	0 1 0 0 0 1	$V_{17}''$	$V_8 + (V_7 - V_8) \times 3250/5700$
10H	0 1 0 0 0 0	$V_{16}''$	$V_8 + (V_7 - V_8) \times 2950/5700$
0FH	0 0 1 1 1 1	$V_{15}''$	$V_8 + (V_7 - V_8) \times 2650/5700$
0EH	0 0 1 1 1 0	$V_{14}''$	$V_8 + (V_7 - V_8) \times 2350/5700$
0DH	0 0 1 1 0 1	$V_{13}''$	$V_8 + (V_7 - V_8) \times 2000/5700$
0CH	0 0 1 1 0 0	$V_{12}''$	$V_8 + (V_7 - V_8) \times 2650/5700$
0BH	0 0 1 0 1 1	$V_{11}''$	$V_8 + (V_7 - V_8) \times 1300/5700$
0AH	0 0 1 0 1 0	$V_{10}''$	$V_8 + (V_7 - V_8) \times 900/5700$
09H	0 0 1 0 0 1	$V_9''$	$V_8 + (V_7 - V_8) \times 500/5700$
08H	0 0 1 0 0 0	$V_8''$	$V_8$
07H	0 0 0 1 1 1	$V_7''$	$V_8 + (V_8 - V_9) \times 4600/5100$
06H	0 0 0 1 1 0	$V_6''$	$V_8 + (V_8 - V_9) \times 4050/5100$
05H	0 0 0 1 0 1	$V_5''$	$V_8 + (V_8 - V_9) \times 3500/5100$
04H	0 0 0 1 0 0	$V_4''$	$V_8 + (V_8 - V_9) \times 2900/5100$
03H	0 0 0 0 1 1	$V_3''$	$V_8 + (V_8 - V_9) \times 2250/5100$
02H	0 0 0 0 1 0	$V_2''$	$V_8 + (V_8 - V_9) \times 1550/5100$
01H	0 0 0 0 0 1	$V_1''$	$V_8 + (V_8 - V_9) \times 1550/5100$
00H	0 0 0 0 0 0	$V_0''$	$V_8$

Table 1. Ladder Resistance Values

Resistor Name	Resistance Value ( $\Omega$ )	Resistor Name	Resistance Value ( $\Omega$ )
r <sub>0</sub>	800	r <sub>32</sub>	100
r <sub>1</sub>	750	r <sub>33</sub>	100
r <sub>2</sub>	700	r <sub>34</sub>	100
r <sub>3</sub>	650	r <sub>35</sub>	100
r <sub>4</sub>	600	r <sub>36</sub>	100
r <sub>5</sub>	550	r <sub>37</sub>	100
r <sub>6</sub>	550	r <sub>38</sub>	100
r <sub>7</sub>	500	r <sub>39</sub>	100
r <sub>8</sub>	500	r <sub>40</sub>	100
r <sub>9</sub>	400	r <sub>41</sub>	100
r <sub>10</sub>	400	r <sub>42</sub>	100
r <sub>11</sub>	350	r <sub>43</sub>	100
r <sub>12</sub>	350	r <sub>44</sub>	100
r <sub>13</sub>	350	r <sub>45</sub>	100
r <sub>14</sub>	300	r <sub>46</sub>	100
r <sub>15</sub>	300	r <sub>47</sub>	100
r <sub>16</sub>	300	r <sub>48</sub>	100
r <sub>17</sub>	250	r <sub>49</sub>	100
r <sub>18</sub>	250	r <sub>50</sub>	100
r <sub>19</sub>	250	r <sub>51</sub>	100
r <sub>20</sub>	200	r <sub>52</sub>	100
r <sub>21</sub>	200	r <sub>53</sub>	150
r <sub>22</sub>	200	r <sub>54</sub>	150
r <sub>23</sub>	150	r <sub>55</sub>	150
r <sub>24</sub>	150	r <sub>56</sub>	200
r <sub>25</sub>	150	r <sub>57</sub>	200
r <sub>26</sub>	150	r <sub>58</sub>	250
r <sub>27</sub>	100	r <sub>59</sub>	250
r <sub>28</sub>	100	r <sub>60</sub>	300
r <sub>29</sub>	100	r <sub>61</sub>	500
r <sub>30</sub>	100	r <sub>62</sub>	800
r <sub>31</sub>	100	Total	15850

