

Features

- 100 MHz gain-bandwidth product
- Gain-of-2 stable
- Low supply current
= 5.2 mA at $V_S = \pm 15V$
- Wide supply range
= $\pm 2V$ to $\pm 18V$ dual-supply
= 2.5V to 36V single-supply
- High slew rate = 275 V/ μs
- Fast settling = 80 ns to 0.1% for a 10V step
- Low differential gain = 0.02% at $A_V = +2$, $R_L = 150\Omega$
- Low differential phase = 0.07° at $A_V = +2$, $R_L = 150\Omega$
- Stable with unlimited capacitive load
- Wide output voltage swing
= $\pm 13.6V$ with $V_S = \pm 15V$,
 $R_L = 1000\Omega$
= 3.8V/0.3V with $V_S = +5V$,
 $R_L = 500\Omega$

Applications

- Video amplifier
- Single-supply amplifier
- Active filters/integrators
- High-speed sample-and-hold
- High-speed signal processing
- ADC/DAC buffer
- Pulse/RF amplifier
- Pin diode receiver
- Log amplifier
- Photo multiplier amplifier
- Difference amplifier

Ordering Information

Part No.	Temp. Range	Package	Outline #
EL2045CN	0°C to +75°C	8-Pin P-DIP	MDP0031
EL2045CS	0°C to +75°C	8-Lead SO	MDP0027

General Description

The EL2045C is a high speed, low power, low cost monolithic operational amplifier built on Elantec's proprietary complementary bipolar process. The EL2045C is gain-of-2 stable and features a 275 V/ μs slew rate and 100 MHz gain-bandwidth product while requiring only 5.2 mA of supply current.

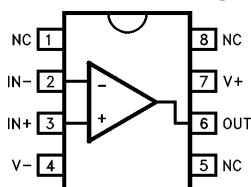
The power supply operating range of the EL2045C is from $\pm 18V$ down to as little as $\pm 2V$. For single-supply operation, the EL2045C operates from 36V down to as little as 2.5V. The excellent power supply operating range of the EL2045C makes it an obvious choice for applications on a single +5V or +3V supply.

The EL2045C also features an extremely wide output voltage swing of $\pm 13.6V$ with $V_S = \pm 15V$ and $R_L = 1000\Omega$. At $\pm 5V$, output voltage swing is a wide $\pm 3.8V$ with $R_L = 500\Omega$ and $\pm 3.2V$ with $R_L = 150\Omega$. Furthermore, for single-supply operation at +5V, output voltage swing is an excellent 0.3V to 3.8V with $R_L = 500\Omega$.

At a gain of +2, the EL2045C has a -3 dB bandwidth of 100 MHz with a phase margin of 50°. It can drive unlimited load capacitance, and because of its conventional voltage-feedback topology, the EL2045C allows the use of reactive or non-linear elements in its feedback network. This versatility combined with low cost and 75 mA of output-current drive makes the EL2045C an ideal choice for price-sensitive applications requiring low power and high speed.

Connection Diagram

DIP and SO Package



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EL2045C

Low-Power 100 MHz Gain-of-2 Stable Operational Amplifier

Absolute Maximum Ratings ($T_A = 25^\circ\text{C}$)

Supply Voltage (V_S)	$\pm 18\text{V}$ or 36V	Power Dissipation (P_D)	See Curves
Peak Output Current (I_{OP})	Short-Circuit Protected	Operating Temperature Range (T_A)	0°C to $+75^\circ\text{C}$
Output Short-Circuit Duration (Note 1)	Infinite	Operating Junction Temperature (T_J)	150°C
Input Voltage (V_{IN})	$\pm V_S$	Storage Temperature (T_{ST})	-65°C to $+150^\circ\text{C}$
Differential Input Voltage (dV_{IN})	$\pm 10\text{V}$		

Important Note:

All parameters having Min/Max specifications are guaranteed. The Test Level column indicates the specific device testing actually performed during production and Quality inspection. Elantec performs most electrical tests using modern high-speed automatic test equipment, specifically the LTX77 Series system. Unless otherwise noted, all tests are pulsed tests, therefore $T_J = T_C = T_A$.

Test Level	Test Procedure
I	100% production tested and QA sample tested per QA test plan QCX0002.
II	100% production tested at $T_A = 25^\circ\text{C}$ and QA sample tested at $T_A = 25^\circ\text{C}$, T_{MAX} and T_{MIN} per QA test plan QCX0002.
III	QA sample tested per QA test plan QCX0002.
IV	Parameter is guaranteed (but not tested) by Design and Characterization Data.
V	Parameter is typical value at $T_A = 25^\circ\text{C}$ for information purposes only.

