

# ADC1610S125

**Single 16-bit ADC 125 Msps  
CMOS or LVDS DDR digital outputs**

Rev. 01 — 28 May 2009

Objective data sheet

## 1. General description

The ADC1610S is a single channel 16-bit Analog-to-Digital Converter (ADC) optimized for high dynamic performances and low power consumption at a sample rate of 125 Msps. Pipelined architecture and output error correction ensure the ADC1610S is accurate enough to guarantee zero missing codes over the entire operating range. Supplied from a single 3 V source, it can handle output logic levels from 1.8 V to 3.3 V in CMOS mode, because of a separate digital output supply. It supports the Low Voltage Differential Signalling (LVDS) Double Data Rate (DDR) output standard. An integrated Serial Peripheral Interface (SPI) allows the user to easily configure the ADC. The device also includes a programmable gain amplifier with a flexible input voltage range. With excellent dynamic performance from the baseband to input frequencies of 170 MHz or more, the ADC1610S is ideal for use in communications, imaging and medical applications.

## 2. Features

- SNR, 73 dB
- SFDR, 90 dBc
- Sample rate, 125 Msps
- 16-bit pipelined ADC core
- Single 3 V supply
- Flexible input voltage range: 1 V to 2 V (p-p) with 6 dB programmable fine gain
- CMOS or LVDS DDR digital outputs
- INL  $\pm 4$  LSB, DNL  $\pm 0.95$  LSB (typical)
- Input bandwidth, 600 MHz
- Power dissipation, 570 mW
- SPI Interface
- Duty cycle stabilizer
- Fast Out-of-Range (OTR) detection
- Offset binary, 2's complement, gray code
- Power-down and Sleep modes
- HVQFN40 package

## 3. Applications

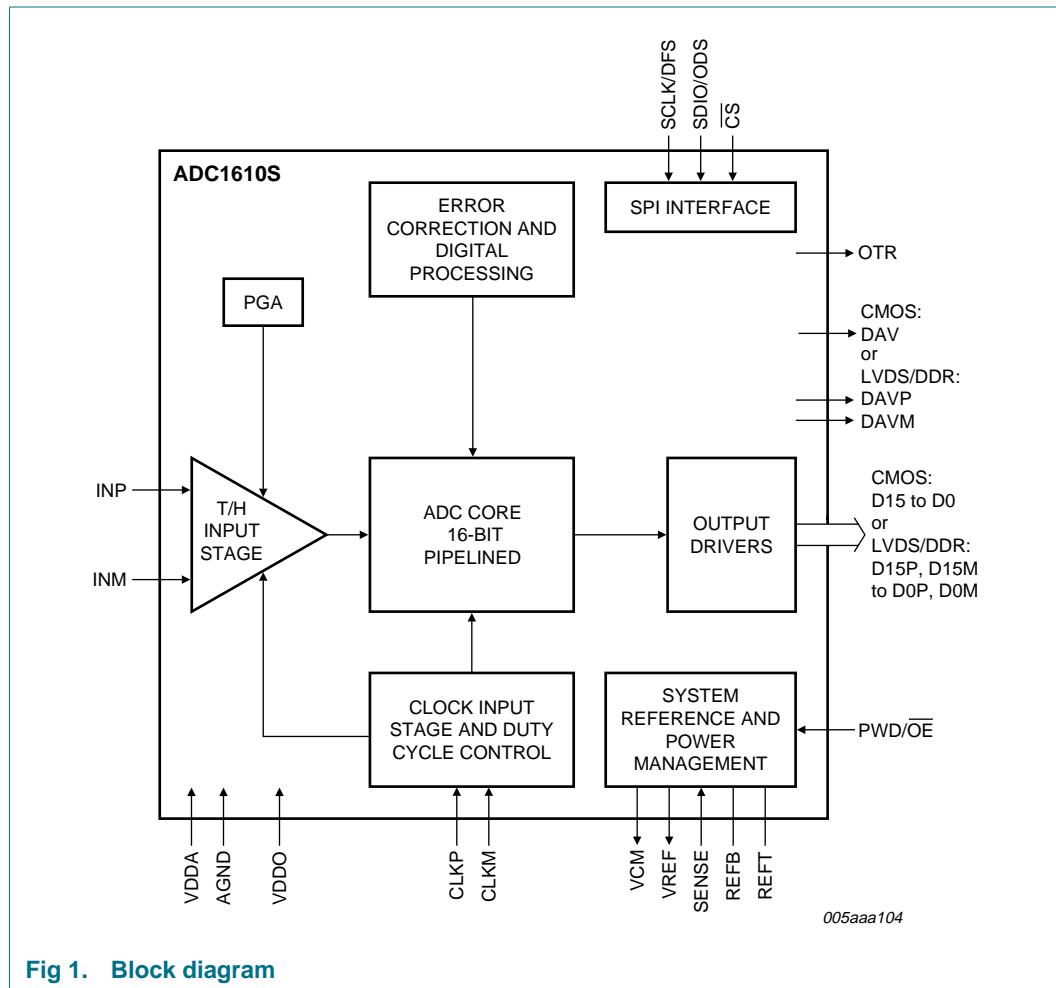
- Wireless and wired broadband communications
- Spectral analysis
- Ultrasound equipment
- Portable instrumentation
- Imaging systems

## 4. Ordering information

**Table 1. Ordering information**

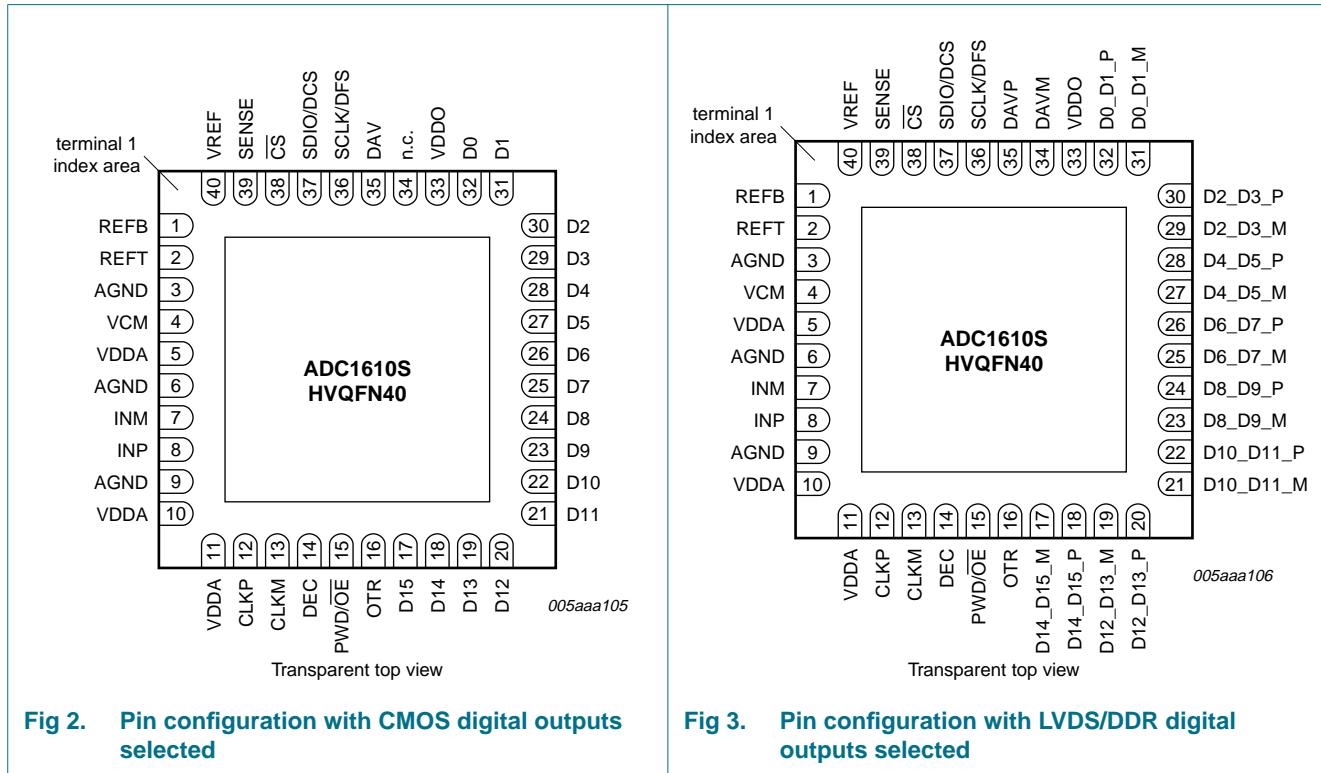
Type number	$f_s$ (Msps)	Package		Version
		Name	Description	
ADC1610S125HN/C1	125	HVQFN40	plastic thermal enhanced very thin quad flat package; no leads; 40 terminals; body $6 \times 6 \times 0.85$ mm	SOT618-1

## 5. Block diagram



## 6. Pinning information

### 6.1 Pinning



### 6.2 Pin description

**Table 2. Pin description (CMOS digital outputs)**

Symbol	Pin	Type <sup>[1]</sup>	Description
REFB	1	O	bottom reference
REFT	2	O	top reference
AGND	3	G	analog ground
VCM	4	O	common-mode output voltage
VDDA	5	P	analog power supply
AGND	6	G	analog ground
INM	7	I	complementary analog input
INP	8	I	analog input
AGND	9	G	analog ground
VDDA	10	P	analog power supply
VDDA	11	P	analog power supply
CLKP	12	I	clock input
CLKM	13	I	complementary clock input
DEC	14	O	regulator decoupling node
PWD/OE	15	I	power-down, active HIGH; output enable, active LOW
OTR	16	O	out-of-range
D15	17		
D14	18		
D13	19		
D12	20		
D11	21		
D10_D11_P	22		
D8_D9_M	23		
D8_D9_P	24		
D6_D7_M	25		
D6_D7_P	26		
D4_D5_M	27		
D4_D5_P	28		
D2_D3_M	29		
D2_D3_P	30		
D0_D1_M	31		
D0_D1_P	32		
VDDO	33		
DAVM	34		
DAVP	35		
SCLK/DFS	36		
SDIO/DCS	37		
CS	38		
SENSE	39		
VREF	40		

**Table 2.** Pin description (CMOS digital outputs)

Symbol	Pin	Type [1]	Description
D15	17	O	data output bit 15 (MSB)
D14	18	O	data output bit 14
D13	19	O	data output bit 13
D12	20	O	data output bit 12
D11	21	O	data output bit 11
D10	22	O	data output bit 10
D9	23	O	data output bit 9
D8	24	O	data output bit 8
D7	25	O	data output bit 7
D6	26	O	data output bit 6
D5	27	O	data output bit 5
D4	28	O	data output bit 4
D3	29	O	data output bit 3
D2	30	O	data output bit 2
D1	31	O	data output bit 1
D0	32	O	data output bit 0 (LSB)
VDDO	33	P	output power supply
n.c.	34	-	not connected
DAV	35	O	data valid output clock
SCLK/DFS	36	I	SPI clock / data format select
SDIO/ODS	37	I/O	SPI data IO / output data standard
CS	38	I	SPI chip select
SENSE	39	I	reference programming pin
VREF	40	I/O	voltage reference input/output

[1] P: power supply; G: ground; I: input; O: output; I/O: input/output.

