1.5 A, 3.3 V Fixed Linear Regulator

The CS52015–3 linear regulator provides 1.5 A @ 3.3 V reference at 1.0 A with an output voltage accuracy of ± 1.5 %.

The regulator is intended for use as a post regulator and microprocessor supply. The fast loop response and low dropout voltage make this regulator ideal for applications where low voltage operation and good transient response are important.

The circuit is designed to operate with dropout voltages less than 1.4 V at 1.5 A output current. The maximum quiescent current is only 10 mA at full load. Device protection includes over-current and thermal shutdown.

The CS52015–3 is pin compatible with the LT1086 family of linear regulators but has lower dropout voltage.

The regulator is available in TO-220, surface mount D^2 , and SOT-223 packages.

Features

- Output Current to 1.5 A
- Output Accuracy to ±1.5% Over Temperature
- Dropout Voltage (typical) 1.05 V @ 1.5 A
- Fast Transient Response
- Fault Protection
 - Current Limit
 - Thermal Shutdown

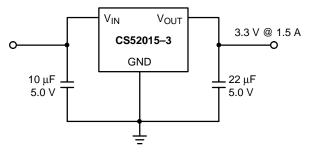


Figure 1. Applications Diagram



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TO-220 THREE LEAD T SUFFIX CASE 221A



D²PAK 3-PIN DP SUFFIX CASE 418E $Tab = V_{OUT}$ Pin 1. GND
2. V_{OUT}
3. V_{IN}



SOT-223 ST SUFFIX CASE 318E

ORDERING INFORMATION*†

Device	Package	Shipping
CS52015-3GT3	TO-220‡	50 Units/Rail
CS52015-3GDP3	D ² PAK‡	50 Units/Rail
CS52015-3GDPR3	D ² PAK‡	750 Tape & Reel
CS52015-3GST3	SOT-223‡	80 Units/Rail
CS52015–3GSTR3	SOT-223‡	2500 Tape & Reel

- *Additional ordering information can be found on page 6 of this data sheet.
- †Consult your local sales representative for other fixed output voltage versions.
- ‡TO-220 are all 3-pin, straight leaded. D²PAK and SOT-223 are all 3-pin.

DEVICE MARKING INFORMATION

See general marking information in the device marking section on page 6 of this data sheet.

MAXIMUM RATINGS*

Parameter	Value	Unit
Supply Voltage, V _{IN}	7.0	V
Operating Temperature Range	-40 to +70	°C
Junction Temperature	150	°C
Storage Temperature Range	-60 to +150	°C
Lead Temperature Soldering: Wave Solder (through hole styles only) Note 1 Reflow (SMD styles only) Note 2	260 Peak 230 Peak	°C °C
ESD Damage Threshold	2.0	kV

- 1. 10 second maximum.
- 2. 60 second maximum above 183°C

ELECTRICAL CHARACTERISTICS (C_{IN} = 10 μ F, C_{OUT} = 22 μ F Tantalum, V_{OUT} + $V_{DROPOUT}$ < V_{IN} < 7.0 V, 0°C \leq $T_A \leq$ 70°C, $T_J \leq$ +150°C, unless otherwise specified, $I_{full\ load}$ = 1.5 A)