DC Electrical Characteristics $V_S = \pm 15\text{V}$, $R_L = 1000\Omega$, unless otherwise specified

Parameter	Description	Condition	Temp	Min	Typ	Max	Test Level	Units
V_{OS}	Input Offset Voltage	$V_S = \pm 15\text{V}$	25°C		0.5	7.0	I	mV
			T_{MIN}, T_{MAX}			9.0	III	mV
TCV_{OS}	Average Offset Voltage Drift	(Note 2)	All		10.0		V	$\mu\text{V}/^\circ\text{C}$
I_B	Input Bias Current	$V_S = \pm 15\text{V}$	25°C		2.8	8.2	I	μA
			T_{MIN}, T_{MAX}			9.2	III	μA
		$V_S = \pm 5\text{V}$	25°C		2.8		V	μA
I_{OS}	Input Offset Current	$V_S = \pm 15\text{V}$	25°C		50	300	I	nA
			T_{MIN}, T_{MAX}			400	III	nA
		$V_S = \pm 5\text{V}$	25°C		50		V	nA
TCI_{OS}	Average Offset Current Drift	(Note 2)	All		0.3		V	$\text{nA}/^\circ\text{C}$
A_{VOL}	Open-Loop Gain	$V_S = \pm 15\text{V}, V_{OUT} = \pm 10\text{V}, R_L = 1000\Omega$	25°C	1500	3000		I	V/V
			T_{MIN}, T_{MAX}	1500			III	V/V
		$V_S = \pm 5\text{V}, V_{OUT} = \pm 2.5\text{V}, R_L = 500\Omega$	25°C		2500		V	V/V
		$V_S = \pm 5\text{V}, V_{OUT} = \pm 2.5\text{V}, R_L = 150\Omega$	25°C		1750		V	V/V
PSRR	Power Supply Rejection Ratio	$V_S = \pm 5\text{V}$ to $\pm 15\text{V}$	25°C	65	85		I	dB
			T_{MIN}, T_{MAX}	60			III	dB

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DC Electrical Characteristics $V_S = \pm 15V$, $R_L = 1000\Omega$, unless otherwise specified — Contd.

Parameter	Description	Condition	Temp	Min	Typ	Max	Test Level	Units
CMRR	Common-Mode Rejection Ratio	$V_{CM} = \pm 12V$, $V_{OUT} = 0V$	25°C	70	95		I	dB
			T_{MIN} , T_{MAX}	70			III	dB
CMIR	Common-Mode Input Range	$V_S = \pm 15V$	25°C		± 14.0		V	V
		$V_S = \pm 5V$	25°C		± 4.2		V	V
		$V_S = +5V$	25°C		4.2/0.1		V	V
V_{OUT}	Output Voltage Swing	$V_S = \pm 15V$, $R_L = 1000\Omega$	25°C	± 13.4	± 13.6		I	V
			T_{MIN} , T_{MAX}	± 13.1			III	V
		$V_S = \pm 15V$, $R_L = 500\Omega$	25°C	± 12.0	± 13.4		I	V
		$V_S = \pm 5V$, $R_L = 500\Omega$	25°C	± 3.4	± 3.8		IV	V
		$V_S = \pm 5V$, $R_L = 150\Omega$	25°C		± 3.2		V	V
		$V_S = +5V$, $R_L = 500\Omega$	25°C	3.6/0.4	3.8/0.3		I	V
			T_{MIN} , T_{MAX}	3.5/0.5			III	V
I_{SC}	Output Short Circuit Current		25°C	40	75		I	mA
			T_{MIN} , T_{MAX}	35			III	mA
I_S	Supply Current	$V_S = \pm 15V$, No Load	25°C		5.2	7	I	mA
			T_{MIN} , T_{MAX}			7.6	III	mA
		$V_S = \pm 5V$, No Load	25°C		5.0		V	mA
R_{IN}	Input Resistance	Differential	25°C		150		V	k Ω
		Common-Mode	25°C		15		V	M Ω
C_{IN}	Input Capacitance	$A_V = +2$ @ 10 MHz	25°C		1.0		V	pF
R_{OUT}	Output Resistance	$A_V = +2$	25°C		50		V	m Ω
PSOR	Power-Supply Operating Range	Dual-Supply	25°C	± 2.0		± 18.0	V	V
		Single-Supply	25°C	2.5		36.0	V	V

Closed-Loop AC Electrical Characteristics

$V_S = \pm 15V$, $A_V = +2$, $R_f = R_g = 1\text{ k}\Omega$, $C_f = 3\text{ pF}$, $R_L = 1000\Omega$ unless otherwise specified

