# 512Mb G-die DDR2 SDRAM Specification 

## 60FBGA \& 84FBGA with Lead-Free and Halogen-Free (RoHS compliant)

[^0]* Samsung Electronics reserves the right to change products or specification without notice.
K4T51043QG
K4T51083QG
K4T51163QG


## Table of Contents

1.0 Ordering Information ..... 4
2.0 Key Features ..... 4
3.0 Package Pinout/Mechanical Dimension \& Addressing ..... 5
3.1 x4 package pinout (Top View) : 60ball FBGA Package ..... 5
$3.2 \times 8$ package pinout (Top View) : 60ball FBGA Package ..... 6
$3.3 \times 16$ package pinout (Top View) : 84ball FBGA Package ..... 7
3.4 FBGA Package Dimension(x4/ x8) ..... 8
3.5 FBGA Package Dimension(x16) ..... 9
4.0 Input/Output Functional Description ..... 10
5.0 DDR2 SDRAM Addressing ..... 11
6.0 Absolute Maximum DC Ratings ..... 12
7.0 AC \& DC Operating Conditions ..... 12
7.1 Recommended DC Operating Conditions (SSTL - 1.8) ..... 12
7.2 Operating Temperature Condition ..... 13
7.3 Input DC Logic Level ..... 13
7.4 Input AC Logic Level ..... 13
7.5 AC Input Test Conditions ..... 13
7.6 Differential input AC logic Level ..... 14
7.7 Differential AC output parameters ..... 14
8.0 ODT DC electrical characteristics ..... 14
9.0 OCD default characteristics ..... 15
10.0 IDD Specification Parameters and Test Conditions ..... 16
11.0 DDR2 SDRAM IDD Spec ..... 18
12.0 Input/Output capacitance ..... 19
13.0 Electrical Characteristics \& AC Timing for DDR2-800/667/533/400 ..... 19
13.1 Refresh Parameters by Device Density ..... 19
13.2 Speed Bins and CL, tRCD, tRP, tRC and tRAS for Corresponding Bin ..... 19
13.3 Timing parameters by speed grade (DDR2-800 and DDR2-667) ..... 20
13.4 Timing parameters by speed grade (DDR2-533 and DDR2-400) ..... 22
14.0 General notes, which may apply for all AC parameters ..... 24
15.0 Specific Notes for dedicated AC parameters ..... 26

## Revision History

| Revision | Month | Year |  |
| :---: | :---: | :---: | :--- |
| 1.0 | August | 2007 | - Initial Release |
| 1.1 | October | 2007 | - Added x16 IDD Specification |
| 1.2 | January | 2008 | - Added x4 Specification |
| 1.21 | February | 2008 | - Typo Correction |
| 1.3 | July | 2008 | - Updated AC timing table with the JEDEC update(JESD79-2E) |
| 1.4 | December | 2008 | - Updated AC/DC operating condition with the JEDEC update(JESD79-2E) |

### 1.0 Ordering Information

| Org. | DDR2-800 5-5-5 | DDR2-800 6-6-6 | DDR2-667 5-5-5 | DDR2-533 4-4-4 | DDR2-400 3-3-3 | Package |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 128Mx4 | K4T51043QG-HC(L)E7 | K4T51043QG-HC(L)F7 | K4T51043QG-HC(L)E6 | K4T51043QG-HC(L)D5 | K4T51043QG-HC(L)CC | 60 FBGA |
| 64Mx8 | K4T51083QG-HC(L)E7 | K4T51083QG-HC(L)F7 | K4T51083QG-HC(L)E6 | K4T51083QG-HC(L)D5 | K4T51083QG-HC(L)CC | 60 FBGA |
| 32Mx16 | K4T51163QG-HC(L)E7 | K4T51163QG-HC(L)F7 | K4T51163QG-HC(L)E6 | K4T51163QG-HC(L)D5 | K4T51163QG-HC(L)CC | 84 FBGA |

Note :

1. Speed bin is in order of CL-tRCD-tRP
2. "H" of Part number(12th digit) stands for Lead-Free, Halogen-Free, and RoHS compliant products.

### 2.0 Key Features

| Speed | DDR2-800 5-5-5 | DDR2-800 6-6-6 | DDR2-667 5-5-5 | DDR2-533 4-4-4 | DDR2-400 3-3-3 | Units |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| CAS Latency | 5 | 6 | 5 | 4 | 3 | tCK |
| $\operatorname{tRCD}(\min )$ | 12.5 | 15 | 15 | 15 | 15 | ns |
| $\operatorname{tRP}(\min )$ | 12.5 | 15 | 15 | 15 | 15 | ns |
| $\operatorname{tRC}(\min )$ | 57.5 | 60 | 60 | $n 0$ | 55 | ns |

- JEDEC standard $\mathrm{V}_{\mathrm{DD}}=1.8 \mathrm{~V} \pm 0.1 \mathrm{~V}$ Power Supply
- $\mathrm{V}_{\mathrm{DDQ}}=1.8 \mathrm{~V} \pm 0.1 \mathrm{~V}$
- $200 \mathrm{MHz} \mathrm{f}_{\mathrm{CK}}$ for $400 \mathrm{Mb} / \mathrm{sec} / \mathrm{pin}, 267 \mathrm{MHz} \mathrm{f}_{\mathrm{CK}}$ for $533 \mathrm{Mb} / \mathrm{sec} /$ pin, $333 \mathrm{MHz} \mathrm{f}_{\mathrm{CK}}$ for $667 \mathrm{Mb} /$ sec/pin, $400 \mathrm{MHz} \mathrm{f}_{\mathrm{CK}}$ for $800 \mathrm{Mb} /$ sec/pin
- 4 Banks
- Posted $\overline{\mathrm{CAS}}$
- Programmable $\overline{\mathrm{CAS}}$ Latency: $3,4,5,6$
- Programmable Additive Latency: 0, 1, 2, 3, 4, 5
- Write Latency(WL) = Read Latency(RL) -1
- Burst Length: 4 , 8(Interleave/Nibble sequential)
- Programmable Sequential / Interleave Burst Mode
- Bi-directional Differential Data-Strobe (Single-ended datastrobe is an optional feature)
- Off-Chip Driver(OCD) Impedance Adjustment
- On Die Termination
- Special Function Support -50ohm ODT
-High Temperature Self-Refresh rate enable
- Average Refresh Period 7.8 us at lower than $\mathrm{T}_{\text {CASE }} 85^{\circ} \mathrm{C}$, 3.9 us at $85^{\circ} \mathrm{C}<\mathrm{T}_{\text {CASE }} \leq 95^{\circ} \mathrm{C}$
- All of products are Lead-Free, Halogen-Free, and RoHS compliant

The 512 Mb DDR2 SDRAM is organized as a $32 \mathrm{Mbit} \times 4 \mathrm{I} / \mathrm{Os} \mathrm{x}$ 4banks or 16Mbit x 8 I/Os x 4banks or 8 Mbit x 16 I/Os x 4 banks device. This synchronous device achieves high speed double-data-rate transfer rates of up to $800 \mathrm{Mb} / \mathrm{sec} / \mathrm{pin}$ (DDR2-800) for general applications.
The chip is designed to comply with the following key DDR2 SDRAM features such as posted CAS with additive latency, write latency = read latency -1, Off-Chip Driver(OCD) impedance adjustment and On Die Termination.
All of the control and address inputs are synchronized with a pair of externally supplied differential clocks. Inputs are latched at the crosspoint of differential clocks (CK rising and $\overline{\mathrm{CK}}$ falling). All I/Os are synchronized with a pair of bidirectional strobes (DQS and $\overline{\mathrm{DQS}}$ ) in a source synchronous fashion. The address bus is used to convey row, column, and bank address information in a RAS/ $\overline{\mathrm{CAS}}$ multiplexing style. For example, $512 \mathrm{Mb}(\mathrm{x} 8)$ device receive 14/10/2 addressing.
The 512 Mb DDR2 device operates with a single $1.8 \mathrm{~V} \pm 0.1 \mathrm{~V}$ power supply and $1.8 \mathrm{~V} \pm 0.1 \mathrm{~V} \mathrm{~V}_{\text {DDQ }}$.
The 512 Mb DDR2 device is available in 60ball FBGAs(x8) and in 84ball FBGAs(x16).

Note : The functionality described and the timing specifications included in this data sheet are for the DLL Enabled mode of operation.
ote : This data sheet is an abstract of full DDR2 specification and does not cover the common features which are described in "Samsung's DDR2 SDRAM Device Operation \& Timing Diagram"
3.1 x4 package pinout (Top View) : 60ball FBGA Package

| 1 | 2 | 3 |  | 7 | 8 | 9 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $V_{D D}$ | NC | $\mathrm{V}_{\text {ss }}$ | A | $\mathrm{V}_{\text {SSO }}$ | $\overline{\text { DQS }}$ | $\mathrm{V}_{\text {DDQ }}$ |
| NC | $\mathrm{V}_{\text {SSQ }}$ | DM | B | DQS | $\mathrm{V}_{\text {SSQ }}$ | NC |
| $\mathrm{V}_{\text {DDQ }}$ | DQ1 | $\mathrm{V}_{\mathrm{DDQ}}$ | C | $\mathrm{V}_{\mathrm{DDQ}}$ | DQ0 | $\mathrm{V}_{\mathrm{DDQ}}$ |
| NC | $\mathrm{V}_{\text {SSQ }}$ | DQ3 | D | DQ2 | $\mathrm{V}_{\text {SSO }}$ | NC |
| $\mathrm{V}_{\text {DDL }}$ | $\mathrm{V}_{\text {REF }}$ | $\mathrm{V}_{\mathrm{SS}}$ | E | $\mathrm{V}_{\text {SSDL }}$ | CK | $V_{D D}$ |
|  | CKE | $\overline{\mathrm{WE}}$ | F | $\overline{\mathrm{RAS}}$ | $\overline{\mathrm{CK}}$ | ODT |
| NC | BAO | BA1 | G | $\overline{\text { CAS }}$ | $\overline{\overline{C S}}$ |  |
|  | A10/AP | A1 | H | A2 | A0 | $V_{D D}$ |
| $\mathrm{V}_{\text {SS }}$ | A3 | A5 | $J$ | A6 | A4 |  |
|  | A7 | A9 | K | A11 | A8 | $\mathrm{V}_{\text {ss }}$ |
| $V_{D D}$ | A12 | NC | L | NC | A13 |  |

Note:

1. Pin B3 has identical capacitance as pin B7.
2. $V_{D D L}$ and $V_{S S D L}$ are power and ground for the DLL.

Ball Locations (x4)

- Populated Ball
+ : Depopulated Ball
Top View (See the balls through the package)

3.2 x8 package pinout (Top View) : 60ball FBGA Package


Note :

1. Pins $B 3$ and $A 2$ have identical capacitance as pins $B 7$ and $A 8$.
2. For a read, when enabled, strobe pair RDQS \& $\overline{R D Q S}$ are identical in function and timing to strobe pair DQS \& $\overline{D Q S}$ and input masking function is disabled.
3. The function of DM or RDQS/RDQS are enabled by EMRS command.
4. $\mathrm{V}_{\mathrm{DDL}}$ and $\mathrm{V}_{\mathrm{SSDL}}$ are power and ground for the DLL.

## Ball Locations (x8)

- Populated Ball
+ : Depopulated Ball
Top View (See the balls through the package)

$3.3 \times 16$ package pinout (Top View) : 84ball FBGA Package

| 1 | 2 | 3 |  | 7 | 8 | 9 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $V_{D D}$ | NC | $\mathrm{V}_{\text {SS }}$ | A | $\mathrm{V}_{\text {SSQ }}$ | $\overline{\text { UDQS }}$ | $\mathrm{V}_{\text {DDQ }}$ |
| DQ14 | $\mathrm{V}_{\text {SSQ }}$ | UDM | B | UDQS | $\mathrm{V}_{\text {SSQ }}$ | DQ15 |
| $\mathrm{V}_{\text {DDQ }}$ | DQ9 | $\mathrm{V}_{\mathrm{DDQ}}$ | C | $\mathrm{V}_{\mathrm{DDQ}}$ | DQ8 | $\mathrm{V}_{\mathrm{DDQ}}$ |
| DQ12 | $\mathrm{V}_{\text {SSQ }}$ | DQ11 | D | DQ10 | $\mathrm{V}_{\text {SSQ }}$ | DQ13 |
| $V_{D D}$ | NC | $\mathrm{V}_{\text {SS }}$ | E | $\mathrm{V}_{\text {SSQ }}$ | $\overline{\text { LDQS }}$ | $\mathrm{V}_{\text {DDQ }}$ |
| DQ6 | $\mathrm{V}_{\text {SSQ }}$ | LDM | F | LDQS | $\mathrm{V}_{\text {SSQ }}$ | DQ7 |
| $V_{\text {DDQ }}$ | DQ1 | $V_{\text {DDQ }}$ | G | $\mathrm{V}_{\text {DDQ }}$ | DQ0 | $\mathrm{V}_{\text {DDQ }}$ |
| DQ4 | $\mathrm{V}_{\text {SSQ }}$ | DQ3 | H | DQ2 | $\mathrm{V}_{\text {SSQ }}$ | DQ5 |
| $\mathrm{V}_{\mathrm{DDL}}$ | $\mathrm{V}_{\text {REF }}$ | $\mathrm{V}_{\text {SS }}$ | J | $\mathrm{V}_{\text {SSDL }}$ | CK | $\mathrm{V}_{\mathrm{DD}}$ |
|  | CKE | $\overline{\mathrm{WE}}$ | K | $\overline{\text { RAS }}$ | $\overline{\mathrm{CK}}$ | ODT |
| NC | BAO | BA1 | L | $\overline{\mathrm{CAS}}$ | $\overline{\mathrm{CS}}$ |  |
|  | A10/AP | A1 | M | A2 | A0 | $\mathrm{V}_{\mathrm{DD}}$ |
| $\mathrm{V}_{\mathrm{SS}}$ | A3 | A5 | N | A6 | A4 |  |
|  | A7 | A9 | P | A11 | A8 | $\mathrm{V}_{\mathrm{SS}}$ |
| $V_{D D}$ | A12 | NC | R | NC | NC |  |

Note :

1. $V_{D D L}$ and $V_{S S D L}$ are power and ground for the DLL.
2. In case of only 8 DQs out of 16 DQs are used, LDQS, LDQSB and DQ0~7 must be used.