## RELATIONSHIP BETWEEN OUTPUT CIRCUIT AND D/A CONVERTER

Data format : 6 bits  $\times$  2 RGBs (6 dots)

Input width : 36 bits (2-pixel data)

$R/\bar{L} = H$  (Right shift)

Output	S <sub>1</sub>	S <sub>2</sub>	S <sub>3</sub>	S <sub>4</sub>	...	S <sub>383</sub>	S <sub>384</sub>
Data	D <sub>00</sub> -D <sub>05</sub>	D <sub>10</sub> -D <sub>15</sub>	D <sub>20</sub> -D <sub>25</sub>	D <sub>30</sub> -D <sub>35</sub>	...	D <sub>40</sub> -D <sub>45</sub>	D <sub>50</sub> -D <sub>55</sub>

$R/\bar{L} = L$  (Left shift)

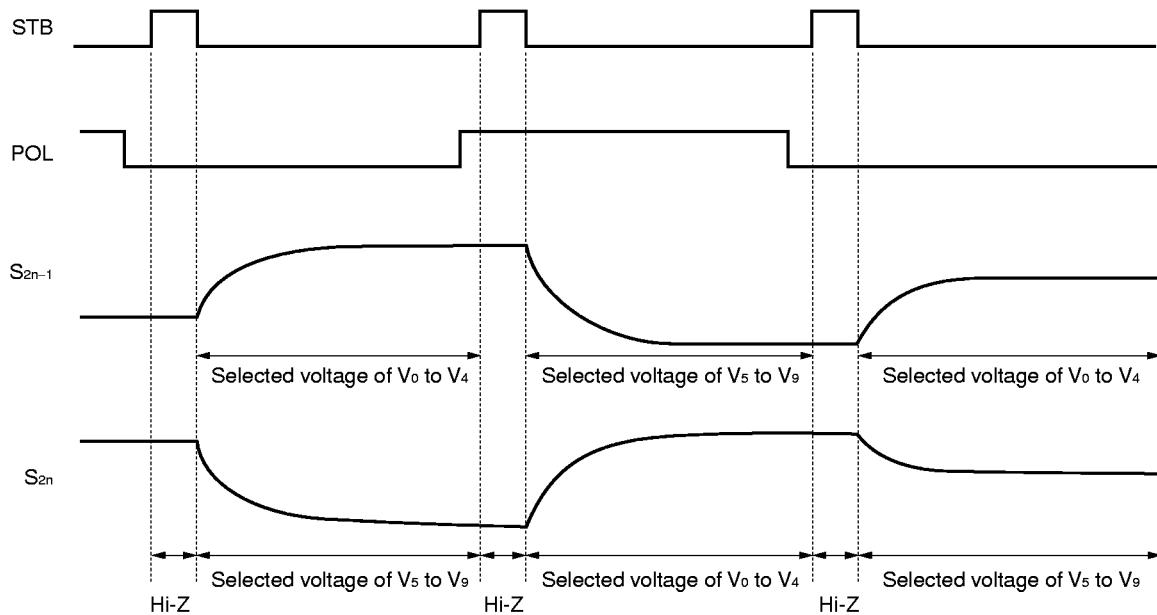
Output	S <sub>1</sub>	S <sub>2</sub>	S <sub>3</sub>	S <sub>4</sub>	...	S <sub>383</sub>	S <sub>384</sub>
Data	D <sub>00</sub> -D <sub>05</sub>	D <sub>10</sub> -D <sub>15</sub>	D <sub>20</sub> -D <sub>25</sub>	D <sub>30</sub> -D <sub>35</sub>	...	D <sub>40</sub> -D <sub>45</sub>	D <sub>50</sub> -D <sub>55</sub>