Parameter	Description	Condition	Temp	Min	Typ	Max	Test Level	Units
BW	– 3 dB Bandwidth ($V_{OUT} = 0.4\text{ V}_{PP}$)	$V_S = \pm 15V$, $A_V = +2$	25°C		100		V	MHz
		$V_S = \pm 15V$, $A_V = -1$	25°C		75		V	MHz
		$V_S = \pm 15V$, $A_V = +5$	25°C		20		V	MHz
		$V_S = \pm 15V$, $A_V = +10$	25°C		10		V	MHz
		$V_S = \pm 15V$, $A_V = +20$	25°C		5		V	MHz
		$V_S = \pm 5V$, $A_V = +2$	25°C		75		V	MHz
GBWP	Gain-Bandwidth Product	$V_S = \pm 15V$	25°C		100		V	MHz
		$V_S = \pm 5V$	25°C		75		V	MHz
PM	Phase Margin	$R_L = 1\text{ k}\Omega$, $C_L = 10\text{ pF}$	25°C		50		V	°

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Closed-Loop AC Electrical Characteristics

$V_S = \pm 15V$, $A_V = +2$, $R_f = R_g = 1\text{ k}\Omega$, $C_f = 3\text{ pF}$, $R_L = 1000\Omega$, unless otherwise specified — Contd.

Parameter	Description	Condition	Temp	Min	Typ	Max	Test Level	Units
SR	Slew Rate (Note 3)	$V_S = \pm 15V$, $R_L = 1000\Omega$	25°C	200	275		I	V/ μs
		$V_S = \pm 5V$, $R_L = 500\Omega$	25°C		200		V	V/ μs
FPBW	Full-Power Bandwidth (Note 4)	$V_S = \pm 15V$	25°C	3.2	4.4		I	MHz
		$V_S = \pm 5V$	25°C		12.7		V	MHz
t_r, t_f	Rise Time, Fall Time	0.1V Output Step	25°C		3.0		V	ns
OS	Overshoot	0.1V Output Step	25°C		20		V	%
t_{PD}	Propagation Delay		25°C		2.5		V	ns
t_s	Settling to $\pm 0.1\%$ ($A_V = +2$)	$V_S = \pm 15V$, 10V Step	25°C		80		V	ns
		$V_S = \pm 5V$, 5V Step	25°C		60		V	ns
dG	Differential Gain (Note 5)	NTSC/PAL	25°C		0.02		V	%
dP	Differential Phase (Note 5)	NTSC/PAL	25°C		0.07		V	°
eN	Input Noise Voltage	10 kHz	25°C		15.0		V	nV/ $\sqrt{\text{Hz}}$
iN	Input Noise Current	10 kHz	25°C		1.50		V	pA/ $\sqrt{\text{Hz}}$
CI STAB	Load Capacitance Stability	$A_V = +2$	25°C		Infinite		V	pF

Note 1: A heat-sink is required to keep junction temperature below absolute maximum when an output is shorted.

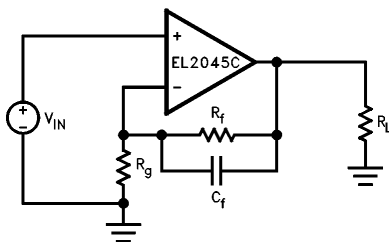
Note 2: Measured from T_{MIN} to T_{MAX} .

Note 3: Slew rate is measured on rising edge.

Note 4: For $V_S = \pm 15V$, $V_{OUT} = 20\text{ V}_{pp}$. For $V_S = \pm 5V$, $V_{OUT} = 5\text{ V}_{pp}$. Full-power bandwidth is based on slew rate measurement using: $FPBW = SR/(2\pi * V_{peak})$.

Note 5: Video Performance measured at $V_S = \pm 15V$, $A_V = +2$ with 2 times normal video level across $R_L = 150\Omega$. This corresponds to standard video levels across a back-terminated 75 Ω load. For other values of R_L , see curves.