**Table 3.** Pin description (LVDS/DDR) digital outputs)

Symbol	Pin [1]	Type [2]	Description
D14_D15_M	17	O	differential output data D14 and D15 multiplexed, complement
D14_D15_P	18	O	differential output data D14 and D15 multiplexed, true
D12_D13_M	19	O	differential output data D12 and D13 multiplexed, complement
D12_D13_P	20	O	differential output data D12 and D13 multiplexed, true
D10_D11_M	21	O	differential output data D10 and D11 multiplexed, complement
D10_D11_P	22	O	differential output data D10 and D11 multiplexed, true
D8_D9_M	23	O	differential output data D8 and D9 multiplexed, complement
D8_D9_P	24	O	differential output data D8 and D9 multiplexed, true
D6_D7_M	25	O	differential output data D6 and D7 multiplexed, complement
D6_D7_P	26	O	differential output data D6 and D7 multiplexed, true
D4_D5_M	27	O	differential output data D4 and D5 multiplexed, complement
D4_D5_P	28	O	differential output data D4 and D5 multiplexed, true
D2_D3_M	29	O	differential output data D2 and D3 multiplexed, complement
D2_D3_P	30	O	differential output data D2 and D3 multiplexed, true
D0_D1_M	31	O	differential output data D0 and D1 multiplexed, complement

**Table 3.** Pin description ...continued (LVDS/DDR) digital outputs)

Symbol	Pin [1]	Type [2]	Description
D0_D1_P	32	O	differential output data D0 and D1 multiplexed, true
DAVM	34	O	data valid output clock, complement
DAVP	35	O	data valid output clock, true

[1] Pins 1 to 16 and pins 36 to 40 are the same for both CMOS and LVDS DDR outputs (see [Table 2](#))

[2] P: power supply; G: ground; I: input; O: output; I/O: input/output.

## 7. Limiting values

**Table 4.** Limiting values

In accordance with the Absolute Maximum Rating System (IEC 60134).

Symbol	Parameter	Conditions	Min	Max	Unit
V <sub>DDA</sub>	analog supply voltage		<tbd>	<tbd>	V
V <sub>DDO</sub>	output supply voltage		<tbd>	<tbd>	V
ΔV <sub>CC</sub>	supply voltage difference	V <sub>DDA</sub> – V <sub>DDO</sub>	<tbd>	<tbd>	V
T <sub>stg</sub>	storage temperature		-55	+125	°C
T <sub>amb</sub>	ambient temperature		-40	+85	°C
T <sub>j</sub>	junction temperature		-	<tbd>	°C

## 8. Thermal characteristics

**Table 5.** Thermal characteristics

Symbol	Parameter	Conditions	Typ	Unit
R <sub>th(j-a)</sub>	thermal resistance from junction to ambient		[1] <tbd>	K/W
R <sub>th(j-c)</sub>	thermal resistance from junction to case		[1] <tbd>	K/W

[1] In compliance with JEDEC test board, in free air.

## 9. Static characteristics

**Table 6. Static characteristics**

Typical values measured at  $V_{DDA} = 3\text{ V}$ ,  $V_{DDO} = 1.8\text{ V}$ ,  $T_{amb} = 25^\circ\text{C}$  and  $C_L = 5\text{ pF}$ ; min. and max. values are across the full temperature range  $T_{amb} = -40^\circ\text{C}$  to  $+85^\circ\text{C}$  at  $V_{DDA} = 3\text{ V}$ ,  $V_{DDO} = 1.8\text{ V}$ ;  $V_{INP} - V_{INM} = -1\text{ dBFS}$ ; internal reference mode; applied to CMOS and LVDS interface; unless otherwise specified.

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
<b>Supplies</b>						
$V_{DDA}$	analog supply voltage		2.85	3.0	3.4	V
$V_{DDO}$	output supply voltage	CMOS mode	1.65	1.8	3.6	V
		LVDS DDR mode	2.85	3.0	3.6	V
$I_{DDA}$	analog supply current	$f_{clk} = 125\text{ Msps}$ ; $f_i = 70\text{ MHz}$	-	185	-	mA
$I_{DDO}$	output supply current	CMOS mode; $f_{clk} = 125\text{ Msps}$ ; $f_i = 70\text{ MHz}$	-	20	-	mA
		LVDS DDR mode: $f_{clk} = 125\text{ Msps}$ ; $f_i = 70\text{ MHz}$	-	45	-	mA
$P$	power dissipation		-	570	-	mW
		Power-down mode	-	2	-	mW
		Sleep mode	-	40	-	mW
<b>Clock inputs: pins CLKP and CLKM</b>						
AC coupled; LVPECL, LVDS and sine wave						
$V_{i(clk)diff}$	differential clock input voltage	peak-to-peak	0.2	0.8	<tbd>	V
<b>LVCMS</b>						
$V_I$	input voltage		$0.3V_{DDA}$	-	$0.7V_{DDA}$	V
<b>Logic Inputs: pin PWD/OE</b>						
$V_{IL}$	LOW-level input voltage		-	0	-	V
		LOW-medium level	-	$0.3V_{DDA}$	-	
		medium-HIGH level	-	$0.6V_{DDA}$	-	
$V_{IH}$	HIGH-level input voltage		-	$V_{DDA}$	-	V
$I_{IL}$	LOW-level input current		<tbd>	-	<tbd>	$\mu\text{A}$
$I_{IH}$	HIGH-level input current		-10	-	+10	$\mu\text{A}$
<b>Serial Peripheral Interface: pins CS, SDIO/ODS, SCLK/DFS</b>						
$V_{IL}$	LOW-level input voltage		0	-	$0.3V_{DDA}$	V
$V_{IH}$	HIGH-level input voltage		$0.7V_{DDA}$	-	$V_{DDA}$	V
$I_{IL}$	LOW-level input current		-10	-	+10	$\mu\text{A}$
$I_{IH}$	HIGH-level input current		-50	-	+50	$\mu\text{A}$
$C_I$	input capacitance		-	4	-	pF
<b>Digital Outputs: CMOS mode - pins D13 to D0, OTR, DAV</b>						
Output levels, $V_{DDO} = 3\text{ V}$						
$V_{OL}$	LOW-level output voltage	$I_{OL} = <\text{tbd}>$	AGND	-	$0.2V_{DDO}$	V
$V_{OH}$	HIGH-level output voltage	$I_{OH} = <\text{tbd}>$	$0.8V_{DDO}$	-	$V_{DDO}$	V
$I_{OL}$	LOW-level output current	3-state; output level = 0 V	-	<tbd>	-	$\mu\text{A}$
$I_{OH}$	HIGH-level output current	3-state; output level = $V_{DDA}$	-	<tbd>	-	$\mu\text{A}$
$C_O$	output capacitance	high impedance; $\overline{OE} = \text{HIGH}$	-	3	-	pF

**Table 6. Static characteristics ...continued**

Typical values measured at  $V_{DDA} = 3$  V,  $V_{DDO} = 1.8$  V,  $T_{amb} = 25^\circ\text{C}$  and  $C_L = 5 \text{ pF}$ ; min. and max. values are across the full temperature range  $T_{amb} = -40^\circ\text{C}$  to  $+85^\circ\text{C}$  at  $V_{DDA} = 3$  V,  $V_{DDO} = 1.8$  V;  $V_{INP} - V_{INM} = -1 \text{ dBFS}$ ; internal reference mode; applied to CMOS and LVDS interface; unless otherwise specified.

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
<b>Output levels, <math>V_{DDO} = 1.8</math> V</b>						
$V_{OL}$	LOW-level output voltage	$I_{OL} = <\text{tbd}>$	AGND	-	$0.2V_{DDO}$	V
$V_{OH}$	HIGH-level output voltage	$I_{OH} = <\text{tbd}>$	$0.8V_{DDO}$	-	$V_{DDO}$	V
<b>Digital Outputs, LVDS mode - pins D13P, D13M to D0P, D0M, DAVP and DAVM</b>						
<b>Output levels, <math>V_{DDO} = 3</math> V only, <math>R_{load} = 100 \Omega</math></b>						
$V_{O(\text{offset})}$	output offset voltage	output buffer current set to 3.5 mA	-	1.2	-	V
$V_{O(\text{dif})}$	differential output voltage	output buffer current set to 3.5 mA	-	350	-	mV
$C_O$	output capacitance		-	$<\text{tbd}>$	-	pF
<b>Analog inputs: pins INP and INM</b>						
$I_I$	Input current		-5	-	+5	$\mu\text{A}$
$R_I$	input resistance		-	$<\text{tbd}>$	-	$\Omega$
$C_I$	input capacitance		-	5	-	pF
$V_{I(\text{cm})}$	common-mode input voltage	$V_{INP} = V_{INM}$	0.9	1.5	2	V
$B_i$	input bandwidth		-	600	-	MHz
$V_{I(\text{dif})}$	differential input voltage	peak-to-peak	1	-	2	V
<b>Common mode output voltage: pin VCM</b>						
$V_{O(\text{cm})}$	common-mode output voltage		-	$0.5V_{DDA}$	-	V
$I_{O(\text{cm})}$	common-mode output current		-	$<\text{tbd}>$	-	$\mu\text{A}$
<b>I/O reference voltage: pin VREF</b>						
$V_{VREF}$	voltage on pin VREF	output	-	0.5 to 1	-	V
		input	0.5	-	1	V
<b>Accuracy</b>						
INL	integral non-linearity		-	$\pm 4$	-	LSB
DNL	differential non-linearity	guaranteed no missing codes	-	$\pm 0.95$	-	LSB
$E_{\text{offset}}$	offset error		-	$\pm 2$	-	mV
$E_G$	gain error		-	$\pm 0.5$	-	%FS
<b>Supply</b>						
PSRR	power supply rejection ratio	100 mV (p-p) on $V_{DDA}$	-	35	-	dBc

## 10. Dynamic characteristics

### 10.1 Dynamic characteristics

**Table 7. Dynamic characteristics**

Typical values measured at  $V_{DDA} = 3\text{ V}$ ,  $V_{DDO} = 1.8\text{ V}$ ,  $T_{amb} = 25^\circ\text{C}$  and  $C_L = 5\text{ pF}$ ; min. and max. values are across the full temperature range  $T_{amb} = -40^\circ\text{C}$  to  $+85^\circ\text{C}$  at  $V_{DDA} = 3\text{ V}$ ,  $V_{DDO} = 1.8\text{ V}$ ;  $V_{INP} - V_{INM} = -1\text{ dBFS}$ ; internal reference mode; applied to CMOS and LVDS interface; unless otherwise specified.