POL	S <sub>2n-1</sub>	S <sub>2n</sub>
L	V <sub>0</sub> to V <sub>4</sub>	V <sub>5</sub> to V <sub>9</sub>
H	V <sub>5</sub> to V <sub>9</sub>	V <sub>0</sub> to V <sub>4</sub>

S<sub>2n-1</sub> (Odd output), S<sub>2n</sub> (Even output) n = 1, 2, ..., 192

## RELATIONSHIP BETWEEN STB, POL, AND OUTPUT WAVEFORM

The output voltage is written to the LCD panel synchronized with the STB rising edge.

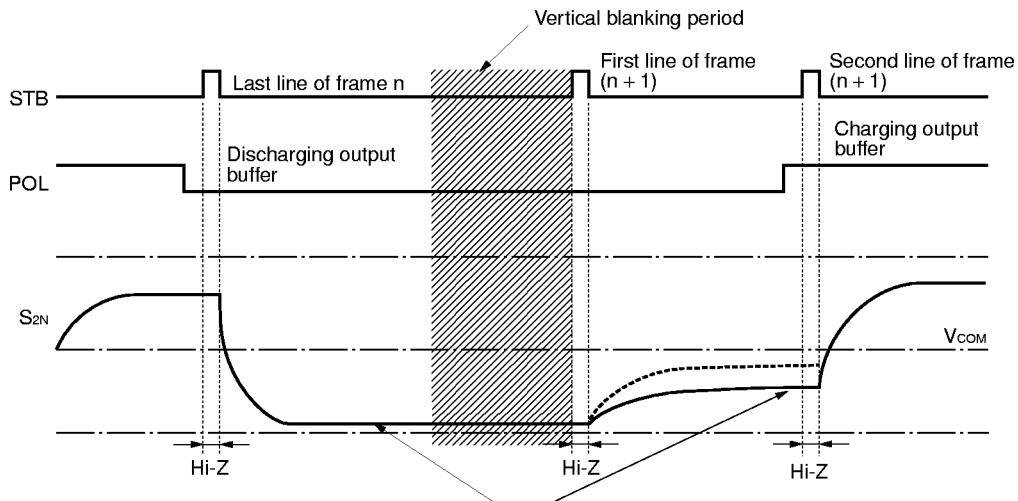


## 7. NOTES ON FRAME INVERSION

The  $\mu$ PD16715 is an IC for dot inversion and inverts dots by alternately using a charging output buffer and a discharging output buffer. Therefore, the output voltage of the first line may not be correctly written because the last line's output polarity of frame n ( $n + 1$ ) and the first line's output polarity are the same (refer to **Figure 3**).