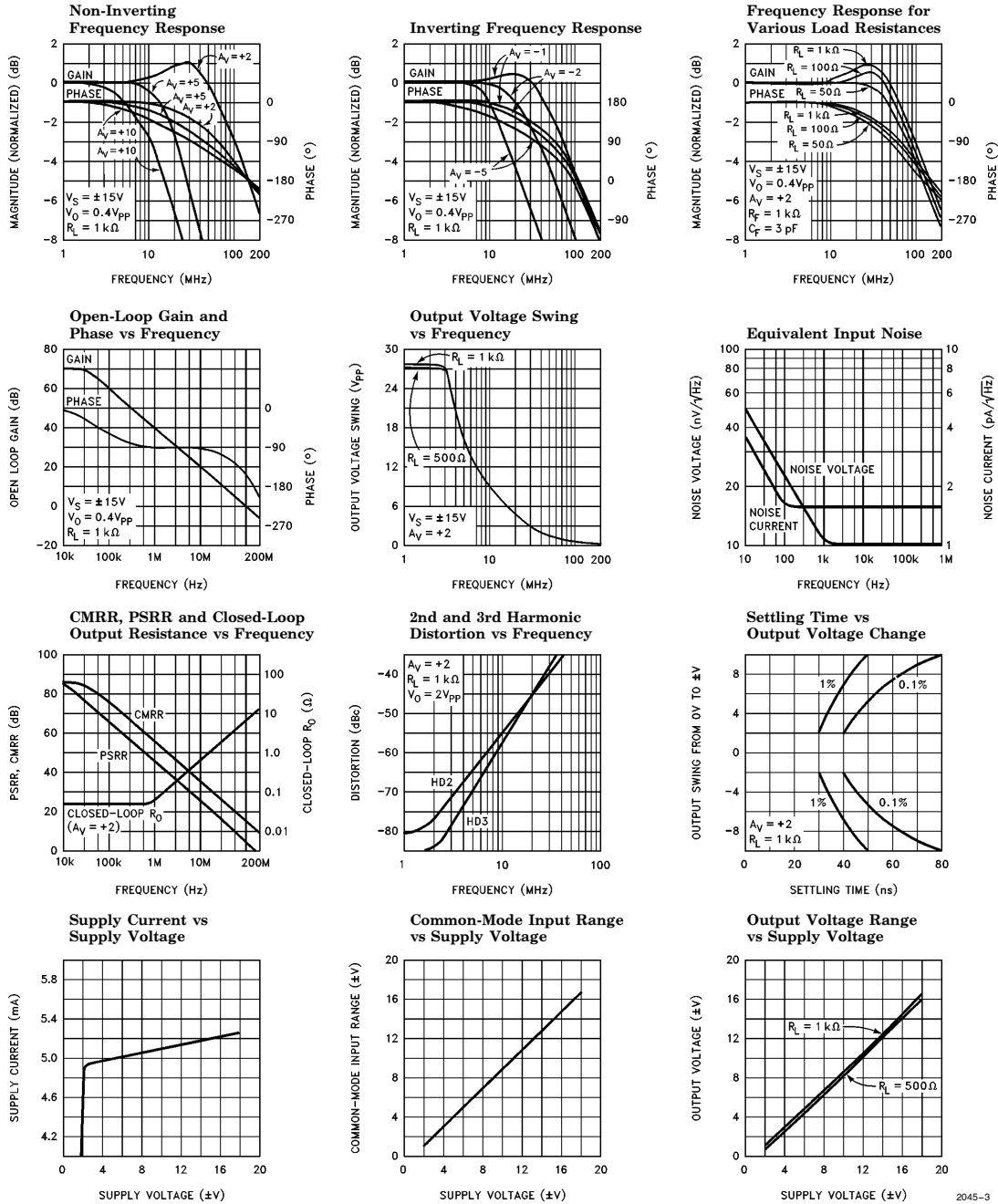
EL2045C Test Circuit



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Typical Performance Curves

($T_A = 25^\circ\text{C}$, $R_f = 1\text{ k}\Omega$, $C_f = 3\text{ pF}$, $R_L = 1000\Omega$, $A_V = +2$ unless otherwise specified)