## Ball Locations (x16)

- Populated Ball
+ : Depopulated Ball

Top View
(See the balls through the package)

3.4 FBGA Package Dimension(x4/x8)

3.5 FBGA Package Dimension(x16)


### 4.0 Input/Output Functional Description

| Symbol | Type | Function |
| :---: | :---: | :---: |
| CK, $\overline{\mathrm{CK}}$ | Input | Clock: CK and $\overline{\mathrm{CK}}$ are differential clock inputs. All address and control input signals are sampled on the crossing of the positive edge of CK and negative edge of $\overline{\mathrm{CK}}$. Output (read) data is referenced to the crossings of CK and $\overline{\mathrm{CK}}$ (both directions of crossing). |
| CKE | Input | Clock Enable: CKE HIGH activates, and CKE Low deactivates, internal clock signals and device input buffers and output drivers. Taking CKE Low provides Precharge Power-Down and Self Refresh operation (all banks idle), or Active Power-Down (row Active in any bank). CKE is synchronous for power down entry and exit, and for self refresh entry. CKE is asynchronous for self refresh exit. After $\mathrm{V}_{\text {REF }}$ has become stable during the power on and initialization swquence, it must be maintained for proper operation of the CKE receiver. For proper self-refresh entry and exit, $\mathrm{V}_{\text {REF }}$ must be maintained to this input. CKE must be maintained high throughout read and write accesses. Input buffers, excluding CK, $\overline{\mathrm{CK}}$, ODT and CKE are disabled during power-down. Input buffers, excluding CKE, are disabled during self refresh. |
| $\overline{\mathrm{CS}}$ | Input | Chip Select: All commands are masked when $\overline{\mathrm{CS}}$ is registered HIGH. $\overline{\mathrm{CS}}$ provides for external Rank selection on systems with multiple Ranks. $\overline{\mathrm{CS}}$ is considered part of the command code. |
| ODT | Input | On Die Termination: ODT (registered HIGH) enables termination resistance internal to the DDR2 SDRAM. When enabled, ODT is only applied to each DQ, DQS, $\overline{\mathrm{DQS}}$, RDQS, $\overline{R D Q S}$, and DM signal for $\mathrm{x} 4 / \mathrm{x} 8$ configurations. For $\times 16$ configuration, ODT is applied to each DQ, UDQS/UDQS, LDQS/ $\overline{L D Q S}$, UDM, and LDM signal. The ODT pin will be ignored if the Extended Mode Register Set(EMRS) is programmed to disable ODT. |
| $\overline{\mathrm{RAS}}, \overline{\mathrm{CAS}}, \overline{\mathrm{WE}}$ | Input | Command Inputs: $\overline{\mathrm{RAS}}, \overline{\mathrm{CAS}}$ and $\overline{\mathrm{WE}}$ (along with $\overline{\mathrm{CS}}$ ) define the command being entered. |
| $\begin{gathered} \text { DM } \\ \text { (UDM), (LDM) } \end{gathered}$ | Input | Input Data Mask: DM is an input mask signal for write data. Input data is masked when DM is sampled HIGH coincident with that input data during a Write access. DM is sampled on both edges of DQS. Although DM pins are input only, the DM loading matches the DQ and DQS loading. For $x 8$ device, the function of DM or RDQS/RDQS is enabled by EMRS command. |
| BA0 - BA1 | Input | Bank Address Inputs: BA0, BA1 and BA2 define to which bank an Active, Read, Write or Precharge command is being applied. Bank address also determines if the mode register or extended mode register is to be accessed during a MRS or EMRS cycle. |
| A0-A13 | Input | Address Inputs: Provided the row address for Active commands and the column address and Auto Precharge bit for Read/Write commands to select one location out of the memory array in the respective bank. A10 is sampled during a Precharge command to determine whether the Precharge applies to one bank (A10 LOW) or all banks (A10 HIGH). If only one bank is to be precharged, the bank is selected by BA0, BA1 and BA2. The address inputs also provide the opcode during Mode Register Set commands. |
| DQ | Input/Output | Data Input/ Output: Bi-directional data bus. |
| DQS, ( $\overline{(\overline{\mathrm{DQS}})}$ (LDQS), ( $\overline{\mathrm{LDQS}})$ (UDQS), ( $\overline{\mathrm{UDQS}})$ (RDQS), (RDQS) | Input/Output | Data Strobe: Output with read data, input with write data. Edge-aligned with read data, centered in write data. For the x16, LDQS corresponds to the data on DQ0-DQ7; UDQS corresponds to the data on DQ8-DQ15. For the x8, an RDQS option using DM pin can be enabled via the EMRS(1) to simplify read timing. The data strobes DQS, LDQS, UDQS, and RDQS may be used in single ended mode or paired with optional complementary signals $\overline{\mathrm{DQS}}, \overline{\mathrm{LDQS}}, \overline{\mathrm{UDQS}}$, and $\overline{R D Q S}$ to provide differential pair signaling to the system during both reads and writes. A control bit at EMRS(1)[A10] enables or disables all complementary data strobe signals. <br> In this data sheet, "differential DQS signals" refers to any of the following with A10 $=0$ of EMRS(1) <br> x4 DQS/DQS <br> x8 DQS/DQS if EMRS(1)[A11] $=0$ <br> x8 DQS/DQS, RDQS/RDQS, x16 LDQS/LDQS and UDQS/UDQS <br> "single-ended DQS signals" refers to any of the following with A10 = 1 of EMRS(1) $\begin{aligned} & \text { x4 DQS } \\ & \text { x } 8 \text { DQS if EMRS(1) }[\mathrm{A} 11]=0 \\ & \text { x8 DQS, RDQS, if EMRS(1) [A11] = } 1 \\ & \text { x16 LDQS and UDQS } \end{aligned}$ |
| NC |  | No Connect : No internal electrical connection is present. |
| $\mathrm{V}_{\mathrm{DD}} / \mathrm{V}_{\mathrm{DDQ}}$ | Supply | Power Supply : $1.8 \mathrm{~V}+/-0.1 \mathrm{~V}$, DQ Power Supply : $1.8 \mathrm{~V}+/-0.1 \mathrm{~V}$ |
| $\mathrm{V}_{\mathrm{SS}} / \mathrm{V}_{\text {SSQ }}$ | Supply | Ground, DQ Ground |
| $\mathrm{V}_{\text {DDL }}$ | Supply | DLL Power Supply : 1.8V +/- 0.1V |
| $\mathrm{V}_{\text {SSDL }}$ | Supply | DLL Ground |
| $\mathrm{V}_{\text {REF }}$ | Supply | Reference voltage |

### 5.0 DDR2 SDRAM Addressing

512Mb

| Configuration | $128 \mathrm{Mb} \times 4$ | 64Mb x 8 | $32 \mathrm{Mb} \times 16$ |
| :---: | :---: | :---: | :---: |
| \# of Banks | 4 | 4 | 4 |
| Bank Address | BA0,BA1 | BA0,BA1 | BA0,BA1 |
| Auto precharge | A10/AP | A10/AP | A10/AP |
| Row Address | A0 ~ A13 | A0 ~ A13 | A0 ~ A12 |
| Column Address | A0 ~ A9, A11 | A0 ~ A9 | A0 ~ A9 |

* Reference information: The following tables are address mapping information for other densities.

256Mb

| Configuration | 64Mb x4 | 32Mb $\times 8$ | $16 \mathrm{Mb} \times 16$ |
| :---: | :---: | :---: | :---: |
| \# of Banks | 4 | 4 | 4 |
| Bank Address | BA0,BA1 | BA0,BA1 | BA0,BA1 |
| Auto precharge | A10/AP | A10/AP | A10/AP |
| Row Address | A0 ~ A12 | A0 ~ A12 | A0 ~ A12 |
| Column Address | A0 ~ A9,A11 | A0 ~ A9 | A0 ~ A8 |

1Gb

| Configuration | 256Mb x4 | $128 \mathrm{Mb} \times 8$ | 64Mb x16 |
| :---: | :---: | :---: | :---: |
| \# of Banks | 8 | 8 | 8 |
| Bank Address | BA0 ~ BA2 | $B A 0 \sim B A 2$ | $B A 0 \sim B A 2$ |
| Auto precharge | A10/AP | A10/AP | A10/AP |
| Row Address | A0 ~ A13 | A0 ~ A13 | A0 ~ A12 |
| Column Address | A0 ~ A9, A11 | A0 ~ A9 | A0 ~ A9 |

2Gb

| Configuration | 512Mb x4 | 256Mb x 8 | 128Mb x16 |
| :---: | :---: | :---: | :---: |
| \# of Banks | 8 | 8 | 8 |
| Bank Address | $B A 0 \sim B A 2$ | BAO ~ BA2 | $B A 0 \sim B A 2$ |
| Auto precharge | A10/AP | A10/AP | A10/AP |
| Row Address | A0 ~ A14 | A0 ~ A14 | A0 ~ A13 |
| Column Address | A0 ~ A9, A11 | A0 ~ A9 | A0 ~ A9 |

4Gb

| Configuration | $1 \mathrm{~Gb} \times 4$ | $512 \mathrm{Mb} \times 8$ | 256Mb x16 |
| :---: | :---: | :---: | :---: |
| \# of Banks | 8 | 8 | 8 |
| Bank Address | BA0 ~ BA2 | BAO ~ BA2 | BA0 ~ BA2 |
| Auto precharge | A10/AP | A10/AP | A10/AP |
| Row Address | A0-A15 | A0-A15 | A0-A14 |
| Column Address/page size | A0-A9,A11 | A0-A9 | A0-A9 |

### 6.0 Absolute Maximum DC Ratings

| Symbol | Parameter | Rating | Units | Notes |
| :---: | :---: | :---: | :---: | :---: |
| $V_{D D}$ | Voltage on $\mathrm{V}_{\text {DD }}$ pin relative to $\mathrm{V}_{S S}$ | - 1.0 V ~ 2.3 V | V | 1 |
| $V_{\text {DDQ }}$ | Voltage on $\mathrm{V}_{\text {DDQ }}$ pin relative to $\mathrm{V}_{S S}$ | - 0.5V ~ 2.3 V | V | 1 |
| $V_{\text {DDL }}$ | Voltage on $\mathrm{V}_{\text {DDL }}$ pin relative to $\mathrm{V}_{\text {SS }}$ | - $0.5 \mathrm{~V} \sim 2.3 \mathrm{~V}$ | V | 1 |
| $\mathrm{V}_{\text {IN }}, \mathrm{V}_{\text {OUT }}$ | Voltage on any pin relative to $\mathrm{V}_{\text {SS }}$ | - $0.5 \mathrm{~V} \sim 2.3 \mathrm{~V}$ | V | 1 |
| $\mathrm{T}_{\text {STG }}$ | Storage Temperature | -55 to +100 | ${ }^{\circ} \mathrm{C}$ | 1, 2 |

Note :

1. Stresses greater than those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. This is a stress rating only and functional operation of the device at these or any other conditions above those indicated in the operational sections of this specification is not implied. Exposure to absolute maximum rating conditions for extended periods may affect reliability.
2. Storage Temperature is the case surface temperature on the center/top side of the DRAM. For the measurement conditions, please refer to JESD51-2 standard.
3. $V_{D D}$ and $V_{D D Q}$ must be within 300 mV of each other at all times; and $V_{R E F}$ must be not greater than $0.6 \times V_{D D Q}$. When $V_{D D}$ and $V_{D D Q}$ and $V_{D D L}$ are less than 500 mV , $\mathrm{V}_{\text {REF }}$ may be equal to or less than 300 mV .
4. Voltage on any input or I/O may not exceed voltage on $V_{\text {DDQ }}$.

### 7.0 AC \& DC Operating Conditions

### 7.1 Recommended DC Operating Conditions (SSTL - 1.8)

| Symbol | Parameter | Rating |  |  | Units | Notes |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | Min. | Typ. | Max. |  |  |
| $V_{\text {DD }}$ | Supply Voltage | 1.7 | 1.8 | 1.9 | V |  |
| $\mathrm{V}_{\text {DDL }}$ | Supply Voltage for DLL | 1.7 | 1.8 | 1.9 | V | 4 |
| $V_{\text {DDQ }}$ | Supply Voltage for Output | 1.7 | 1.8 | 1.9 | V | 4 |
| $\mathrm{V}_{\text {REF }}$ | Input Reference Voltage | $0.49 * \mathrm{~V}_{\text {DDQ }}$ | $0.50 * V_{\text {DDQ }}$ | $0.51{ }^{*} V_{\text {DDQ }}$ | mV | 1,2 |
| $V_{\text {TT }}$ | Termination Voltage | $\mathrm{V}_{\text {REF }}-0.04$ | $V_{\text {REF }}$ | $\mathrm{V}_{\text {REF }}+0.04$ | V | 3 |

Note : There is no specific device $V_{D D}$ supply voltage requirement for SSTL-1.8 compliance. However under all conditions $V_{D D Q}$ must be less than or equal to $V_{D D}$.

1. The value of $V_{R E F}$ may be selected by the user to provide optimum noise margin in the system. Typically the value of $V_{R E F}$ is expected to be about 0.5 $x V_{D D Q}$ of the transmitting device and $V_{R E F}$ is expected to track variations in $V_{D D Q}$.
2. Peak to peak $A C$ noise on $V_{\text {REF }}$ may not exceed $+/-2 \% V_{\text {REF }}(D C)$.
3. $\mathrm{V}_{\mathrm{TT}}$ of transmitting device must track $\mathrm{V}_{\text {REF }}$ of receiving device.
4. $A C$ parameters are measured with $V_{D D}, V_{D D Q}$ and $V_{D D L}$ tied together.

### 7.2 Operating Temperature Condition

| Symbol | Parameter | Rating | Units | Notes |
| :---: | :---: | :---: | :---: | :---: |
| $\mathrm{T}_{\text {OPER }}$ | Operating Temperature | 0 to 95 | ${ }^{\circ} \mathrm{C}$ | 1,2 |

Note :

1. Operating Temperature is the case surface temperature on the center/top side of the DRAM. For the measurement conditions, please refer to JESD51.2 standard.
2. At $85-95^{\circ} \mathrm{C}$ operation temperature range, doubling refresh commands in frequency to a 32 ms period ( $\mathrm{tREFI}=3.9$ us ) is required, and to enter to self refresh mode at this temperature range, an EMRS command is required to change internal refresh rate.

### 7.3 Input DC Logic Level

| Symbol | Parameter | Min. | Max. | Units | Notes |
| :---: | :---: | :---: | :---: | :---: | :---: |
| $\mathrm{V}_{\mathrm{IH}}(\mathrm{DC})$ | DC input logic high | $\mathrm{V}_{\mathrm{REF}}+0.125$ | $\mathrm{~V}_{\mathrm{DDQ}}+0.3$ | V |  |
| $\mathrm{~V}_{\mathrm{IL}}(\mathrm{DC})$ | DC input logic low | -0.3 | $\mathrm{~V}_{\text {REF }}-0.125$ | V |  |

### 7.4 Input AC Logic Level

| Symbol | Parameter | DDR2-400, DDR2-533 |  | DDR2-667, DDR2-800 |  | Units |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | Min. | Max. | Min. | Max. |  |
| $\mathrm{V}_{\mathrm{IH}}(\mathrm{AC})$ | AC input logic high | $\mathrm{V}_{\text {REF }}+0.250$ | $\mathrm{V}_{\text {DDQ }}+\mathrm{V}_{\text {PEAK }}$ | $\mathrm{V}_{\text {REF }}+0.200$ | $\mathrm{V}_{\text {DDQ }}+\mathrm{V}_{\text {PEAK }}$ | V |
| $\mathrm{V}_{\text {IL }}(\mathrm{AC})$ | AC input logic low | $\mathrm{V}_{\text {SSQ }}-\mathrm{V}_{\text {PEAK }}$ | $\mathrm{V}_{\text {REF }}-0.250$ | $\mathrm{V}_{\text {SSQ }}-\mathrm{V}_{\text {PEAK }}$ | $V_{\text {REF }}-0.200$ | V |

Note :

1. For information related to $V_{\text {PEAK }}$ value, Refer to overshoot/undershoot specification in device operation and timing datasheet; maximum peak amplitude allowed for overshoot and undershoot.

### 7.5 AC Input Test Conditions

| Symbol | Condition | Value | Units | Notes |
| :---: | :---: | :---: | :---: | :---: |
| $V_{\text {REF }}$ | Input reference voltage | $0.5^{*} V_{D D Q}$ | V | 1 |
| $\mathrm{~V}_{\text {SWING(MAX) }}$ | Input signal maximum peak to peak swing | 1.0 | V | 1 |
| SLEW | Input signal minimum slew rate | 1.0 | $\mathrm{~V} / \mathrm{ns}$ |  |

Note :

1. Input waveform timing is referenced to the input signal crossing through the $\mathrm{V}_{\mathrm{IH} / \mathrm{LL}}(A C)$ level applied to the device under test.
2. The input signal minimum slew rate is to be maintained over the range from $V_{R E F}$ to $V_{I H}(A C)$ min for rising edges and the range from $V_{R E F}$ to $V_{I L}(A C)$ max for falling edges as shown in the below figure.
3. $A C$ timings are referenced with input waveforms switching from $V_{I L}(A C)$ to $V_{I H}(A C)$ on the positive transitions and $V_{I H}(A C)$ to $V_{I L}(A C)$ on the negative transitions.

< AC Input Test Signal Waveform >

### 7.6 Differential input AC logic Level

| Symbol | Parameter | Min. | Max. | Units |
| :---: | :---: | :---: | :---: | :---: |
| $\mathrm{V}_{\mathrm{ID}}(\mathrm{AC})$ | AC differential input voltage | 0.5 | $\mathrm{~V}_{\mathrm{DDQ}}$ | V |
| $\mathrm{V}_{\mathrm{IX}}(\mathrm{AC})$ | AC differential cross point voltage | $0.5^{*} \mathrm{~V}_{\mathrm{DDQ}}-0.175$ | $0.5^{*} \mathrm{~V}_{\mathrm{DDQ}}+0.175$ | V |

Note:

1. $\mathrm{V}_{I D}(\mathrm{AC})$ specifies the input differential voltage $\left|\mathrm{V}_{T R}-\mathrm{V}_{\mathrm{CP}}\right|$ required for switching, where $\mathrm{V}_{T R}$ is the true input signal (such as $C K, D Q S, L D Q S$ or UDQS) and $\mathrm{V}_{\mathrm{CP}}$ is the complementary input signal (such as $\overline{\mathrm{CK}}, \overline{\mathrm{DQS}}, \overline{\mathrm{LDQS}}$ or $\overline{\mathrm{UDQS}}$ ). The minimum value is equal to $\mathrm{V}_{\mathrm{IH}}(\mathrm{AC})-\mathrm{V}_{\mathrm{IL}}(\mathrm{AC})$.
2. The typical value of $V_{I X}(A C)$ is expected to be about $0.5^{*} V_{D D Q}$ of the transmitting device and $V_{I X}(A C)$ is expected to track variations in $V_{D D Q}$. $V_{I X}(A C)$ indicates the voltage at which differential input signals must cross.
3. For information related to $V_{\text {PEAK }}$ value, Refer to overshoot/undershoot specification in device operation and timing datasheet; maximum peak amplitude allowed for overshoot and undershoot.