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
<b>Analog signal processing</b>						
$\alpha_{2H}$	second harmonic level	$f_i = 3\text{ MHz}$	-	95	-	dBc
		$f_i = 30\text{ MHz}$	-	92	-	dBc
		$f_i = 70\text{ MHz}$	-	89	-	dBc
		$f_i = 170\text{ MHz}$	-	88	-	dBc
$\alpha_{3H}$	third harmonic level	$f_i = 3\text{ MHz}$	-	90	-	dBc
		$f_i = 30\text{ MHz}$	-	89	-	dBc
		$f_i = 70\text{ MHz}$	-	88	-	dBc
		$f_i = 170\text{ MHz}$	-	87	-	dBc
THD	total harmonic distortion	$f_i = 3\text{ MHz}$	-	87	-	dBc
		$f_i = 30\text{ MHz}$	-	86	-	dBc
		$f_i = 70\text{ MHz}$	-	83	-	dBc
		$f_i = 170\text{ MHz}$	-	79	-	dBc
ENOB	effective number of bits	$f_i = 3\text{ MHz}$	-	11.8	-	bits
		$f_i = 30\text{ MHz}$	-	11.7	-	bits
		$f_i = 70\text{ MHz}$	-	11.6	-	bits
		$f_i = 170\text{ MHz}$	-	11.5	-	bits
SNR	signal-to-noise ratio	$f_i = 3\text{ MHz}$	-	73	-	dBFS
		$f_i = 30\text{ MHz}$	-	72.2	-	dBFS
		$f_i = 70\text{ MHz}$	-	71.6	-	dBFS
		$f_i = 170\text{ MHz}$	-	71	-	dBFS
	at $-20\text{ dBFS}$					
		$f_i = 3\text{ MHz}$	-	74	-	dBFS
		$f_i = 30\text{ MHz}$	-	73.5	-	dBFS
		$f_i = 70\text{ MHz}$	-	73.4	-	dBFS
		$f_i = 170\text{ MHz}$	-	73.3	-	dBFS
SFDR	spurious-free dynamic range	$f_i = 3\text{ MHz}$	-	90	-	dBc
		$f_i = 30\text{ MHz}$	-	89	-	dBc
		$f_i = 70\text{ MHz}$	-	88	-	dBc
		$f_i = 170\text{ MHz}$	-	87	-	dBc
IMD	Intermodulation distortion	$f_i = 3\text{ MHz}$	-	93	-	dBc
		$f_i = 30\text{ MHz}$	-	92	-	dBc
		$f_i = 70\text{ MHz}$	-	91	-	dBc
		$f_i = 170\text{ MHz}$	-	90	-	dBc

## 10.2 Clock and digital output timing

**Table 8. Dynamic characteristics**

Typical values measured at  $V_{DDA} = 3\text{ V}$ ,  $V_{DDO} = 1.8\text{ V}$ ,  $T_{amb} = 25^\circ\text{C}$  and  $C_L = 5\text{ pF}$ ; min. and max. values are across the full temperature range  $T_{amb} = -40^\circ\text{C}$  to  $+85^\circ\text{C}$  at  $V_{DDA} = 3\text{ V}$ ,  $V_{DDO} = 1.8\text{ V}$ ;  $V_{INP} - V_{INM} = -1\text{ dBFS}$ ; unless otherwise specified.

Symbol	Parameter	Conditions	Min	Typ	Max	Unit	
<b>Clock timing input: pins CLKP and CLKM</b>							
$f_{clk}$	clock frequency		60	-	125	MHz	
$t_{lat(data)}$	data latency time		-	14	-	clk/cy	
$\delta_{clk}$	clock duty cycle	$DCS\_EN = 1$	30	50	70	%	
		$DCS\_EN = 0$	45	50	55	%	
$t_{d(s)}$	sampling delay time		-	0.8	-	ns	
$t_{wake}$	wake-up time		-	<tbd>	-	ns	
<b>CMOS mode timing output: pins D13 to D0 and DAV</b>							
$t_{PD}$	propagation delay	DATA	-	3.9	-	ns	
		DAV	-	4.2	-	ns	
$t_{su}$	set-up time		-	4.3	-	ns	
$t_h$	hold time		-	3.5	-	ns	
$t_r$	rise time	DATA	[1]	0.5	-	2.4	ns
		DAV		0.5	-	2.4	ns
$t_f$	fall time	DATA	[1]	0.5	-	2.4	ns
<b>LVDS DDR mode timing output: pins D13P, D13M to D0P, D0M, DAVP and DAVM</b>							
$t_{PD}$	propagation delay	DATA	-	3.9	-	ns	
		DAV	-	4.2	-	ns	

[1] Measured between 20 % to 80 % of  $V_{DDO}$ ; rise time measured from  $-50\text{ mV}$  to  $+50\text{ mV}$ ; fall time measured from  $+50\text{mV}$  to  $-50\text{mV}$ .

## 10.3 SPI timings

**Table 9. Characteristics**

Typical values measured at  $V_{DDA} = 3\text{ V}$ ,  $V_{DDO} = 1.8\text{ V}$ ,  $T_{amb} = 25^\circ\text{C}$  and  $C_L = 5\text{ pF}$ .

Min. and max. values are across the full temperature range  $T_{amb} = -40^\circ\text{C}$  to  $+85^\circ\text{C}$  at  $V_{DDA} = 3\text{V}$ ,  $V_{DDO} = 1.8\text{ V}$

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
<b>SPI timings</b>						
$t_w(SCLK)$	SCLK pulse width		40	-	-	ns
$t_w(SCLKH)$	SCLK HIGH pulse width		16	-	-	ns
$t_w(SCLKL)$	SCLK LOW pulse width		16	-	-	ns
$t_{su}$	set-up time	data to SCLKH	5	-	-	ns
		$\bar{CS}$ to SCLKH	5	-	-	ns
$t_h$	hold time	data to SCLKH	2	-	-	ns
		$\bar{CS}$ to SCLKH	2	-	-	ns
$f_{clk(max)}$	maximum clock frequency		-	-	25	MHz

## 11. Application information

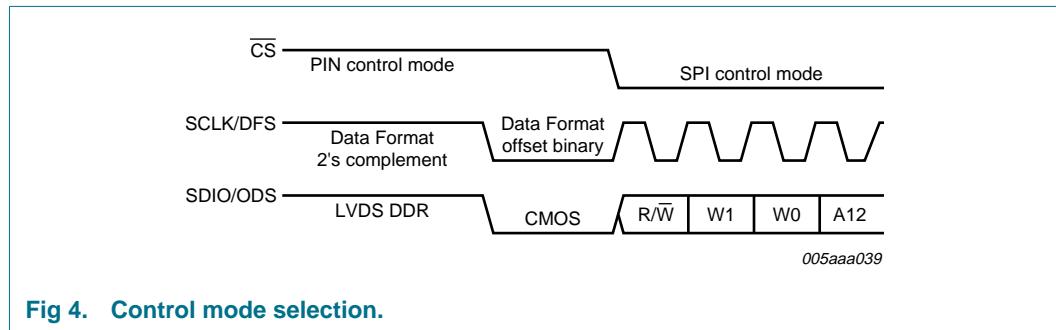
### 11.1 Device control

The ADC1610S can be controlled via the SPI (SPI control mode) or directly via the I/O pins (PIN control mode).

#### 11.1.1 SPI and PIN control modes

The device enters PIN control mode at power-up, and remains in this mode as long as pin  $\overline{CS}$  is held HIGH. In PIN control mode, the SPI pins SDIO,  $\overline{CS}$  and SCLK are used as static control pins. SPI settings are ignored.

SPI control mode is enabled by forcing pin  $\overline{CS}$  LOW. It is not possible to toggle between PIN control and SPI control modes. Once SPI control mode has been enabled, the device will remain in this mode until it is powered down. The transition from PIN control mode to SPI control mode is illustrated in [Figure 4](#).



**Fig 4. Control mode selection.**

When the device enters SPI control mode, the output data standard (CMOS or LVDS DDR) is not determined by the state of the relevant SPI control bit (LVDS/CMOS; see [Table 21](#)), but by the level on pin SDIO at the instant a transition is triggered by a falling edge on  $\overline{CS}$  (SDIO = LOW = CMOS).

#### 11.1.2 Operating mode selection

The active ADC1610S operating mode (Power-up, Power-down or Sleep) can be selected via the SPI interface (see [Table 18](#)) or using pins PWD and  $\overline{OE}$  in PIN control mode, as described in [Table 10](#).

**Table 10. Operating mode selection via pin PWD/ $\overline{OE}$**

PWD/ $\overline{OE}$	Operating mode	Output high-Z
0	Power-down	yes
$1/3V_{DDA}$	Sleep	yes
$2/3V_{DDA}$	Power-up	yes
$V_{DDA}$	Power-up	no

#### 11.1.3 Selecting the output data standard

The output data standard (CMOS or LVDS DDR) can be selected via the SPI interface (see [Table 21](#)) or using pin ODS in PIN control mode. LVDS DDR is selected when ODS is HIGH, otherwise CMOS is selected.

#### 11.1.4 Selecting the output data format

The output data format can be selected via the SPI interface (offset binary, 2's complement or gray code; see [Table 21](#)) or using pin DFS in PIN control mode (offset binary or 2's complement). Offset binary is selected when DFS is LOW. When DFS is HIGH, 2's complement is selected.

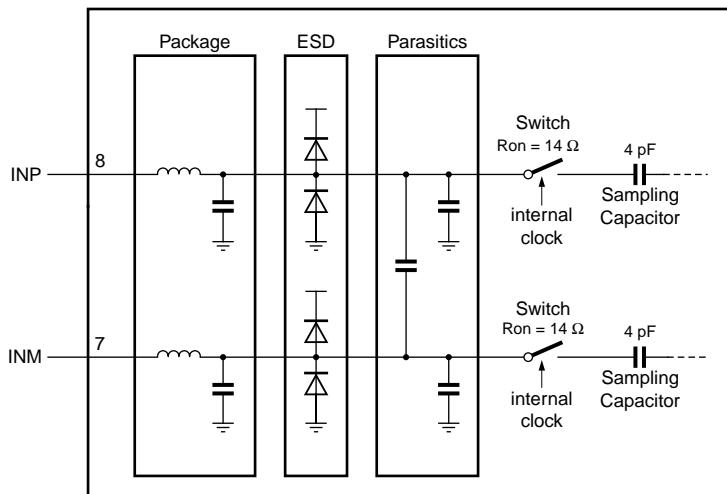
### 11.2 Analog inputs

#### 11.2.1 Input stage

The analog input of the ADC1610S supports a differential or a single-ended input drive. Optimal performance is achieved using differential inputs with the common-mode input voltage ( $V_{I(cm)}$ ) on pins INP and INM set to  $0.5V_{DDA}$ .

The full scale analog input voltage range is configurable between 1 V (p-p) and 2 V (p-p) via a programmable internal reference (see [Section 11.3](#) and [Table 20](#) further details).

The equivalent circuit of the sample and hold input stage, including ESD protection and circuit and package parasitics, is shown in [Figure 5](#).



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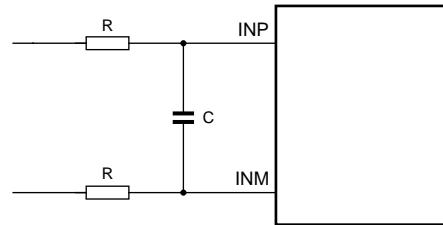
**Fig 5. Input sampling circuit**

The sample phase occurs when the internal clock (derived from the clock signal on pin CLKP/CLKM) is HIGH. The voltage is then held on the sampling capacitors. When the clock signal goes LOW, the stage enters the hold phase and the voltage information is transmitted to the ADC core.

#### 11.2.2 Anti-kickback circuitry

Anti-kickback circuitry (R-C filter in [Figure 6](#)) is needed to counteract the effects of charge injection generated by the sampling capacitance.

The RC filter is also used to filter noise from the signal before it reaches the sampling stage. The value of the capacitor should be chosen to maximize noise attenuation without degrading the settling time excessively.



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**Fig 6. Anti-kickback circuit**

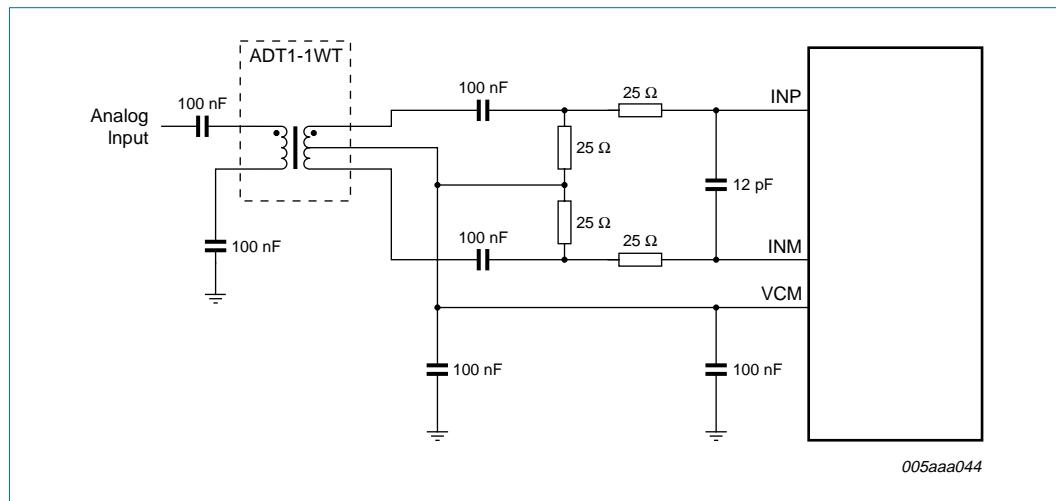
The component values are determined by the input frequency and should be selected so as not to affect the input bandwidth.