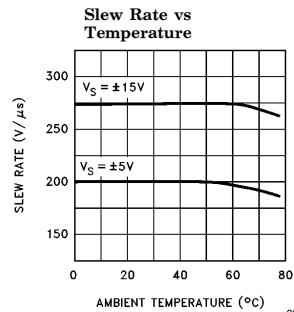
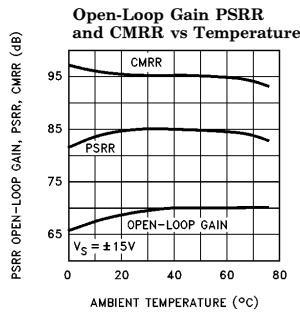
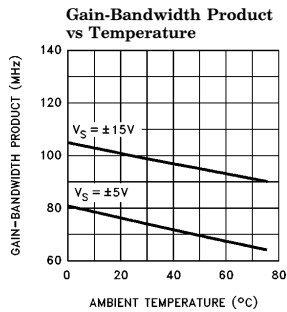
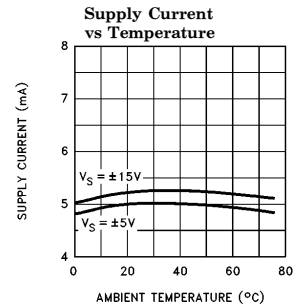
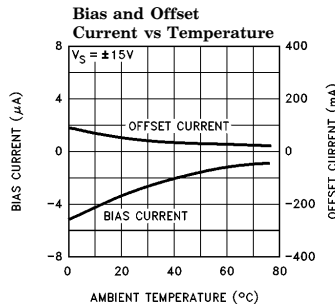
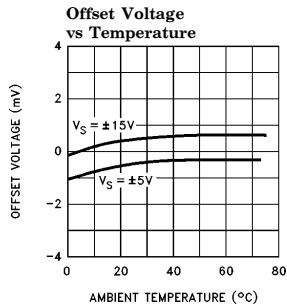
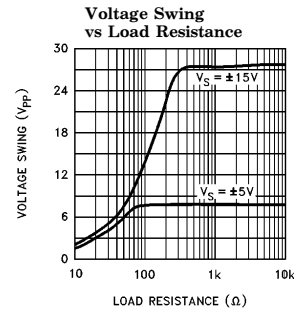
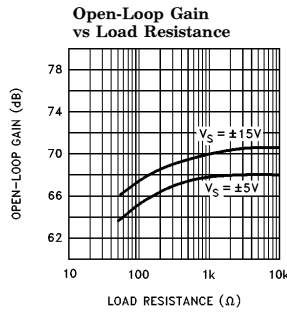
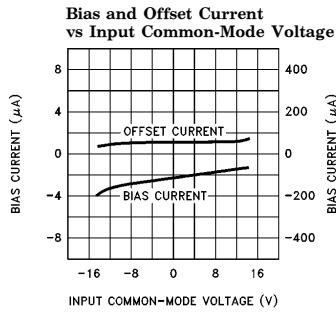
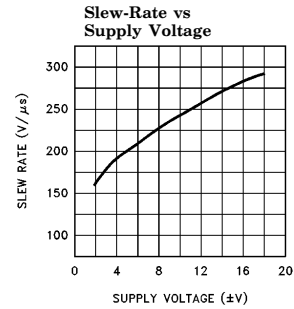
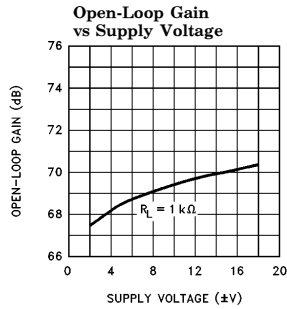
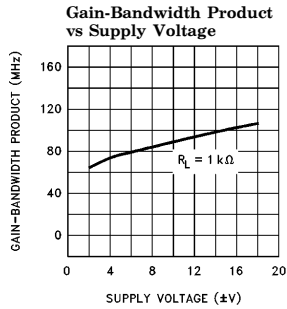
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EL2045C

Low-Power 100 MHz Gain-of-2 Stable Operational Amplifier

Typical Performance Curves

($T_A = 25^\circ\text{C}$, $R_f = 1\text{ k}\Omega$, $C_f = 3\text{ pF}$, $R_L = 1000\Omega$, $A_V = +2$ unless otherwise specified) — Contd.



2045-4

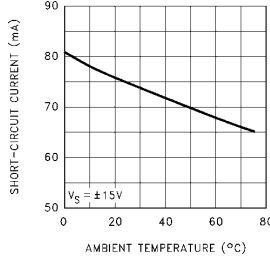
EL2045C

Low-Power 100 MHz Gain-of-2 Stable Operational Amplifier

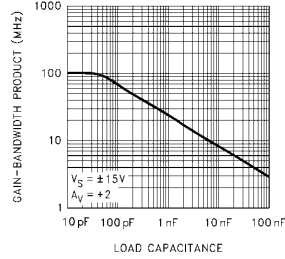
Typical Performance Curves

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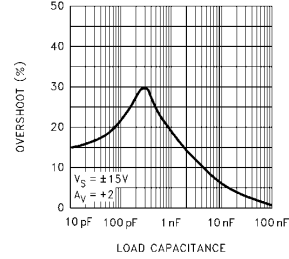
Short-Circuit Current
vs Temperature



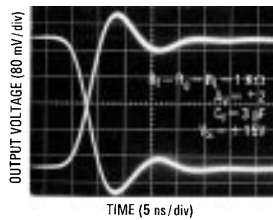
Gain-Bandwidth Product
vs Load Capacitance