### 7.7 Differential AC output parameters

| Symbol | Parameter | Min. | Max. | Units |
| :---: | :---: | :---: | :---: | :---: |
| $\mathrm{V}_{\mathrm{OX}}(\mathrm{AC})$ | NC differential cross point voltage | $0.5^{*} \mathrm{~V}_{\mathrm{DDQ}}-0.125$ | $0.5^{*} \mathrm{~V}_{\mathrm{DDQ}}+0.125$ | V |

Note:

1. The typical value of $\mathrm{V}_{\mathrm{OX}}(\mathrm{AC})$ is expected to be about 0.5 * $\mathrm{V}_{\mathrm{DDQ}}$ of the transmitting device and $\mathrm{V}_{\mathrm{OX}}(\mathrm{AC})$ is expected to track variations in $\mathrm{V}_{\mathrm{DDQ}}$. $\mathrm{V}_{\mathrm{OX}}(\mathrm{AC})$ indicates the voltage at which differential output signals must cross.

### 8.0 ODT DC electrical characteristics

| PARAMETER/CONDITION | SYMBOL | MIN | NOM | MAX | UNITS |
| :--- | :---: | :---: | :---: | :---: | :---: |
| NOTES |  |  |  |  |  |
| Rtt effective impedance value for EMRS(A6,A2)=0,1; 75 ohm | Rtt1(eff) | 60 | 75 | 90 | ohm |
| Rtt effective impedance value for EMRS(A6,A2) $=1,0 ; 150$ ohm | Rtt2(eff) | 120 | 150 | 180 | ohm |
| Rtt effective impedance value for EMRS(A6,A2)=1,1; 50 ohm | Rtt3(eff) | 40 | 50 | 60 | ohm |
| Deviation of VM with respect to $\mathrm{V}_{\text {DDQ }} / 2$ | delta $\mathrm{V}_{\mathrm{M}}$ | -6 | 1 |  |  |

Note :

1. Test condition for Rtt measurements

Measurement Definition for $R t t(e f f)$ : Apply $\mathrm{V}_{\mathrm{IH}}(\mathrm{AC})$ and $\mathrm{V}_{\mathrm{IL}}(\mathrm{AC})$ to test pin separately, then measure current $\mathrm{I}\left(\mathrm{V}_{\mathrm{IH}}(\mathrm{AC})\right.$ ) and $\mathrm{I}\left(\mathrm{V}_{\mathrm{IL}}(\mathrm{AC})\right)$ respectively. $V_{I H}(A C), V_{I L}(A C)$, and $V_{D D Q}$ values defined in SSTL_18

$$
\operatorname{Rtt}(\text { eff })=\frac{\mathrm{V}_{\mathrm{IH}}(\mathrm{AC})-\mathrm{V}_{\mathrm{IL}}(\mathrm{AC})}{\mathrm{I}\left(\mathrm{~V}_{\mathrm{IH}}(\mathrm{AC})\right)-\mathrm{I}\left(\mathrm{~V}_{\mathrm{IL}}(\mathrm{AC})\right)}
$$

$$
\text { delta } V_{M}=\left(\begin{array}{ll}
\frac{2 \times V_{M}}{V_{D D Q}} & -1
\end{array}\right) \times 100 \%
$$

Measurement Definition for $\mathrm{V}_{\mathrm{M}}$ : Measure voltage $\left(\mathrm{V}_{\mathrm{M}}\right)$ at test pin (midpoint) with no load.

K4T51043QG
K4T51083QG
K4T51163QG

### 9.0 OCD default characteristics

| Description | Parameter | Min | Nom | Max | Unit | Notes |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Output impedance |  | 18ohm at norminal condition See full strength default driver characteristics on device operation specification |  |  | ohms | 1,2 |
| Output impedance step size for OCD calibration |  | 0 |  | 1.5 | ohms | 6 |
| Pull-up and pull-down mismatch |  | 0 |  | 4 | ohms | 1,2,3 |
| Output slew rate | Sout | 1.5 |  | 5 | V/ns | 1,4,5,6,7,8 |

## Note :

1. Absolute Specifications $\left(0^{\circ} \mathrm{C} \leq \mathrm{T}_{\mathrm{CASE}} \leq+95^{\circ} \mathrm{C} ; \mathrm{V}_{\mathrm{DD}}=+1.8 \mathrm{~V} \pm 0.1 \mathrm{~V}, \mathrm{~V}_{\mathrm{DDQ}}=+1.8 \mathrm{~V} \pm 0.1 \mathrm{~V}\right)$
2. Impedance measurement condition for output source dc current: $\mathrm{V}_{\mathrm{DDQ}}=1.7 \mathrm{~V}$; VOUT $=1420 \mathrm{mV} ;\left(\mathrm{V}_{\mathrm{OUT}} \mathrm{V}_{\mathrm{DDQ}}\right) /$ /oh must be less than 23.4 ohms for values of $V_{\text {OUT }}$ between $V_{D D Q}$ and $V_{D D Q}-280 \mathrm{mV}$. Impedance measurement condition for output sink dc current: $V_{D D Q}=1.7 \mathrm{~V} ; \mathrm{V}_{\mathrm{OUT}}=280 \mathrm{mV} ; \mathrm{V}_{\mathrm{OUT}} /$ Iol must be less than 23.4 ohms for values of VOUT between OV and 280 mV .
3. Mismatch is absolute value between pull-up and pull-dn, both are measured at same temperature and voltage.
4. Slew rate measured from $\mathrm{V}_{\mathrm{IL}}(\mathrm{AC})$ to $\mathrm{V}_{\mathrm{IH}}(\mathrm{AC})$.
5. The absolute value of the slew rate as measured from $D C$ to $D C$ is equal to or greater than the slew rate as measured from $A C$ to $A C$. This is guaranteed by design and characterization.
6. This represents the step size when the OCD is near 18 ohms at nominal conditions across all process and represents only the DRAM uncertainty.

Output slew rate load :

7. DRAM output slew rate specification applies to $400 \mathrm{Mb} / \mathrm{sec} / \mathrm{pin}, 533 \mathrm{Mb} / \mathrm{sec} / \mathrm{pin}, 667 \mathrm{Mb} / \mathrm{sec} / \mathrm{pin}$ and $800 \mathrm{Mb} / \mathrm{sec} / \mathrm{pin}$ speed bins.
8. Timing skew due to DRAM output slew rate mis-match between DQS / $\overline{\mathrm{DQS}}$ and associated DQs is included in tDQSQ and tQHS specification.

### 10.0 IDD Specification Parameters and Test Conditions

(IDD values are for full operating range of Voltage and Temperature, Notes 1-5)

| Symbol | Proposed Conditions |  | Units | Notes |
| :---: | :---: | :---: | :---: | :---: |
| IDD0 | Operating one bank active-precharge current; tCK = tCK(IDD), tRC = tRC(IDD), tRAS = tRASmin(IDD); CKE is HIGH, $\overline{\mathrm{CS}}$ is HIGH between valid commands; Address bus inputs are SWITCHING; Data bus inputs are SWITCHING |  | mA |  |
| IDD1 | Operating one bank active-read-precharge current; IOUT $=0 \mathrm{~mA} ; \mathrm{BL}=4, \mathrm{CL}=\mathrm{CL}($ IDD $), \mathrm{AL}=0 ; \mathrm{tCK}=\mathrm{tCK}(\mathrm{IDD}), \mathrm{tRC}=\mathrm{tRC}$ (IDD), tRAS $=\mathrm{tRASmin}(\mathrm{IDD}), \operatorname{tRCD}=$ tRCD(IDD); CKE is HIGH, $\overline{C S}$ is HIGH between valid commands; Address businputs are SWITCHING; Data pattern is same as IDD4W |  | mA |  |
| IDD2P | Precharge power-down current; <br> All banks idle; $\mathrm{CCK}=\mathrm{tCK}$ (IDD); CKE is LOW; Other control and address bus inputs are STABLE; Data bus inputs are FLOATING |  | mA |  |
| IDD2Q | Precharge quiet standby current; <br> All banks ide; tCK = tCK(IDD); CKE is HIGH, $\overline{\mathrm{CS}}$ is HIGH; Other control and address bus inputsare STABLE; Data bus inputs are FLOATING |  | mA |  |
| IDD2N | Precharge standby current; <br> All banks idle; tCK = tCK(IDD); CKE is HIGH, $\overline{\mathrm{CS}}$ is HIGH; Other control and address bus inputs are SWITCHING; Data bus inputs are SWITCHING |  | mA |  |
| IDD3P | Active power-down current; <br> All banks open; tCK = tCK(IDD); CKE is LOW; Other control and address bus inputs are STABLE; Data bus inputs are FLOATING | Fast PDN Exit MRS(12) $=0$ | mA |  |
|  |  | Slow PDN Exit MRS(12) = 1 | mA |  |
| IDD3N | Active standby current; <br> All banks open; $\mathrm{tCK}=\mathrm{tCK}($ IDD $)$, tRAS $=\operatorname{tRASmax}(I D D), \operatorname{tRP}=\operatorname{tRP}(I D D)$; CKE is HIGH, $\overline{\mathrm{CS}}$ is HIGH between valid commands; Other control and address bus inputs are SWITCHING; Data bus inputs are SWITCHING |  | mA |  |
| IDD4W | Operating burst write current; <br> All banks open, Continuous burst writes; $\mathrm{BL}=4, \mathrm{CL}=\mathrm{CL}(\mathrm{IDD}), \mathrm{AL}=0$; $\mathrm{tCK}=\mathrm{tCK}($ IDD $), \mathrm{tRAS}=\mathrm{tRASmax}($ IDD $), \operatorname{tRP}$ $=\operatorname{tRP}(I D D)$; CKE is HIGH, $\overline{\mathrm{CS}}$ is HIGH between valid commands; Address bus inputs are SWITCHING; Data bus inputs are SWITCHING |  | mA |  |
| IDD4R | Operating burst read current; <br> All banks open, Continuous burst reads, $I O U T=0 \mathrm{~mA} ; \mathrm{BL}=4, \mathrm{CL}=\mathrm{CL}(I D D), \mathrm{AL}=0$; $\mathrm{tCK}=\mathrm{tCK}$ (IDD), tRAS $=\mathrm{tRAS}-$ $\max ($ IDD $), ~ t R P=t R P(I D D)$; CKE is HIGH, $\overline{\mathrm{CS}}$ is HIGH between valid commands; Address bus inputs are SWITCHING; Data pattern is same as IDD4W |  | mA |  |
| IDD5B | Burst auto refresh current; <br> tCK = tCK(IDD); Refresh command at every tRFC(IDD) interval; CKE is HIGH, $\overline{\mathrm{CS}}$ is HIGH between valid commands; Other control and address bus inputs are SWITCHING; Data bus inputs are SWITCHING |  | mA |  |
| IDD6 | Self refresh current; <br> CK and $\overline{\mathrm{CK}}$ at 0 V ; CKE $\leq 0.2 \mathrm{~V}$; Other control and address bus inputs are FLOATING; Data bus inputs are FLOATING | Normal | mA |  |
|  |  | Low Power | mA |  |
| IDD7 | Operating bank interleave read current; <br> All bank interleaving reads, IOUT = 0mA; BL = 4, CL = CL(IDD), AL = tRCD(IDD)-1*tCK(IDD); tCK = tCK(IDD), tRC $=\mathrm{tRC}(I D D), \mathrm{tRRD}=\mathrm{tRRD}(I D D), \mathrm{tFAW}=\mathrm{tFAW}(I D D), \mathrm{tRCD}=1^{*} \mathrm{tCK}(I D D)$; CKE is $\mathrm{HIGH}, \overline{\mathrm{CS}}$ is HIGH between valid commands; Address bus inputs are STABLE during DESELECTs; Data pattern is same as IDD4R; Refer to the following page for detailed timing conditions |  | mA |  |

## Note :

1. IDD specifications are tested after the device is properly initialized
2. Input slew rate is specified by AC Parametric Test Condition
3. IDD parameters are specified with ODT disabled.
4. Data bus consists of DQ, DM, DQS, $\overline{\mathrm{DQS}}$, RDQS, $\overline{R D Q S}$, LDQS, $\overline{\mathrm{LDQS}}$, UDQS, and $\overline{\mathrm{UDQS}}$. IDD values must be met with all combinations of EMRS bits 10 and 11.
5. Definitions for IDD

LOW is defined as $V_{I N} \leq V_{\text {IL }} A C$ (max)
HIGH is defined as $V_{I N} \geq V_{I H} A C$ (min)
STABLE is defined as inputs stable at a HIGH or LOW level
FLOATING is defined as inputs at $\mathrm{V}_{\text {REF }}=\mathrm{V}_{\mathrm{DDQ}} / 2$
SWITCHING is defined as:
inputs changing between HIGH and LOW every other clock cycle (once per two clocks) for address and control signals, and
inputs changing between HIGH and LOW every other data transfer (once per clock) for DQ signals not including masks or strobes.

For purposes of IDD testing, the following parameters are utilized

|  | DDR2-800 | DDR2-800 | DDR2-667 | DDR2-533 | DDR2-400 | Units |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Parameter | 5-5-5 | 6-6-6 | 5-5-5 | 4-4-4 | 3-3-3 |  |
| CL(IDD) | 5 | 6 | 5 | 4 | 3 | tCK |
| tRCD(IDD) | 12.5 | 15 | 15 | 15 | 15 | ns |
| tRC(IDD) | 57.5 | 60 | 60 | 60 | 55 | ns |
| tRRD(IDD)-x4/x8 | 7.5 | 7.5 | 7.5 | 7.5 | 7.5 | ns |
| tRRD(IDD)-x16 | 10 | 10 | 10 | 10 | 10 | ns |
| tCK(IDD) | 2.5 | 2.5 | 3 | 3.75 | 5 | ns |
| tRASmin(IDD) | 45 | 45 | 45 | 45 | 40 | ns |
| tRP(IDD) | 12.5 | 15 | 15 | 15 | 15 | ns |
| tRFC(IDD) | 105 | 105 | 105 | 105 | 105 | ns |

## Detailed IDD7

The detailed timings are shown below for IDD7.
Legend: A = Active; RA = Read with Autoprecharge; D = Deselect

## IDD7: Operating Current: All Bank Interleave Read operation

All banks are being interleaved at minimum tRC(IDD) without violating tRRD(IDD) and tFAW(IDD) using a burst length of 4 . Control and address bus inputs are STABLE during DESELECTs. IOUT $=0 \mathrm{~mA}$

Timing Patterns for 4 bank devices $\times 4 / \times 8 / \times 16$
-DDR2-400 3/3/3
A0 RA0 A1 RA1 A2 RA2 A3 RA3 D D D
-DDR2-533 4/4/4
A0 RA0 D A1 RA1 D A2 RA2 D A3 RA3 D D D D D
-DDR2-667 5/5/5
A0 RA0 D D A1 RA1 D D A2 RA2 D D A3 RA3 D D D D D D
-DDR2-667 4/4/4
A0 RA0 D D A1 RA1 D D A2 RA2 D D A3 RA3 D D D D D
-DDR2-800 6/6/6
A0 RA0 D D A1 RA1 D D A2 RA2 D D A3 RA3 D D D D D D D D D
-DDR2-800 5/5/5
A0 RA0 D D A1 RA1 D D A2 RA2 D D A3 RA3 D D D D D D D D D