**Table 11. RC coupling versus input frequency - recommended values**

Input frequency	R	C
3 MHz	25 ohms	12 pF
70 MHz	12 ohms	8 pF
170 MHz	12 ohms	8 pF

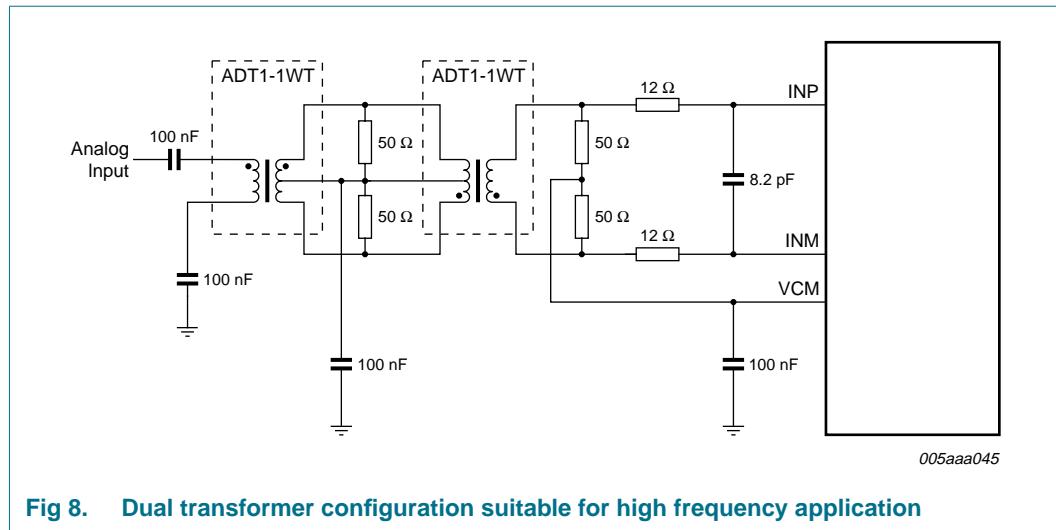
### 11.2.3 Transformer

The configuration of the transformer circuit is determined by the input frequency. The configuration shown in [Figure 7](#) would be suitable for a baseband application.



**Fig 7. Single transformer configuration suitable for baseband applications**

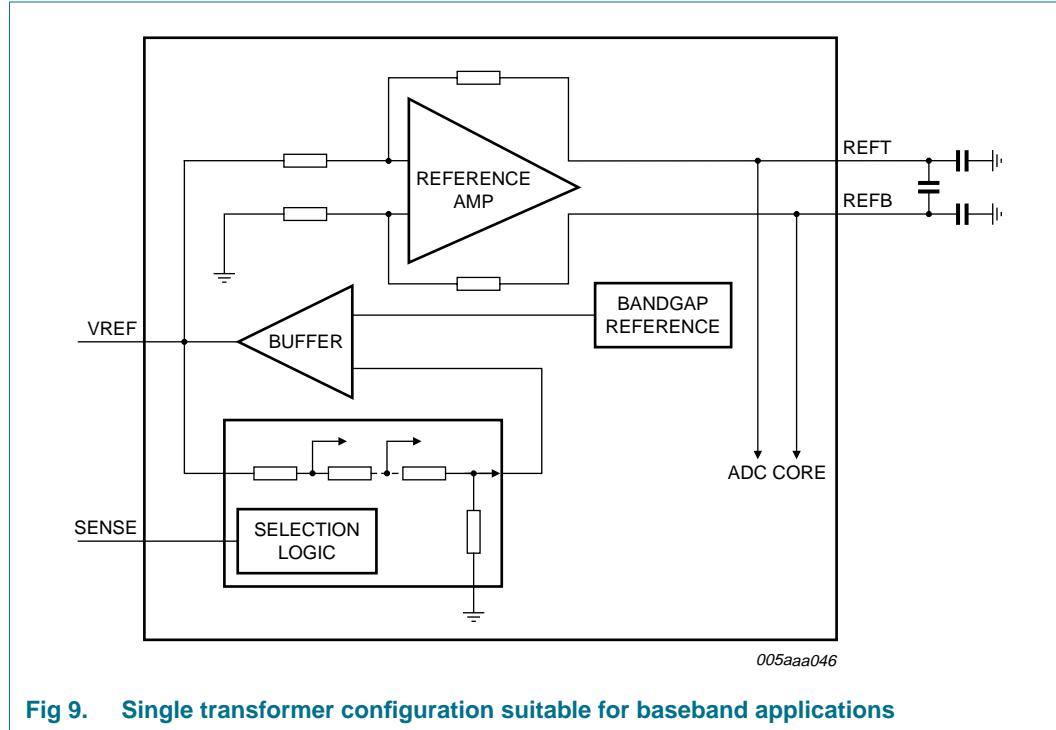
The configuration shown in [Figure 8](#) is recommended for high frequency applications. In both cases, the choice of transformer will be a compromise between cost and performance.



### 11.3 System reference and power management

#### 11.3.1 Internal/external references

The ADC1610S has a stable and accurate built-in internal reference voltage. This reference voltage can be set internally, externally or via the SPI (programmable in 1 dB steps between 0 dB and -6 dB via control bits INTREF when bit INTREF\_EN = 1; see [Table 20](#)). The equivalent reference circuit is shown in [Figure 9](#).



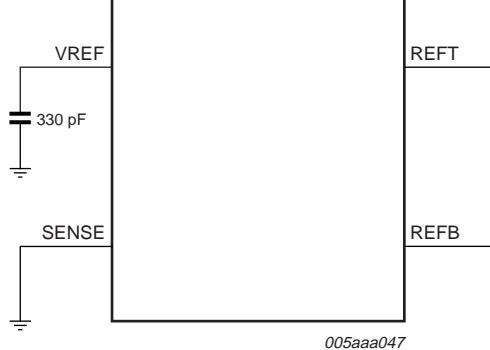
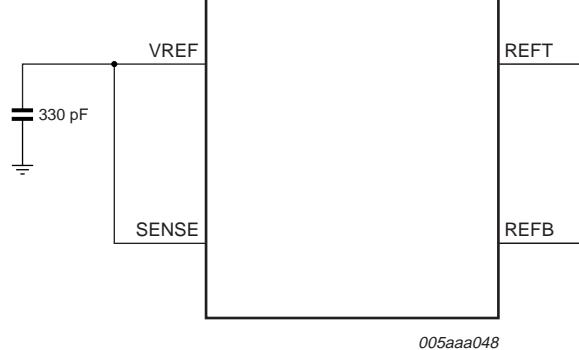
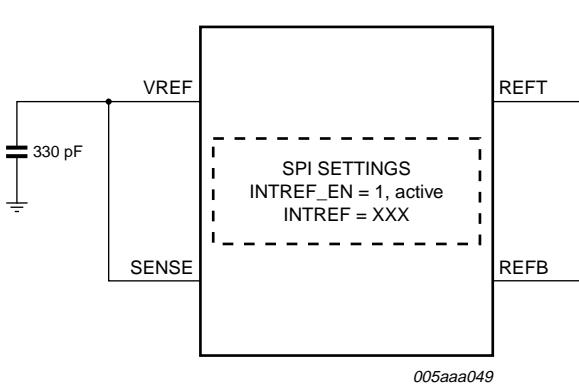
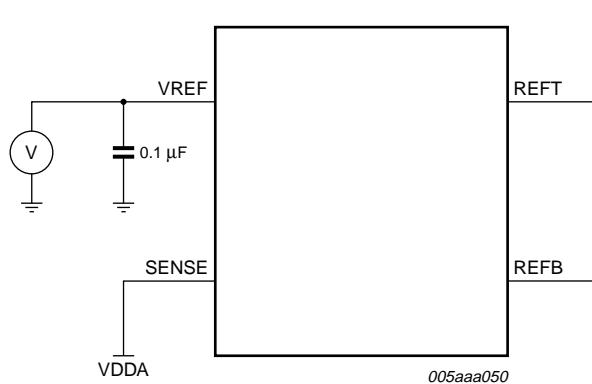
If bit INTREF\_EN is set to 0, the reference voltage will be determined either internally or externally as detailed in [Table 12](#).

**Table 12. Reference selection**

Selection	SPI bit INTREF_EN	SENSE pin	VREF pin	full scale (p-p)
internal	0	AGND	330 pF capacitor to AGND	2 V
internal	0		pin VREF connected to pin SENSE and via a 330 pF capacitor to AGND	1 V
external	0	$V_{DDA}$	external voltage between 0.5 V and 1 V <sup>[1]</sup>	1 V to 2 V
internal via SPI	1		pin VREF connected to pin SENSE and via 330 pF capacitor to AGND	1 V to 2 V

[1] The voltage on pin VREF is doubled internally to generate the internal reference voltage.

[Figure 10](#) to [Figure 13](#) illustrate how to connect the SENSE and VREF pins to select the required reference voltage source.

**Fig 10. Internal reference, 2 V (p-p) full scale****Fig 11. Internal reference, 1 V (p-p) full scale****Fig 12. Internal reference via SPI, 1 V to 2 V (p-p) full scale****Fig 13. External reference, 1 V to 2 V (p-p) full scale**

### 11.3.2 Gain control

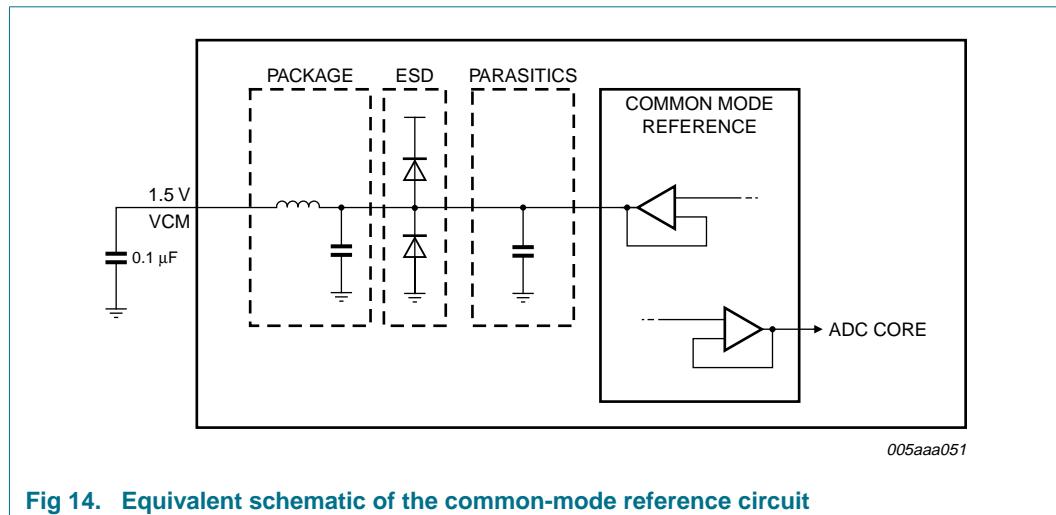
The gain is programmable between 0 dB to –6 dB in 1 dB steps via the SPI (see [Table 20](#)). This makes it possible to improve the Spurious-Free Dynamic Range (SFDR) of the ADC1610S. The corresponding full scale input voltage range varies between 2 V (p-p) and 1 V (p-p), as shown in [Table 13](#):

**Table 13. Reference SPI Gain Control**

INTREF	Gain	full scale (p-p)
000	0 dB	2 V
001	–1 dB	1.78 V
010	–2 dB	1.59 V
011	–3 dB	1.42 V
100	–4 dB	1.26 V
101	–5 dB	1.12 V
110	–6 dB	1 V
111	reserved	x

### 11.3.3 Common-mode output voltage ( $V_{O(cm)}$ )

A 0.1  $\mu$ F filter capacitor should be connected between pin VCM and ground to ensure a low-noise common-mode output voltage. When AC-coupled, pin VCM can then be used to set the common-mode reference for the analog inputs, for instance via a transformer middle point.

