

Rail-to-Rail Input/Output, 10 MHz Op Amps

Features

- Rail-to-Rail Input/Output
- Wide Bandwidth: 10 MHz (typ.)
- Low Noise: 8.7 nV/√Hz, at 10 kHz (typ.)
- Low Offset Voltage:
 - Industrial Temperature: ±500 µV (max.)
- Extended Temperature: ±250 µV (max.)
- Mid-Supply V_{RFF}: MCP6021 and MCP6023
- Low Supply Current: 1 mA (typ.)
- Total Harmonic Distortion: 0.00053% (typ., G = 1)
- Unity Gain Stable
- Power Supply Range: 2.5V to 5.5V
- Temperature Range:
 - Industrial: -40°C to +85°C
 - Extended: -40°C to +125°C

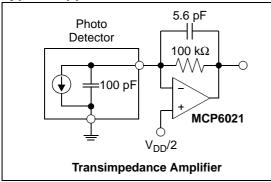
Typical Applications

- Automotive
- Driving A/D Converters
- Multi-Pole Active Filters
- Barcode Scanners
- Audio Processing
- Communications
- DAC Buffer
- Test Equipment
- Medical Instrumentation

Available Tools

- SPICE Macro Model (at www.microchip.com)
- FilterLab[®] software (at www.microchip.com)

Typical Application



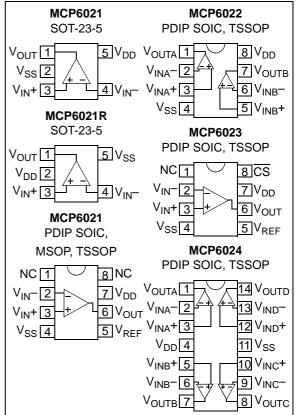
Description

The MCP6021, MCP6021R, MCP6022, MCP6023 and MCP6024 from Microchip Technology Inc. are rail-to-rail input and output op amps with high performance. Key specifications include: wide bandwidth (10 MHz), low noise (8.7 nV/ \sqrt{Hz}), low input offset voltage and low distortion (0.00053% THD+N). The MCP6023 also offers a Chip Select pin (CS) that gives power savings when the part is not in use.

The single MCP6021 and MCP6021R are available in SOT-23-5. The single MCP6021, single MCP6023 and dual MCP6022 are available in 8-lead PDIP, SOIC and TSSOP. The Extended Temperature single MCP6021 is available in 8-lead MSOP. The quad MCP6024 is offered in 14-lead PDIP, SOIC and TSSOP packages.

The MCP6021/1R/2/3/4 family is available in Industrial and Extended temperature ranges. It has a power supply range of 2.5V to 5.5V.

Package Types



1.0 ELECTRICAL CHARACTERISTICS

Absolute Maximum Ratings †

$V_{DD} - V_{SS}$	7.0V
All Inputs and Outputs V _{SS} - 0.3V	' to V _{DD} + 0.3V
Difference Input Voltage	V _{DD} – V _{SS}
Output Short Circuit Current	continuous
Current at Input Pins	±2 mA
Current at Output and Supply Pins	±30 mA
Storage Temperature6	5°C to +150°C
Junction Temperature	+150°C
ESD Protection on all pins (HBM; MM)	≥2 kV; 200V

† Notice: Stresses above those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. This is a stress rating only and functional operation of the device at those or any other conditions above those indicated in the operational listings of this specification is not implied. Exposure to maximum rating conditions for extended periods may affect device reliability.

DC ELECTRICAL CHARACTERISTICS

Electrical Specifications: Unless otherwise indicated, $T_A = +25^{\circ}C$, $V_{DD} = +2.5V$ to +5.5V, $V_{SS} = GND$, $V_{CM} = V_{DD}/2$, $V_{OUT} \approx V_{DD}/2$ and $R_L = 10 \text{ k}\Omega$ to $V_{DD}/2$.

Parameters	Sym	Min	Тур	Max	Units	Conditions
Input Offset						
Input Offset Voltage:						
Industrial Temperature Parts	V _{OS}	-500	_	+500	μV	$V_{CM} = 0V$
Extended Temperature Parts	V _{OS}	-250	_	+250	μV	V _{CM} = 0V, V _{DD} = 5.0V
Extended Temperature Parts	V _{OS}	-2.5	—	+2.5	mV	$V_{CM} = 0V, V_{DD} = 5.0V$ $T_{A} = -40^{\circ}C$ to +125°C
Input Offset Voltage Temperature Drift	$\Delta V_{OS} / \Delta T_A$	_	±3.5	_	µV/°C	T _A = -40°C to +125°C
Power Supply Rejection Ratio	PSRR	74	90	_	dB	$V_{CM} = 0V$
Input Current and Impedance						
Input Bias Current	I _B	_	1	_	pА	
Industrial Temperature Parts	I _B		30	150	pА	T _A = +85°C
Extended Temperature Parts	I _B	—	640	5,000	pА	T _A = +125°C
Input Offset Current	I _{OS}	_	±1	_	pА	
Common-Mode Input Impedance	Z _{CM}	_	10 ¹³ 6	_	Ω pF	
Differential Input Impedance	Z _{DIFF}	_	10 ¹³ 3	—	Ω pF	
Common-Mode						
Common-Mode Input Range	V _{CMR}	V _{SS} -0.3	—	V _{DD} +0.3	V	
Common-Mode Rejection Ratio	CMRR	74	90	_	dB	V_{DD} = 5V, V_{CM} = -0.3V to 5.3V
	CMRR	70	85	_	dB	V_{DD} = 5V, V_{CM} = 3.0V to 5.3V
	CMRR	74	90	—	dB	V_{DD} = 5V, V_{CM} = -0.3V to 3.0V
Voltage Reference (MCP6021 and MC	P6023 only)					
V _{REF} Accuracy (V _{REF} – V _{DD} /2)	V _{REF_ACC}	-50	—	+50	mV	
V _{REF} Temperature Drift	$\Delta V_{REF} / \Delta T_A$	_	±100	_	µV/°C	T _A = -40°C to +125°C
Open-Loop Gain						
DC Open-Loop Gain (Large Signal)	A _{OL}	90	110	—	dB	$V_{CM} = 0V$, $V_{OUT} = V_{SS}+0.3V$ to $V_{DD}-0.3V$
Output						
Maximum Output Voltage Swing	V _{OL} , V _{OH}	V _{SS} +15	—	V _{DD} -20	mV	0.5V output overdrive
Output Short Circuit Current	I _{SC}	_	±30	_	mA	$V_{DD} = 2.5 V$
	I _{SC}		±22	_	mA	V _{DD} = 5.5V
Power Supply						
Supply Voltage	Vs	2.5	—	5.5	V	
Quiescent Current per Amplifier	ا _م	0.5	1.0	1.35	mA	$I_{O} = 0$

AC ELECTRICAL CHARACTERISTICS

Electrical Specifications: Unless otherwise indicated, $T_A = +25^{\circ}C$, $V_{DD} = +2.5V$ to +5.5V, $V_{SS} = GND$, $V_{CM} = V_{DD}/2$, $V_{OUT}\approx V_{DD}/2,~R_L~$ = 10 k Ω to $V_{DD}/2$ and C_L = 60 pF. Parameters Max Units Conditions Sym Min Тур AC Response Gain Bandwidth Product GBWP 10 MHz _ _ Phase Margin at Unity-Gain ΡM _ 65 _ G = +1 Settling Time, 0.2% 250 $G = +1, V_{OUT} = 100 \text{ mV}_{p-p}$ ns **t**SETTLE Slew Rate SR 7.0 V/µs ____ _ **Total Harmonic Distortion Plus Noise** $V_{OUT} = 0.25V$ to 3.25V (1.75V ± 1.50V_{PK}), f = 1 kHz, G = +1 V/VTHD+N 0.00053 % $V_{DD} = 5.0V, BW = 22 \text{ kHz}$ $V_{OUT} = 0.25V$ to 3.25V (1.75V ± 1.50V_{PK}), $f = 1 \text{ kHz}, G = +1 \text{ V/V}, R_L = 600\Omega$ THD+N ____ 0.00064 ____ % V_{DD} = 5.0V, BW = 22 kHz f = 1 kHz, G = +1 V/VTHD+N _ 0.0014 _ % $V_{OUT} = 4V_{P-P}, V_{DD} = 5.0V, BW = 22 \text{ kHz}$ f = 1 kHz, G = +10 V/V THD+N 0.0009 % $V_{OUT} = 4V_{P-P}, V_{DD} = 5.0V, BW = 22 \text{ kHz}$ f = 1 kHz, G = +100 V/V THD+N 0.005 % $V_{OUT} = 4V_{P-P}, V_{DD} = 5.0V, BW = 22 \text{ kHz}$ Noise Input Noise Voltage E_{ni} 2.9 ____ μ Vp-p f = 0.1 Hz to 10 Hz Input Noise Voltage Density 8.7 nV/\sqrt{Hz} f = 10 kHz e_{ni} Input Noise Current Density fA/\sqrt{Hz} f = 1 kHz____ 3 _ i_{ni}