K4T51043QG
K4T51083QG
K4T51163QG
DDR2 SDRAM

### 11.0 DDR2 SDRAM IDD Spec

| Symbol | 128Mx4 (K4T51043QG) |  |  |  |  |  |  |  |  |  | Unit | Notes |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | 800@CL=5 |  | 800@CL=6 |  | 667@CL=5 |  | 533@CL=4 |  | 400@CL=3 |  |  |  |
|  | CE7 | LE7 | CF7 | LF7 | CE6 | LE6 | CD5 | LD5 | CCC | LCC |  |  |
| IDD0 | 80 |  | 80 |  | 75 |  | 75 |  | 70 |  | mA |  |
| IDD1 | 90 |  | 90 |  | 85 |  | 85 |  | 85 |  | mA |  |
| IDD2P | 8 | 5 | 8 | 5 | 8 | 5 | 8 | 4.5 | 8 | 4.5 | mA |  |
| IDD2Q | 35 |  | 35 |  | 35 |  | 30 |  | 30 |  | mA |  |
| IDD2N | 40 |  | 40 |  | 40 |  | 35 |  | 35 |  | mA |  |
| IDD3P-F | 30 |  | 30 |  | 30 |  | 30 |  | 30 |  | mA |  |
| IDD3P-S | 12 |  | 12 |  | 12 |  | 12 |  | 12 |  | mA |  |
| IDD3N | 55 |  | 55 |  | 55 |  | 50 |  | 50 |  | mA |  |
| IDD4W | 115 |  | 115 |  | 100 |  | 75 |  | 75 |  | mA |  |
| IDD4R | 125 |  | 125 |  | 110 |  | 85 |  | 85 |  | mA |  |
| IDD5 | 115 |  | 115 |  | 110 |  | 105 |  | 105 |  | mA |  |
| IDD6 | 8 | 4 | 8 | 4 | 8 | 4 | 8 | 4 | 8 | 4 | mA |  |
| IDD7 | 210 |  | 210 |  | 175 |  | 175 |  | 175 |  | mA |  |
| Symbol | 64Mx8 (K4T51083QG) |  |  |  |  |  |  |  |  |  | Unit | Notes |
|  | 800@CL=5 |  | 800@CL=6 |  | 667@CL=5 |  | 533@CL=4 |  | 400@CL=3 |  |  |  |
|  | CE7 | LE7 | CF7 | LF7 | CE6 | LE6 | CD5 | LD5 | CCC | LCC |  |  |
| IDD0 | 85 |  | 85 |  | 75 |  | 75 |  | 70 |  | mA |  |
| IDD1 | 95 |  | 95 |  | 90 |  | 85 |  | 85 |  | mA |  |
| IDD2P | 8 | 5 | 8 | 5 | 8 | 5 | 8 | 4.5 | 8 | 4.5 | mA |  |
| IDD2Q | 35 |  | 35 |  | 35 |  | 30 |  | 30 |  | mA |  |
| IDD2N | 40 |  | 40 |  | 40 |  | 35 |  | 35 |  | mA |  |
| IDD3P-F | 30 |  | 30 |  | 30 |  | 30 |  | 30 |  | mA |  |
| IDD3P-S | 12 |  | 12 |  | 12 |  | 12 |  | 12 |  | mA |  |
| IDD3N | 60 |  | 60 |  | 55 |  | 50 |  | 50 |  | mA |  |
| IDD4W | 110 |  | 110 |  | 100 |  | 85 |  | 80 |  | mA |  |
| IDD4R | 140 |  | 140 |  | 130 |  | 105 |  | 95 |  | mA |  |
| IDD5 | 110 |  | 110 |  | 105 |  | 105 |  | 100 |  | mA |  |
| IDD6 | 8 | 4 | 8 | 4 | 8 | 4 | 8 | 4 | 8 | 4 | mA |  |
| IDD7 | 210 |  | 210 |  | 175 |  | 175 |  | 175 |  | mA |  |
| Symbol | 32Mx16 (K4T51163QG) |  |  |  |  |  |  |  |  |  | Unit | Notes |
|  | 800@CL=5 |  | 800@CL=6 |  | 667@CL=5 |  | 533@CL=4 |  | 400@CL=3 |  |  |  |
|  | CE7 | LE7 | CF7 | LF7 | CE6 | LE6 | CD5 | LD5 | CCC | LCC |  |  |
| IDD0 | 95 |  | 95 |  | 90 |  | 90 |  | 90 |  | mA |  |
| IDD1 | 115 |  | 115 |  | 110 |  | 105 |  | 105 |  | mA |  |
| IDD2P | 8 | 5 | 8 | 5 | 8 | 5 | 8 | 4.5 | 8 | 4.5 | mA |  |
| IDD2Q | 35 |  | 35 |  | 35 |  | 30 |  | 30 |  | mA |  |
| IDD2N | 40 |  | 40 |  | 40 |  | 35 |  | 35 |  | mA |  |
| IDD3P-F | 30 |  | 30 |  | 30 |  | 30 |  | 30 |  | mA |  |
| IDD3P-S | 12 |  | 12 |  | 12 |  | 12 |  | 12 |  | mA |  |
| IDD3N | 60 |  | 60 |  |  |  |  |  |  |  | mA |  |
| IDD4W |  |  |  |  |  |  |  |  |  |  | mA |  |
| IDD4R |  |  |  |  |  |  |  |  |  |  | mA |  |
| IDD5 |  |  |  |  |  |  |  |  |  |  | mA |  |
| IDD6 | 8 | 4 | 8 | 4 | 8 | 4 | 8 | 4 | 8 | 4 | mA |  |
| IDD7 |  |  |  |  |  |  |  |  |  |  | mA |  |

### 12.0 Input/Output capacitance

| Parameter | Symbol | DDR2-400 DDR2-533 |  | DDR2-667 |  | DDR2-800 |  | Units |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | Min | Max | Min | Max | Min | Max |  |
| Input capacitance, CK and $\overline{\mathrm{CK}}$ | CCK | 1.0 | 2.0 | 1.0 | 2.0 | 1.0 | 2.0 | pF |
| Input capacitance delta, CK and $\overline{\mathrm{CK}}$ | CDCK | X | 0.25 | X | 0.25 | X | 0.25 | pF |
| Input capacitance, all other input-only pins | Cl | 1.0 | 2.0 | 1.0 | 2.0 | 1.0 | 1.75 | pF |
| Input capacitance delta, all other input-only pins | CDI | x | 0.25 | X | 0.25 | x | 0.25 | pF |
| Input/output capacitance, DQ, DM, DQS, $\overline{\text { DQS }}$ | ClO | 2.5 | 4.0 | 2.5 | 3.5 | 2.5 | 3.5 | pF |
| Input/output capacitance delta, DQ, DM, DQS, $\overline{\text { DQS }}$ | CDIO | X | 0.5 | X | 0.5 | X | 0.5 | pF |

13.0 Electrical Characteristics \& AC Timing for DDR2-800/667/533/400
$\left(0^{\circ} \mathrm{C} \leq \mathrm{T}_{\mathrm{OPER}} \leq 95^{\circ} \mathrm{C} ; \mathrm{V}_{\mathrm{DDQ}}=1.8 \mathrm{~V} \pm 0.1 \mathrm{~V} ; \mathrm{V}_{\mathrm{DD}}=1.8 \mathrm{~V} \pm 0.1 \mathrm{~V}\right)$
13.1 Refresh Parameters by Device Density

| Parameter |  | Symbol | 256Mb | 512Mb | 1Gb | 2Gb | 4Gb | Units |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Refresh to active/Refresh command time | tRFC |  | 75 | 105 | 127.5 | 195 | 327.5 | ns |
| Average periodic refresh interval | tREFI | $0^{\circ} \mathrm{C} \leq \mathrm{T}_{\text {CASE }} \leq 85^{\circ} \mathrm{C}$ | 7.8 | 7.8 | 7.8 | 7.8 | 7.8 | $\mu \mathrm{s}$ |
|  |  | $85^{\circ} \mathrm{C}<\mathrm{T}_{\text {CASE }} \leq 95^{\circ} \mathrm{C}$ | 3.9 | 3.9 | 3.9 | 3.9 | 3.9 | $\mu \mathrm{s}$ |

13.2 Speed Bins and CL, tRCD, tRP, tRC and tRAS for Corresponding Bin

| Speed | DDR | 0(E7) | DDR | 0(F7) | DDR | 7(E6) | DDR | 3(D5) | DDR | (CC) | Units |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Bin (CL - tRCD - tRP) | 5-5-5 |  | 6-6-6 |  | 5-5-5 |  | 4-4-4 |  | 3-3-3 |  |  |
| Parameter | min | max | min | max | min | max | min | max | min | max |  |
| tCK, CL=3 | 5 | 8 | - | - | 5 | 8 | 5 | 8 | 5 | 8 | ns |
| tCK, CL=4 | 3.75 | 8 | 3.75 | 8 | 3.75 | 8 | 3.75 | 8 | 5 | 8 | ns |
| tCK, CL=5 | 2.5 | 8 | 3 | 8 | 3 | 8 | 3.75 | 8 | - | - | ns |
| tCK, CL=6 | - | - | 2.5 | 8 | - | - | - | - | - | - | ns |
| tRCD | 12.5 | - | 15 | - | 15 | - | 15 | - | 15 | - | ns |
| tRP | 12.5 | - | 15 | - | 15 | - | 15 | - | 15 | - | ns |
| tRC | 57.5 | - | 60 | - | 60 | - | 60 | - | 55 | - | ns |
| tRAS | 45 | 70000 | 45 | 70000 | 45 | 70000 | 45 | 70000 | 40 | 70000 | ns |

### 13.3 Timing parameters by speed grade (DDR2-800 and DDR2-667)

(For information related to the entries in this table, refer to both the general notes and the specific notes following this table.)

| Parameter | Symbol | DDR2-800 |  | DDR2-667 |  | Units | Notes |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | min | max | min | max |  |  |
| DQ output access time from CK/ $\overline{C K}$ | tAC | -400 | 400 | -450 | 450 | ps | 40 |
| DQS output access time from CK/ $\overline{\mathrm{CK}}$ | tDQSCK | -350 | 350 | -400 | 400 | ps | 40 |
| Average clock HIGH pulse width | tCH(avg) | 0.48 | 0.52 | 0.48 | 0.52 | tCK(avg) | 35,36 |
| Average clock LOW pulse width | tCL(avg) | 0.48 | 0.52 | 0.48 | 0.52 | tCK(avg) | 35,36 |
| CK half pulse period | tHP | $\begin{gathered} \mathrm{Min}(\mathrm{tCL}(\mathrm{abs}), \\ \mathrm{tCH}(\mathrm{abs})) \end{gathered}$ | x | $\begin{gathered} \operatorname{Min}(\mathrm{tCL}(\mathrm{abs}), \\ \mathrm{tCH}(\mathrm{abs})) \end{gathered}$ | x | ps | 37 |
| Average clock period | tCK(avg) | 2500 | 8000 | 3000 | 8000 | ps | 35,36 |
| DQ and DM input hold time | tDH(base) | 125 | x | 175 | x | ps | 6,7,8,21,28,31 |
| DQ and DM input setup time | tDS(base) | 50 | x | 100 | x | ps | 6,7,8,20,28,31 |
| Control \& Address input pulse width for each input | tIPW | 0.6 | x | 0.6 | x | tCK(avg) |  |
| DQ and DM input pulse width for each input | tDIPW | 0.35 | x | 0.35 | x | tCK(avg) |  |
| Data-out high-impedance time from $\mathrm{CK} / \overline{\mathrm{CK}}$ | tHZ | x | tAC(max) | x | tAC(max) | ps | 18,40 |
| DQS/ $\overline{\text { DQS }}$ low-impedance time from CK/ $/ \overline{C K}$ | tLZ(DQS) | tAC(min) | tAC(max) | tAC(min) | tAC(max) | ps | 18,40 |
| DQ low-impedance time from CK/ $\overline{C K}$ | tLZ(DQ) | $2^{*}$ tAC(min) | tAC(max) | $2^{*}$ tAC(min) | tAC(max) | ps | 18,40 |
| DQS-DQ skew for DQS and associated DQ signals | tDQSQ | x | 200 | x | 240 | ps | 13 |
| DQ hold skew factor | tQHS | x | 300 | x | 340 | ps | 38 |
| DQ/DQS output hold time from DQS | tQH | tHP - tQHS | x | tHP - tQHS | x | ps | 39 |
| DQS latching rising transitions to associated clock edges | tDQSS | - 0.25 | 0.25 | -0.25 | 0.25 | tCK(avg) | 30 |
| DQS input HIGH pulse width | tDQSH | 0.35 | x | 0.35 | x | tCK(avg) |  |
| DQS input LOW pulse width | tDQSL | 0.35 | x | 0.35 | x | tCK(avg) |  |
| DQS falling edge to CK setup time | tDSS | 0.2 | x | 0.2 | x | tCK(avg) | 30 |
| DQS falling edge hold time from CK | tDSH | 0.2 | x | 0.2 | x | tCK(avg) | 30 |
| Mode register set command cycle time | tMRD | 2 | x | 2 | x | nCK |  |
| MRS command to ODT update delay | tMOD | 0 | 12 | 0 | 12 | ns | 32 |
| Write postamble | tWPST | 0.4 | 0.6 | 0.4 | 0.6 | tCK(avg) | 10 |
| Write preamble | tWPRE | 0.35 | x | 0.35 | x | tCK(avg) |  |
| Address and control input hold time | tIH(base) | 250 | x | 275 | x | ps | 5,7,9,23,29 |
| Address and control input setup time | tIS(base) | 175 | x | 200 | x | ps | 5,7,9,22,29 |
| Read preamble | tRPRE | 0.9 | 1.1 | 0.9 | 1.1 | tCK(avg) | 19,41 |
| Read postamble | tRPST | 0.4 | 0.6 | 0.4 | 0.6 | tCK(avg) | 19,42 |
| Activate to activate command period for 1 KB page size products | tRRD | 7.5 | $x$ | 7.5 | $x$ | ns | 4,32 |
| Activate to activate command period for 2KB page size products | tRRD | 10 | x | 10 | x | ns | 4,32 |


| Parameter | Symbol | DDR2-800 |  | DDR2-667 |  | Units | Notes |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | min | max | min | max |  |  |
| Four Activate Window for 1KB page size products | tFAW | 35 | X | 37.5 | x | ns | 32 |
| Four Activate Window for 2KB page size products | tFAW | 45 | x | 50 | x | ns | 32 |
| $\overline{\mathrm{CAS}}$ to $\overline{\mathrm{CAS}}$ command delay | tCCD | 2 | x | 2 | x | nCK |  |
| Write recovery time | tWR | 15 | x | 15 | x | ns | 32 |
| Auto precharge write recovery + precharge time | tDAL | WR + tnRP | x | WR + tnRP | x | nCK | 33 |
| Internal write to read command delay | tWTR | 7.5 | x | 7.5 | x | ns | 24,32 |
| Internal read to precharge command delay | tRTP | 7.5 | x | 7.5 | x | ns | 3,32 |
| Exit self refresh to a non-read command | tXSNR | tRFC + 10 | x | tRFC + 10 | x | ns | 32 |
| Exit self refresh to a read command | tXSRD | 200 | x | 200 | x | nCK |  |
| Exit precharge power down to any command | tXP | 2 | x | 2 | x | nCK |  |
| Exit active power down to read command | tXARD | 2 | x | 2 | x | nCK | 1 |
| Exit active power down to read command (slow exit, lower power) | tXARDS | 8 - AL | x | 7 - AL | x | nCK | 1,2 |
| CKE minimum pulse width (HIGH and LOW pulse width) | tCKE | 3 | x | 3 | x | nCK | 27 |
| ODT turn-on delay | tAOND | 2 | 2 | 2 | 2 | nCK | 16 |
| ODT turn-on | tAON | tAC(min) | tAC(max)+0.7 | tAC(min) | tAC(max)+0.7 | ns | 6,16,40 |
| ODT turn-on (Power-Down mode) | tAONPD | tAC(min)+2 | $\begin{gathered} \hline 2^{\star} \mathrm{tCK}(\mathrm{avg}) \\ +\mathrm{tAC}(\max )+1 \end{gathered}$ | tAC(min)+2 | $\begin{gathered} \quad 2^{\star} \mathrm{tCK}(\mathrm{avg}) \\ +\mathrm{tAC}(\max )+1 \end{gathered}$ | ns |  |
| ODT turn-off delay | tAOFD | 2.5 | 2.5 | 2.5 | 2.5 | nCK | 17,45 |
| ODT turn-off | tAOF | tAC(min) | tAC(max)+0.6 | tAC(min) | tAC(max)+0.6 | ns | 17,43,45 |
| ODT turn-off (Power-Down mode) | tAOFPD | tAC(min)+2 | $\begin{aligned} & 2.5^{*} \mathrm{tCK}(\mathrm{avg}) \\ & +\mathrm{tAC}(\max )+1 \end{aligned}$ | $\mathrm{tAC}($ min $)+2$ | $\begin{aligned} & 2.5^{*} \mathrm{tCK}(\mathrm{avg}) \\ & +\mathrm{tAC}(\max )+1 \end{aligned}$ | ns |  |
| ODT to power down entry latency | tANPD | 3 | $x$ | 3 | x | nCK |  |
| ODT power down exit latency | tAXPD | 8 | x | 8 | x | nCK |  |
| OCD drive mode output delay | tOIT | 0 | 12 | 0 | 12 | ns | 32 |
| Minimum time clocks remains ON after CKE asynchronously drops LOW | tDelay | $\begin{gathered} \text { tIS }+\mathrm{tCK}(\mathrm{avg}) \\ +\mathrm{tIH} \end{gathered}$ | x | $\begin{gathered} \mathrm{tIS}+\mathrm{tCK}(\mathrm{avg}) \\ +\mathrm{tIH} \end{gathered}$ | x | ns | 15 |