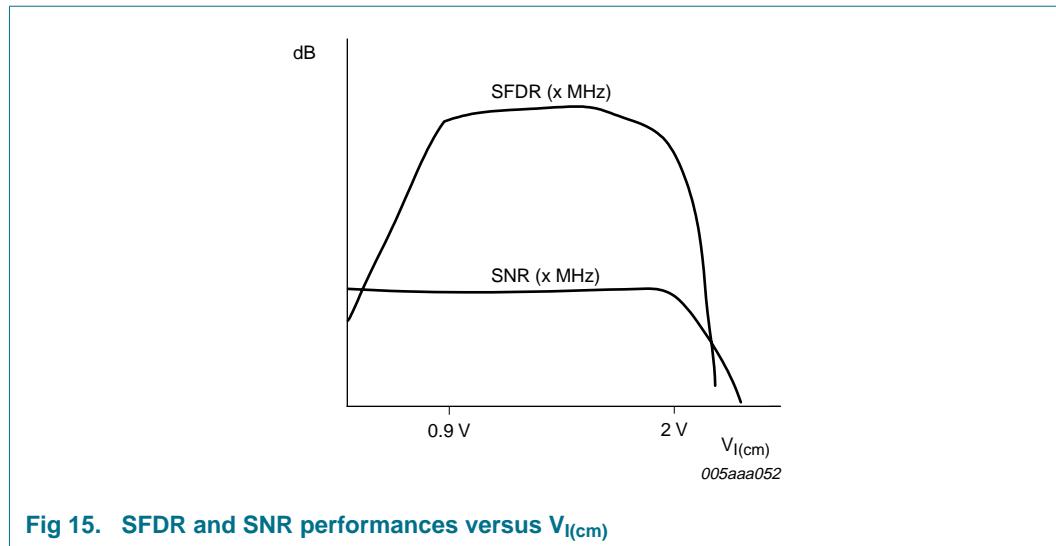


**Fig 14. Equivalent schematic of the common-mode reference circuit**

### 11.3.4 Biasing

The common-mode input voltage ( $V_{I(cm)}$ ) on pins INP and INM should be set externally to 0.5V<sub>DDA</sub> for optimal performance and should always be between 0.9 V and 2 V.

The graph in [Figure 15](#) illustrates how the SFDR and SNR characteristics vary with changes in the common-mode input voltage.

Fig 15. SFDR and SNR performances versus  $V_I(cm)$ 

## 11.4 Clock input

### 11.4.1 Drive modes

The ADC1610S can be driven differentially (SINE, LVPECL or LVDS) without the performance being affected by the choice of configuration. It can also be driven by a single-ended LVCMS signal connected to pin CLKP (CLKM should be connected to ground via a capacitor) or CLKM (CLKP should be connected to ground via a capacitor).

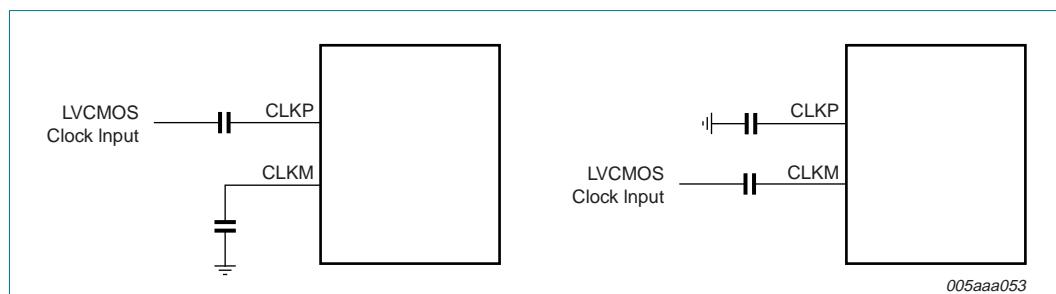


Fig 16. LVCMS Single-ended clock input

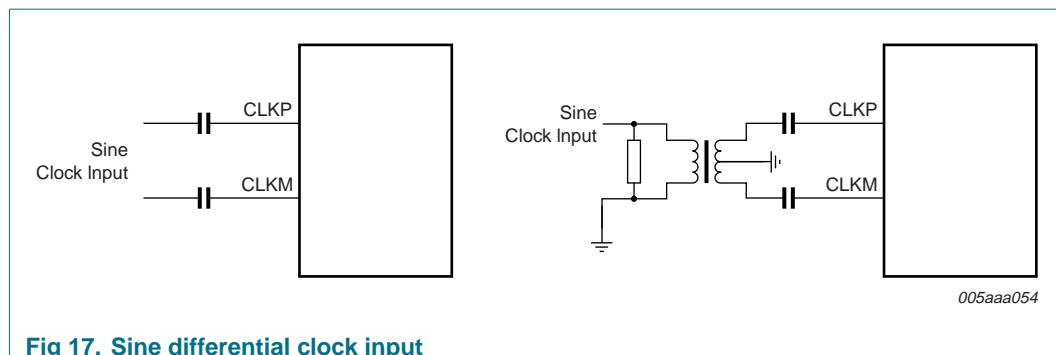
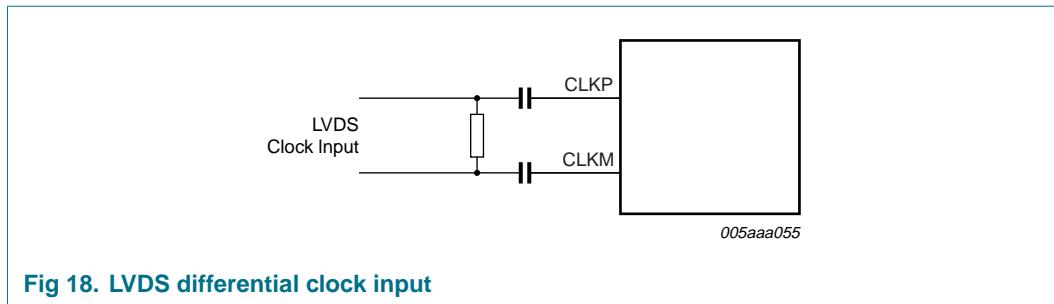
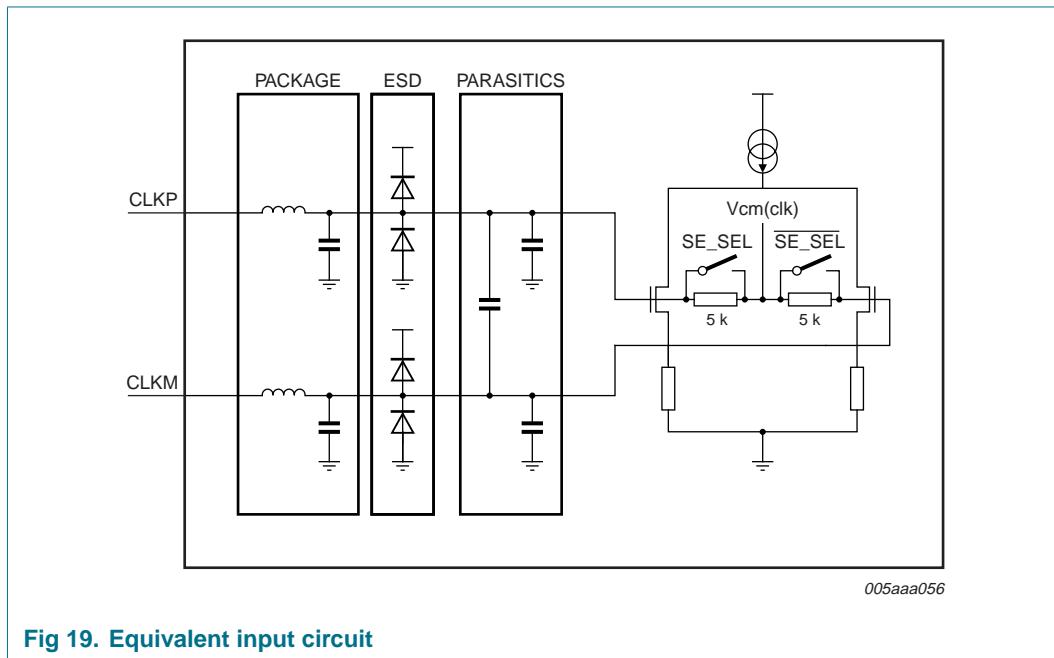


Fig 17. Sine differential clock input



#### 11.4.2 Equivalent input circuit

The equivalent circuit of the input clock buffer is shown in [Figure 19](#). The common-mode voltage of the differential input stage is set via internal  $5\text{ k}\Omega$  resistors.



Single-ended or differential clock inputs can be selected via the SPI interface (see [Table 19](#)). If single-ended is enabled, the input pin (CLKM or CLKP) is selected via control bit SE\_SEL.

If single-ended is implemented without setting SE\_SEL to the appropriate value, the unused pin should be connected to ground via a capacitor.

#### 11.4.3 Duty cycle stabilizer

The duty cycle stabilizer can improve the overall performances of the ADC by compensating the duty cycle of the input clock signal. When the duty cycle stabilizer is active (bit DCS\_EN = 1; see [Table 19](#)), the circuit can handle signals with duty cycles of between 30 % and 70 % (typical). When the duty cycle stabilizer is disabled (DCS\_EN = 0), the input clock signal should have a duty cycle of between 45 % and 55 %.

#### 11.4.4 Clock input divider

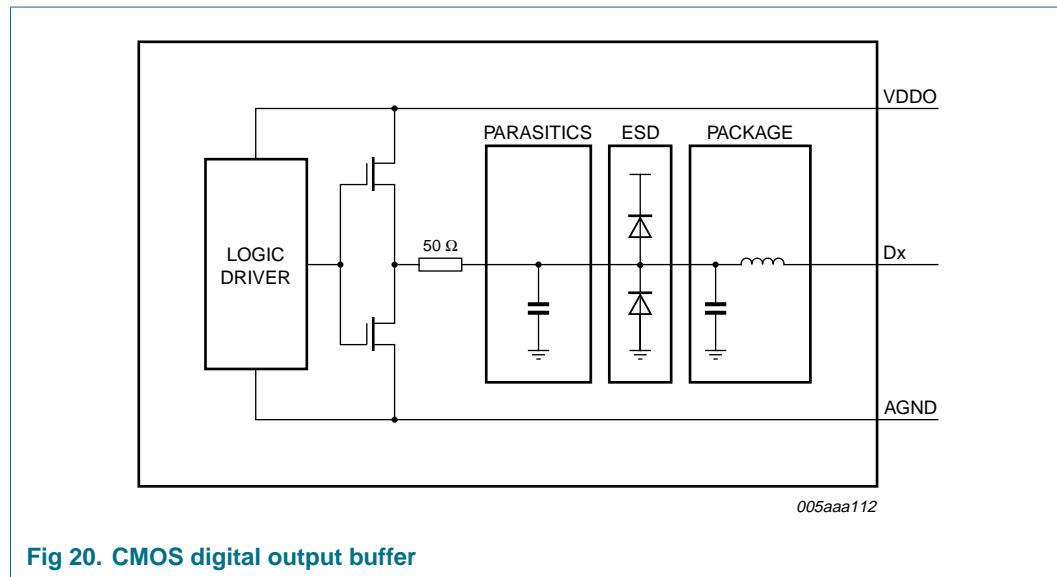
The ADC1610S contains an input clock divider that divides the incoming clock by a factor of 2 (when bit CLKDIV = 1; see [Table 19](#)). This feature allows the user to deliver a higher clock frequency with better jitter performance, leading to a better SNR result once acquisition has been performed.