MCP6023 CHIP SELECT (CS) ELECTRICAL CHARACTERISTICS

Parameters	Sym	Min	Тур	Max	Units	Conditions
CS Low Specifications						
CS Logic Threshold, Low	V _{IL}	V _{SS}		0.2 V _{DD}	V	
CS Input Current, Low	I _{CSL}	-1.0	0.01	_	μA	$\overline{\text{CS}} = \text{V}_{\text{SS}}$
CS High Specifications						
CS Logic Threshold, High	V _{IH}	0.8 V _{DD}		V _{DD}	V	
CS Input Current, High	I _{CSH}		0.01	2.0	μA	$\overline{\text{CS}} = \text{V}_{\text{DD}}$
GND Current	I _{SS}	-2	-0.05	_	μA	$\overline{\text{CS}} = \text{V}_{\text{DD}}$
Amplifier Output Leakage	I _{O(LEAK)}		0.01	_	μA	$\overline{\text{CS}} = \text{V}_{\text{DD}}$
CS Dynamic Specifications						
CS Low to Amplifier Output Turn-on Time	t _{ON}		2	10	μs	$\frac{G}{CS} = +1, V_{IN} = V_{SS},$ $\frac{G}{CS} = 0.2V_{DD} \text{ to } V_{OUT} = 0.45V_{DD} \text{ time}$
CS High to Amplifier Output High-Z Time	t _{OFF}	—	0.01	—	μs	$\frac{G = +1, V_{IN} = V_{SS},}{CS = 0.8V_{DD} \text{ to } V_{OUT} = 0.05V_{DD} \text{ time}}$
Hysteresis	V _{HYST}	_	0.6	_	V	V _{DD} = 5.0V, Internal Switch

TEMPERATURE CHARACTERISTICS

Parameters	Sym	Min	Тур	Max	Units	Conditions
Temperature Ranges						
Industrial Temperature Range	T _A	-40	_	+85	°C	
Extended Temperature Range	T _A	-40	_	+125	°C	
Operating Temperature Range	T _A	-40	_	+125	°C	Note 1
Storage Temperature Range	T _A	-65	_	+150	°C	
Thermal Package Resistances						
Thermal Resistance, 5L-SOT-23	θ_{JA}	_	256	_	°C/W	
Thermal Resistance, 8L-PDIP	θ_{JA}	_	85	—	°C/W	
Thermal Resistance, 8L-SOIC	θ_{JA}	_	163	—	°C/W	
Thermal Resistance, 8L-MSOP	θ_{JA}	_	206	_	°C/W	
Thermal Resistance, 8L-TSSOP	θ_{JA}	_	124	_	°C/W	
Thermal Resistance, 14L-PDIP	θ_{JA}	_	70	—	°C/W	
Thermal Resistance, 14L-SOIC	θ_{JA}	_	120	—	°C/W	
Thermal Resistance, 14L-TSSOP	θ_{JA}	_	100	_	°C/W	

Note 1: The industrial temperature devices operate over this extended temperature range, but with reduced performance. In any case, the internal junction temperature (T_J) must not exceed the absolute maximum specification of 150°C.

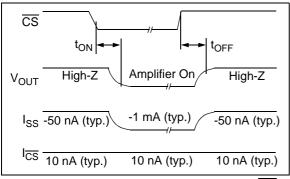
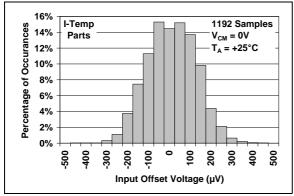
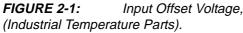


FIGURE 1-1: Timing diagram for the \overline{CS} pin on the MCP6023.

2.0 TYPICAL PERFORMANCE CURVES

Note: The graphs and tables provided following this note are a statistical summary based on a limited number of samples and are provided for informational purposes only. The performance characteristics listed herein are not tested or guaranteed. In some graphs or tables, the data presented may be outside the specified operating range (e.g., outside specified power supply range) and therefore outside the warranted range.





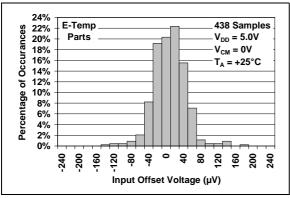


FIGURE 2-2: Input Offset Voltage, (Extended Temperature Parts).

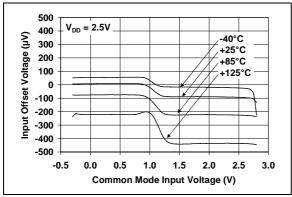


FIGURE 2-3: Input Offset Voltage vs. Common Mode Input Voltage with $V_{DD} = 2.5V$.

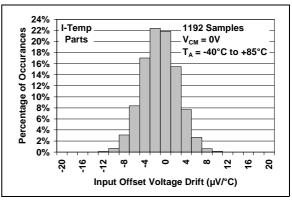


FIGURE 2-4: Input Offset Voltage Drift, (Industrial Temperature Parts).

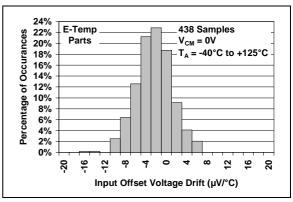


FIGURE 2-5: Input Offset Voltage Drift, (Extended Temperature Parts).

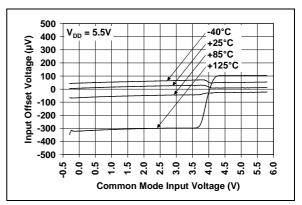


FIGURE 2-6: Input Offset Voltage vs. Common Mode Input Voltage with $V_{DD} = 5.5V$.

Note: Unless otherwise indicated, $T_A = +25^{\circ}C$, $V_{DD} = +2.5V$ to +5.5V, $V_{SS} = GND$, $V_{CM} = V_{DD}/2$, $V_{OUT} \approx V_{DD}/2$, $R_L = 10 \text{ k}\Omega$ to $V_{DD}/2$ and $C_L = 60 \text{ pF}$.

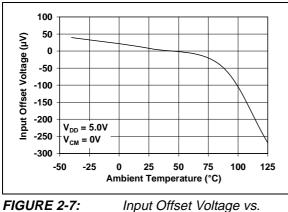


FIGURE 2-7: Temperature.

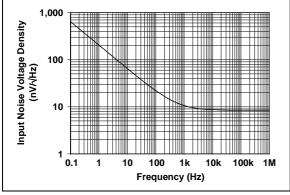
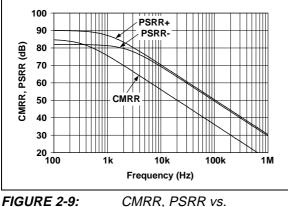


FIGURE 2-8: I vs. Frequency.

Input Noise Voltage Density



Frequency.

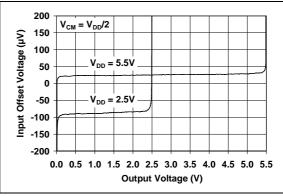


FIGURE 2-10: Input Offset Voltage vs. Output Voltage.

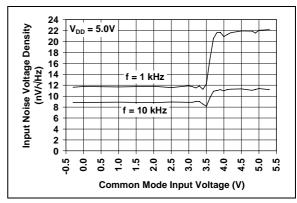


FIGURE 2-11: Input Noise Voltage Density vs. Common Mode Input Voltage.

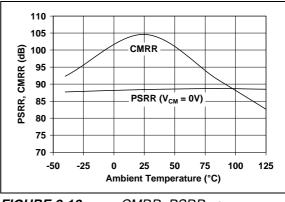


FIGURE 2-12: C Temperature.