### 13.4 Timing parameters by speed grade (DDR2-533 and DDR2-400)

(For information related to the entries in this table, refer to both the general notes and the specific notes following this table.)

| Parameter | Symbol | DDR2-533 |  | DDR2-400 |  | Units | Notes |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | min | max | min | max |  |  |
| DQ output access time from CK/ $\overline{C K}$ | tAC | -500 | 500 | -600 | 600 | ps |  |
| DQS output access time from CK/CK | tDQSCK | -450 | 450 | -500 | 500 | ps |  |
| CK HIGH pulse width | tCH | 0.45 | 0.55 | 0.45 | 0.55 | tCK |  |
| CK LOW pulse width | tCL | 0.45 | 0.55 | 0.45 | 0.55 | tCK |  |
| CK half pulse period | tHP | Min(tCL, tCH) | x | Min(tCL, tCH) | x | ps | 11,12 |
| Clock cycle time, CL=x | tCK | 3750 | 8000 | 5000 | 8000 | ps | 15 |
| DQ and DM input hold time (differential strobe) | tDH(base) | 225 | x | 275 | x | ps | 6,7,8,21,28 |
| DQ and DM input setup time (differential strobe) | tDS(base) | 100 | x | 150 | x | ps | 6,7,8,20,28 |
| DQ and DM input hold time (single-ended strobe) | tDH1 (base) | -25 | x | 25 | x | ps | 6,7,8,26 |
| DQ and DM input setup time (single-ended strobe) | tDS1(base) | -25 | $x$ | 25 | x | ps | 6,7,8,25 |
| Control \& Address input pulse width for each input | tIPW | 0.6 | x | 0.6 | x | tCK |  |
| DQ and DM input pulse width for each input | tDIPW | 0.35 | x | 0.35 | x | tCK |  |
| Data-out high-impedance time from CK/ $\overline{\mathrm{CK}}$ | tHZ | x | tAC(max) | $\times 14$ | tAC(max) | ps | 18 |
| DQS(//̄QS) low-impedance time from CK/ $\overline{\mathrm{CK}}$ | tLZ(DQS) | tAC(min) | tAC(max) | tAC(min) | tAC(max) | ps | 18 |
| DQ low-impedance time from CK/ $\overline{C K}$ | tLZ(DQ) | $2^{*}$ tAC(min) | tAC(max) | 2* tAC(min) | tAC(max) | ps | 18 |
| DQS-DQ skew for DQS and associated DQ signals | tDQSQ | x | 300 | x | 350 | ps | 13 |
| DQ hold skew factor | tQHS | x | 400 | x | 450 | ps | 12 |
| DQ/DQS output hold time from DQS | tQH | tHP - tQ ${ }^{\text {d }}$ | x | tHP - tQHS | x | ps |  |
| DQS latching rising transitions to associated clock edges | tDQSS | -0.25 | 0.25 | -0.25 | 0.25 | tCK |  |
| DQS input HIGH pulse width | tDQSH | 0.35 | x | 0.35 | x | tCK |  |
| DQS input LOW pulse width | tDQSL | 0.35 | x | 0.35 | x | tCK |  |
| DQS falling edge to CK setup time | tDSS | 0.2 | x | 0.2 | x | tCK |  |
| DQS falling edge hold time from CK | tDSH | 0.2 | x | 0.2 | $x$ | tCK |  |
| Mode register set command cycle time | tMRD | 2 | x | 2 | x | tCK |  |
| MRS command to ODT update delay | tMOD | 0 | 12 | 0 | 12 | ns |  |
| Write postamble | tWPST | 0.4 | 0.6 | 0.4 | 0.6 | tCK | 10 |
| Write preamble | tWPRE | 0.35 | x | 0.35 | $x$ | tCK |  |
| Address and control input hold time | tIH(base) | 375 | x | 475 | x | ps | 5,7,9,23 |
| Address and control input setup time | tIS(base) | 250 | x | 350 | x | ps | 5,7,9,22 |
| Read preamble | tRPRE | 0.9 | 1.1 | 0.9 | 1.1 | tCK | 19 |
| Read postamble | tRPST | 0.4 | 0.6 | 0.4 | 0.6 | tCK | 19 |
| Active to active command period for 1 KB page size products | tRRD | 7.5 | x | 7.5 | x | ns | 4 |
| Active to active command period for 2KB page size products | tRRD | 10 | x | 10 | x | ns | 4 |


| Parameter | Symbol | DDR2-533 |  | DDR2-400 |  | Units | Notes |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | min | max | min | max |  |  |
| Four Activate Window for 1KB page size products | tFAW | 37.5 | x | 37.5 | x | ns |  |
| Four Activate Window for 2KB page size products | tFAW | 50 | x | 50 | x | ns |  |
| $\overline{\mathrm{CAS}}$ to $\overline{\mathrm{CAS}}$ command delay | tCCD | 2 | x | 2 | x | tCK |  |
| Write recovery time | tWR | 15 | x | 15 | x | ns |  |
| Auto precharge write recovery + precharge time | tDAL | WR+tRP | x | WR+tRP | $x$ | tCK | 14 |
| Internal write to read command delay | tWTR | 7.5 | x | 10 | x | ns | 24 |
| Internal read to precharge command delay | tRTP | 7.5 | x | 7.5 | x | ns | 3 |
| Exit self refresh to a non-read command | tXSNR | tRFC + 10 | x | tRFC + 10 | x | ns |  |
| Exit self refresh to a read command | tXSRD | 200 | x | 200 | x | tCK |  |
| Exit precharge power down to any non-read command | tXP | 2 | x | 2 | x | tCK |  |
| Exit active power down to read command | tXARD | 2 | x | 2 | x | tCK | 1 |
| Exit active power down to read command (slow exit, lower power) | tXARDS | 6 - AL | x | 6 - AL | x | tCK | 1,2 |
| CKE minimum pulse width (HIGH and LOW pulse width) | tCKE | 3 | x | 3 | x | tCK | 27 |
| ODT turn-on delay | tAOND | 2 | 2 | 2 | 2 | tCK | 16 |
| ODT turn-on | tAON | tAC(min) | tAC(max) +1 | tAC(min) | tAC(max) +1 | ns | 16 |
| ODT turn-on (Power-Down mode) | tAONPD | tAC(min)+2 | $\begin{gathered} 2 \mathrm{tCK}+ \\ \mathrm{tAC}(\max )+1 \end{gathered}$ | tAC(min)+2 | $\begin{gathered} 2 \mathrm{tCK}+ \\ \mathrm{tAC}(\max )+1 \end{gathered}$ | ns |  |
| ODT turn-off delay | tAOFD | 2.5 | 2.5 | 2.5 | 2.5 | tCK | 17,44 |
| ODT turn-off | tAOF | tAC(min) | $\begin{gathered} \hline \mathrm{tAC}(\max ) \\ +0.6 \end{gathered}$ | tAC(min) | $\begin{gathered} \mathrm{tAC}(\max ) \\ +0.6 \end{gathered}$ | ns | 17,44 |
| ODT turn-off (Power-Down mode) | tAOFPD | tAC(min)+2 | $\begin{gathered} 2.5 \mathrm{tCK}+ \\ \mathrm{tAC}(\max )+1 \end{gathered}$ | tAC(min)+2 | $\begin{gathered} 2.5 \mathrm{tCK}+ \\ \mathrm{tAC}(\max )+1 \end{gathered}$ | ns |  |
| ODT to power down entry latency | tANPD | 3 | $x$ | 3 | x | tCK |  |
| ODT power down exit latency | tAXPD | 8 | X | 8 | x | tCK |  |
| OCD drive mode output delay | tOIT | 0 | 12 | 0 | 12 | ns | 32 |
| Minimum time clocks remains ON after CKE asynchronously drops LOW | tDelay | tIS+tCK+tIH | x | tIS+tCK+tIH | x | ns | 15 |

### 14.0 General notes, which may apply for all AC parameters

## 1. DDR2 SDRAM AC timing reference load

Figure 1 represents the timing reference load used in defining the relevant timing parameters of the part. It is not intended to be either a precise repre sentation of the typical system environment or a depiction of the actual load presented by a production tester. System designers will use IBIS or other simulation tools to correlate the timing reference load to a system environment. Manufacturers will correlate to their production test conditions (generally a coaxial transmission line terminated at the tester electronics).


Figure 1 - AC Timing Reference Load
The output timing reference voltage level for single ended signals is the crosspoint with $\mathrm{V}_{\mathrm{TT}}$. The output timing reference voltage level for differential signals is the crosspoint of the true (e.g. DQS) and the complement (e.g. $\overline{\mathrm{DQS}}$ ) signal.

## 2. Slew Rate Measurement Levels

a) Output slew rate for falling and rising edges is measured between $\mathrm{V}_{\mathrm{TT}}-250 \mathrm{mV}$ and $\mathrm{V}_{\mathrm{TT}}+250 \mathrm{mV}$ for single ended signals. For differential signals (e.g. DQS - $\overline{\mathrm{DQS}}$ ) output slew rate is measured between DQS $-\overline{\mathrm{DQS}}=-500 \mathrm{mV}$ and $\mathrm{DQS}-\overline{\mathrm{DQS}}=+500 \mathrm{mV}$. Output slew rate is guaranteed by design, but is not necessarily tested on each device.
b) Input slew rate for single ended signals is measured from $V_{R E F}(D C)$ to $V_{I H}(A C)$, min for rising edges and from $V_{R E F}(D C)$ to $V_{I L}(A C)$, max for falling edges. For differential signals (e.g. CK $-\overline{\mathrm{CK}}$ ) slew rate for rising edges is measured from $\mathrm{CK}-\overline{\mathrm{CK}}=-250 \mathrm{mV}$ to $\mathrm{CK}-\overline{\mathrm{CK}}=+500 \mathrm{mV}(+250 \mathrm{mV}$ to 500 mV for falling edges).
c) $\mathrm{V}_{I D}$ is the magnitude of the difference between the input voltage on CK and the input voltage on $\overline{\mathrm{CK}}$, or between DQS and $\overline{\mathrm{DQS}}$ for differential strobe.

## 3. DDR2 SDRAM output slew rate test load

Output slew rate is characterized under the test conditions as shown in Figure 2.


Figure 2 - Slew Rate Test Load

## 4. Differential data strobe

DDR2 SDRAM pin timings are specified for either single ended mode or differential mode depending on the setting of the EMRS "Enable DQS" mode bit; timing advantages of differential mode are realized in system design. The method by which the DDR2 SDRAM pin timings are measured is mode dependent. In single ended mode, timing relationships are measured relative to the rising or falling edges of DQS crossing at $\mathrm{V}_{\mathrm{REF}}$. In differential mode, these timing relationships are measured relative to the crosspoint of DQS and its complement, $\overline{D Q S}$. This distinction in timing methods is guaranteed by design and characterization. Note that when differential data strobe mode is disabled via the EMRS, the complementary pin, $\overline{\mathrm{DQS}}$, must be tied externally to $\mathrm{V}_{\mathrm{SS}}$ through a $20 \Omega$ to $10 \mathrm{k} \Omega$ resistor to insure proper operation.


Figure 3 - Data Input (Write) Timing


Figure 4 - Data Output (Read) Timing
5. AC timings are for linear signal transitions. See Specific Notes on derating for other signal transitions.
6. All voltages are referenced to $\mathrm{V}_{\mathrm{SS}}$.
7. These parameters guarantee device behavior, but they are not necessarily tested on each device. They may be guaranteed by device design or tester correlation.
8. Tests for AC timing, IDD, and electrical (AC and DC) characteristics, may be conducted at nominal reference/supply voltage levels, but the related specifications and device operation are guaranteed for the full voltage range specified.

### 15.0 Specific Notes for dedicated AC parameters

1. User can choose which active power down exit timing to use via MRS (bit 12). tXARD is expected to be used for fast active power down exit timing tXARDS is expected to be used for slow active power down exit timing.
2. $\mathrm{AL}=$ Additive Latency.
3. This is a minimum requirement. Minimum read to precharge timing is $A L+B L / 2$ provided that the tRTP and tRAS(min) have been satisfied.
4. A minimum of two clocks $(2 x t C K$ or $2 x n C K)$ is required irrespective of operating frequency.
5. Timings are specified with command/address input slew rate of $1.0 \mathrm{~V} / \mathrm{ns}$.
6. Timings are specified with DQs, DM, and DQS's (DQS/RDQS in single ended mode) input slew rate of $1.0 \mathrm{~V} / \mathrm{ns}$.
7. Timings are specified with $\mathrm{CK} / \overline{\mathrm{CK}}$ differential slew rate of $2.0 \mathrm{~V} / \mathrm{ns}$. Timings are guaranteed for DQS signals with a differential slew rate of $2.0 \mathrm{~V} / \mathrm{ns}$ in differential strobe mode and a slew rate of $1.0 \mathrm{~V} / \mathrm{ns}$ in single ended mode.
8. Data setup and hold time derating.

Table 1 - DDR2-400/533 tDS/tDH derating with differential data strobe

| $\Delta t D S,{ }_{\text {d }}$ (DH Derating Values of DDR2-400, DDR2-533 (ALL units in 'ps', the note applies to entire Table) |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | DQS, $\overline{\text { DQS }}$ Differential Slew Rate |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |
|  |  | 4.0 V/ns |  | 3.0 V/ns |  | 2.0 V/ns |  | $1.8 \mathrm{~V} / \mathrm{ns}$ |  | $1.6 \mathrm{~V} / \mathrm{ns}$ |  | 1.4V/ns |  | 1.2V/ns |  | $1.0 \mathrm{~V} / \mathrm{ns}$ |  | $0.8 \mathrm{~V} / \mathrm{ns}$ |  |
|  |  | $\Delta$ tDS | $\Delta \mathrm{tDH}$ | $\Delta$ tDS | $\Delta$ tDH | $\Delta$ tDS | $\Delta \mathrm{tDH}$ | $\Delta$ tDS | $\Delta$ tDH | $\Delta$ tDS | $\Delta \mathrm{tDH}$ | $\Delta$ tDS | $\Delta \mathrm{tDH}$ | $\Delta$ tDS | $\Delta \mathrm{tDH}$ | $\Delta$ tDS | $\Delta \mathrm{tDH}$ | $\Delta$ tDS | $\Delta$ tDH |
| DQ <br> Siew <br> rate <br> V/ns | 2.0 | 125 | 45 | 125 | 45 | 125 | 45 | - | - | - | - | - | - | - | - | - | - | - | - |
|  | 1.5 | 83 | 21 | 83 | 21 | 83 | 21 | 95 | 33 | - | - | - | - | - | - | - | - | - | - |
|  | 1.0 | 0 | 0 | 0 | 0 | 0 | 0 | 12 | 12 | 24 | 24 | - | - | - | - | - | - | - | - |
|  | 0.9 | - | - | -11 | -14 | -11 | -14 | 1 | -2 | 13 | 10 | 25 | 22 | - | - | - | - | - | - |
|  | 0.8 | - | - | - | - | -25 | -31 | -13 | -19 | -1 | -7 | 11 | 5 | 23 | 17 | - | - | - | - |
|  | 0.7 | - | - | - | - | - | - | -31 | -42 | -19 | -30 | -7 | -18 | 5 | -6 | 17 | 6 | - | - |
|  | 0.6 | - | - | - | - | - | - | - | - | -43 | -59 | -31 | -47 | -19 | -35 | -7 | -23 | 5 | -11 |
|  | 0.5 | - | - | - | - | - | - | - | - | - | - | -74 | -89 | -62 | -77 | -50 | -65 | -38 | -53 |
|  | 0.4 | - | - | - | - | - | - | - | - | - | - | - | - | -127 | -140 | -115 | -128 | -103 | -116 |