Characteristic	Test Conditions	Min	Тур	Max	Unit		
Fixed Output Voltage							
Output Voltage (Notes 3 and 4)	$V_{IN} - V_{OUT} = 1.5 \text{ V};$ $0 \le I_{OUT} \le 1.5 \text{ A}$	3.250 (-1.5%)	3.300	3.350 (+1.5%)	V		
Line Regulation	$2.0 \text{ V} \le V_{IN} - V_{OUT} \le 3.7 \text{ V}; I_{OUT} = 10 \text{ mA}$	_	0.02	0.20	%		
Load Regulation (Notes 3 and 4)	$V_{IN} - V_{OUT} = 2.0 \text{ V}; 10 \text{ mA} \le I_{OUT} \le 1.5 \text{ A}$	-	0.04	0.4	%		
Dropout Voltage (Note 5)	I _{OUT} = 1.5 A	-	1.05	1.4	V		
Current Limit	V _{IN} – V _{OUT} = 3.0 V	1.6	3.1	-	Α		
Quiescent Current	I _{OUT} = 10 mA	_	5.0	10	mA		
Thermal Regulation (Note 6)	30 ms Pulse, T _A = 25°C	_	0.002	0.020	%/W		
Ripple Rejection (Note 6)	f = 120 Hz; I _{OUT} = 1.5 A; V _{IN} – V _{OUT} = 3.0 V; V _{RIPPLE} = 1.0 V _{PP}	-	80	-	dB		
Thermal Shutdown (Note 7)	-	150	180	210	°C		
Thermal Shutdown Hysteresis (Note 7)	-	-	25	-	°C		

- 3. Load regulation and output voltage are measured at a constant junction temperature by low duty cycle pulse testing. Changes in output voltage due to temperature changes must be taken into account seperately.
- 4. Specifications apply for an external Kelvin sense connection at a point on the output pin 1/4" from the bottom of the package.
- 5. Dropout voltage is a measurement of the minimum input/output differential at full load.
- 6. Guaranteed by design, not 100% tested in production.
- 7. Thermal shutdown is 100% functionally tested in production.

PACKAGE PIN DESCRIPTION

Pack	Package Pin Number				
TO-220	D ² PAK	SOT-223	Pin Symbol	Function	
1	1	1	GND	Ground connection.	
2	2	2	V _{OUT}	Regulated output voltage (case).	
3	3	3	V _{IN}	Input voltage.	

^{*}The maximum package power dissipation must be observed.