CMRR, PSRR vs.

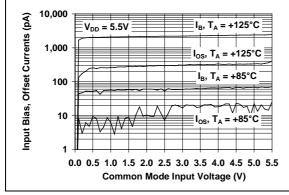


FIGURE 2-13: Input Bias, Offset Currents vs. Common Mode Input Voltage.

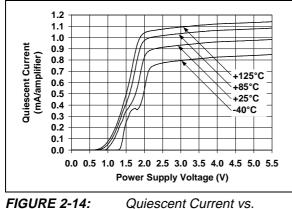


FIGURE 2-14: Supply Voltage.

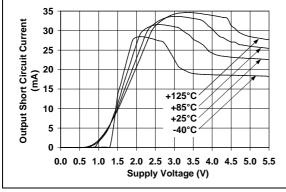


FIGURE 2-15: Output Short-Circuit Current vs. Supply Voltage.

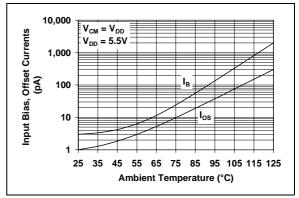


FIGURE 2-16: Input Bias, Offset Currents vs. Temperature.

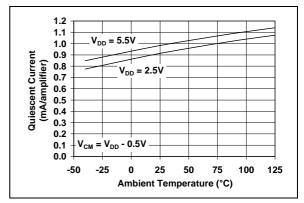


FIGURE 2-17: Quiescent Current vs. Temperature.

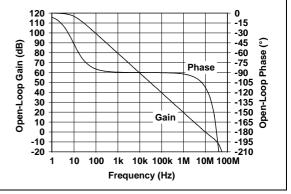


FIGURE 2-18: Open-Loop Gain, Phase vs. Frequency.

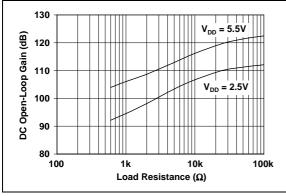
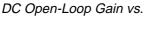


FIGURE 2-19: DC C Load Resistance.



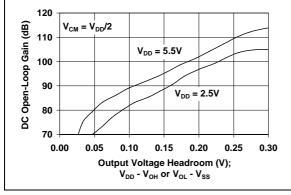


FIGURE 2-20: Small Signal DC Open-Loop Gain vs. Output Voltage Headroom.

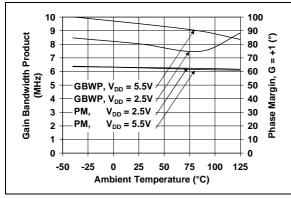


FIGURE 2-21: Gain Bandwidth Product, Phase Margin vs. Temperature.

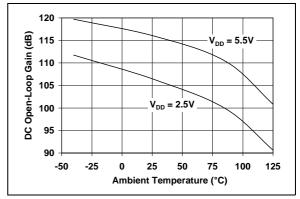


FIGURE 2-22: DC Open-Loop Gain vs. Temperature.

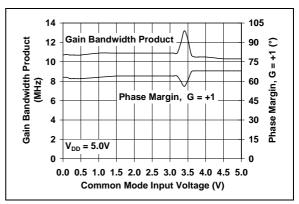


FIGURE 2-23: Gain Bandwidth Product, Phase Margin vs. Common Mode Input Voltage.

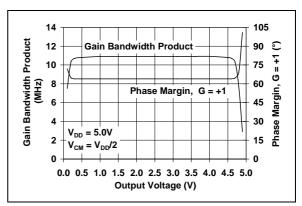


FIGURE 2-24: Gain Bandwidth Product, Phase Margin vs. Output Voltage.

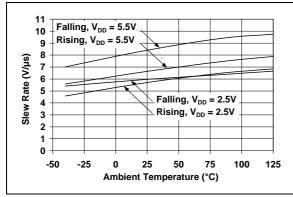


FIGURE 2-25: Slew Rate vs. Temperature.

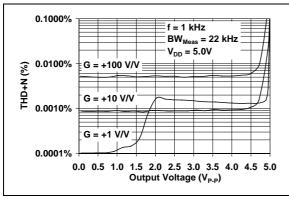


FIGURE 2-26: Total Harmonic Distortion plus Noise vs. Output Voltage with f = 1 kHz.

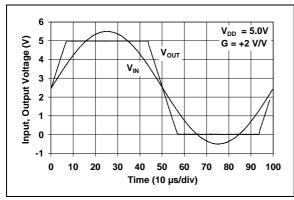


FIGURE 2-27: The MCP6021/1R/2/3/4 family shows no phase reversal under overdrive.

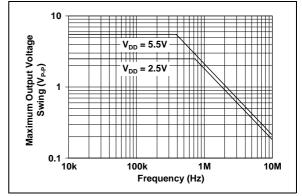


FIGURE 2-28: Maximum Output Voltage Swing vs. Frequency.

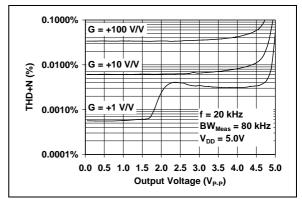


FIGURE 2-29: Total Harmonic Distortion plus Noise vs. Output Voltage with f = 20 kHz.

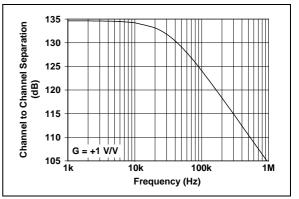


FIGURE 2-30: Channel-to-Channel Separation vs. Frequency (MCP6022 and MCP6024 only).

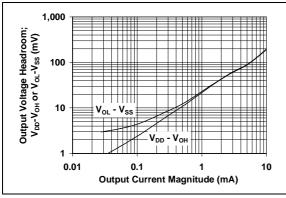
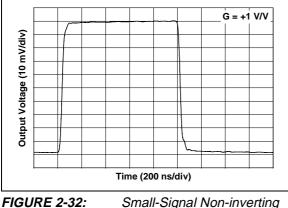
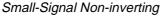


FIGURE 2-31: Output Voltage Headroom vs. Output Current.



Pulse Response.



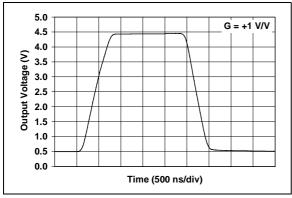


FIGURE 2-33: Pulse Response.

Large-Signal Non-inverting

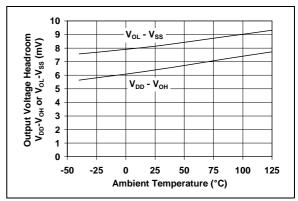


FIGURE 2-34: Output Voltage Headroom vs. Temperature.

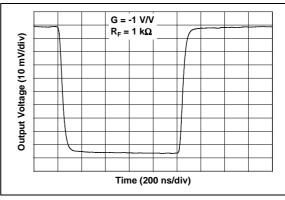


FIGURE 2-35: Small-Signal Inverting Pulse Response.

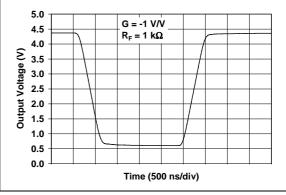


FIGURE 2-36: Response.

Large-Signal Inverting Pulse

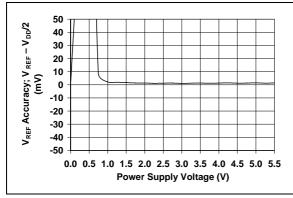


FIGURE 2-37: V_{REF} Accuracy vs. Supply Voltage (MCP6021 and MCP6023 only).

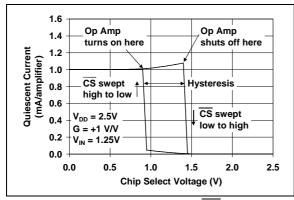


FIGURE 2-38: Chip Select (CS) Hysteresis (MCP6023 only) with $V_{DD} = 2.5V$.

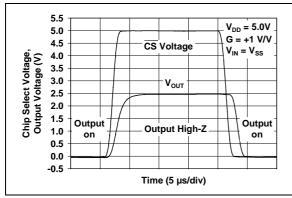


FIGURE 2-39: Chip Select (\overline{CS}) to Amplifier Output Response Time (MCP6023 only).