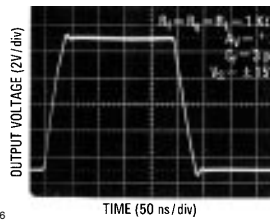
Overshoot vs
Load Capacitance



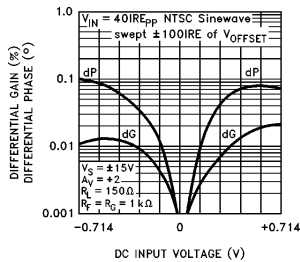
Small-Signal
Step Response



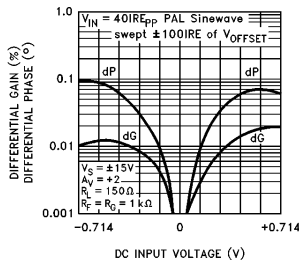
Large-Signal
Step Response



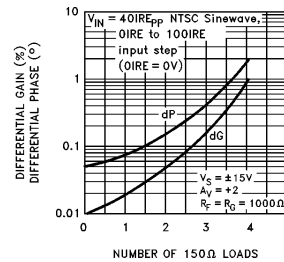
Differential Gain and
Phase vs DC Input
Offset at 3.58 MHz



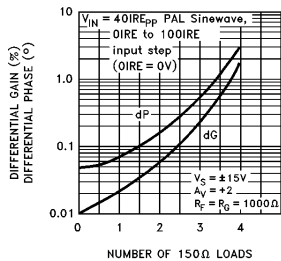
Differential Gain and
Phase vs DC Input
Offset at 4.43 MHz



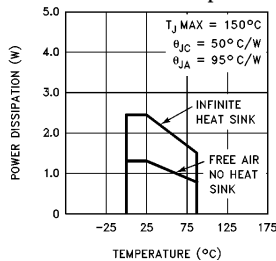
Differential Gain and
Phase vs Number of
150Ω Loads at 3.58 MHz



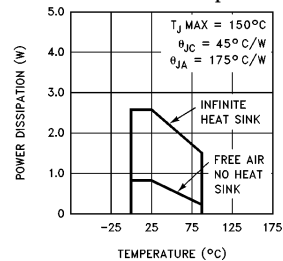
Differential Gain and
Phase vs Number of
150Ω Loads at 4.43 MHz



8-Pin Plastic DIP
Maximum Power Dissipation
vs Ambient Temperature



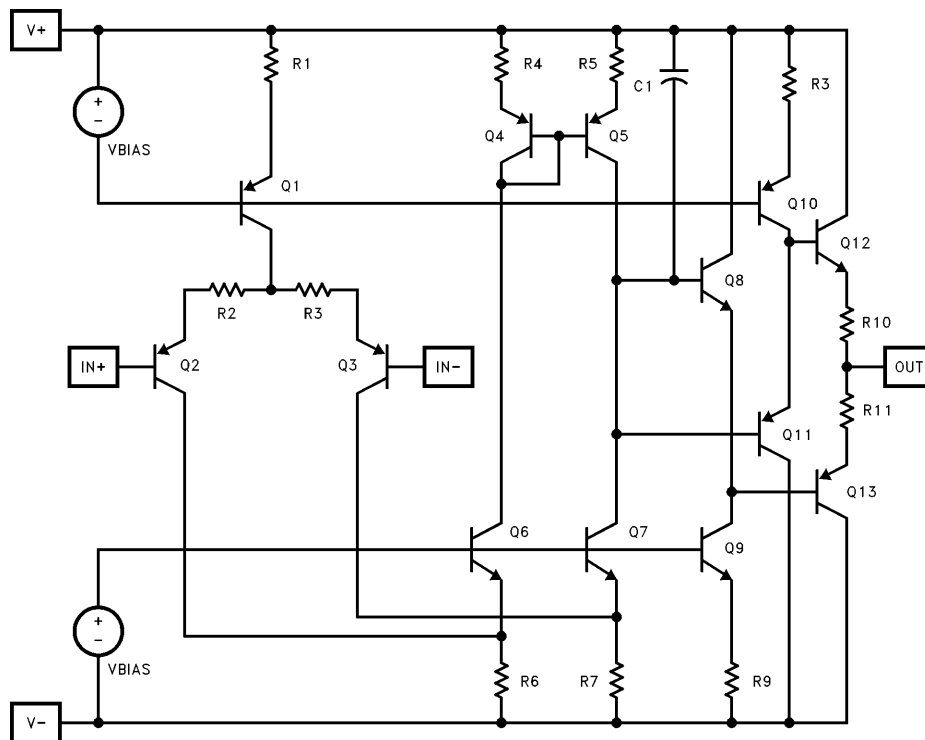
8-Lead SO
Maximum Power Dissipation
vs Ambient Temperature



EL2045C

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Simplified Schematic

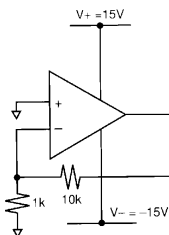


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EL2045C

Low-Power 100 MHz Gain-of-2 Stable Operational Amplifier

Burn-In Circuit



All Packages Use the Same Schematic

Applications Information

Product Description

The EL2045C is a low-power wideband, gain-of-2 stable monolithic operational amplifier built on Elantec's proprietary high-speed complementary bipolar process. The EL2045C uses a classical voltage-feedback topology which allows it to be used in a variety of applications where current-feedback amplifiers are not appropriate because of restrictions placed upon the feedback element used with the amplifier. The conventional topology of the EL2045C allows, for example, a capacitor to be placed in the feedback path, making it an excellent choice for applications such as active filters, sample-and-holds, or integrators. Similarly, because of the ability to use diodes in the feedback network, the EL2045C is an excellent choice for applications such as fast log amplifiers.