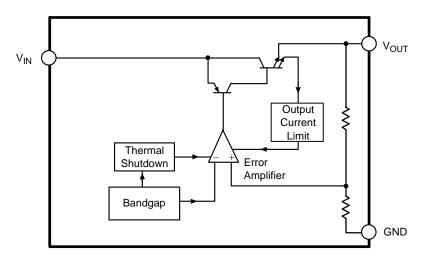


Figure 2. Block Diagram

TYPICAL PERFORMANCE CHARACTERISTICS

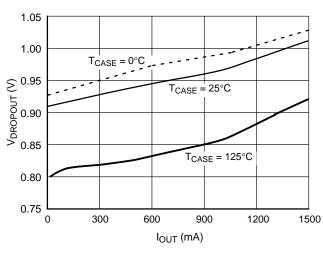


Figure 3. Dropout Voltage vs. Output Current

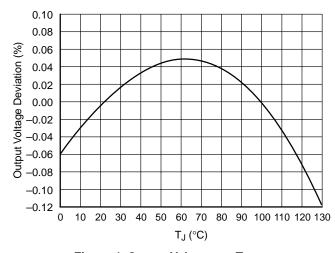


Figure 4. Output Voltage vs. Temperature

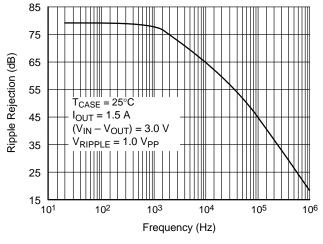


Figure 5. Ripple Rejection vs. Frequency

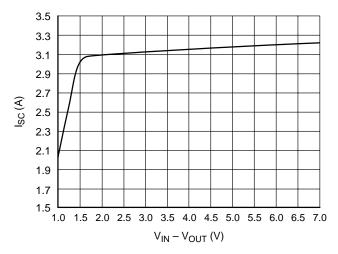


Figure 6. Short Circuit Current vs. $\label{eq:Vin-Vout} V_{IN} - V_{OUT}$

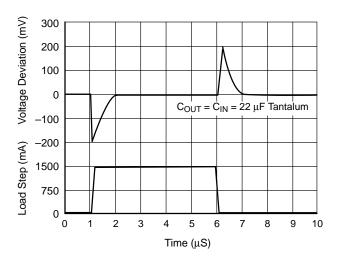


Figure 7. Transient Response

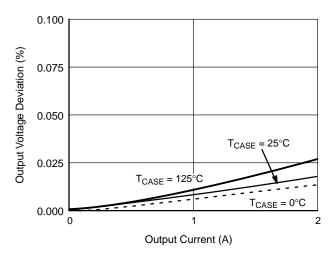


Figure 8. Load Regulation vs. Output Current

APPLICATIONS INFORMATION

The CS52015–3 linear regulator provides a 3.3 V output voltage at currents up to 1.5 A. The regulator is protected against overcurrent conditions and includes thermal shutdown.

The CS52015–3 has a composite PNP–NPN output transistor and requires an output capacitor for stability. A detailed procedure for selecting this capacitor is included in the Stability Considerations section.

Stability Considerations

The output compensation capacitor helps determine three main characteristics of a linear regulator: start—up delay, load transient response, and loop stability.

The capacitor value and type is based on cost, availability, size and temperature constraints. A tantalum or aluminum electrolytic capacitor is best, since a film or ceramic capacitor with almost zero ESR can cause instability. The aluminum electrolytic capacitor is the least expensive solution. However, when the circuit operates at low temperatures, both the value and ESR of the capacitor will vary considerably. The capacitor manufacturer's data sheet provides this information.

A 22 μ F tantalum capacitor will work for most applications, but with high current regulators such as the CS52015–3 the transient response and stability improve with higher values of capacitance. The majority of applications for this regulator involve large changes in load current so the output capacitor must supply the instantaneous load current. The ESR of the output capacitor causes an immediate drop in output voltage given by:

$$\Delta V = \Delta I \times ESR$$

For microprocessor applications it is customary to use an output capacitor network consisting of several tantalum and ceramic capacitors in parallel. This reduces the overall ESR and reduces the instantaneous output voltage drop under transient load conditions. The output capacitor network should be as close to the load as possible for the best results.

Protection Diodes

When large external capacitors are used with a linear regulator it is sometimes necessary to add protection diodes. If the input voltage of the regulator gets shorted, the output capacitor will discharge into the output of the regulator. The discharge current depends on the value of the capacitor, the output voltage and the rate at which $V_{\rm IN}$ drops. In the CS52015–3 linear regulator, the discharge path is through a large junction and protection diodes are not usually needed. If the regulator is used with large values of output capacitance and the input voltage is instantaneously shorted to ground, damage can occur. In this case, a diode connected as shown in Figure 9 is recommended.

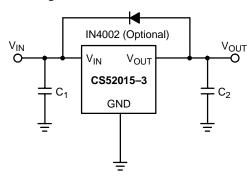


Figure 9. Protection Diode Scheme for Large Output Capacitors

Output Voltage Sensing

Since the CS52015–3 is a three terminal regulator, it is not possible to provide true remote load sensing. Load regulation is limited by the resistance of the conductors connecting the regulator to the load. For best results the regulator should be connected as shown in Figure 10.

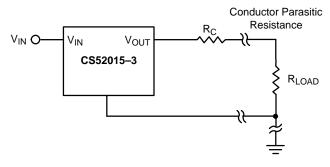


Figure 10. Conductor Parasitic Resistance Effects
Can Be Minimized With the Above Grounding
Scheme For Fixed Output Regulators

Calculating Power Dissipation and Heat Sink Requirements

The CS52015–3 linear regulator includes thermal shutdown and current limit circuitry to protect the device. High power regulators such as these usually operate at high junction temperatures so it is important to calculate the power dissipation and junction temperatures accurately to ensure that an adequate heat sink is used.

The case is connected to V_{OUT} on the CS52015–3, electrical isolation may be required for some applications. Thermal compound should always be used with high current regulators such as these.