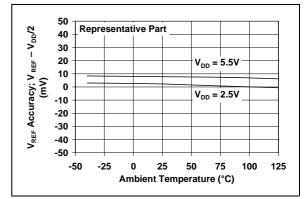


FIGURE 2-40: V_{REF} Accuracy vs.</sub> Temperature (MCP6021 and MCP6023 only).

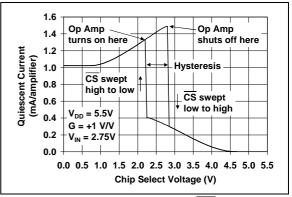


FIGURE 2-41: Chip Select (\overline{CS}) Hysteresis (MCP6023 only) with $V_{DD} = 5.5V$.

3.0 PIN DESCRIPTIONS

Descriptions of the pins are listed in Table 3-1.

IADLE 3-	I. FINF		ADLE				
MCP6021 (PDIP, SOIC, MSOP, TSSOP) (Note 1)	MCP6021 (SOT-23-5) (Note 1)	MCP6021R (SOT-23-5) (Note 2)	MCP6022	MCP6023	MCP6024	Symbol	Description
6	1	1	1	6	1	V _{OUT} , V _{OUTA}	Analog Output (op amp A)
2	4	4	2	2	2	V _{IN} -, V _{INA} -	Inverting Input (op amp A)
3	3	3	3	3	3	V _{IN} +, V _{INA} +	Non-inverting Input (op amp A)
7	5	2	8	7	4	V _{DD}	Positive Power Supply
—	—	—	5	—	5	V _{INB} +	Non-inverting Input (op amp B)
—	—	—	6	—	6	V _{INB} –	Inverting Input (op amp B)
_	—	—	7	—	7	V _{OUTB}	Analog Output (op amp B)
_	—	—	—	—	8	V _{OUTC}	Analog Output (op amp C)
—	—	—	—	—	9	V _{INC} -	Inverting Input (op amp C)
—	—	—	—	—	10	V _{INC} +	Non-inverting Input (op amp C)
4	2	5	4	4	11	V _{SS}	Negative Power Supply
—	—	—	—	—	12	V _{IND} +	Non-inverting Input (op amp D)
—	—	—	—	—	13	V _{IND} -	Inverting Input (op amp D)
_	—	—	—	_	14	V _{OUTD}	Analog Output (op amp D)
5	—		—	5	—	V _{REF}	Reference Voltage
_	_	_	_	8	_	CS	Chip Select
1, 8	_		_	1	_	NC	No Internal Connection

TABLE 3-1: PIN FUNCTION TABLE

Note 1: The MCP6021 in the 8-pin MSOP package is only available for E-temp (Extended Temperature) parts. The MCP6021 in the 8-pin TSSOP package is only available for I-temp (Industrial Temperature) parts.

2: The MCP6021R is only available in the 5-pin SOT-23 package, and for E-temp (Extended Temperature) parts.

3.1 Analog Outputs

The op amp output pins are low-impedance voltage sources.

3.2 Analog Inputs

The op amp non-inverting and inverting inputs are highimpedance CMOS inputs with low bias currents.

3.3 V_{REF} Output (MCP6021 and MCP6023)

Mid-supply reference voltage provided by the single op amps (except in SOT-23-5 package). This is an unbuffered, resistor voltage divider internal to the part.

3.4 CS Digital Input

This is a CMOS, Schmitt-triggered input that places the part into a low power mode of operation.

3.5 Power Supply (V_{SS} and V_{DD})

The positive power supply pin (V_{DD}) is 2.5V to 5.5V higher than the negative power supply pin (V_{SS}). For normal operation, the other pins are at voltages between V_{SS} and V_{DD}.

Typically, these parts are used in a single (positive) supply configuration. In this case, V_{SS} is connected to ground and V_{DD} is connected to the supply. V_{DD} will need a local bypass capacitor (typically 0.01 μF to 0.1 μF) within 2 mm of the V_{DD} pin. These parts need to use a bulk capacitor (typically 1 μF or larger) within 100 mm of the V_{DD} pin; it can be shared with nearby analog parts.

4.0 APPLICATIONS INFORMATION

The MCP6021/1R/2/3/4 family of operational amplifiers are fabricated on Microchip's state-of-the-art CMOS process. They are unity-gain stable and suitable for a wide range of general-purpose applications.

4.1 Rail-to-Rail Input

The MCP6021/1R/2/3/4 amplifier family is designed to not exhibit phase inversion when the input pins exceed the supply voltages. Figure 2-27 shows an input voltage exceeding both supplies with no resulting phase inversion.

The input stage of the MCP6021/1R/2/3/4 family of devices uses two differential input stages in parallel; one operates at low common-mode input voltage (V_{CM}), while the other operates at high V_{CM}. With this topology, the device operates with V_{CM} up to 0.3V past either supply rail (V_{SS} – 0.3V to V_{DD} + 0.3V) at +25°C. The amplifier input behaves linearly as long as V_{CM} is kept within the specified V_{CMR} limits. The input offset voltage is measured at both V_{CM} = V_{SS} – 0.3V and V_{DD} + 0.3V to ensure proper operation.

Input voltages that exceed the input voltage range (V_{CMR}) can cause excessive current to flow in or out of the input pins. Current beyond ±2 mA introduces possible reliability problems. Thus, applications that exceed this rating must externally limit the input current with an input resistor (R_{IN}), as shown in Figure 4-1.

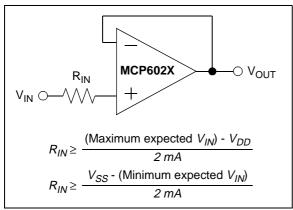


FIGURE 4-1: R_{IN} limits the current flow into an input pin.

Total Harmonic Distortion Plus Noise (THD+N) can be affected by the common mode input voltage (V_{CM}). As shown in Figure 2-3 and Figure 2-6, the input offset voltage (V_{OS}) is affected by the change from the NMOS to the PMOS input differential pairs. This change in V_{OS} will increase the distortion if the input voltage includes this transition region. This transition occurs between V_{DD} – 1.0V and V_{DD} – 2.0V, depending on V_{DD} and temperature.

4.2 Rail-to-Rail Output

The Maximum Output Voltage Swing is the maximum swing possible under a particular output load. According to the specification table, the output can reach within 20 mV of either supply rail when $R_L = 10 \ k\Omega$. See Figure 2-31 and Figure 2-34 for more information concerning typical performance.

4.3 Capacitive Loads

Driving large capacitive loads can cause stability problems for voltage feedback op amps. As the load capacitance increases, the feedback loop's phase margin decreases, and the closed loop bandwidth is reduced. This produces gain-peaking in the frequency response, with overshoot and ringing in the step response.

When driving large capacitive loads with these op amps (e.g., > 60 pF when G = +1), a small series resistor at the output (R_{ISO} in Figure 4-2) improves the feedback loop's phase margin (stability) by making the load resistive at higher frequencies. The bandwidth will be generally lower than the bandwidth with no capacitive load.

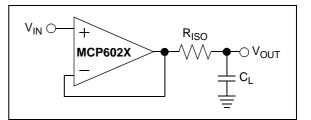


FIGURE 4-2: Output resistor R_{ISO} stabilizes large capacitive loads.

Figure 4-3 gives recommended R_{ISO} values for different capacitive loads and gains. The x-axis is the normalized load capacitance (C_L/G_N), where G_N is the circuit's noise gain. For non-inverting gains, G_N and the Signal Gain are equal. For inverting gains, G_N is 1+|Signal Gain| (e.g., -1 V/V gives $G_N = +2$ V/V).

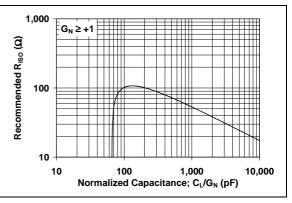


FIGURE 4-3: Recommended R_{ISO} values for capacitive loads.

After selecting R_{ISO} for your circuit, double-check the resulting frequency response peaking and step response overshoot. Modify R_{ISO}'s value until the response is reasonable. Evaluation on the bench and simulations with the MCP6021/1R/2/3/4 Spice macro model are helpful.