Table 2 - DDR2-667/800 tDS/tDH derating with differential data strobe

| $\Delta t D S,{ }_{\text {d }}$ (DH Derating Values for DDR2-667, DDR2-800 (ALL units in 'ps', the note applies to entire Table) |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | DQS, $\overline{\text { DQS }}$ Differential Slew Rate |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |
|  |  | $4.0 \mathrm{~V} / \mathrm{ns}$ |  | $3.0 \mathrm{~V} / \mathrm{ns}$ |  | $2.0 \mathrm{~V} / \mathrm{ns}$ |  | $1.8 \mathrm{~V} / \mathrm{ns}$ |  | $1.6 \mathrm{~V} / \mathrm{ns}$ |  | 1.4V/ns |  | $1.2 \mathrm{~V} / \mathrm{ns}$ |  | 1.0V/ns |  | $0.8 \mathrm{~V} / \mathrm{ns}$ |  |
|  |  | $\Delta$ tDS | $\Delta \mathrm{tDH}$ | $\Delta$ tDS | $\Delta \mathrm{tDH}$ | $\Delta$ tDS | $\Delta \mathrm{tDH}$ | $\Delta$ tDS | $\Delta$ tDH | $\Delta$ tDS | $\Delta$ tDH | $\Delta$ tDS | $\Delta$ tDH | $\Delta$ tDS | $\Delta \mathrm{tDH}$ | $\Delta t D S$ | $\Delta$ tDH | $\Delta$ tDS | $\Delta \mathrm{tDH}$ |
| DQ <br> Slew <br> rate <br> V/ns | 2.0 | 100 | 45 | 100 | 45 | 100 | 45 | - | - | - | - | - | - | - | - | - | - | - | - |
|  | 1.5 | 67 | 21 | 67 | 21 | 67 | 21 | 79 | 33 | - | - | - | - | - | - | - | - | - | - |
|  | 1.0 | 0 | 0 | 0 | 0 | 0 | 0 | 12 | 12 | 24 | 24 | - | - | - | - | - | - | - | - |
|  | 0.9 | - | - | -5 | -14 | -5 | -14 | 7 | -2 | 19 | 10 | 31 | 22 | - | - | - | - | - | - |
|  | 0.8 | - | - | - | - | -13 | -31 | -1 | -19 | 11 | -7 | 23 | 5 | 35 | 17 | - | - | - | - |
|  | 0.7 | - | - | - | - | - | - | -10 | -42 | 2 | -30 | 14 | -18 | 26 | -6 | 38 | 6 | - | - |
|  | 0.6 | - | - | - | - | - | - | - | - | -10 | -59 | 2 | -47 | 14 | -35 | 26 | -23 | 38 | -11 |
|  | 0.5 | - | - | - | - | - | - | - | - | - | - | -24 | -89 | -12 | -77 | 0 | -65 | 12 | -53 |
|  | 0.4 | - | - | - | - | - | - | - | - | - | - | - | - | -52 | -140 | -40 | -128 | -28 | -116 |

Table 3 - DDR2-400/533 tDS1/tDH1 derating with single-ended data strobe

| $\Delta$ tDS1, $\Delta$ tDH1 Derating Values for DDR2-400, DDR2-533(All units in 'ps'; the note applies to the entire table) |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | DQS Single-ended Slew Rate |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |
|  |  | 2.0 V/ns |  | $1.5 \mathrm{~V} / \mathrm{ns}$ |  | $1.0 \mathrm{~V} / \mathrm{ns}$ |  | $0.9 \mathrm{~V} / \mathrm{ns}$ |  | $0.8 \mathrm{~V} / \mathrm{ns}$ |  | $0.7 \mathrm{~V} / \mathrm{ns}$ |  | $0.6 \mathrm{~V} / \mathrm{ns}$ |  | $0.5 \mathrm{~V} / \mathrm{ns}$ |  | $0.4 \mathrm{~V} / \mathrm{ns}$ |  |
|  |  | $\begin{array}{\|c} \Delta \mathrm{tDS} \\ 1 \end{array}$ | $\underset{1}{\Delta \mathrm{tDH}}$ | $\Delta \underset{1}{\Delta t D S}$ | $\underset{1}{\Delta t D H}$ | $\Delta{ }_{1}$ | $\underset{1}{\Delta t D H}$ | $\Delta_{1}$ | $\underset{1}{\Delta t D H}$ | $\operatorname{\Delta tDS}_{1}$ | $\underset{1}{\Delta t D H}$ | $\underset{1}{\Delta t D S}$ | $\underset{1}{\Delta t D H}$ | $\Delta t \mathrm{tDS}$ | $\Delta_{1}$ | $\stackrel{\Delta t D S}{1}$ | $\underset{1}{\Delta \mathrm{tDH}}$ | $\underset{1}{\Delta t D S}$ | $\underset{1}{\Delta t_{1}}$ |
|  | 2.0 | 188 | 188 | 167 | 146 | 125 | 63 | - | - | - | - | - | - | - | - | - | - | - | - |
|  | 1.5 | 146 | 167 | 125 | 125 | 83 | 42 | 81 | 43 | - | - | - | - | - | - | - | - | - | - |
|  | 1.0 | 63 | 125 | 42 | 83 | 0 | 0 | -2 | 1 | -7 | -13 | - | - | - | - | - | - | - | - |
| DQ | 0.9 | - | - | 31 | 69 | -11 | -14 | -13 | -13 | -18 | -27 | -29 | -45 | - | - | - | - | - | - |
| Slew | 0.8 | - | - | - | - | -25 | -31 | -27 | -30 | -32 | -44 | -43 | -62 | -60 | -86 | - | - | - | - |
| V/ns | 0.7 | - | - | - | - | - | - | -45 | -53 | -50 | -67 | -61 | -85 | -78 | -109 | -108 | -152 | - | - |
|  | 0.6 | - | - | - | - | - | - | - | - | -74 | -96 | -85 | -114 | -102 | -138 | -138 | -181 | -183 | -246 |
|  | 0.5 | - | - | - | - | - | - | - | - | - | - | -128 | -156 | -145 | -180 | -175 | -223 | -226 | -288 |
|  | 0.4 | - | - | - | - | - | - | - | - | - | - | - | - | -210 | -243 | -240 | -286 | -291 | -351 |

For all input signals the total tDS (setup time) and tDH (hold time) required is calculated by adding the data sheet tDS(base) and tDH(base) value to the $\Delta t D S$ and $\Delta t D H$ derating value respectively. Example: $t D S$ (total setup time) $=t D S$ (base) $+\Delta t D S$.

Setup (tDS) nominal slew rate for a rising signal is defined as the slew rate between the last crossing of $V_{\text {REF }}(D C)$ and the first crossing of $V_{I H}(A C)$ min. Setup (tDS) nominal slew rate for a falling signal is defined as the slew rate between the last crossing of $V_{\text {REF }}(D C)$ and the first crossing of $V_{I L}(A C)$ max. If the actual signal is always earlier than the nominal slew rate line between shaded ' $V_{R E F}(D C)$ to ac region', use nominal slew rate for derating value (See Figure 5 for differential data strobe and Figure 6 for single-ended data strobe.) If the actual signal is later than the nominal slew rate line anywhere between shaded ' $\mathrm{V}_{\mathrm{REF}}(\mathrm{DC})$ to ac region', the slew rate of a tangent line to the actual signal from the ac level to dc level is used for derating value (see Figure 7 for differential data strobe and Figure 8 for single-ended data strobe)

Hold (tDH) nominal slew rate for a rising signal is defined as the slew rate between the last crossing of $\mathrm{V}_{\mathrm{IL}}(\mathrm{DC})$ max and the first crossing of $\mathrm{V}_{\text {REF }}(\mathrm{DC})$. Hold (tDH) nominal slew rate for a falling signal is defined as the slew rate between the last crossing of $\mathrm{V}_{\mathrm{IH}}(\mathrm{DC})$ min and the first crossing of $\mathrm{V}_{\mathrm{REF}}(\mathrm{DC})$. If the actual signal is always later than the nominal slew rate line between shaded 'dc level to $V_{\text {REF }}(D C)$ region', use nominal slew rate for derating value (see Figure 9 for differential data strobe and Figure 10 for single-ended data strobe) If the actual signal is earlier than the nominal slew rate line anywhere between shaded 'dc to $V_{R E F}(D C)$ region', the slew rate of a tangent line to the actual signal from the dc level to $V_{\text {REF }}(D C)$ level is used for derating value (see Figure 11 for differential data strobe and Figure 12 for single-ended data strobe)

Although for slow slew rates the total setup time might be negative (i.e. a valid input signal will not have reached $\mathrm{V}_{\mathrm{IH} / \mathrm{IL}}(\mathrm{AC})$ at the time of the rising clock transition) a valid input signal is still required to complete the transition and reach $\mathrm{V}_{\mathrm{IH} / \mathrm{IL}}(\mathrm{AC})$.

For slew rates in between the values listed in Tables 1, 2 and 3, the derating values may obtained by linear interpolation.
These values are typically not subject to production test. They are verified by design and characterization.

$\begin{gathered}\text { Setup Slew Rate } \\ \text { Falling Signal }\end{gathered}=\frac{V_{\text {REF }}(D C)-V_{\text {IL }}(A C) \max }{\Delta T F}$
$\underset{\text { Retup Slew Rate }}{\text { Sising Signal }}=\frac{\mathrm{V}_{\mathrm{IH}}(\mathrm{AC}) \text { min }-\mathrm{V}_{\mathrm{REF}}(\mathrm{DC})}{\Delta \mathrm{TR}}$

Figure 5 - Illustration of nominal slew rate for tDS (differential DQS, $\overline{\mathrm{DQS}}$ )


Note : DQS signal must be monotonic between $\mathrm{V}_{\mathrm{IL}}(\mathrm{AC}) \max$ and $\mathrm{V}_{\mathrm{IH}}(\mathrm{AC}) \min$.
Figure 6 - Illustration of nominal slew rate for tDS (single-ended DQS)


Setup Slew Rate tangent line[ $\left.\mathrm{V}_{\mathrm{REF}}(\mathrm{DC})-\mathrm{V}_{\mathrm{IL}}(\mathrm{AC}) \mathrm{max}\right]$ Falling Signal $=\frac{\Delta T F}{}$

Figure 7 - Illustration of tangent line for tDS (differential DQS, $\overline{\mathrm{DQS}}$ )


Setup Slew Rate tangent line[ $\left.\mathrm{V}_{\mathrm{REF}}(\mathrm{DC})-\mathrm{V}_{\mathrm{IL}}(\mathrm{AC}) \mathrm{max}\right]$
Falling Signal $=$ $\qquad$ $\Delta T F$

Note : DQS signal must be monotonic between $\mathrm{V}_{\mathrm{IL}}(\mathrm{DC}) \max$ and $\mathrm{V}_{\mathrm{IH}}(\mathrm{DC}) \min$.
Figure 8 - Illustration of tangent line for tDS (single-ended DQS)

$\underset{\text { Hold Slew Rate }}{\text { Rising Signal }}=\frac{\mathrm{V}_{\text {REF }}(\mathrm{DC})-\mathrm{V}_{\mathrm{IL}}(\mathrm{DC}) m a x}{\Delta \mathrm{TR}}$
$\begin{gathered}\text { Hold Slew Rate } \\ \text { Falling Signal }\end{gathered}=\frac{\mathrm{V}_{\mathrm{IH}}(\mathrm{DC}) \mathrm{min}-\mathrm{V}_{\mathrm{REF}}(\mathrm{DC})}{\Delta \mathrm{TF}}$

Figure 9 - Illustration of nominal slew rate for tDH (differential DQS, $\overline{\mathrm{DQS}}$ )


$$
\begin{gathered}
\text { Hold Slew Rate } \\
\text { Rising Signal }
\end{gathered}=\frac{\mathrm{V}_{\mathrm{REF}}(\mathrm{DC})-\mathrm{V}_{\mathrm{IL}}(\mathrm{DC}) m a x}{\Delta \mathrm{TR}} \quad \begin{aligned}
& \text { Hold Slew Rate } \\
& \text { Falling Signal }
\end{aligned}=\frac{\mathrm{V}_{\mathrm{IH}}(\mathrm{DC}) \min -\mathrm{V}_{\mathrm{REF}}(\mathrm{DC})}{\Delta \mathrm{TF}}
$$

Note : DQS signal must be monotonic between $\mathrm{V}_{\mathrm{IL}}(\mathrm{DC}) \max$ and $\mathrm{V}_{\mathrm{IH}}(\mathrm{DC}) \mathrm{min}$.
Figure 10 - Illustration of nominal slew rate for tDH (single-ended DQS)


Figure 11 - Illustration of tangent line for tDH (differential DQS, $\overline{\mathrm{DQS}}$ )


Note : DQS signal must be monotonic between $\mathrm{V}_{\mathrm{IL}}(\mathrm{DC}) \max$ and $\mathrm{V}_{\mathrm{IH}}(\mathrm{DC}) \mathrm{min}$.
Figure 12 - Illustration of tangent line for tDH (single-ended DQS)
9. tIS and tIH (input setup and hold) derating

Table 4 - Derating values for DDR2-400, DDR2-533

| $\Delta$ tIS, $\Delta$ tIH Derating Values for DDR2-400, DDR2-533 |  |  |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | CK, $\overline{\mathrm{CK}}$ Differential Slew Rate |  |  |  |  |  |  |  |
|  |  | 2.0 V/ns |  | 1.5 V/ns |  | 1.0 V/ns |  | Units | Notes |
|  |  | $\Delta$ tIS | $\Delta$ tIH | $\Delta$ tIS | $\Delta \mathrm{tIH}$ | $\Delta$ tIS | $\Delta t \mathrm{H}$ |  |  |
| $\begin{aligned} & \text { Command/ } \\ & \text { Address Slew } \\ & \text { rate(V/ns) } \end{aligned}$ | 4.0 | +187 | +94 | +217 | +124 | +247 | +154 | ps | 1 |
|  | 3.5 | +179 | +89 | +209 | +119 | +239 | +149 | ps | 1 |
|  | 3.0 | +167 | +83 | +197 | +113 | +227 | +143 | ps | 1 |
|  | 2.5 | +150 | +75 | +180 | +105 | +210 | +135 | ps | 1 |
|  | 2.0 | +125 | +45 | +155 | +75 | +185 | +105 | ps | 1 |
|  | 1.5 | +83 | +21 | +113 | +51 | +143 | +81 | ps | 1 |
|  | 1.0 | 0 | 0 | +30 | +30 | +60 | +60 | ps | 1 |
|  | 0.9 | -11 | -14 | +19 | +16 | +49 | +46 | ps | 1 |
|  | 0.8 | -25 | -31 | +5 | -1 | +35 | +29 | ps | 1 |
|  | 0.7 | -43 | -54 | -13 | -24 | +17 | +6 | ps | 1 |
|  | 0.6 | -67 | -83 | -37 | -53 | -7 | -23 | ps | 1 |
|  | 0.5 | -110 | -125 | -80 | -95 | -50 | -65 | ps | 1 |
|  | 0.4 | -175 | -188 | -145 | -158 | -115 | -128 | ps | 1 |
|  | 0.3 | -285 | -292 | -255 | -262 | -225 | -232 | ps | 1 |
|  | 0.25 | -350 | -375 | -320 | -345 | -290 | -315 | ps | 1 |
|  | 0.2 | -525 | -500 | -495 | -470 | -465 | -440 | ps | 1 |
|  | 0.15 | -800 | -708 | -770 | -678 | -740 | -648 | ps | 1 |
|  | 0.1 | -1450 | -1125 | -1420 | -1095 | -1390 | -1065 | ps | 1 |

Table 5 - Derating values for DDR2-667, DDR2-800

| $\Delta t I S$ and $\Delta$ tIH Derating Values for DDR2-667, DDR2-800 |  |  |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | CK, $\overline{\mathrm{CK}}$ Differential Slew Rate |  |  |  |  |  |  |  |
|  |  | 2.0 V/ns |  | 1.5 V/ns |  | 1.0 V/ns |  | Units | Notes |
|  |  | $\Delta$ tIS | $\Delta \mathrm{tIH}$ | $\Delta$ tIS | $\Delta t \mathrm{l}$ | $\Delta$ tIS | $\Delta \mathrm{tIH}$ |  |  |
| Command/ Address Slew rate(V/ns) | 4.0 | +150 | +94 | +180 | +124 | +210 | +154 | ps | 1 |
|  | 3.5 | +143 | +89 | +173 | +119 | +203 | +149 | ps | 1 |
|  | 3.0 | +133 | +83 | +163 | +113 | +193 | +143 | ps | 1 |
|  | 2.5 | +120 | +75 | +150 | +105 | +180 | +135 | ps | 1 |
|  | 2.0 | +100 | +45 | +130 | +75 | +160 | +105 | ps | 1 |
|  | 1.5 | +67 | +21 | +97 | +51 | +127 | +81 | ps | 1 |
|  | 1.0 | 0 | 0 | +30 | +30 | +60 | +60 | ps | 1 |
|  | 0.9 | -5 | -14 | +25 | +16 | +55 | +46 | ps | 1 |
|  | 0.8 | -13 | -31 | +17 | -1 | +47 | +29 | ps | 1 |
|  | 0.7 | -22 | -54 | +8 | -24 | +38 | +6 | ps | 1 |
|  | 0.6 | -34 | -83 | -4 | -53 | +26 | -23 | ps | 1 |
|  | 0.5 | -60 | -125 | -30 | -95 | 0 | -65 | ps | 1 |
|  | 0.4 | -100 | -188 | -70 | -158 | -40 | -128 | ps | 1 |
|  | 0.3 | -168 | -292 | -138 | -262 | -108 | -232 | ps | 1 |
|  | 0.25 | -200 | -375 | -170 | -345 | -140 | -315 | ps | 1 |
|  | 0.2 | -325 | -500 | -295 | -470 | -265 | -440 | ps | 1 |
|  | 0.15 | -517 | -708 | -487 | -678 | -457 | -648 | ps | 1 |
|  | 0.1 | -1000 | -1125 | -970 | -1095 | -940 | -1065 | ps | 1 |

For all input signals the total tIS (setup time) and tIH (hold time) required is calculated by adding the data sheet tIS(base) and tIH(base) value to the $\Delta$ tIS and $\Delta \mathrm{tIH}$ derating value respectively. Example: tIS (total setup time) $=\mathrm{tIS}$ (base) $+\Delta \mathrm{tIS}$

Setup (tIS) nominal slew rate for a rising signal is defined as the slew rate between the last crossing of $\mathrm{V}_{\text {REF }}(\mathrm{DC})$ and the first crossing of $\mathrm{V}_{\mathrm{IH}}(\mathrm{AC})$ min. Setup (tIS) nominal slew rate for a falling signal is defined as the slew rate between the last crossing of $\mathrm{V}_{\text {REF }}(\mathrm{DC})$ and the first crossing of $\mathrm{V}_{\mathrm{IL}}(\mathrm{AC})$ max. If the actual signal is always earlier than the nominal slew rate line between shaded ' $V_{R E F}(D C)$ to ac region', use nominal slew rate for derating value (see Figure 13). If the actual signal is later than the nominal slew rate line anywhere between shaded ' $V_{R E F}(D C)$ to ac region', the slew rate of a tangent line to the actual signal from the ac level to dc level is used for derating value (see Figure 14).