The thermal characteristics of an IC depend on the following four factors:

- 1. Maximum Ambient Temperature T_A (°C)
- 2. Power dissipation P_D (Watts)
- 3. Maximum junction temperature T_J (°C)
- 4. Thermal resistance junction to ambient R_{ΘJA} (°C/W)

These four are related by the equation

$$T_{J} = T_{A} + P_{D} \times R_{\Theta J A} \tag{1}$$

The maximum ambient temperature and the power dissipation are determined by the design while the maximum junction temperature and the thermal resistance depend on the manufacturer and the package type.

The maximum power dissipation for a regulator is:

$$PD(max) = {VIN(max) - VOUT(min)}IOUT(max) + VIN(max)IQ$$
(2)

where:

V_{IN(max)} is the maximum input voltage,

V_{OUT(min)} is the minimum output voltage,

 $I_{OUT(max)}$ is the maximum output current, for the application

I_O is the maximum quiescent current at I_{OUT(max)}.

A heat sink effectively increases the surface area of the package to improve the flow of heat away from the IC and into the surrounding air.

Each material in the heat flow path between the IC and the outside environment has a thermal resistance. Like series electrical resistances, these resistances are summed to determine $R_{\Theta JA}$, the total thermal resistance between the junction and the surrounding air.

- Thermal Resistance of the junction to case, R_{ΘJC} (°C/W)
- 2. Thermal Resistance of the case to Heat Sink, $R_{\Theta CS}$ (°C/W)
- 3. Thermal Resistance of the Heat Sink to the ambient air, $R_{\Theta SA}$ (°C/W)

These are connected by the equation:

$$R_{\Theta,JA} = R_{\Theta,JC} + R_{\Theta,CS} + R_{\Theta,SA}$$
 (3)

The value for $R_{\Theta JA}$ is calculated using equation (3) and the result can be substituted in equation (1).

The value for $R_{\Theta JC}$ is 3.5°C/W for a given package type based on an average die size. For a high current regulator such as the CS52015–3 the majority of the heat is generated in the power transistor section. The value for $R_{\Theta SA}$ depends on the heat sink type, while $R_{\Theta CS}$ depends on factors such as package type, heat sink interface (is an insulator and thermal grease used?), and the contact area between the heat sink and the package. Once these calculations are complete, the maximum permissible value of $R_{\Theta JA}$ can be calculated and the proper heat sink selected. For further discussion on heat sink selection, see application note "Thermal Management," document number AND8036/D, available through the Literature Distribution Center or via our website at http://onsemi.com.

ADDITIONAL ORDERING INFORMATION

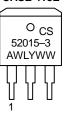
Orderable Part Number	Туре	Description
CS52015-3GT3	1.5 A, 3.3 V Output	TO-220 THREE LEAD, STRAIGHT
CS52015-3GDP3	1.5 A, 3.3 V Output	D ² PAK 3–PIN
CS52015-3GDPR3	1.5 A, 3.3 V Output	D ² PAK 3–PIN (Tape & Reel)
CS52015-3GST3	1.5 A, 3.3 V Output	SOT-223
CS52015-3GSTR3	1.5 A, 3.3 V Output	SOT-223 (Tape & Reel)

MARKING DIAGRAMS









SOT-223 ST SUFFIX CASE 318E

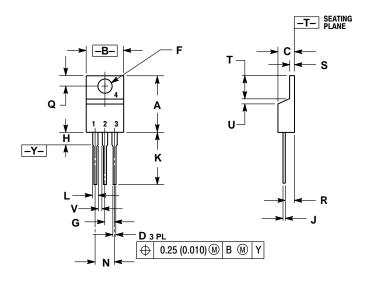


A = Assembly Location

WL, L = Wafer Lot YY, Y = Year WW, W = Work Week

PACKAGE DIMENSIONS

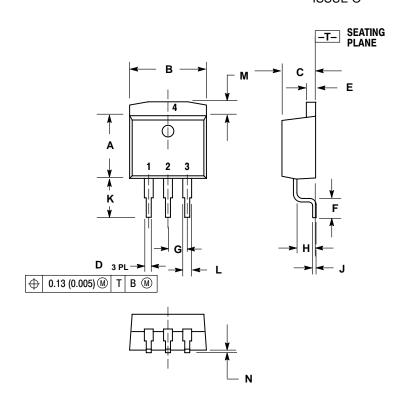
TO-220 **THREE LEAD** T SUFFIX CASE 221A-08 **ISSUE AA**



- NOTES:
 1. DIMENSIONING AND TOLERANCING PER ANSI Y14.5M, 1982.
 2. CONTROLLING DIMENSION: INCH.