4.4 Gain Peaking

Figure 2-35 and Figure 2-36 use $R_{\rm F}$ = 1 k Ω to avoid (frequency response) gain peaking and (step response) overshoot. The capacitance to ground at the inverting input (C_G) is the op amp's common mode input capacitance plus board parasitic capacitance. C_G is in parallel with R_G, which causes an increase in gain at high frequencies for non-inverting gains greater than 1 V/V (unity gain). C_G also reduces the phase margin of the feedback loop for both non-inverting and inverting gains.

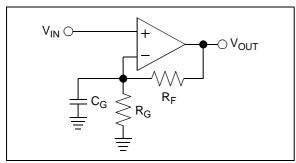


FIGURE 4-4: Non-inverting gain circuit with parasitic capacitance.

The largest value of R_F in Figure 4-4 that should be used is a function of noise gain (see G_N in **Section 4.3** "**Capacitive Loads**") and C_G. Figure 4-5 shows results for various conditions. Other compensation techniques may be used, but they tend to be more complicated to the design.

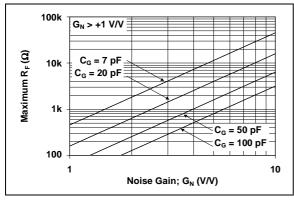


FIGURE 4-5: Non-inverting gain circuit with parasitic capacitance.

4.5 MCP6023 Chip Select (CS)

The MCP6023 is a single amplifier with chip select (CS). When CS is high, the supply current is less than 10 nA (typ) and travels from the CS pin to V_{SS} , with the amplifier output being put into a high-impedance state. When CS is low, the amplifier is enabled. If CS is left floating, the amplifier may not operate properly. Figure 1-1 and Figure 2-39 show the output voltage and supply current response to a CS pulse.

4.6 MCP6021 and MCP6023 Reference Voltage

The single op amps (MCP6021 and MCP6023), not in the SOT-23-5 package, have an internal mid-supply reference voltage connected to the V_{REF} pin (see Figure 4-6). The MCP6021 has $\overline{\text{CS}}$ internally tied to V_{SS}, which always keeps the op amp on and always provides a mid-supply reference. With the MCP6023, taking the $\overline{\text{CS}}$ pin high conserves power by shutting down both the op amp and the V_{REF} circuitry. Taking the $\overline{\text{CS}}$ pin low turns on the op amp and V_{REF} circuitry.

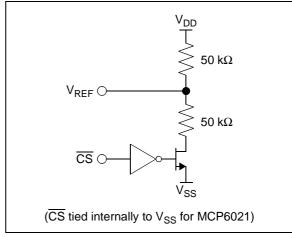


FIGURE 4-6: Simplified internal V_{REF} circuit (MCP6021 and MCP6023 only).

See Figure 4-7 for a non-inverting gain circuit using the internal mid-supply reference. The DC-blocking capacitor (C_B) also reduces noise by coupling the op amp input to the source.

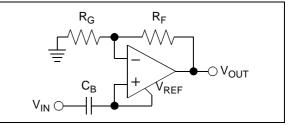


FIGURE 4-7: Non-inverting gain circuit using V_{REF} (MCP6021 and MCP6023 only).

To use the internal mid-supply reference for an inverting gain circuit, connect the V_{REF} pin to the non-inverting input, as shown in Figure 4-8. The capacitor C_B helps reduce power supply noise on the output.

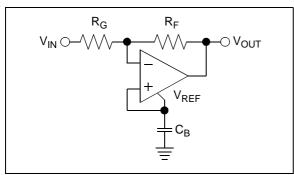


FIGURE 4-8: Inverting gain circuit using V_{REF} (MCP6021 and MCP6023 only).

If you don't need the mid-supply reference, leave the $V_{\mathsf{REF}}\xspace$ pin open.

4.7 Supply Bypass

With this family of operational amplifiers, the power supply pin (V_{DD} for single supply) should have a local bypass capacitor (i.e., 0.01 μ F to 0.1 μ F) within 2 mm for good, high-frequency performance. It also needs a bulk capacitor (i.e., 1 μ F or larger) within 100 mm to provide large, slow currents. This bulk capacitor can be shared with nearby analog parts.

4.8 Unused Op Amps

An unused op amp in a quad package (MCP6024) should be configured as shown in Figure 4-9. These circuits prevent the output from toggling and causing crosstalk. Circuit A can use any reference voltage between the supplies, provides a buffered DC voltage, and minimizes the supply current draw of the unused op amp. Circuit B uses the minimum number of components and operates as a comparator; it may draw more current.

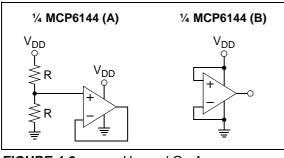


FIGURE 4-9:

Unused Op Amps.

4.9 PCB Surface Leakage

In applications where low input bias current is critical, PCB (printed circuit board) surface-leakage effects need to be considered. Surface leakage is caused by humidity, dust or other contamination on the board. Under low humidity conditions, a typical resistance between nearby traces is $10^{12}\Omega$. A 5V difference would cause 5 pA of current to flow, which is greater than the MCP6021/1R/2/3/4 family's bias current at +25°C (1 pA, typ).

The easiest way to reduce surface leakage is to use a guard ring around sensitive pins (or traces). The guard ring is biased at the same voltage as the sensitive pin. Figure 4-10 shows an example of this type of layout.

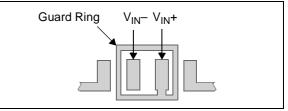


FIGURE 4-10: Example Guard Ring Layout.

- 1. Non-inverting Gain and Unity-Gain Buffer.
 - Connect the guard ring to the inverting input pin (V_{IN}-); this biases the guard ring to the common mode input voltage.
 - b) Connect the non-inverting pin (V_{IN}+) to the input with a wire that does not touch the PCB surface.
- 2. Inverting (Figure 4-10) and Transimpedance Gain Amplifiers (convert current to voltage, such as photo detectors).
 - a) Connect the guard ring to the non-inverting input pin (V_{IN}+). This biases the guard ring to the same reference voltage as the op amp's input (e.g., V_{DD}/2 or ground).
 - b) Connect the inverting pin (V_{IN}-) to the input with a wire that does not touch the PCB surface.

4.10 High Speed PCB Layout

Due to their speed capabilities, a little extra care in the PCB (Printed Circuit Board) layout can make a significant difference in the performance of these op amps. Good PC board layout techniques will help you achieve the performance shown in **Section 1.0 "Electrical Characteristics"** and **Section 2.0 "Typical Performance Curves"**, while also helping you minimize EMC (Electro-Magnetic Compatibility) issues.

Use a solid ground plane and connect the bypass local capacitor(s) to this plane with minimal length traces. This cuts down inductive and capacitive crosstalk.

Separate digital from analog, low speed from high speed and low power from high power. This will reduce interference.

Keep sensitive traces short and straight. Separating them from interfering components and traces. This is especially important for high-frequency (low rise-time) signals.

Sometimes it helps to place guard traces next to victim traces. They should be on both sides of the victim trace, and as close as possible. Connect the guard trace to ground plane at both ends, and in the middle for long traces.

Use coax cables (or low inductance wiring) to route signal and power to and from the PCB.

4.11 Typical Applications

4.11.1 A/D CONVERTER DRIVER AND ANTI-ALIASING FILTER

Figure 4-11 shows a third-order Butterworth filter that can be used as an A/D converter driver. It has a bandwidth of 20 kHz and a reasonable step response. It will work well for conversion rates of 80 ksps and greater (it has 29 dB attenuation at 60 kHz).

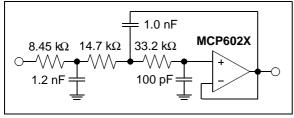


FIGURE 4-11: A/D converter driver and anti-aliasing filter with a 20 kHz cutoff frequency.

This filter can easily be adjusted to another bandwidth by multiplying all capacitors by the same factor. Alternatively, the resistors can all be scaled by another common factor to adjust the bandwidth.

4.11.2 OPTICAL DETECTOR AMPLIFIER

Figure 4-12 shows the MCP6021 op amp used as a transimpedance amplifier in a photo detector circuit. The photo detector looks like a capacitive current source, so the 100 k Ω resistor gains the input signal to a reasonable level. The 5.6 pF capacitor stabilizes this circuit and produces a flat frequency response with a bandwidth of 370 kHz.