### 11.5 Digital outputs

#### 11.5.1 Digital output buffers: CMOS mode

The digital output buffers can be configured as CMOS by setting bit LVDS/CMOS to 0 (see [Table 21](#)).

Each digital output has a dedicated output buffer. The equivalent circuit of the CMOS digital output buffer is shown in [Figure 20](#). The buffer is powered by a separate AGND/V<sub>DDO</sub> to ensure 1.8 V to 3.4 V compatibility and is isolated from the ADC core. Each buffer can be loaded by a maximum of 10 pF.

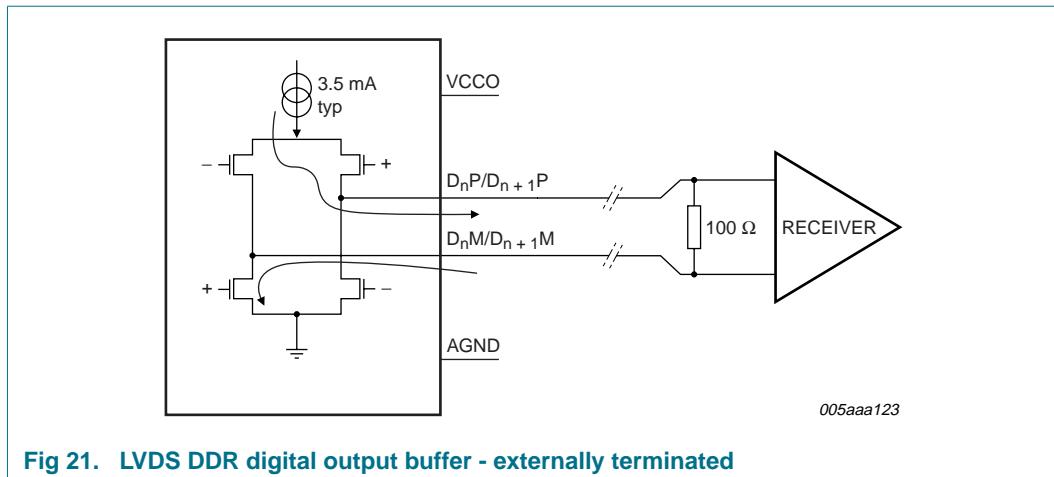


**Fig 20. CMOS digital output buffer**

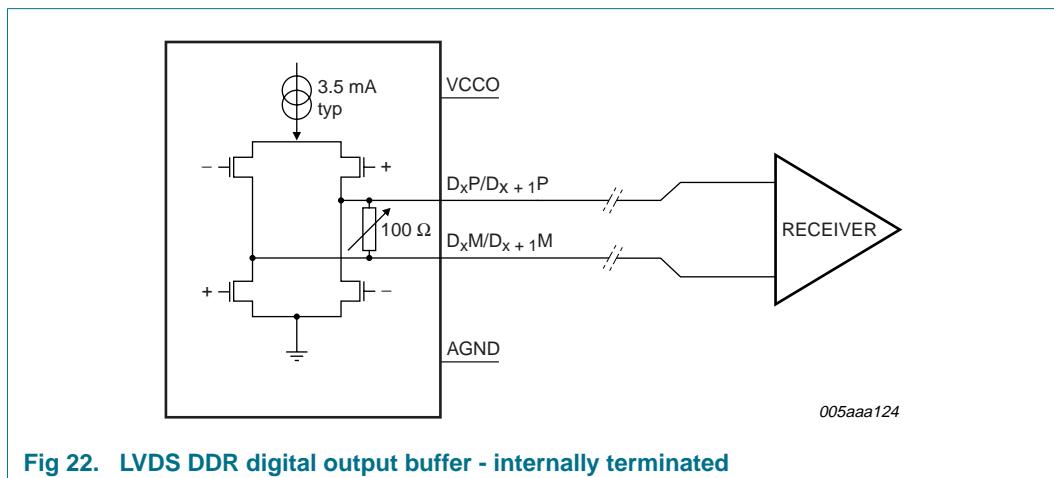
The output resistance is 50 Ω and is the combination of the an internal resistor and the equivalent output resistance of the buffer. There is no need for an external damping resistor. The drive strength of both data and DAV buffers can be programmed via the SPI in order to adjust the rise and fall times of the output digital signals (see [Table 28](#)):

#### 11.5.2 Digital output buffers: LVDS DDR mode

The digital output buffers can be configured as LVDS DDR by setting bit LVDS/CMOS to 1 (see [Table 21](#)).

**Fig 21.** LVDS DDR digital output buffer - externally terminated

Each output should be terminated externally with a  $100\ \Omega$  resistor (typical) at the receiver side ([Figure 21](#)) or internally via SPI control bits LVDS\_INTTER (see [Figure 22](#) and [Table 30](#)).

**Fig 22.** LVDS DDR digital output buffer - internally terminated

The default LVDS DDR output buffer current is set to 3.5 mA. It can be programmed via the SPI (bits DAVI and DATA1; see [Table 29](#)) in order to adjust the output logic voltage levels.

### 11.5.3 DData Valid (DAV) output clock

A DSV output clock signal is provided that can be used to capture the data delivered by the ADC1610S. Detailed timing diagrams for CMOS and LVDS DDR modes are provided in [Figure 23](#) and [Figure 24](#) respectively.

### 11.5.4 Out-of-Range (OTR)

An out-of-range signal is provided on pin OTR. By default, pin OTR goes HIGH fourteen clock cycles after an OTR event has occurred. The OTR response can be speeded up by enabling Fast OTR (bit FASTOTR = 1; see [Table 27](#)). When Fast OTR is enabled, OTR goes HIGH four clock cycles after the OTR event. The Fast OTR detection threshold (below full scale) can be programmed via bits FASTOTR\_DET.

### 11.5.5 Digital offset

By default, the ADC1610S delivers output code that corresponds to the analog input. However it is possible to add a digital offset to the output code via the SPI (bits DIG\_OFFSET; see [Table 23](#)).

### 11.5.6 Test patterns

For test purposes, the ADC1610S can be configured to transmit one of a number of predefined test patterns (via bits TESTPAT\_SEL; see [Table 24](#)). A custom test pattern can be defined by the user (TESTPAT\_USER; see [Table 25](#) and [Table 26](#)) and is selected when TESTPAT\_SEL = 101. The selected test pattern will be transmitted regardless of the analog input.

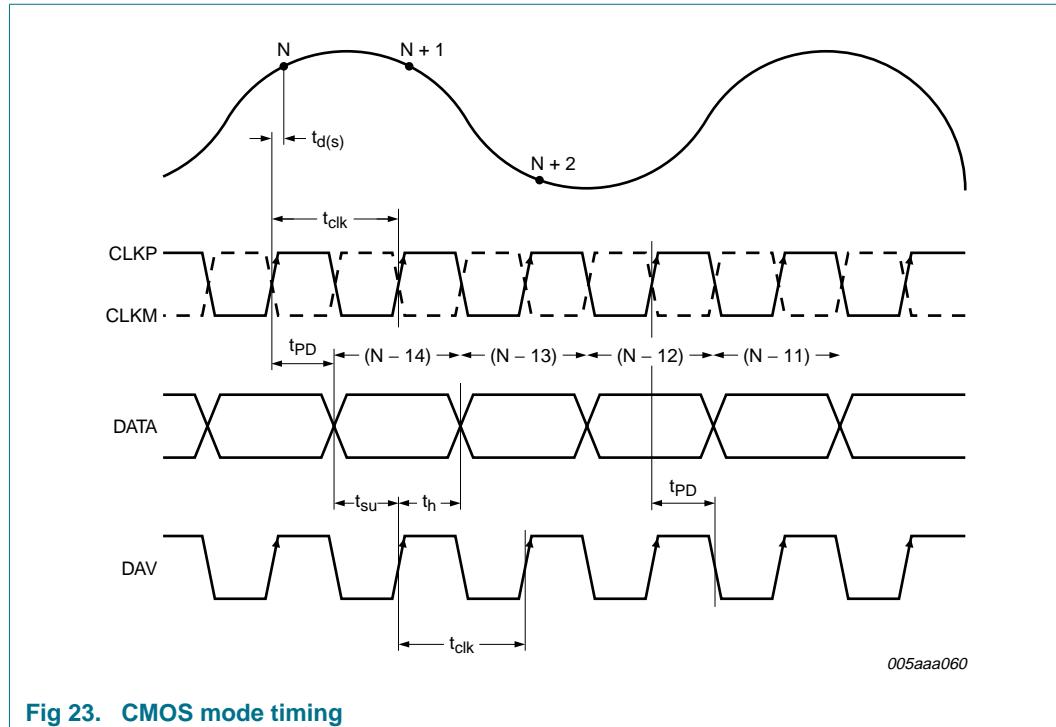
### 11.5.7 Output codes versus input voltage