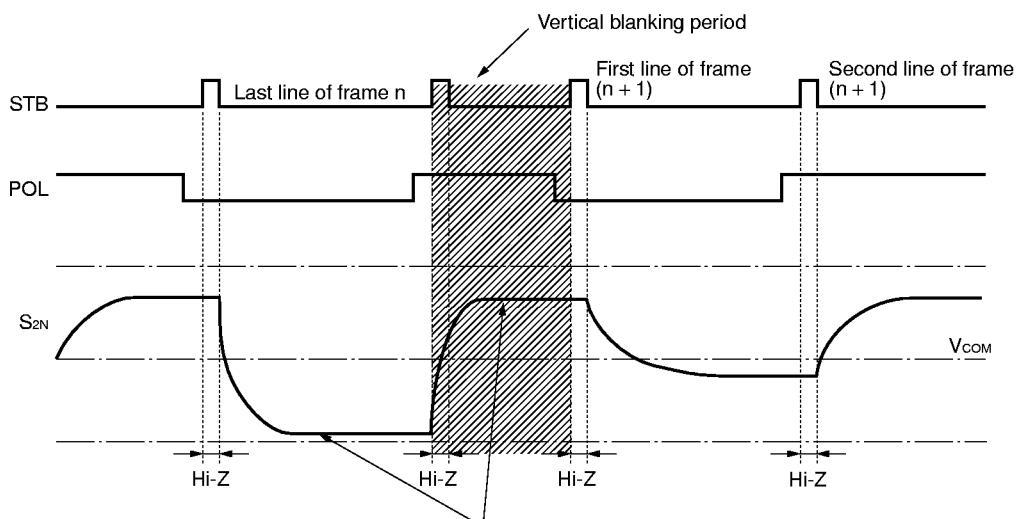
Consequently, polarity inversion and write operation must be performed between frames (vertical blanking period) in order to invert (clear) the polarity of the wiring level of the liquid crystal panel by using the last line output of the previous frame (refer to **Figure 4**).

**Figure 3**



If the write voltage of the first line of the last ( $n + 1$ ) frame is greater than the write voltage of the last line of frame n, the targeted voltage cannot be correctly written with the discharging buffer with a low charging capability.

**Figure 4**



Because data of negative polarity is to be written on the first line of frame  $(n + 1)$ , write data of positive polarity in advance during the vertical blanking period.

## 8. ELECTRIC SPECIFICATION

### Absolute Maximum Ratings ( $T_A = 25^\circ\text{C}$ , $V_{SS1} = V_{SS2} = 0 \text{ V}$ )

Parameter	Symbol	Rating	Unit
Logic Part Supply Voltage	$V_{DD1}$	-0.3 to +6.5	V
Driver Part Supply Voltage	$V_{DD2}$	-0.3 to +15.0	V
Logic Part Input Voltage	$V_{I1}$	-0.3 to $V_{DD1} + 0.3$	V
Driver Part Input Voltage	$V_{I2}$	-0.3 to $V_{DD2} + 0.3$	V
Logic Part Output Voltage	$V_{O1}$	-0.3 to $V_{DD1} + 0.3$	V
Driver Part Output Voltage	$V_{O2}$	-0.3 to $V_{DD2} + 0.3$	V
Power Dissipation	$P_d$	200	mW
Operating Temperature Range	$T_A$	-10 to +75	°C
Storage Temperature Range	$T_{stg}$	-55 to +125	°C

### Recommended Operating Condition ( $T_A = -10$ to $+75^\circ\text{C}$ , $V_{SS1} = V_{SS2} = 0 \text{ V}$ )

Parameter	Symbol	MIN.	TYP.	MAX.	Unit
Logic Part Supply Voltage	$V_{DD1}$	3.0	3.3	3.6	V
Driver Part Supply Voltage	$V_{DD2}$	9.0		13.5	V
High-Level Input Voltage	$V_{IH}$	$0.7V_{DD1}$		$V_{DD1}$	V
Low-Level Input Voltage	$V_{IL}$	0		$0.3V_{DD1}$	V
$\gamma$ -Corrected Voltage	$V_o$ to $V_g$	0.1		$V_{DD2} - 0.1$	V
Driver Part Output Voltage	$V_o$	0.1		$V_{DD2} - 0.1$	V
Maximum Clock Frequency	$f_{max.}$	55			MHz
Output Load Capacitance	$C_L$			150	pF

### Electrical Specifications ( $T_A = -10$ to $+75^\circ\text{C}$ , $V_{DD1} = 3.3 \pm 0.3 \text{ V}$ , $V_{DD2} = 9.0 \text{ V}$ to $13.5 \text{ V}$ , $V_{SS1} = V_{SS2} = 0 \text{ V}$ )