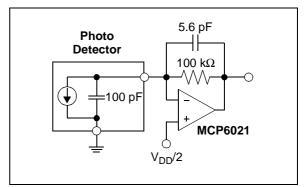


FIGURE 4-12: Transimpedance Amplifier for an Optical Detector.

5.0 DESIGN TOOLS

Microchip provides the basic design tools needed for the MCP6021/1R/2/3/4 family of op amps.

5.1 SPICE Macro Model

The latest SPICE macro model available for the MCP6021/1R/2/3/4 op amps is on Microchip's web site at www.microchip.com. This model is intended as an initial design tool that works well in the op amp's linear region of operation at room temperature. Within the macro model file is information on its capabilities.

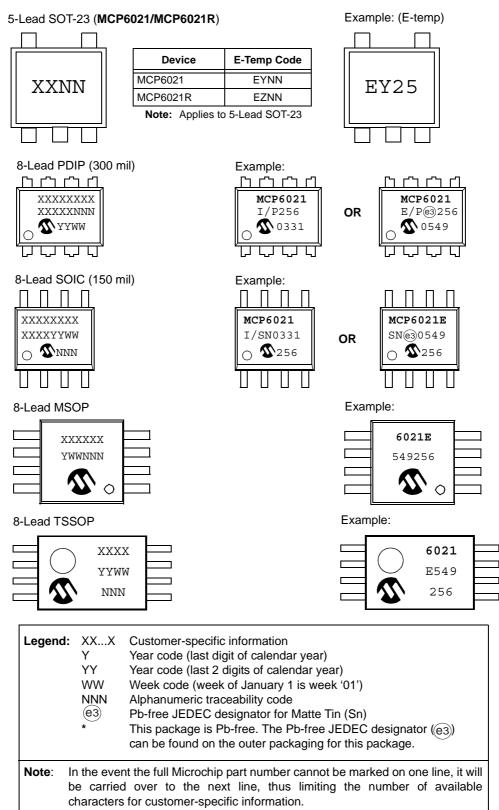
Bench testing is a very important part of any design and cannot be replaced with simulations. Also, simulation results using this macro model need to be validated by comparing them to the data sheet specifications and characteristic curves.

5.2 FilterLab[®] Software

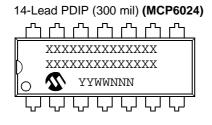
Microchip's FilterLab[®] software is an innovative tool that simplifies analog active filter (using op amps) design. It is available free of charge from our web site at www.microchip.com. The FilterLab software tool provides full schematic diagrams of the filter circuit with component values. It also outputs the filter circuit in SPICE format, which can be used with the macro model to simulate actual filter performance.

6.0 PACKAGING INFORMATION

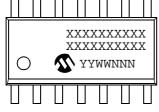
6.1 Package Marking Information

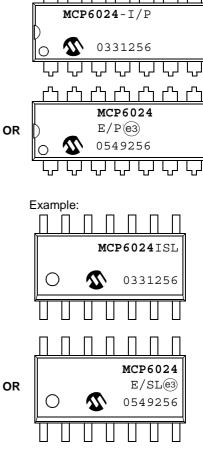


Package Marking Information (Continued)

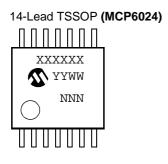


14-Lead SOIC (150 mil) (MCP6024)



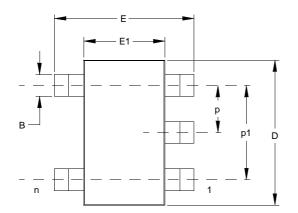


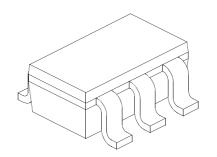
Example:

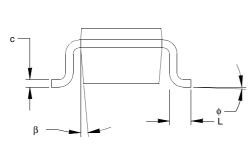


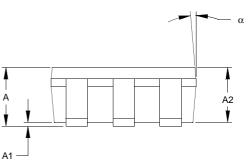
Example: 6024E 0331 256

5-Lead Plastic Small Outline Transistor (OT) (SOT-23)









	Units		INCHES*		MILLIMETERS		
Dimension Lin	nits	MIN	NOM	MAX	MIN	NOM	MAX
Number of Pins	n		5			5	
Pitch	р		.038			0.95	
Outside lead pitch (basic)	p1		.075			1.90	
Overall Height	Α	.035	.046	.057	0.90	1.18	1.45
Molded Package Thickness	A2	.035	.043	.051	0.90	1.10	1.30
Standoff	A1	.000	.003	.006	0.00	0.08	0.15
Overall Width	E	.102	.110	.118	2.60	2.80	3.00
Molded Package Width	E1	.059	.064	.069	1.50	1.63	1.75
Overall Length	D	.110	.116	.122	2.80	2.95	3.10
Foot Length	L	.014	.018	.022	0.35	0.45	0.55
Foot Angle	f	0	5	10	0	5	10
Lead Thickness	С	.004	.006	.008	0.09	0.15	0.20
Lead Width	В	.014	.017	.020	0.35	0.43	0.50
Mold Draft Angle Top	а	0	5	10	0	5	10
Mold Draft Angle Bottom	b	0	5	10	0	5	10

* Controlling Parameter

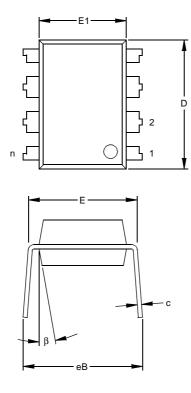
Notes:

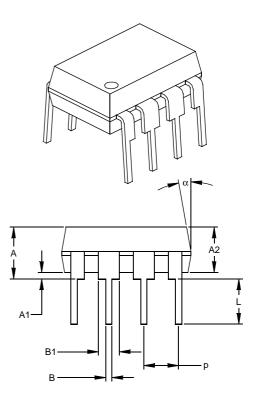
Dimensions D and E1 do not include mold flash or protrusions. Mold flash or protrusions shall not exceed .005" (0.127mm) per side. EIAJ Equivalent: SC-74A

Drawing No. C04-091

Revised 09-12-05

8-Lead Plastic Dual In-line (P) – 300 mil (PDIP)



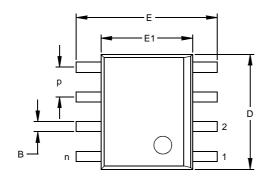


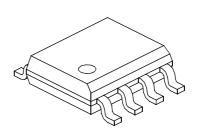
	Units		INCHES*		N	IILLIMETERS	
Dimensi	on Limits	MIN	NOM	MAX	MIN	NOM	MAX
Number of Pins	n		8			8	
Pitch	р		.100			2.54	
Top to Seating Plane	Α	.140	.155	.170	3.56	3.94	4.32
Molded Package Thickness	A2	.115	.130	.145	2.92	3.30	3.68
Base to Seating Plane	A1	.015			0.38		
Shoulder to Shoulder Width	Е	.300	.313	.325	7.62	7.94	8.26
Molded Package Width	E1	.240	.250	.260	6.10	6.35	6.60
Overall Length	D	.360	.373	.385	9.14	9.46	9.78
Tip to Seating Plane	L	.125	.130	.135	3.18	3.30	3.43
Lead Thickness	С	.008	.012	.015	0.20	0.29	0.38
Upper Lead Width	B1	.045	.058	.070	1.14	1.46	1.78
Lower Lead Width	В	.014	.018	.022	0.36	0.46	0.56
Overall Row Spacing	§ eB	.310	.370	.430	7.87	9.40	10.92
Mold Draft Angle Top	α	5	10	15	5	10	15
Mold Draft Angle Bottom	β	5	10	15	5	10	15

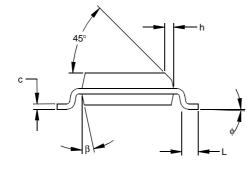
* Controlling Parameter § Significant Characteristic

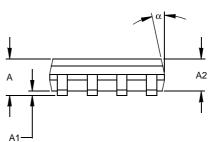
Notes: Dimensions D and E1 do not include mold flash or protrusions. Mold flash or protrusions shall not exceed .010" (0.254mm) per side. JEDEC Equivalent: MS-001 Drawing No. C04-018