**Table 14. Output codes**

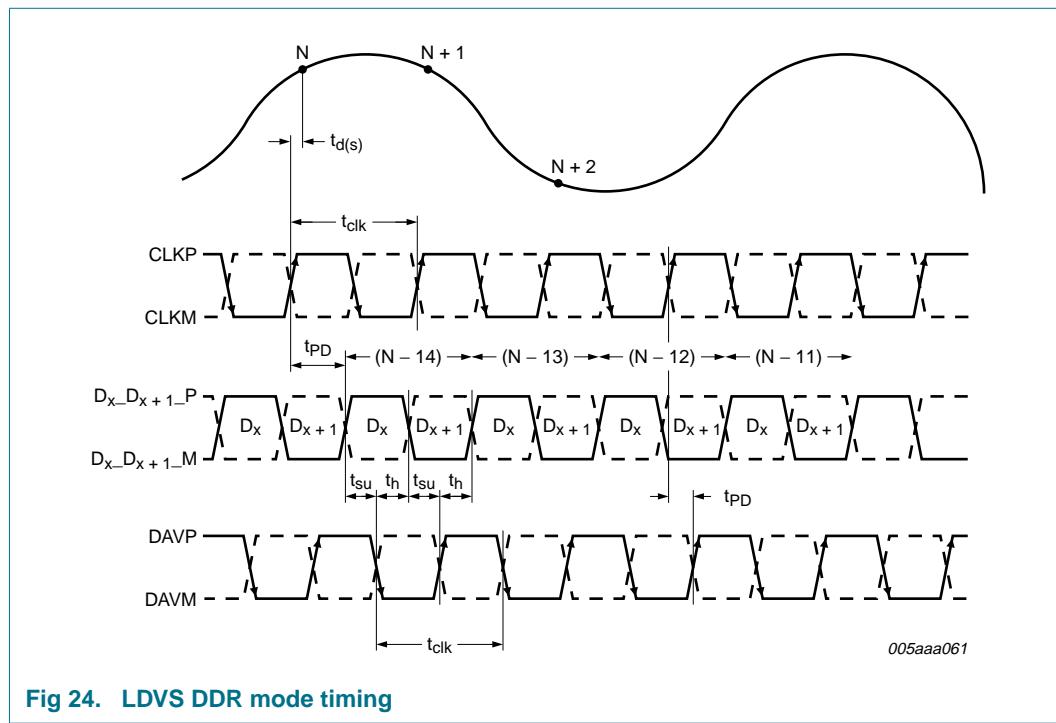
$V_{INP} - V_{INM}$	Offset binary	2's complement	OTR pin
< -1	0000 0000 0000 0000	1000 0000 0000 0000	1
-1	0000 0000 0000 0000	1000 0000 0000 0000	0
-0.99996948	0000 0000 0000 0001	1000 0000 0000 0001	0
-0.99993896	0000 0000 0000 0010	1000 0000 0000 0010	0
-0.99990845	0000 0000 0000 0011	1000 0000 0000 0011	0
-0.99987793	0000 0000 0000 0100	1000 0000 0000 0100	0
....	....	....	0
-0.00006104	01 1111 1111 1110	1111 1111 1111 1110	0
-0.00003052	01 1111 1111 1111	1111 1111 1111 1111	0
0	<b>10 0000 0000 0000</b>	<b>0000 0000 0000 0000</b>	0
+0.00003052	10 0000 0000 0001	0000 0000 0000 0001	0
+0.00006104	10 0000 0000 0010	0000 0000 0000 0010	0
....	....	....	0
+0.99987793	1111 1111 1111 1011	0111 1111 1111 1011	0
+0.99990845	1111 1111 1111 1100	0111 1111 1111 1100	0
+0.99993896	1111 1111 1111 1101	0111 1111 1111 1101	0
+0.99996948	1111 1111 1111 1110	0111 1111 1111 1110	0
+1	1111 1111 1111 1111	0111 1111 1111 1111	0
> +1	1111 1111 1111 1111	0111 1111 1111 1111	1

## 11.6 Timings summary

### 11.6.1 CMOS mode timings



### 11.6.2 LVDS DDR mode timing



## 11.7 Serial Peripheral Interface (SPI)

### 11.7.1 Register description

The ADC1610S serial interface is a synchronous serial communications port that allows for easy interfacing with many commonly-used microprocessors. It provides access to the registers that control the operation of the chip.

This interface is configured as a 3-wire type (SDIO as bidirectional pin)

Pin SCLK is the serial clock input and  $\overline{CS}$  is the chip select pin.

Each read/write operation is initiated by a LOW level on  $\overline{CS}$ . A minimum of three bytes will be transmitted (two instruction bytes and at least one data byte). The number of data bytes is determined by the value of bits W1 and W2 (see [Table 16](#)).

**Table 15. Instruction bytes for the SPI**

Bit	MSB								LSB
	7	6	5	4	3	2	1	0	
Description	R/W <sup>[1]</sup>	W1 <sup>[2]</sup>	W0 <sup>[2]</sup>	A12	A11	A10	A9	A8	
	A7	A6	A5	A4	A3	A2	A1	A0	

[1] Bit R/W indicates whether it is a read (1) or a write (0) operation.

[2] Bits W1 and W0 indicate the number of bytes to be transferred after the instruction byte (see [Table 16](#)).

**Table 16. Number of data bytes to be transferred after the instruction bytes**

W1	W0	Number of bytes transmitted
0	0	1 byte
0	1	2 bytes
1	0	3 bytes
1	1	4 bytes or more

Bits A12 to A0 indicate the address of the register being accessed. In the case of a multiple byte transfer, this address is the first register to be accessed. An address counter is incremented to access subsequent addresses.

The steps involved in a data transfer are as follows:

1. A falling edge on  $\overline{CS}$  in combination with a rising edge on SCLK determine the start of communications.
2. The first phase is the transfer of the 2-byte instruction.
3. The second phase is the transfer of the data which can vary in length but will always be a multiple of 8 bits. The MSB is always sent first (for instruction and data bytes).
4. A rising edge on  $\overline{CS}$  indicates the end of data transmission.

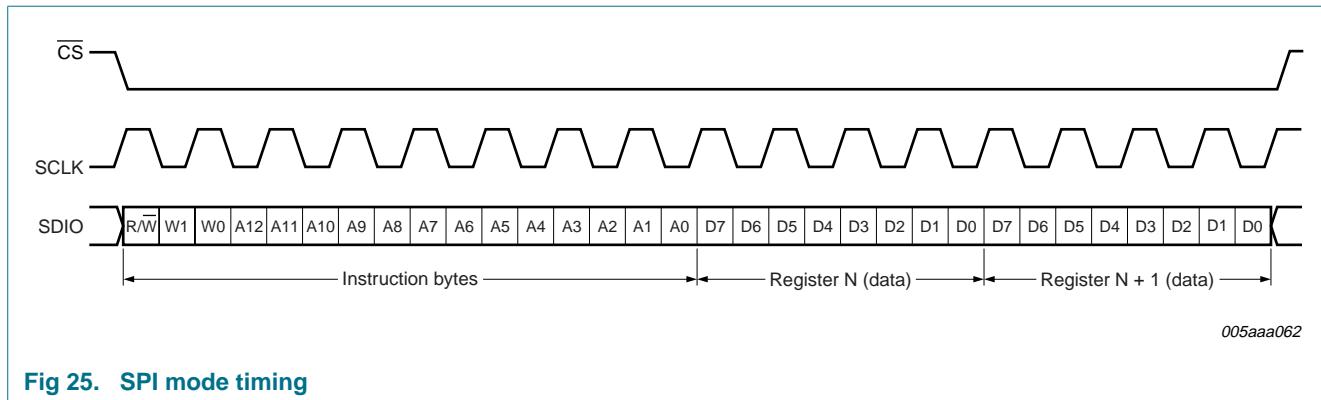


Fig 25. SPI mode timing

### 11.7.2 Default modes at start-up

During circuit initialization, it does not matter which output data standard has been selected. At power-up, the device defaults to PIN control mode.

A falling edge on **CS** will trigger a transition to SPI control mode. When the ADC1610S enters SPI control mode, the output data standard (CMOS/LVDS DDR) is determined by the level on pin SDIO (see [Figure 26](#)). Once in SPI control mode, the output data standard can be changed via bit LVDS/CMOS in [Table 21](#).

When the ADC1610S enters SPI control mode, the output data format (2's complement or offset binary) is determined by the level on pin SCLK (grey code can only be selected via the SPI). Once in SPI control mode, the output data format can be changed via bit **DATA\_FORMAT** in [Table 21](#).

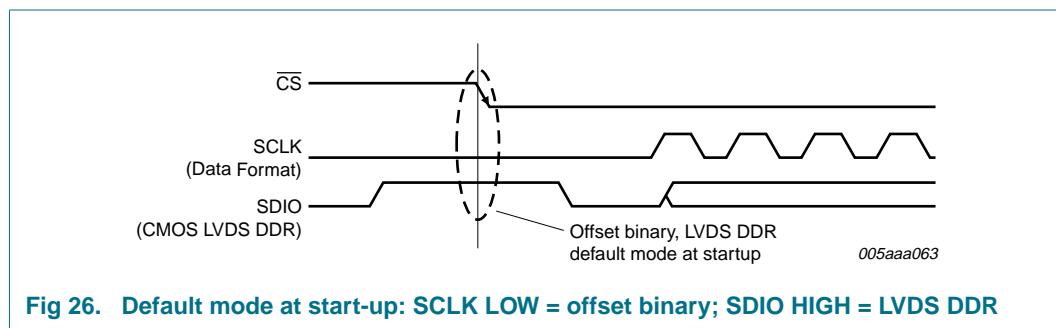


Fig 26. Default mode at start-up: SCLK LOW = offset binary; SDIO HIGH = LVDS DDR

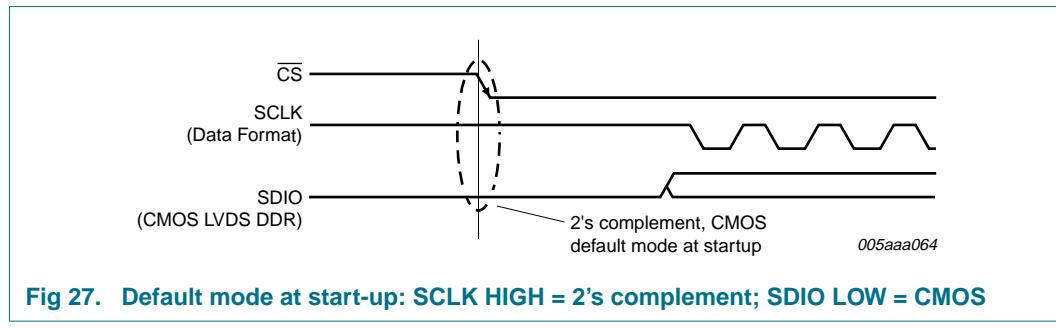


Fig 27. Default mode at start-up: SCLK HIGH = 2's complement; SDIO LOW = CMOS

### 11.7.3 Register allocation map

**Table 17. Register allocation map**

Addr Hex	Register name	R/W	Bit definition								Default Bin
			Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	
0005	Reset and operating mode	R/W	SW_RST	-	-	-	-	-	OP_MODE	0000 0000	
0006	Clock	R/W	-	-	-	SE_SEL	DIFF/SE	-	CLKDIV	DCS_EN	0000 0001
0008	Internal reference	R/W	-	-	-	-	INTREF_EN	INTREF		0000 0000	
0011	Output data standard.	R/W	-	-	-	LVDS/CMOS	OUTBUF	-	DATA_FORMAT	0000 0000	
0012	Output clock	R/W	-	-	-	-	DAVINV	DAVPHASE		0000 1110	
0013	Offset	R/W	-	-	DIG_OFFSET					0000 0000	
0014	Test pattern 1	R/W	-	-	-	-	-	TESTPAT_SEL		0000 0000	
0015	Test pattern 2	R/W	TESTPAT_USER							0000 0000	
0016	Test pattern 2	R/W	TESTPAT_USER							0000 0000	
0017	Fast OTR	R/W	-	-	-	-	FASTOTR	FASTOTR_DET		0000 0000	
0020	CMOS output	R/W	-	-	-	-	DAV_DRV	DATA_DRV		0000 1110	
0021	LVDS DDR O/P 1	R/W	-	-	DAVI_x2_EN	DAVI		DATAI_x 2_EN		0000 0000	
0022	LVDS DDR O/P 2	R/W	-	-	-	-	BIT/BYTE_WISE	LVDS_INTTER		0000 0000	

**Table 18. Reset and operating mode control register (address 0005h) bit description**

Bit	Symbol	Access	Value	Description
7	SW_RST	R/W		reset digital section
			0	no reset
			1	performs a reset of the digital section
6 to 2 reserved				
1 to 0	OP_MODE	R/W		operating mode
			00	Normal (Power-up)
			01	Power-down
			10	Sleep
			11	Normal (Power-up)

**Table 19. Clock control register (address 0006h) bit description**

Bit	Symbol	Access	Value	Description
7 to 5	reserved			
4	SE_SEL	R/W		single-ended clock input pin select
			0	CLKM
			1	CLKP
3	DIFF/SE	R/W		differential/single ended clock input select
			0	fully differential
			1	single-ended
2	reserved			
1	CLKDIV	R/W		clock input divide by 2
			0	disabled
			1	enabled
0	DCS_EN	R/W		duty cycle stabilizer
			0	disabled
			1	enabled