	INCHES		MILLIN	IETERS
DIM	MIN MAX		MIN	MAX
Α	0.560	0.625	14.23	15.87
В	0.380	0.420	9.66	10.66
С	0.140	0.190	3.56	4.82
D	0.025	0.035	0.64	0.89
F	0.139	0.155	3.53	3.93
G	0.100	BSC	2.54 BSC	
Н		0.280		7.11
J	0.012	0.045	0.31	1.14
K	0.500	0.580	12.70	14.73
L	0.045	0.060	1.15	1.52
N	0.200	BSC	5.08 BSC	
Q	0.100	0.135	2.54	3.42
R	0.080	0.115	2.04	2.92
S	0.020	0.055	0.51	1.39
T	0.235	0.255	5.97	6.47
U	0.000	0.050	0.00	1.27
٧	0.045		1.15	

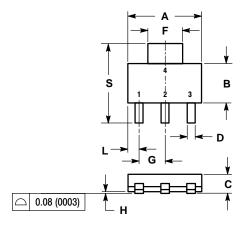
D²PAK 3-PIN **DP SUFFIX** CASE 418E-01 ISSUE O

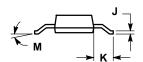


- NOTES:
 1. DIMENSIONS AND TOLERANCING PER ANSI Y14.5M, 1982.
 2. CONTROLLING DIMENSION: INCH.

	INCHES		MILLIN	IETERS
DIM	MIN	MAX	MIN	MAX
Α	0.326	0.336	8.28	8.53
В	0.396	0.406	10.05	10.31
С	0.170	0.180	4.31	4.57
D	0.026	0.036	0.66	0.91
Е	0.045	0.055	1.14	1.40
F	0.090	0.110	2.29	2.79
G	0.10	0 BSC	2.54 BSC	
Н	0.098	0.108	2.49	2.74
J	0.018	0.025	0.46	0.64
K	0.204	0.214	5.18	5.44
L	0.045	0.055	1.14	1.40
M	0.055	0.066	1.40	1.68
N	0.000	0.004	0.00	0.10

SOT-223 ST SUFFIX CASE 318E-04 ISSUE K





NOTES:

- DIMENSIONING AND TOLERANCING PER ANSI Y14.5M. 1982.
- CONTROLLING DIMENSION: INCH.

	INCHES		MILLIMETER	
DIM	MIN MAX		MIN	MAX
Α	0.249	0.263	6.30	6.70
В	0.130	0.145	3.30	3.70
С	0.060	0.068	1.50	1.75
D	0.024 0.035		0.60	0.89
F	0.115	0.126	2.90	3.20
G	0.087	0.094	2.20	2.40
Н	0.0008	0.0040	0.020	0.100
J	0.009	0.014	0.24	0.35
K	0.060	0.078	1.50	2.00
L	0.033	0.041	0.85	1.05
M	0°	10 °	0 °	10 °
S	0.264	0.287	6.70	7.30

PACKAGE THERMAL DATA

Parameter		TO-220 THREE LEAD	D ² PAK 3–PIN	SOT-223	Unit
$R_{\Theta JC}$	Typical	3.5	3.5	15	°C/W
$R_{\Theta JA}$	Typical	50	10–50*	156	°C/W

^{*} Depending on thermal properties of substrate. $R_{\Theta,JA} = R_{\Theta,JC} + R_{\Theta,CA}$

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