8-Lead Plastic Small Outline (SN) - Narrow, 150 mil (SOIC)









	Units		INCHES*		N	1ILLIMETERS	
Dimer	nsion Limits	MIN	NOM	MAX	MIN	NOM	MAX
Number of Pins	n		8			8	
Pitch	р		.050			1.27	
Overall Height	А	.053	.061	.069	1.35	1.55	1.75
Molded Package Thickness	A2	.052	.056	.061	1.32	1.42	1.55
Standoff §	A1	.004	.007	.010	0.10	0.18	0.25
Overall Width	E	.228	.237	.244	5.79	6.02	6.20
Molded Package Width	E1	.146	.154	.157	3.71	3.91	3.99
Overall Length	D	.189	.193	.197	4.80	4.90	5.00
Chamfer Distance	h	.010	.015	.020	0.25	0.38	0.51
Foot Length	L	.019	.025	.030	0.48	0.62	0.76
Foot Angle	φ	0	4	8	0	4	8
Lead Thickness	С	.008	.009	.010	0.20	0.23	0.25
Lead Width	В	.013	.017	.020	0.33	0.42	0.51
Mold Draft Angle Top	α	0	12	15	0	12	15
Mold Draft Angle Bottom	β	0	12	15	0	12	15
* Controlling Parameter							

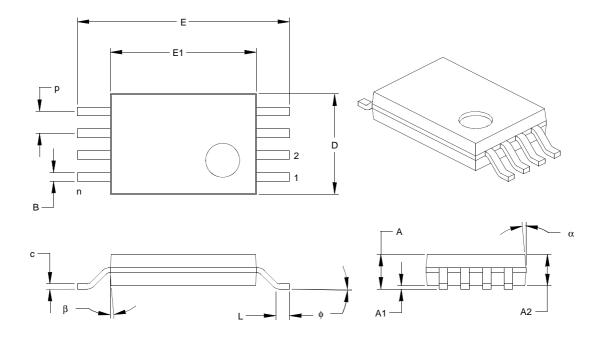
* Controlling Parameter § Significant Characteristic

Notes:

Dimensions D and E1 do not include mold flash or protrusions. Mold flash or protrusions shall not exceed .010" (0.254mm) per side. JEDEC Equivalent: MS-012

Drawing No. C04-057

8-Lead Plastic Thin Shrink Small Outline (ST) – 4.4 mm (TSSOP)



	Units	Units INCHES				LLIMETERS*	
Dimension	Limits	MIN	NOM	MAX	MIN	NOM	MAX
Number of Pins	n		8			8	
Pitch	р		.026			0.65	
Overall Height	А	.039	.041	.043	1.00	1.05	1.10
Molded Package Thickness	A2	.033	.035	.037	0.85	0.90	0.95
Standoff	A1	.002	.004	.006	0.05	0.10	0.15
Overall Width	E	.246	.251	.256	6.25	6.38	6.50
Molded Package Width	E1	.169	.173	.177	4.30	4.40	4.50
Molded Package Length	D	.114	.118	.122	2.90	3.00	3.10
Foot Length	L	.020	.024	.028	0.50	0.60	0.70
Foot Angle	φ	0°	4°	8°	0°	4°	8
Lead Thickness	С	.004	.006	.008	0.09	0.15	0.20
Lead Width	В	.007	.010	.012	0.19	0.25	0.30
Mold Draft Angle Top	α	0°	5°	10°	0°	5°	10°
Mold Draft Angle Bottom	β	0°	5°	10°	0°	5°	10°

* Controlling Parameter

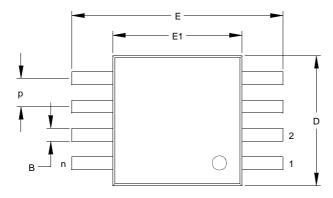
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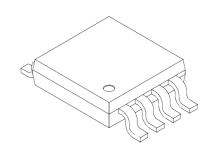
Dimensions D and E1 do not include mold flash or protrusions. Mold flash or protrusions shall not exceed .005" (0.127mm) per side. JEDEC Equivalent: MO-153

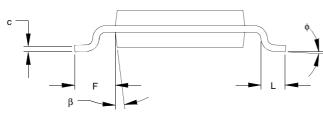
Drawing No. C04-086

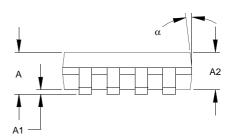
Revised 07-21-05

8-Lead Plastic Micro Small Outline Package (MS) (MSOP)









	Units	INCHES			MI	LLIMETERS*	
Dimension Limit	s	MIN	NOM	MAX	MIN	NOM	MAX
Number of Pins	n		8			8	
Pitch	р	.026 BSC			0.65 BSC		
Overall Height	А	-	-	.043	-	-	1.10
Molded Package Thickness	A2	.030	.033	.037	0.75	0.85	0.95
Standoff	A1	.000	-	.006	0.00	-	0.15
Overall Width	E	.193 BSC			4.90 BSC		
Molded Package Width	E1	.118 BSC				3.00 BSC	
Overall Length	D		.118 BSC		3.00 BSC		
Foot Length	L	.016	.024	.031	0.40	0.60	0.80
Footprint (Reference)	F		.037 REF		0.95 REF		
Foot Angle	φ	0°	-	8°	0°	-	8°
Lead Thickness	С	.003	.006	.009	0.08	-	0.23
Lead Width	В	.009	.012	.016	0.22	-	0.40
Mold Draft Angle Top	α	5°	-	15°	5°	-	15°
Mold Draft Angle Bottom	β	5°	-	15°	5°	-	15°

* Controlling Parameter

Notes:

Dimensions D and E1 do not include mold flash or protrusions. Mold flash or protrusions shall not exceed .010" (0.254mm) per side. BSC: Basic Dimension. Theoretically exact value shown without tolerances.

See ASME Y14.5M

REF: Reference Dimension, usually without tolerance, for information purposes only.

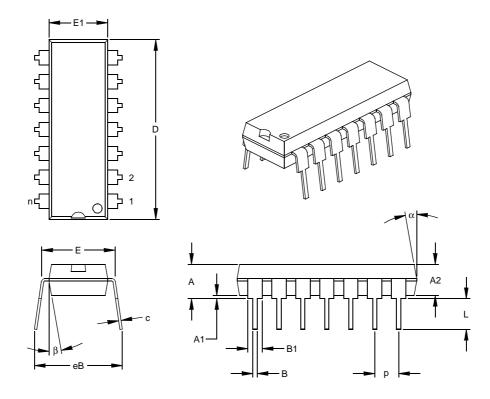
See ASME Y14.5M

JEDEC Equivalent: MO-187

Drawing No. C04-111

Revised 07-21-05

14-Lead Plastic Dual In-line (P) – 300 mil (PDIP)

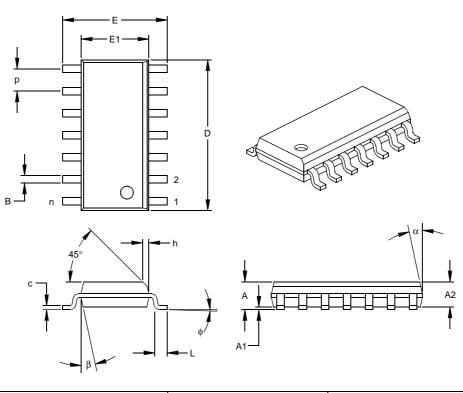


	Units		INCHES*		N	IILLIMETERS	
Dimensi	Dimension Limits			MAX	MIN	NOM	MAX
Number of Pins	n		14			14	
Pitch	р		.100			2.54	
Top to Seating Plane	А	.140	.155	.170	3.56	3.94	4.32
Molded Package Thickness	A2	.115	.130	.145	2.92	3.30	3.68
Base to Seating Plane	A1	.015			0.38		
Shoulder to Shoulder Width	E	.300	.313	.325	7.62	7.94	8.26
Molded Package Width	E1	.240	.250	.260	6.10	6.35	6.60
Overall Length	D	.740	.750	.760	18.80	19.05	19.30
Tip to Seating Plane	L	.125	.130	.135	3.18	3.30	3.43
Lead Thickness	С	.008	.012	.015	0.20	0.29	0.38
Upper Lead Width	B1	.045	.058	.070	1.14	1.46	1.78
Lower Lead Width	В	.014	.018	.022	0.36	0.46	0.56
Overall Row Spacing	§ eB	.310	.370	.430	7.87	9.40	10.92
Mold Draft Angle Top	α	5	10	15	5	10	15
Mold Draft Angle Bottom	β	5	10	15	5	10	15
* Controlling Parameter							