Hold (tIH) nominal slew rate for a rising signal is defined as the slew rate between the last crossing of $\mathrm{V}_{\mathrm{IL}}(\mathrm{DC})$ max and the first crossing of $\mathrm{V}_{\mathrm{REF}}(\mathrm{DC})$. Hold (tIH) nominal slew rate for a falling signal is defined as the slew rate between the last crossing of $\mathrm{V}_{\mathrm{IH}}(\mathrm{DC})$ min and the first crossing of $\mathrm{V}_{\text {REF }}(\mathrm{DC})$. If the actual signal is always later than the nominal slewrate line between shaded 'dc to $V_{R E F}(D C)$ region', use nominal slew rate for derating value (see Figure 15). If the actual signal is earlier than the nominal slew rate line anywhere between shaded 'dc to $V_{R E F}(D C)$ region', the slew rate of a tangent line to the actual signal from the dc level to $\mathrm{V}_{\mathrm{REF}}(\mathrm{DC})$ level is used for derating value (see Figure 16).

Although for slow slew rates the total setup time might be negative (i.e. a valid input signal will not have reached $\mathrm{V}_{\text {IH/IL }}(\mathrm{AC})$ at the time of the rising clock transition) a valid input signal is still required to complete the transition and reach $\mathrm{V}_{\mathrm{IH} / \mathrm{IL}}(\mathrm{AC})$.

For slew rates in between the values listed in Tables 4 and 5, the derating values may obtained by linear interpolation.
These values are typically not subject to production test. They are verified by design and characterization.


Figure 13 - Illustration of nominal slew rate for tIS

$\underset{\text { Setup Slew Rate }}{\text { Falling Signal }}=\frac{\text { tangent line }\left[\mathrm{V}_{\mathrm{REF}}(\mathrm{DC})-\mathrm{V}_{\mathrm{IL}}(\mathrm{AC}) \mathrm{max}\right]}{\Delta \mathrm{TF}}$

Figure 14 - Illustration of tangent line for tIS


Figure 15 - Illustration of nominal slew rate for tlH


Figure 16 - Illustration of tangent line for tIH
10. The maximum limit for this parameter is not a device limit. The device will operate with a greater value for this parameter, but system performance (bus turnaround) will degrade accordingly.
11. MIN ( tCL, tCH) refers to the smaller of the actual clock LOW time and the actual clock HIGH time as provided to the device (i.e. this value can be greater than the minimum specification limits for tCL and tCH ). For example, tCL and tCH are $=50 \%$ of the period, less the half period jitter ( $\mathrm{tJIT}(\mathrm{HP})$ ) of the clock source, and less the half period jitter due to crosstalk ( tJIT(crosstalk)) into the clock traces.
12. $\mathrm{tQH}=\mathrm{tHP}-\mathrm{tQHS}$, where :
tHP = minimum half clock period for any given cycle and is defined by clock HIGH or clock LOW (tCH, tCL).
tQHS accounts for:

1) The pulse duration distortion of on-chip clock circuits; and
2) The worst case push-out of DQS on one transition followed by the worst case pull-in of DQ on the next transition, both of which are, separately, due to data pin skew and output pattern effects, and p-channel to n-channel variation of the output drivers.
13. tDQSQ: Consists of data pin skew and output pattern effects, and $p$-channel to $n$-channel variation of the output drivers as well as output slew rate mismatch between DQS/ $\overline{\mathrm{DQS}}$ and associated $D Q$ in any given cycle.
14. $\operatorname{tDAL}=W R+R U\{\operatorname{tRP}[n s] / \operatorname{tCK}[n s]\}$, where $R U$ stands for round up.

WR refers to the tWR parameter stored in the MRS. For tRP, if the result of the division is not already an integer, round up to the next highest integer. tCK refers to the application clock period.

Example: For DDR533 at $\mathrm{tCK}=3.75 \mathrm{~ns}$ with WR programmed to 4 clocks.
tDAL $=4+(15 \mathrm{~ns} / 3.75 \mathrm{~ns})$ clocks $=4+(4)$ clocks $=8$ clocks.
15. The clock frequency is allowed to change during self refresh mode or precharge power-down mode.
16. ODT turn on time min is when the device leaves high impedance and ODT resistance begins to turn on. ODT turn on time max is when the ODT resistance is fully on. Both are measured from tAOND, which is interpreted differently per speed bin. For DDR2-400/533, tAOND is 10 ns ( $=2 \times 5 \mathrm{~ns}$ ) after the clock edge that registered a first ODT HIGH if tCK $=5 \mathrm{~ns}$. For DDR2-667/800, tAOND is 2 clock cycles after the clock edge that registered a first ODT HIGH counting the actual input clock edges.
17. ODT turn off time min is when the device starts to turn off ODT resistance. ODT turn off time max is when the bus is in high impedance. Both are measured from tAOFD, which is interpreted differently per speed bin. For DDR2-400/533, tAOFD is $12.5 \mathrm{~ns}(=2.5 \times 5 \mathrm{~ns})$ after the clock edge that registered a first ODT LOW if $\mathrm{tCK}=5 \mathrm{~ns}$. For DDR2-667/800, if $\mathrm{tCK}(\mathrm{avg})=3 \mathrm{~ns}$ is assumed, tAOFD is 1.5 ns ( $=0.5 \times 3 \mathrm{~ns}$ ) after the second trailing clock edge counting from the clock edge that registered a first ODT LOW and by counting the actual input clock edges.
18. $t H Z$ and $t L Z$ transitions occur in the same access time as valid data transitions. These parameters are referenced to a specific voltage level which specifies when the device output is no longer driving (tHZ), or begins driving (tLZ). Figure 17 shows a method to calculate the point when device is no longer driving (tHZ), or beginsdriving (tLZ) by measuring the signal at two different voltages. The actual voltage measurement points are not critical as long as the calculation is consistent. $t L Z(D Q)$ refers to $t L Z$ of the DQS and $t L Z(D Q S)$ refers to $t L Z$ of the (U/L/R)DQS and $\overline{(U / L / R) D Q S ~ e a c h ~ t r e a t e d ~ a s ~}$ single-ended signal.
19. tRPST end point and tRPRE begin point are not referenced to a specific voltage level but specify when the device output is no longer driving (tRPST), or begins driving (tRPRE). Figure 17 shows a method to calculate these points when the device is no longer driving (tRPST), or begins driving (tRPRE) by measuring the signal at two different voltages. The actual voltage measurement points are not critical as long as the calculation is consistent.

$\mathrm{tHZ}, \mathrm{tRPST}$ end point $=2^{*}$ T1-T2

tLZ,tRPRE begin point $=2 * T 1-\mathrm{T} 2$

Figure 17 - Method for calculating transitions and endpoints
20. Input waveform timing tDS with differential data strobe enabled $M R[b i t 10]=0$, is referenced from the input signal crossing at the $V_{I H}(A C)$ level to the differential data strobe crosspoint for a rising signal, and from the input signal crossing at the $\mathrm{V}_{\mathrm{IL}}(\mathrm{AC})$ level to the differential data strobe crosspoint for a falling signal applied to the device under test. DQS, $\overline{\mathrm{DQS}}$ signals must be monotonic between $\mathrm{V}_{\mathrm{IL}}(\mathrm{DC}) \max$ and $\mathrm{V}_{\mathrm{IH}}(\mathrm{DC}) \mathrm{min}$. See Figure 18.
21. Input waveform timing $t \mathrm{DH}$ with differential data strobe enabled $\operatorname{MR}[$ bit10] $=0$, is referenced from the differential data strobe crosspoint to the input signal crossing at the $\mathrm{V}_{\mathrm{IH}}(\mathrm{DC})$ level for a falling signal and from the differential data strobe crosspoint to the input signal crossing at the $\mathrm{V}_{\mathrm{IL}}$ ( DC ) level for a rising signal applied to the device under test. DQS, $\overline{\mathrm{DQS}}$ signals must be monotonic between $\mathrm{V}_{\mathrm{IL}}(\mathrm{DC}) \max$ and $\mathrm{V}_{\mathrm{IH}}(\mathrm{DC}) \mathrm{min}$. See Figure 18.


Figure 18 - Differential input waveform timing - tDS and tDH
22. Input waveform timing is referenced from the input signal crossing at the $V_{I H}(A C)$ level for a rising signal and $V_{I L}$ (AC) for a falling signal applied to the device under test. See Figure 19.
23. Input waveform timing is referenced from the input signal crossing at the $V_{I L}(D C)$ level for a rising signal and $V_{I H}(D C)$ for a falling signal applied to the device under test. See Figure 19.


Figure 19 - Differential input waveform timing - tIS and tIH
24. tWTR is at lease two clocks ( $2 \times \mathrm{tCK}$ or 2 xnCK ) independent of operation frequency.
25. Input waveform timing with single-ended data strobe enabled MR[bit10] $=1$, is referenced from the input signal crossing at the $\mathrm{V}_{\mathrm{IH}}(\mathrm{AC})$ level to the single-ended data strobe crossing $\mathrm{V}_{\mathrm{IH} / \mathrm{L}}(\mathrm{DC})$ at the start of its transition for a rising signal, and from the input signal crossing at the $\mathrm{V}_{\mathrm{IL}}(\mathrm{AC})$ level to the single-ended data strobe crossing $\mathrm{V}_{\mathrm{IH} / \mathrm{L}}(\mathrm{DC})$ at the start of its transition for a falling signal applied to the device under test. The DQS signal must be monotonic between $\mathrm{V}_{\mathrm{IL}}(\mathrm{DC})$ max and $\mathrm{V}_{\mathrm{IH}}(\mathrm{DC})$ min.
26. Input waveform timing with single-ended data strobe enabled $M R[b i t 10]=1$, is referenced from the input signal crossing at the $V_{1 H}(D C)$ level to the single-ended data strobe crossing $\mathrm{V}_{\mathrm{IH} / \mathrm{L}}(\mathrm{AC})$ at the end of its transition for a rising signal, and from the input signal crossing at the $\mathrm{V}_{\mathrm{IL}}(\mathrm{DC})$ level to the single-ended data strobe crossing $\mathrm{V}_{\mathrm{IH} / L}(\mathrm{AC})$ at the end of its transition for a falling signal applied to the device under test. The DQS signal must be monotonic between $\mathrm{V}_{\mathrm{IL}}(\mathrm{DC})$ max and $\mathrm{V}_{\mathrm{IH}}(\mathrm{DC})$ min.
27. tCKEmin of 3 clocks means CKE must be registered on three consecutive positive clock edges. CKE must remain at the valid input level the entire time it takes to achieve the 3 clocks of registration. Thus, after any CKE transition, CKE may not transition from its valid level during the time period of tIS + $2 \mathrm{xtCK}+\mathrm{tIH}$.
28. If $t D S$ or $\operatorname{tDH}$ is violated, data corruption may occur and the data must be re-written with valid data before a valid READ can be executed.
29. These parameters are measured from a command/address signal (CKE, $\overline{C S}, \overline{R A S}, \overline{C A S}, \overline{W E}, O D T, B A 0, A 0, A 1$, etc.) transition edge to its respective clock signal (CK/CK) crossing. The spec values are not affected by the amount of clock jitter applied (i.e. $\mathrm{tIIT}(\mathrm{per})$, $\mathrm{tIIT}(\mathrm{cc})$, etc.), as the setup and hold are relative to the clock signal crossing that latches the command/address. That is, these parameters should be met whether clock jitter is present or not.
30. These parameters are measured from a data strobe signal ((L/U/R)DQS/ $\overline{\mathrm{DQS}}$ ) crossing to its respective clock signal (CK/ $\overline{\mathrm{CK}})$ crossing. The spec values are not affected by the amount of clock jitter applied (i.e. JIT (per), $\mathrm{tJIT}(\mathrm{cc})$, etc.), as these are relative to the clock signal crossing. That is, these parameters should be met whether clock jitter is present or not.
31. These parameters are measured from a data signal ((L/U)DM, (L/U)DQ0, (L/U)DQ1, etc.) transition edge to its respective data strobe signal ((L/U/ R)DQS/ $\overline{\mathrm{DQS}}$ ) crossing.
32. For these parameters, the DDR2 SDRAM device is characterized and verified to support tnPARAM $=R U\{t P A R A M / \operatorname{tCK}(a v g)\}$, which is in clock cycles, assuming all input clock jitter specifications are satisfied.
For example, the device will support $\operatorname{tnRP}=\operatorname{RU\{ tRP} / \operatorname{tCK}(\mathrm{avg})\}$, which is in clock cycles, if all input clock jitter specifications are met. This means: For DDR2-667 5-5-5, of which $\operatorname{tRP}=15$ ns, the device will support $\operatorname{tnRP}=\operatorname{RU}\{\operatorname{tRP} / \operatorname{tCK}(\mathrm{avg})\}=5$, i.e. as long as the input clock jitter specifications are met, Precharge command at Tm and Active command at $\mathrm{Tm}+5$ is valid even if ( $\mathrm{Tm}+5-\mathrm{Tm}$ ) is less than 15 ns due to input clock jitter.
33. $\operatorname{tDAL}[\mathrm{nCK}]=\mathrm{WR}[\mathrm{nCK}]+\operatorname{tnRP}[\mathrm{nCK}]=\mathrm{WR}+\mathrm{RU}\{\operatorname{tRP}[\mathrm{ps}] / \operatorname{tCK}(\mathrm{avg})[\mathrm{ps}]\}$, where WR is the value programmed in the mode register set.
34. New units, 'tCK(avg)' and 'nCK', are introduced in DDR2-667 and DDR2-800. Unit 'tCK(avg)' represents the actual tCK(avg) of the input clock under operation. Unit 'nCK' represents one clock cycle of the input clock, counting the actual clock edges.
Note that in DDR2-400 and DDR2-533, 'tCK' is used for both concepts.
ex) tXP = $2[\mathrm{nCK}]$ means; if Power Down exit is registered at Tm , an Active command may be registered at $\mathrm{Tm}+2$, even if $(\mathrm{Tm}+2-\mathrm{Tm})$ is 2 x tCK (avg) $+\operatorname{tERR}(2$ per $)$,min.
35. Input clock jitter spec parameter. These parameters and the ones in the table below are referred to as 'input clock jitter spec parameters' and these parameters apply to DDR2-667 and DDR2-800 only. The jitter specified is a random jitter meeting a Gaussian distribution.