Parameter	Symbol	Condition	MIN.	TYP.	MAX.	Unit
Input Leak Current	$I_{IL}$				$\pm 1.0$	$\mu\text{A}$
High-Level Output Voltage	$V_{OH}$	STHR (STHL), $I_{OH} = -1.0 \text{ mA}$	$V_{DD1} - 0.1$			V
Low-Level Output Voltage	$V_{OL}$	STHR (STHL), $I_{OL} = +1.0 \text{ mA}$			0.1	V
$\gamma$ -Corrected Supply Current	$I_\gamma$	$V_{DD2} = 13 \text{ V}$ , $V_{O-4}$ $= V_5 - V_g = 6.0 \text{ V}$	$V_o, V_5$	0.39	0.8	$\text{mA}$
			$V_4, V_g$	0.39	0.8	$\text{mA}$
Driver Output Current	$I_{VOH}$	$V_x = 10.0 \text{ V}$ , $V_{OUT} = 1.0 \text{ V}$			-0.3	$\text{mA}$
	$I_{VOL}$	$V_x = 1.0 \text{ V}$ , $V_{OUT} = 10.0 \text{ V}$	<sup>Note</sup> 0.3			$\text{mA}$

**Note**  $V_x$  refers to the output voltage of analog output pins S<sub>1</sub> to S<sub>384</sub>.

$V_{OUT}$  refers to the voltage applied to analog output pins S<sub>1</sub> to S<sub>384</sub>.

**Electrical Specifications ( $T_A = -10$  to  $+75^\circ\text{C}$ ,  $V_{DD1} = 3.3 \pm 0.3$  V,  $V_{DD2} = 9.0$  V to  $13.5$  V,  $V_{SS1} = V_{SS2} = 0$  V)**

Parameter	Symbol	Condition	MIN.	TYP.	MAX.	Unit
Output Voltage Deviation <small>Note 1</small>	$\Delta V_O$	Input data			$\pm 20$	mV
Average Output Voltage Variation <small>Note 2</small>	$\Delta V_{AV}$	Input data		$\pm 10$		mV
Output Voltage Range	$V_O$	Input data	$V_{SS2} +0.1$		$V_{DD2} -0.1$	V
Logic Part Dynamic Current Consumption	$I_{DD1}$	$V_{DD1} = 3.6$ V, $T_A = 25^\circ\text{C}$		1.5	8	mA
Driver Part Dynamic Current Consumption 1 <small>Notes 3, 4</small>	$I_{DD2}$	$V_{DD1} = 3.0$ V, $V_{DD2} = 13.5$ V No leads, $T_A = 25^\circ\text{C}$		3.5	8	mA

- Notes**
1. The output voltage deviation refers to the voltage difference between adjoining output pins when the display data is the same (within the chip).
  2. The average output voltage variation refers to the average output voltage difference between chips. The average output voltage refers to the average voltage between chips when the display data is the same.
  3. The STB cycle is defined to be  $20 \mu\text{s}$  at  $f_{CLK} = 33$  MHz. The TYP. values refer to an all black or all white input pattern. The MAX. value refers to the measured values in the dot checkerboard input pattern.
  4. Refers to the current consumption per driver when cascades are connected under the assumption of XGA single-sided mounting (8 units).

**Switching Characteristics ( $T_A = -10$  to  $+75^\circ\text{C}$ ,  $V_{DD1} = 3.3 \pm 0.3$  V,  $V_{DD2} = 9.0$  V to  $13.5$  V,  $V_{SS1} = V_{SS2} = 0$  V)**