* Controlling Parameter § Significant Characteristic

Notes: Notes: Dimensions D and E1 do not include mold flash or protrusions. Mold flash or protrusions shall not exceed .010" (0.254mm) per side. JEDEC Equivalent: MS-001 Drawing No. C04-005

14-Lead Plastic Small Outline (SL) – Narrow, 150 mil (SOIC)



on Limits	MIN	NOM			MILLIMETERS	
n			MAX	MIN	NOM	MAX
		14			14	
р		.050			1.27	
А	.053	.061	.069	1.35	1.55	1.75
A2	.052	.056	.061	1.32	1.42	1.55
A1	.004	.007	.010	0.10	0.18	0.25
Е	.228	.236	.244	5.79	5.99	6.20
E1	.150	.154	.157	3.81	3.90	3.99
D	.337	.342	.347	8.56	8.69	8.81
h	.010	.015	.020	0.25	0.38	0.51
L	.016	.033	.050	0.41	0.84	1.27
¢	0	4	8	0	4	8
С	.008	.009	.010	0.20	0.23	0.25
В	.014	.017	.020	0.36	0.42	0.51
α	0	12	15	0	12	15
β	0	12	15	0	12	15
	A A2 A1 E D h L φ c B α	p A .053 A2 .052 A1 .004 E .228 E1 .150 D .337 h .010 L .016 φ 0 c .008 B .014 α 0	P .050 A .053 .061 A2 .052 .056 A1 .004 .007 E .228 .236 E1 .150 .154 D .337 .342 h .010 .015 L .016 .033 φ 0 4 c .008 .009 B .014 .017 α 0 12	P .050 A .053 .061 .069 A2 .052 .056 .061 A1 .004 .007 .010 E .228 .236 .244 E1 .150 .154 .157 D .337 .342 .347 h .010 .015 .020 L .016 .033 .050 φ 0 4 8 c .008 .009 .010 B .014 .017 .020 α 0 12 15	$\begin{array}{c c c c c c c c c c c c c c c c c c c $	$\begin{array}{c c c c c c c c c c c c c c c c c c c $

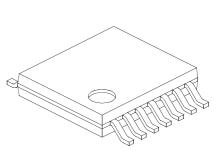
* Controlling Parameter § Significant Characteristic

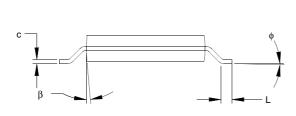
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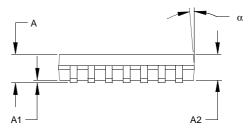
Dimensions D and E1 do not include mold flash or protrusions. Mold flash or protrusions shall not exceed .010" (0.254mm) per side. JEDEC Equivalent: MS-012

Drawing No. C04-065

14-Lead Plastic Thin Shrink Small Outline (ST) – 4.4 mm (TSSOP)







	Units	INCHES			MILLIMETERS*		
Dimension Limit	S	MIN	NOM	MAX	MIN	NOM	MAX
Number of Pins	n	14			14		
Pitch	р	.026 BSC			0.65 BSC		
Overall Height	А	.039	.041	.043	1.00	1.05	1.10
Molded Package Thickness	A2	.033	.035	.037	0.85	0.90	0.95
Standoff	A1	.002	.004	.006	0.05	0.10	0.15
Overall Width	E	.246	.251	.256	6.25	6.38	6.50
Molded Package Width	E1	.169	.173	.177	4.30	4.40	4.50
Molded Package Length	D	.193	.197	.201	4.90	5.00	5.10
Foot Length	L	.020	.024	.028	0.50	0.60	0.70
Foot Angle	φ	0°	4°	8°	0°	4°	8°
Lead Thickness	С	.004	.006	.008	0.09	0.15	0.20
Lead Width	В	.007	.010	.012	0.19	0.25	0.30
Mold Draft Angle Top	α	12° REF			12° REF		
Mold Draft Angle Bottom	β	12° REF			12° REF		

* Controlling Parameter

Notes:

Dimensions D and E1 do not include mold fla sh or protrusions. Mold flash or protrusions shall not exceed .005" (0.127mm) per side. BSC: Basic Dimension. Theoretically exact value shown without tolerances.

See ASME Y14.5M

REF: Reference Dimension, usually without tole rance, for information purposes only.

See ASME Y14.5M

JEDEC Equivalent: MO-153 AB-1

Drawing No. C04-087

Revised: 08-17-05

NOTES:

APPENDIX A: REVISION HISTORY

Revision C (March 2006)

The following is the list of modifications:

- 1. Added SOT-23-5 package option for single op amps MCP6021 and MCP6021R (E-temp only).
- 2. Added MSOP-8 package option for E-temp single op amp (MCP6021).
- 3. Corrected package drawing on front page for dual op amp (MCP6022).
- Clarified spec conditions (I_{SC}, PM and THD+N) in Section 2.0 "Typical Performance Curves".
- 5. Added Section 3.0 "Pin Descriptions".
- 6. Updated **Section 4.0** "**Applications information**" for THD+N, unused op amps, and gain peaking discussions.
- 7. Corrected and updated package marking information in Section 6.0 "Packaging Information".
- 8. Added Appendix A: "REVISION HISTORY".

Revision B (November 2003)

• Second Release of this Document

Revision A (November 2001)

• Original Release of this Document

NOTES:

PRODUCT IDENTIFICATION SYSTEM

To order or obtain information, e.g., on pricing or delivery, refer to the factory or the listed sales office.

<u>PART NO. X /XX</u>		Е	Examples:			
Device Temp		Package	a)) MCP6021T-E/OT:	Tape and Reel, Extended temperature, 5LD SOT-23.	
			b)) MCP6021-E/P:	Extended temperature, 8LD PDIP.	
Device:	MCP6021 MCP6021T	Single Op Amp	c)	MCP6021-E/SN:	Extended temperature, 8LD SOIC.	
		(Tape and Reel for SOT-23, SOIC, TSSOP, MSOP) & Single Op Amp T Single Op Amp) MCP6021RT-E/O	T:Tape and Reel, Extended temperature, 5LD SOT-23.	
	MCP6022 MCP6022T	(Tape and Reel for SOT-23) Dual Op Amp Dual Op Amp	a)) MCP6022-I/P:	Industrial temperature, 8LD PDIP.	
	MCP6023	(Tape and Reel for <u>SO</u> IC and TSSOP) Single Op Amp w/ CS	b)		Extended temperature, 8LD PDIP.	
	MCP6023T MCP6024 MCP6024T	Single Op Amp w/ CS (Tape and Reel for SOIC and TSSOP) Quad Op Amp Quad Op Amp	c)	MCP6022T-E/ST:	Tape and Reel, Extended temperature, 8LD TSSOP.	
	MCF 00241	(Tape and Reel for SOIC and TSSOP)	a)) MCP6023-I/P:	Industrial temperature, 8LD PDIP.	
Temperature Range:		C to +85°C	b)) MCP6023-E/P:	Extended temperature, 8LD PDIP.	
		C to +125°C	c)	MCP6023-E/SN:	Extended temperature, 8LD SOIC.	
Package:		OT = Plastic Small Outline Transistor (SOT-23), 5-lead (MCP6021, E-Temp; MCP6021R, E-Temp)) MCP6024-I/SL:	Industrial temperature, 14LD SOIC.	
		tic MSOP, 8-lead P6021, E-Temp)	b)) MCP6024-E/SL:	Extended temperature, 14LD SOIC.	
	SN = Plas SL = Plas ST = Plas (MC	tic DIP (300 mil Body), 8-lead, 14-lead tic SOIC (150mil Body), 8-lead tic SOIC (150 mil Body), 14-lead tic TSSOP, 8-lead P6021,I-Temp; MCP6022, I-Temp, E-Temp; '6023, I-Temp, E-Temp;)	c)	MCP6024T-E/ST:	Tape and Reel, Extended temperature, 14LD TSSOP.	
		tic TSSOP, 14-lead				

NOTES:

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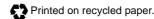
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