Single-Supply Operation

The EL2045C has been designed to have a wide input and output voltage range. This design also makes the EL2045C an excellent choice for single-supply operation. Using a single positive supply, the lower input voltage range is within 100 mV of ground ($R_L = 500\Omega$), and the lower output voltage range is within 300 mV of ground. Upper input voltage range reaches 4.2V, and output voltage range reaches 3.8V with a 5V supply and $R_L = 500\Omega$. This results in a 3.5V output swing on a single 5V supply. This wide output voltage range also allows single-supply operation with a supply voltage as high as 36V or as low as 2.5V. On a single 2.5V supply, the EL2045C still has 1V of output swing.

Gain-Bandwidth Product and the -3 dB Bandwidth

The EL2045C has a gain-bandwidth product of 100 MHz while using only 5.2 mA of supply current. For gains greater than 4, its closed-loop -3 dB bandwidth is approximately equal to the gain-bandwidth product divided by the noise gain of the circuit. For gains less than 4, higher-order poles in the amplifier's transfer function contribute to even higher closed loop bandwidths. For example, the EL2045C has a -3 dB bandwidth of 100 MHz at a gain of +2, dropping to 20 MHz at a gain of +5. It is important to note that the EL2045C has been designed so that this "extra" bandwidth in low-gain applications does not come at the expense of stability. As seen in the typical performance curves, the EL2045C in a gain of +2 only exhibits 1.0 dB of peaking with a 1000 Ω load.

Video Performance

An industry-standard method of measuring the video distortion of a component such as the EL2045C is to measure the amount of differential gain (dG) and differential phase (dP) that it introduces. To make these measurements, a 0.286 V_{PP} (40 IRE) signal is applied to the device with 0V DC offset (0 IRE) at either 3.58 MHz for NTSC or 4.43 MHz for PAL. A second measurement is then made at 0.714V DC offset (100 IRE). Differential gain is a measure of the change in amplitude of the sine wave, and is measured in percent. Differential phase is a measure of the change in phase, and is measured in degrees.

For signal transmission and distribution, a back-terminated cable (75 Ω in series at the drive end, and 75 Ω to ground at the receiving end) is preferred since the impedance match at both ends will absorb any reflections. However, when double termination is used, the received signal is halved; therefore a gain of 2 configuration is typically used to compensate for the attenuation.

The EL2045C has been designed as an economical solution for applications requiring low video distortion. It has been thoroughly characterized

EL2045C

Low-Power 100 MHz Gain-of-2 Stable Operational Amplifier

Applications Information — Contd.

for video performance in the topology described above, and the results have been included as typical dG and dP specifications and as typical performance curves. In a gain of +2, driving 150 Ω , with standard video test levels at the input, the EL2045C exhibits dG and dP of only 0.02% and 0.07° at NTSC and PAL. Because dG and dP can vary with different DC offsets, the video performance of the EL2045C has been characterized over the entire DC offset range from -0.714V to +0.714V. For more information, refer to the curves of dG and dP vs DC Input Offset.

The output drive capability of the EL2045C allows it to drive up to 2 back-terminated loads with good video performance. For more demanding applications such as greater output drive or better video distortion, a number of alternatives such as the EL2120, EL400, or EL2074 should be considered.

Output Drive Capability

The EL2045C has been designed to drive low impedance loads. It can easily drive 6 V_{pp} into a 150 Ω load. This high output drive capability makes the EL2045C an ideal choice for RF, IF and video applications. Furthermore, the current drive of the EL2045C remains a minimum of 35 mA at low temperatures. The EL2045C is current-limited at the output, allowing it to withstand shorts to ground. However, power dissipation with the output shorted can be in excess of the power-dissipation capabilities of the package.

Capacitive Loads

For ease of use, the EL2045C has been designed to drive any capacitive load. However, the EL2045C remains stable by automatically reducing its gain-bandwidth product as capacitive load increases. Therefore, for maximum bandwidth, capacitive loads should be reduced as much as possible or isolated via a series output resistor (R_s). Similarly, coax lines can be driven, but best AC performance is obtained when they are terminated with their characteristic impedance so that the capacitance of the coaxial cable will not add to the capacitive load seen by the amplifier. Al-

though stable with all capacitive loads, some peaking still occurs as load capacitance increases. A series resistor at the output of the EL2045C can be used to reduce this peaking and further improve stability.