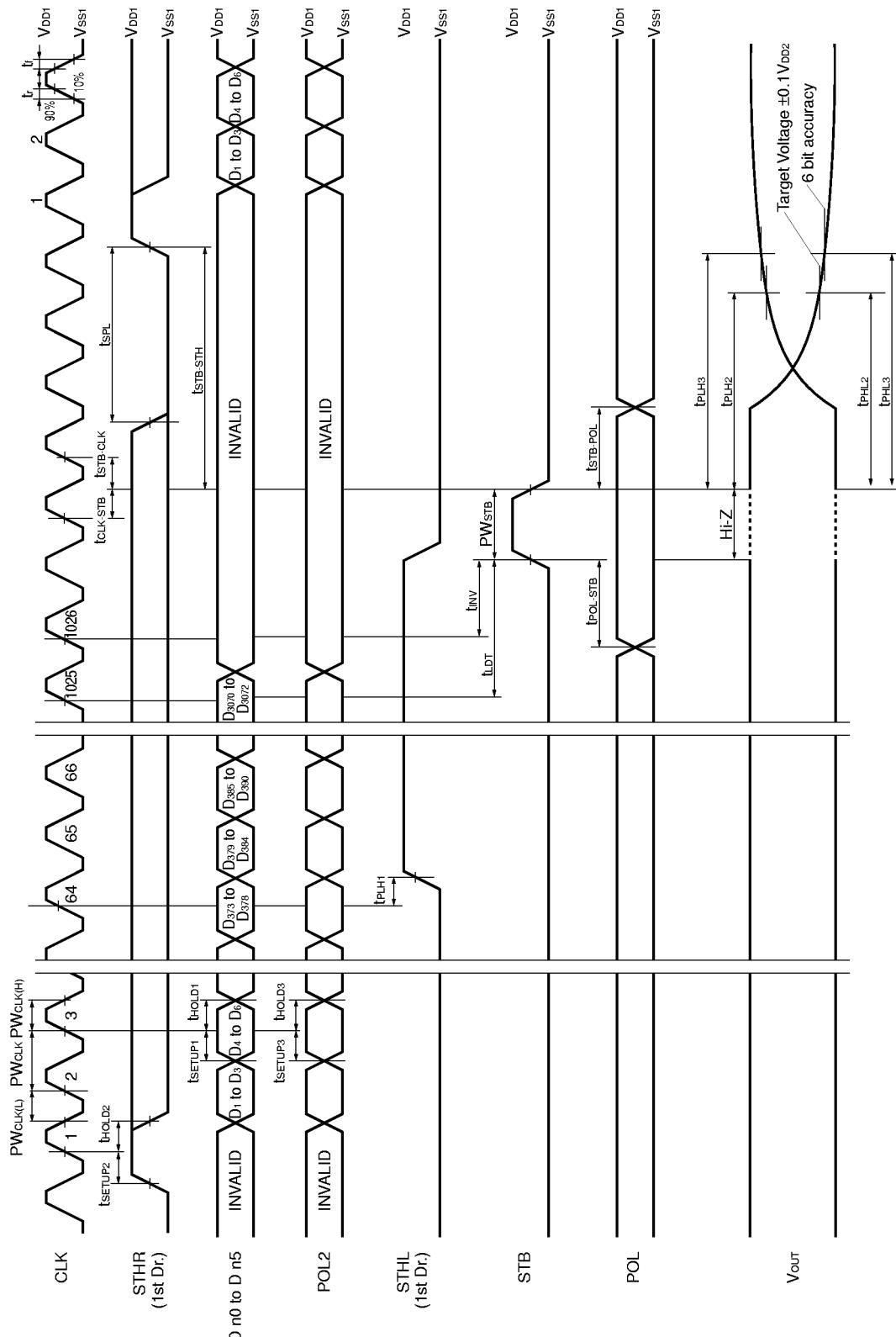
Parameter	Symbol	Condition	MIN.	TYP.	MAX.	Unit
Start Pulse Delay Time	$t_{PLH1}$	$C_L = 20$ pF			14	ns
Driver Output Delay Time 1	$t_{PLH2}$	$C_L = 50$ pF, $R_L = 50$ k $\Omega$			11	$\mu\text{s}$
Driver Output Delay Time 2	$t_{PLH3}$	$C_L = 50$ pF, $R_L = 50$ k $\Omega$			17	$\mu\text{s}$
Driver Output Delay Time 3	$t_{PHL2}$	$C_L = 50$ pF, $R_L = 50$ k $\Omega$			11	$\mu\text{s}$
Driver Output Delay Time 4	$t_{PHL3}$	$C_L = 50$ pF, $R_L = 50$ k $\Omega$			17	$\mu\text{s}$
Input Capacitance 1	$C_{i1}$	STHR (STHL) excluded, $T_A = 25^\circ\text{C}$			15	pF
Input Capacitance 2	$C_{i2}$	STHR (STHL), $T_A = 25^\circ\text{C}$			15	pF