**Table 20. Internal reference control register (address 0008h) bit description**

Bit	Symbol	Access	Value	Description
7 to 4	reserved			
3	INTREF_EN	R/W		programmable internal reference enable
			0	disable
			1	active
2 to 0	INTREF	R/W		programmable internal reference
			000	0 dB (FS = 2 V)
			001	-1 dB (FS = 1.78 V)
			010	-2 dB (FS = 1.59 V)
			011	-3 dB (FS = 1.42 V)
			100	-4 dB (FS = 1.26 V)
			101	-5 dB (FS = 1.12 V)
			110	-6 dB (FS = 1 V)
			111	reserved

**Table 21. Output data standard control register (address 0011h) bit description**

Bit	Symbol	Access	Value	Description
7 to 5	reserved			
4	LVDS/CMOS	R/W		output data standard: LVDS DDR or CMOS
			0	CMOS
			1	LVDS DDR
3	OUTBUF	R/W		output buffers enable
			0	output enabled
			1	output disabled (high Z)
2	reserved			

**Table 21.** Output data standard control register (address 0011h) bit description ...continued

Bit	Symbol	Access	Value	Description
1 to 0	DATA_FORMAT	R/W		output data format
			00	offset binary
			01	2's complement
			10	gray code
			11	offset binary

**Table 22.** Output clock register (address 0012h) bit description

Bit	Symbol	Access	Value	Description
7 to 4	reserved			
3	DAVINV	R/W		output clock data valid (DAV) polarity
			0	normal
			1	inverted
2 to 0	DAVPHASE	R/W		DAV phase select
			000	output clock shifted (ahead) by 3 ns
			001	output clock shifted (ahead) by 2.5 ns
			010	output clock shifted (ahead) by 2 ns
			011	output clock shifted (ahead) by 1.5 ns
			100	output clock shifted (ahead) by 1 ns
			101	output clock shifted (ahead) by 0.5 ns
			110	default value as defined in timing section
			111	output clock shifted (delayed) by 0.5 ns

**Table 23.** Offset register (address 0013h) bit description

Bit	Symbol	Access	Value	Description
7 to 6	reset			
5 to 0	DIG_OFFSET	R/W		digital offset adjustment
			011111	+31 LSB
			...	...
			000000	0
			...	...
			100000	-32 LSB

**Table 24.** Test pattern register 1 (address 0014h) bit description

Bit	Symbol	Access	Value	Description
7 to 3	reserved			
2 to 0	TESTPAT_SEL	R/W		digital test pattern select
			000	off
			001	mid scale
			010	-FS
			011	+FS
			100	toggle '1111..1111'/'0000..0000'
			101	custom test pattern
			110	'1010..1010.'
			111	'010..1010'

**Table 25.** Test pattern register 2 (address 0015h) bit description

Bit	Symbol	Access	Value	Description
7 to 0	TESTPAT_USER	R/W		custom digital test pattern (bits 15 to 8)

**Table 26.** Test pattern register 3 (address 0016h) bit description

Bit	Symbol	Access	Value	Description
7 to 0	TESTPAT_USER	R/W		custom digital test pattern (bits 7 to 0)

**Table 27.** Fast OTR register (address 0017h) bit description

Bit	Symbol	Access	Value	Description
7 to 4	reset			
3	FASTOTR	R/W		fast Out-of-Range (OTR) detection
			0	disabled
			1	enabled
2 to 0	FASTOTR_DET	R/W		set fast OTR detect level
			000	-20.56 dB
			001	-16.12 dB
			010	-11.02 dB
			011	-7.82 dB
			100	-5.49 dB
			101	-3.66 dB
			110	-2.14 dB
			111	-0.86 dB

**Table 28. CMOS output register (address 0020h) bit description**

Bit	Symbol	Access	Value	Description
7 to 4	reserved			
3 to 2	DAV_DRV	R/W		drive strength for DAV CMOS output buffer
			00	low
			01	medium
			10	high
			11	very high
1 to 0	DATA_DRV	R/W		drive strength for DATA CMOS output buffer
			00	low
			01	medium
			10	high
			11	very high

**Table 29. LVDS DDR output register 1 (address 0021h) bit description**

Bit	Symbol	Access	Value	Description
7 to 6				
5	DAVI_x2_EN	R/W		double LVDS current for DAV LVDS buffer
			0	disabled
			1	enabled
4 to 3	DAVI	R/W		LVDS current for DAV LVDS buffer
			00	3.5 mA
			01	4.5 mA
			10	1.25 mA
			11	2.5 mA
2	DATAI_x2_EN	R/W		double LVDS current for DATA LVDS buffer
			0	disabled
			1	enabled
1 to 0	DATAI	R/W		LVDS current for DATA LVDS buffer
			00	3.5 mA
			01	4.5 mA
			10	1.25 mA
			11	2.5 mA

**Table 30. LVDS DDR output register 2 (address 0022h) bit description**

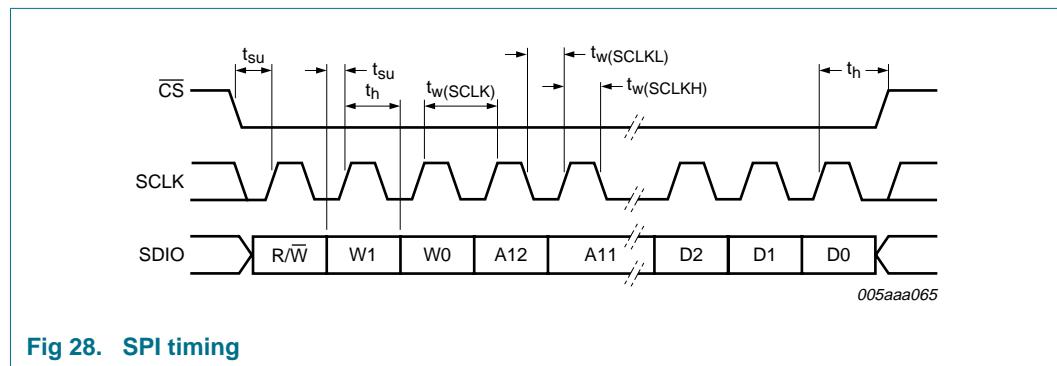
Bit	Symbol	Access	Value	Description
7 to 4	reserved			
3	BIT/BYTE_WISE	R/W		DDR mode for LVDS output
			0	bit wise (even data bits output on DAV rising edge / odd data bits output on DAV falling edge)
			1	byte wise (MSB data bits output on DAV rising edge / LSB data bits output on DAV falling edge)

**Table 30. LVDS DDR output register 2 (address 0022h) bit description ...continued**

Bit	Symbol	Access	Value	Description
2 to 0	LVDS_INTTER	R/W		internal termination for LVDS buffer (DAV and DATA)
			000	no internal termination
			001	300 Ω
			010	180 Ω
			011	110 Ω
			100	150 Ω
			101	100 Ω
			110	81 Ω
			111	60 Ω

#### 11.7.4 Serial timing interface

SPI timing is shown in [Figure 28](#).

**Fig 28. SPI timing**

SPI timing characteristics are detailed in [Table 9](#).

## 12. Package outline

HVQFN40: plastic thermal enhanced very thin quad flat package; no leads;  
40 terminals; body 6 x 6 x 0.85 mm

SOT618-1

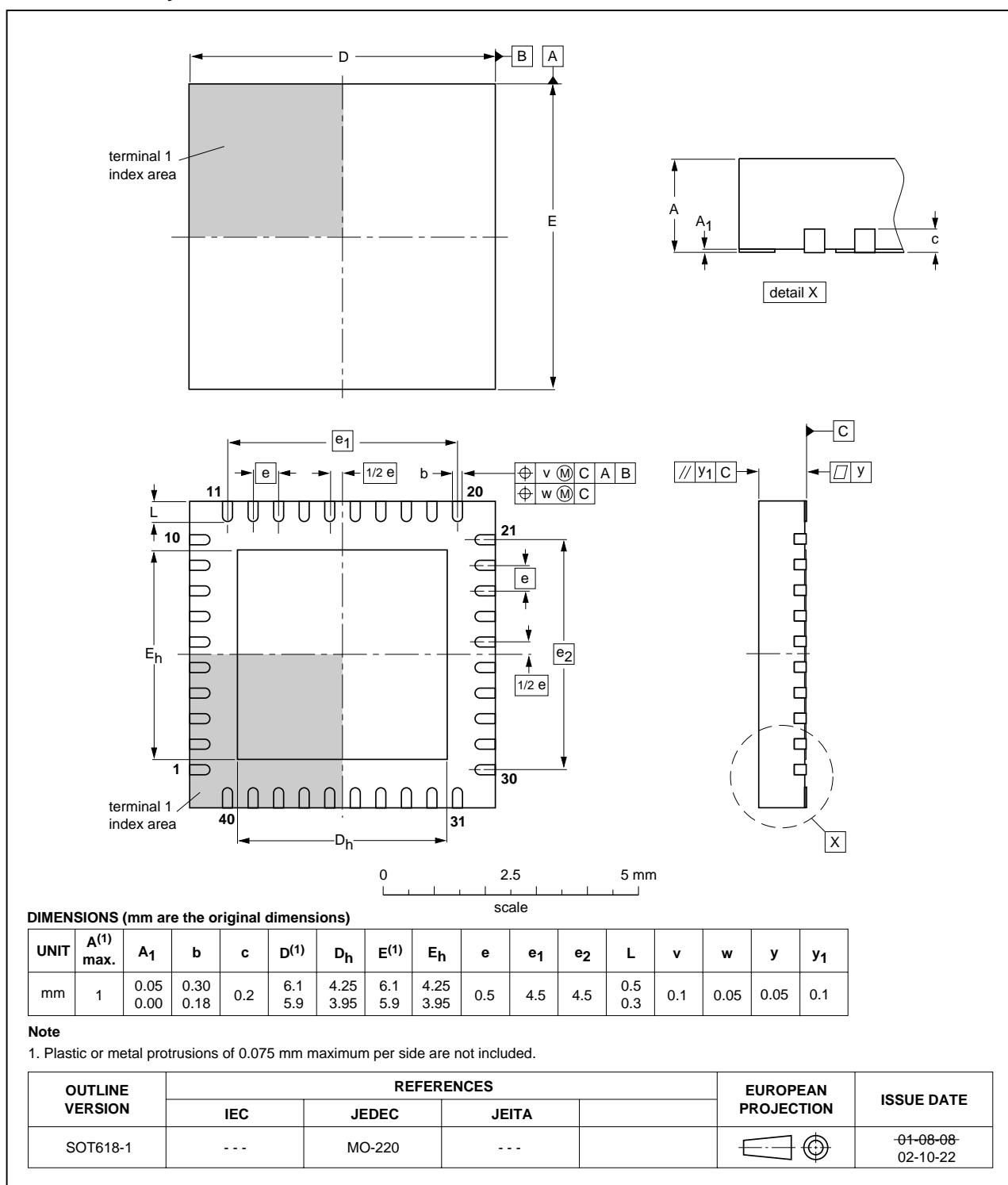


Fig 29. Package outline SOT618-1 (HVQFN40)

## 13. Revision history

**Table 31. Revision history**

Document ID	Release date	Data sheet status	Change notice	Supersedes
ADC1610S125_1	20090528	Objective data sheet	-	-

## 14. Legal information

### 14.1 Data sheet status

Document status <sup>[1][2]</sup>	Product status <sup>[3]</sup>	Definition
Objective [short] data sheet	Development	This document contains data from the objective specification for product development.
Preliminary [short] data sheet	Qualification	This document contains data from the preliminary specification.
Product [short] data sheet	Production	This document contains the product specification.

[1] Please consult the most recently issued document before initiating or completing a design.

[2] The term 'short data sheet' is explained in section "Definitions".

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