| Parameter | Symbol | DDR2-667 |  | DDR2-800 |  | units | Notes |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | Min | Max | Min | Max |  |  |
| Clock period jitter | tJIT(per) | -125 | 125 | -100 | 100 | ps | 35 |
| Clock period jitter during DLL locking period | tJIT(per,lck) | -100 | 100 | -80 | 80 | ps | 35 |
| Cycle to cycle clock period jitter | tJIT(cc) | -250 | 250 | -200 | 200 | ps | 35 |
| Cycle to cycle clock period jitter during DLL locking period | tJIT(cc,lck) | -200 | 200 | -160 | 160 | ps | 35 |
| Cumulative error across 2 cycles | tERR(2per) | -175 | 175 | -150 | 150 | ps | 35 |
| Cumulative error across 3 cycles | tERR(3per) | -225 | 225 | -175 | 175 | ps | 35 |
| Cumulative error across 4 cycles | tERR(4per) | -250 | 250 | -200 | 200 | ps | 35 |
| Cumulative error across 5 cycles | tERR(5per) | -250 | 250 | -200 | 200 | ps | 35 |
| Cumulative error across n cycles, $\mathrm{n}=6 \ldots 10$, inclusive | tERR(6-10per) | -350 | 350 | -300 | 300 | ps | 35 |
| Cumulative error across n cycles, $\mathrm{n}=11 \ldots 50$, inclusive | tERR(11-50per) | -450 | 450 | -450 | 450 | ps | 35 |
| Duty cycle jitter | tJIT(duty) | -125 | 125 | -100 | 100 | ps | 35 |

## Definitions :

## - tCK(avg)

tCK(avg) is calculated as the average clock period across any consecutive 200 cycle window.

$$
\begin{gathered}
\operatorname{tCK}(\operatorname{avg})=\left(\sum_{j=1}^{\mathrm{N}} \mathrm{tCK}_{j}\right) / \mathrm{N} \\
\text { where } \quad N=200
\end{gathered}
$$

- tCH(avg) and tCL(avg)
$\mathrm{tCH}(\mathrm{avg})$ is defined as the average HIGH pulse width, as calculated across any consecutive 200 HIGH pulses.

$$
\begin{gathered}
\mathrm{tCH}(\mathrm{avg})=\left(\sum_{j=1}^{\mathrm{N}} \mathrm{tCH}_{j}\right) /(\mathrm{N} x \operatorname{tCK}(\mathrm{avg})) \\
\text { where } \quad N=200
\end{gathered}
$$

$\mathrm{tCL}(\mathrm{avg})$ is defined as the average LOW pulse width, as calculated across any consecutive 200 LOW pulses.

$$
\begin{gathered}
\operatorname{tCL}(\operatorname{avg})=\left(\sum_{j=1}^{\mathrm{N}} \mathrm{tCL}_{j}\right) /(\mathrm{N} \times \mathrm{tCK}(\mathrm{avg})) \\
\text { where } \quad N=200
\end{gathered}
$$

- tJIT(duty)
tJIT (duty) is defined as the cumulative set of tCH jitter and tCL jitter. tCH jitter is the largest deviation of any single tCH from tCH (avg). tCL jitter is the largest deviation of any single tCL from tCL(avg).
tJIT(duty) $=\operatorname{Min} / \max$ of $\{\mathrm{tJIT}(\mathrm{CH}), \mathrm{tJIT}(\mathrm{CL})\}$
where,
$\mathrm{tJIT}(\mathrm{CH})=\{\mathrm{tCHi}-\mathrm{tCH}($ avg $)$ where $\mathrm{i}=1$ to 200$\}$
$\mathrm{tJIT}(\mathrm{CL})=\{\mathrm{tCLi}-\mathrm{tCL}(\mathrm{avg})$ where $\mathrm{i}=1$ to 200\}
- tJIT(per), tJIT(per,lck)
tJIT (per) is defined as the largest deviation of any single tCK from tCK(avg).
tJIT(per) $=$ Min/max of $\{t C K i-\mathrm{tCK}($ avg $)$ where $\mathrm{i}=1$ to 200\}
tJIT(per) defines the single period jitter when the DLL is already locked.
tJIT(per,lck) uses the same definition for single period jitter, during the DLL locking period only.
tJIT (per) and tJIT (per,lck) are not guaranteed through final production testing.
- tJIT(cc), tJIT(cc,lck)
tJIT (cc) is defined as the difference in clock period between two consecutive clock cycles : tJIT(cc) $=$ Max of $\left|\mathrm{tCK} \mathrm{K}_{\mathrm{i}+1}-\mathrm{tCKi}\right|$
tJIT(cc) defines the cycle to cycle jitter when the DLL is already locked.
tJIT(cc,Ick) uses the same definition for cycle to cycle jitter, during the DLL locking period only.
$\mathrm{tJIT}(\mathrm{cc})$ and $\mathrm{tJIT}(\mathrm{cc}, \mathrm{lck})$ are not guaranteed through final production testing.
- tERR(2per), tERR (3per), tERR (4per), tERR (5per), tERR (6-10per) and tERR (11-50per)
tERR is defined as the cumulative error across multiple consecutive cycles from tCK(avg).

$$
\begin{aligned}
& \operatorname{tERR}(\mathrm{nper})=\binom{\mathrm{i}+\mathrm{n}-1}{\sum_{j=1} \mathrm{tCK}_{j}}-\mathrm{nxtCK}(\mathrm{avg}) \\
& \text { where }\left(\begin{array}{lll}
n=2 & \text { for } & \operatorname{tERR}(2 \mathrm{per}) \\
n=3 & \text { for } & \operatorname{tERR}(3 p e r) \\
n=4 & \text { for } & \operatorname{tERR}(4 \mathrm{per}) \\
n=5 & \text { for } & \operatorname{tERR}(5 p e r) \\
6 \leq \mathrm{n} \leq 10 & \text { for } & \operatorname{tERR}(6-10 \mathrm{per}) \\
11 \leq \mathrm{n} \leq 50 & \text { for } & \operatorname{tERR}(11-50 \mathrm{per})
\end{array}\right.
\end{aligned}
$$

36. These parameters are specified per their average values, however it is understood that the following relationship between the average timing and the absolute instantaneous timing holds at all times. (Min and max of SPEC values are to be used for calculations in the table below.)

| Parameter | Symbol | Min | Max | Units |
| :---: | :---: | :---: | :---: | :---: |
| Absolute clock Period | tCK(abs) | tCK(avg), min + tJIT(per), min | tCK(avg),max + tJIT(per), max | ps |
| Absolute clock HIGH pulse width | tCH(abs) | tCH(avg), min xtCK(avg), min + tJIT(duty),min | tCH(avg), max x tCK(avg), max + tJIT(duty), max | ps |
| Absolute clock LOW pulse width | tCL(abs) | tCL(avg), min $x$ tCK(avg), min + tJIT(duty),min | $\begin{gathered} \text { tCL(avg),max } x \text { tCK(avg), max }+ \\ \text { tJIT(duty), max } \end{gathered}$ | ps |

## Example: For DDR2-667, tCH(abs), $\min =(0.48 \times 3000 \mathrm{ps})-125 \mathrm{ps}=1315 \mathrm{ps}$

37. tHP is the minimum of the absolute half period of the actual input clock. tHP is an input parameter but not an input specification parameter. It is used in conjunction with tQHS to derive the DRAM output timing tQH. The value to be used for tQH calculation is determined by the following equation; $\mathrm{tHP}=\operatorname{Min}(\mathrm{tCH}(\mathrm{abs}), \mathrm{tCL}(\mathrm{abs}))$,
where,
$\mathrm{tCH}(\mathrm{abs})$ is the minimum of the actual instantaneous clock HIGH time;
$\mathrm{tCL}(\mathrm{abs})$ is the minimum of the actual instantaneous clock LOW time;
38. tQHS accounts for:
1) The pulse duration distortion of on-chip clock circuits, which represents how well the actual tHP at the input is transferred to the output; and
2) The worst case push-out of DQS on one transition followed by the worst case pull-in of DQ on the next transition, both of which are independent of each other, due to data pin skew, output pattern effects, and $p$-channel to $n$-channel variation of the output drivers
39. $\mathrm{tQH}=\mathrm{tHP}-\mathrm{tQHS}$, where:
tHP is the minimum of the absolute half period of the actual input clock; and tQHS is the specification value under the max column.
\{The less half-pulse width distortion present, the larger the tQH value is; and the larger the valid data eye will be.\}
Examples:
1) If the system provides tHP of 1315 ps into a DDR2-667 SDRAM, the DRAM provides tQH of 975 ps minimum.
2) If the system provides tHP of 1420 ps into a DDR2-667 SDRAM, the DRAM provides tQH of 1080 ps minimum.
40. When the device is operated with input clock jitter, this parameter needs to be derated by the actual $\operatorname{tERR}(6-10$ per) of the input clock. (output deratings are relative to the SDRAM input clock.)
For example, if the measured jitter into a DDR2-667 SDRAM has tERR(6-10per), min $=-272 \mathrm{ps}$ and $\operatorname{tERR}(6-10 \mathrm{per})$, max $=+293 \mathrm{ps}$, then tDQSCK, min(derated) $=$ tDQSCK, min $-\operatorname{tERR}(6-10 \mathrm{per}), \max =-400 \mathrm{ps}-293 \mathrm{ps}=-693 \mathrm{ps}$ and $\mathrm{tDQSCK}, \max ($ derated $)=\mathrm{tDQSCK}, \max -\mathrm{tERR}(6-$ 10per), $\mathrm{min}=400 \mathrm{ps}+272 \mathrm{ps}=+672 \mathrm{ps}$. Similarly, $t L Z(D Q)$ for DDR2-667 derates to tLZ(DQ),min(derated) $=-900 \mathrm{ps}-293 \mathrm{ps}=-1193 \mathrm{ps}$ and $\mathrm{tLZ}(\mathrm{DQ}), \max ($ derated $)=450 \mathrm{ps}+272 \mathrm{ps}=+722 \mathrm{ps}$.
41. When the device is operated with input clock jitter, this parameter needs to be derated by the actual tJIT(per) of the input clock. (output deratings are relative to the SDRAM input clock.)
For example, if the measured jitter into a DDR2-667 SDRAM has $\mathrm{tJIT}(\mathrm{per})$, $\mathrm{min}=-72 \mathrm{ps}$ and $\mathrm{tJIT}(\mathrm{per})$, max $=+93 \mathrm{ps}$, then tRPRE, min(derated) $=$ tRPRE, $\min +\mathrm{tJIT}($ per $), \min =0.9 \times \mathrm{tCK}(\mathrm{avg})-72 \mathrm{ps}=+2178 \mathrm{ps}$ and $\mathrm{tRPRE}, \max ($ derated $)=\mathrm{tRPRE}, \max +\mathrm{tJIT}(\mathrm{per}), \max =1.1 \mathrm{xtCK}(\mathrm{avg})+93 \mathrm{ps}=$ +2843 ps .
42. When the device is operated with input clock jitter, this parameter needs to be derated by the actual tJIT(duty) of the input clock. (output deratings are relative to the SDRAM input clock.)
For example, if the measured jitter into a DDR2-667 SDRAM has tJIT(duty), min $=-72 \mathrm{ps}$ and tJIT (duty), max $=+93 \mathrm{ps}$, then tRPST , min(derated) $=$ tRPST, min +tJIT (duty), $\min =0.4 \mathrm{xtCK}(\mathrm{avg})-72 \mathrm{ps}=+928 \mathrm{ps}$ and $\mathrm{tRPST}, \max ($ derated $)=\mathrm{tRPST}, \max +\mathrm{tJIT}($ duty $), \max =0.6 \mathrm{xtCK}(\mathrm{avg})+93 \mathrm{ps}=+$ 1592 ps.
43. When the device is operated with input clock jitter, this parameter needs to be derated by $\{-\operatorname{tJIT}($ duty $), \max -\operatorname{tERR}(6-10 \mathrm{per})$, max $\}$ and $\{-$ tJIT(duty), min - tERR(6-10per), min \} of the actual input clock. (output deratings are relative to the SDRAM input clock.)
For example, if the measured jitter into a DDR2-667 SDRAM has tERR(6-10per), min $=-272 \mathrm{ps}$, tERR(6-10per), max $=+293 \mathrm{ps}, \mathrm{tJIT}(\mathrm{duty})$, min $=-$ 106 ps and tJIT(duty), max $=+94 \mathrm{ps}$, then tAOF, min(derated) $=$ tAOF, min $+\{-\mathrm{tJIT}$ (duty), max $-\operatorname{tERR}(6-10 \mathrm{per}), \max \}=-450 \mathrm{ps}+\{-94 \mathrm{ps}-293 \mathrm{ps}\}=$ -837 ps and $\mathrm{tAOF}, \max ($ derated $)=\mathrm{tAOF}, \max +\{-\mathrm{tJIT}($ duty $), \min -\operatorname{tERR}(6-10 \mathrm{per}), \min \}=1050 \mathrm{ps}+\{106 \mathrm{ps}+272 \mathrm{ps}\}=+1428 \mathrm{ps}$.
44. For tAOFD of DDR2-400/533, the $1 / 2$ clock of tCK in the $2.5 \times$ tCK assumes a tCH, input clock HIGH pulse width of 0.5 relative to tCK. tAOF, min and tAOF,max should each be derated by the same amount as the actual amount of tCH offset present at the DRAM input with respect to 0.5 .
For example, if an input clock has a worst case $t C H$ of 0.45 , the $t A O F$, min should be derated by subtracting $0.05 \times \mathrm{tCK}$ from it, whereas if an input clock has a worst case tCH of 0.55 , the tAOF, max should be derated by adding 0.05 xtCK to it. Therefore, we have;
tAOF, min (derated) $=\mathrm{tAC}, \min -[0.5-\operatorname{Min}(0.5, \mathrm{tCH}, \min )] \times \mathrm{tCK}$
tAOF, max (derated) $=\mathrm{tAC}, \max +0.6+[\operatorname{Max}(0.5, \mathrm{tCH}, \max )-0.5] \times \mathrm{tCK}$
or
$\mathrm{tAOF}, \min ($ derated $)=\operatorname{Min}(\mathrm{tAC}, \min , \mathrm{tAC}, \min -[0.5-\mathrm{tCH}, \min ] x \mathrm{tCK})$
tAOF, max (derated) $=0.6+\operatorname{Max}(\mathrm{tAC}, \max , \mathrm{tAC}, \max +[\mathrm{tCH}, \max -0.5] x \mathrm{tCK})$
where tCH , min and $\mathrm{tCH}, \max$ are the minimum and maximum of tCH actually measured at the DRAM input balls.
45. For tAOFD of DDR2-667/800, the $1 / 2$ clock of $n C K$ in the $2.5 \times n C K$ assumes a tCH(avg), average input clock HIGH pulse width of 0.5 relative to tCK (avg). tAOF, min and tAOF, max should each be derated by the same amount as the actual amount of tCH (avg) offset present at the DRAM input with respect to 0.5 .
For example, if an input clock has a worst case $\mathrm{tCH}(\mathrm{avg})$ of 0.48 , the tAOF , min should be derated by subtracting $0.02 \mathrm{xtCK}(\mathrm{avg})$ from it, whereas if an input clock has a worst case $\mathrm{tCH}(\mathrm{avg})$ of 0.52 , the tAOF , max should be derated by adding $0.02 \times \mathrm{tCK}(\mathrm{avg})$ to it. Therefore, we have;
tAOF, $\min ($ derated $)=\mathrm{tAC}, \min -[0.5-\operatorname{Min}(0.5, \mathrm{tCH}(\mathrm{avg}), \min )] x \operatorname{tCK}(\mathrm{avg})$
$\mathrm{tAOF}, \max ($ derated $)=\mathrm{tAC}, \max +0.6+[\operatorname{Max}(0.5, \mathrm{tCH}(\mathrm{avg}), \max )-0.5] \times \mathrm{tCK}(\mathrm{avg})$
tAOF, min (derated) $=\operatorname{Min}(\mathrm{tAC}, \min , \mathrm{tAC}, \min -[0.5-\mathrm{tCH}(\mathrm{avg}), \min ] \times \mathrm{tCK}(\mathrm{avg}))$
$\mathrm{tAOF}, \max ($ derated $)=0.6+\operatorname{Max}(\mathrm{tAC}, \max , \mathrm{tAC}, \max +[\mathrm{tCH}(\mathrm{avg}), \max -0.5] \times \mathrm{tCK}(\mathrm{avg}))$
where $\mathrm{tCH}(\mathrm{avg})$, min and $\mathrm{tCH}(\mathrm{avg})$, max are the minimum and maximum of $\mathrm{tCH}(\mathrm{avg})$ actually measured at the DRAM input balls.
Note : that these deratings are in addition to the tAOF derating per input clock jitter, i.e. tJIT(duty) and tERR(6-10per). However tAC values used in the equations shown above are from the timing parameter table and are not derated. Thus the final derated values for tAOF are;
tAOF, min(derated_final) $=$ tAOF, min(derated) $+\{-$ tJIT(duty), max - tERR(6-10per), max $\}$
tAOF, max(derated_final) $=$ tAOF, $\max ($ derated $)+\{-\operatorname{tJI}($ duty $), \min -\operatorname{tERR}(6-10$ per $), \min \}$

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