**Timing Requirement ( $T_A = -10$  to  $+75^\circ\text{C}$ ,  $V_{DD1} = 3.3 \pm 0.3$  V,  $V_{SS1} = V_{SS2} = 0$  V,  $t_r = t_f = 8.0$  ns)**

Parameter	Symbol	Condition	MIN.	TYP.	MAX.	Unit
Clock Pulse Width	$PW_{CLK}$		18			ns
Clock Pulse Low Period	$PW_{CLK(H)}$		5			ns
Clock Pulse High Period	$PW_{CLK(L)}$		5			ns
Data Setup Time	$t_{SETUP1}$		5			ns
Data Hold Time	$t_{HOLD1}$		5			ns
Start Pulse Setup Time	$t_{SETUP2}$		4			ns
Start Pulse Hold Time	$t_{HOLD2}$		5			ns
POL2 Setup Time	$t_{SETUP3}$		5			ns
POL2 Hold Time	$t_{HOLD3}$		5			ns
STB Pulse Width	$PW_{STB}$		500			ns
Data Invalid Period	$t_{INV}$		1			CLK
Last Data Timing	$t_{LDT}$		2			CLK
CLK - STB Time	$t_{CLK-STB}$	$CLK \uparrow \rightarrow STB \downarrow$	5			ns
STB - CLK Time	$t_{STB-CLK}$	$STB \downarrow \rightarrow CLK \downarrow$	5			ns
Time Between STB and Start Pulse	$t_{STB-SPH}$	$STB \downarrow \rightarrow STHR(L) \uparrow$	50			ns
POL - STB Time	$t_{POL-STB}$	$POL \uparrow \text{ or } \downarrow \rightarrow STB \uparrow$	-7			ns
STB - POL Time	$t_{STB-POL}$	$STB \downarrow \rightarrow POL \downarrow \text{ or } \uparrow$	9			ns

## 9. Switching Characteristics Waveform ( $R/L = H$ )

Unless otherwise specified, the input level is defined to be  $V_{ILH} = 0.5V_{DD1}$ .



## 10. RECOMMENDED MOUNTING CONDITIONS

When mounting this product, please make sure that the following recommended conditions be satisfied.

For packaging methods and conditions other than those recommended below, please contact NEC sales personnel.

Mounting Condition	Mounting Method	Condition
Thermocompression	Soldering	Heating tool 300 to 350°C, heating for 2 to 3 sec; pressure 100 g (per solder)
	ACF (Adhesive Conductive Film)	Temporary bonding 70 to 100°C; pressure 3 to 8 kg/cm <sup>2</sup> ; time 3 to 5 sec. Real bonding 165 to 180°C; pressure 25 to 45 kg/cm <sup>2</sup> , time 30 to 40 secs. (When using the anisotropy conductive film SUMIZAC1003 of Sumitomo Bakelite, Ltd)

**Caution** To find out the detailed conditions for packaging the ACF part, please contact the ACF manufacturing company. Be sure to avoid using two or more packaging methods at a time.

### Reference

NEC Semiconductor Device Reliability/Quality Control System (C10983E)

Quality Grades to NEC's Semiconductor Devices (C11531E)

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