Printed-Circuit Layout

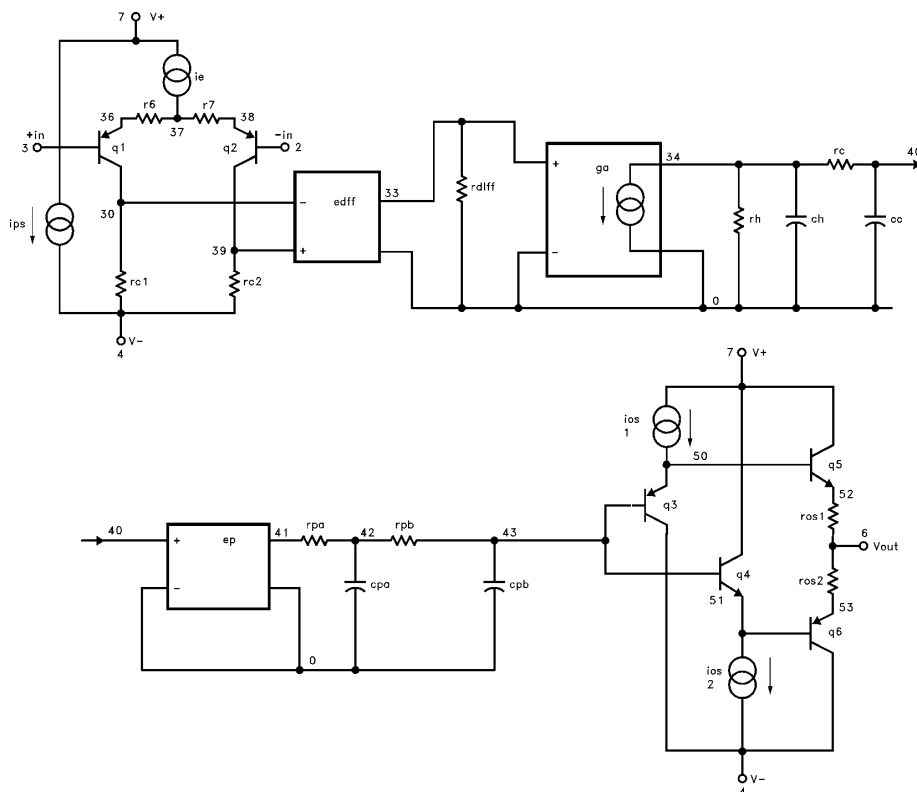
The EL2045C is well behaved, and easy to apply in most applications. However, a few simple techniques will help assure rapid, high quality results. As with any high-frequency device, good PCB layout is necessary for optimum performance. Ground-plane construction is highly recommended, as is good power supply bypassing. A 0.1 μ F ceramic capacitor is recommended for bypassing both supplies. Lead lengths should be as short as possible, and bypass capacitors should be as close to the device pins as possible. For good AC performance, parasitic capacitances should be kept to a minimum at both inputs and at the output. Resistor values should be kept under 5 k Ω because of the RC time constants associated with the parasitic capacitance. Metal-film and carbon resistors are both acceptable, use of wire-wound resistors is not recommended because of their parasitic inductance. Similarly, capacitors should be low-inductance for best performance.

The EL2045C Macromodel

This macromodel has been developed to assist the user in simulating the EL2045C with surrounding circuitry. It has been developed for the PSPICE simulator (copyrighted by the Microsim Corporation), and may need to be rearranged for other simulators. It approximates DC, AC, and transient response for resistive loads, but does not accurately model capacitive loading. This model is slightly more complicated than the models used for low-frequency op-amps, but it is much more accurate for AC analysis.

The model does not simulate these characteristics accurately:

noise	non-linearities
settling-time	temperature effects
CMRR	manufacturing variations
PSRR	

EL2045C**Low-Power 100 MHz Gain-of-2 Stable Operational Amplifier****EL2045C Macromodel — Contd.****EL2045C Model**

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General Disclaimer

Specifications contained in this data sheet are in effect as of the publication date shown. Elantec, Inc. reserves the right to make changes in the circuitry or specifications contained herein at any time without notice. Elantec, Inc. assumes no responsibility for the use of any circuits described herein and makes no representations that they are free from patent infringement.

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HIGH PERFORMANCE ANALOG INTEGRATED CIRCUITS

Elantec, Inc.

1996 Tarob Court

Milpitas, CA 95035

Telephone: (408) 945-1323

(800) 333-6314

Fax: (408) 945-9305

European Office: 44-71-482-4596

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