

FPF2163/4/5

Full Function Load Switch with Adjustable Current Limit

Features

- 1.8 to 5.5V Input Voltage Range
- Controlled Turn-On
- 0.15-1.5A Adjustable Current Limit
- Undervoltage Lockout
- Thermal Shutdown
- $2\mu\text{A}$ Shutdown Current
- Auto Restart
- Fast Current limit Response Time
 - 5 μs to Moderate Over Currents
 - 30ns to Hard Shorts
- Fault Blanking
- Reverse Current Blocking
- RoHS Compliant

Applications

- PDAs
- Cell Phones
- GPS Devices
- MP3 Players
- Digital Cameras
- Peripheral Ports
- Hot Swap Supplies

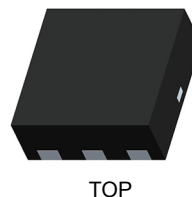
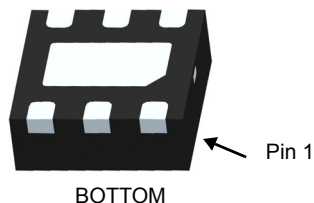


General Description

The FPF2163, FPF2164, and FPF2165 is a series of load switches which provides full protection to systems and loads which may encounter large current conditions. These devices contain a 0.12 Ω current-limited P-channel MOSFET which can operate over an input voltage range of 1.8-5.5V. Internally, current is prevented from flowing when the MOSFET is off and the output voltage is higher than the input voltage. Switch control is by a logic input (ON) capable of interfacing directly with low voltage control signals. Each part contains thermal shutdown protection which shuts off the switch to prevent damage to the part when a continuous over-current condition causes excessive heating.

When the switch current reaches the current limit, the parts operate in a constant-current mode to prohibit excessive currents from causing damage. For the FPF2163 and FPF2164, if the constant current condition still persists after 30ms, these parts will shut off the switch and pull the fault signal pin (FLAGB) low. The FPF2163 has an auto-restart feature which will turn the switch on again after 450ms if the ON pin is still active. The FPF2164 does not have this auto-restart feature so the switch will remain off until the ON pin is cycled. The FPF2165 will not turn off after a current limit fault, but will rather remain in the constant current mode indefinitely. The minimum current limit is 150mA.

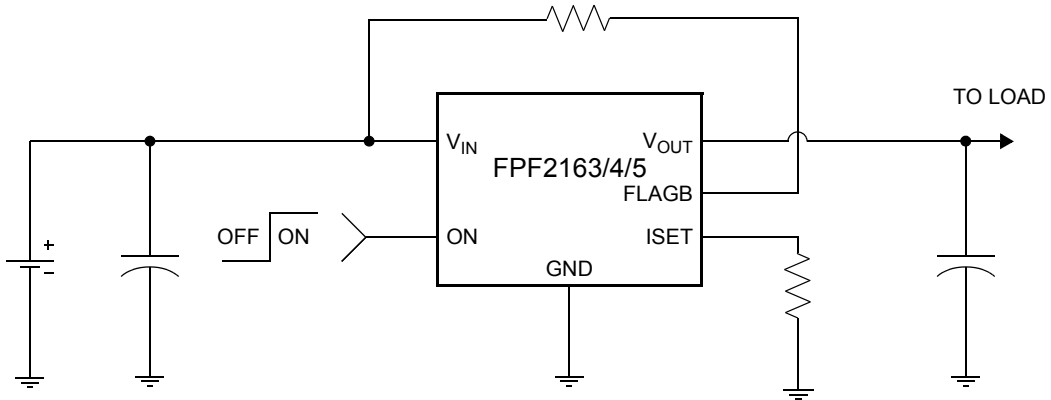
These parts are available in a space-saving 6 pin 2X2 MLP package.



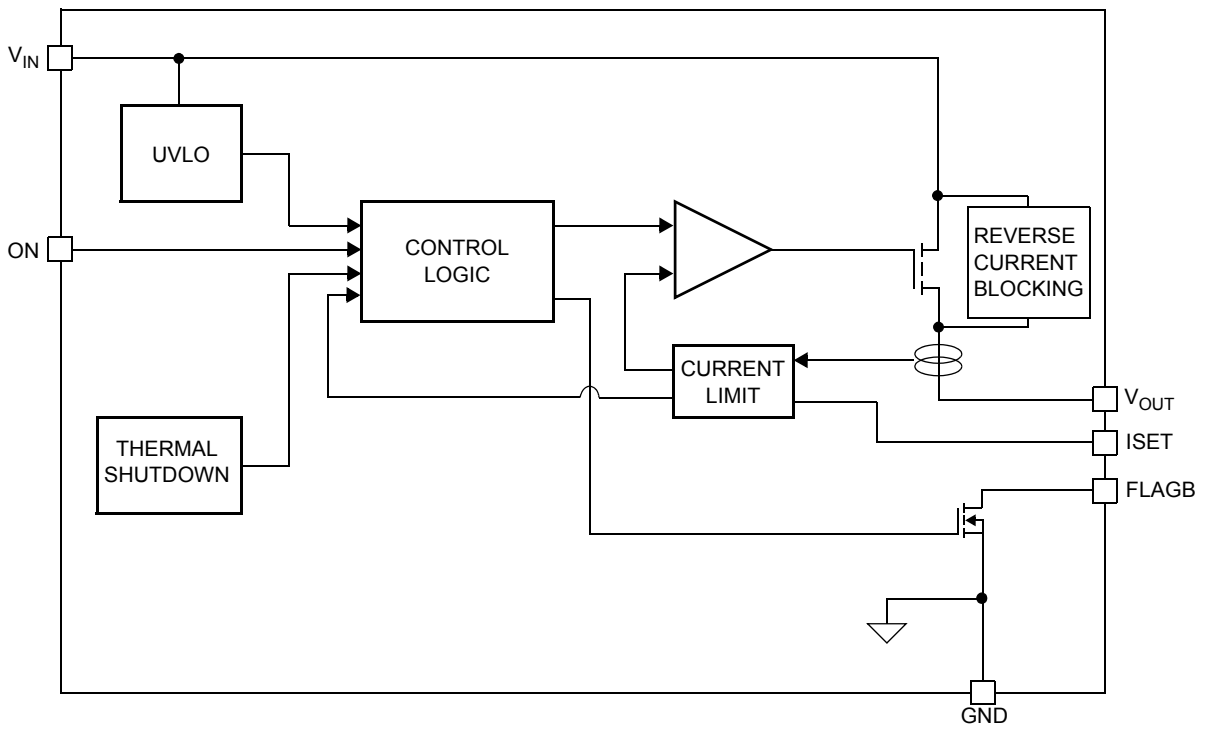
Ordering Information

| Part | Current Limit [mA] | Current Limit Blanking Time [ms] | Auto-Restart Time [ms] | ON Pin Activity | Top Mark |
|---------|--------------------|----------------------------------|------------------------|-----------------|----------|
| FPF2163 | 150-1500 | 15/30/60 | 225/450/900 | Active HI | 163 |
| FPF2164 | 150-1500 | 15/30/60 | NA | Active HI | 164 |
| FPF2165 | 150-1500 | 0 | NA | Active HI | 165 |

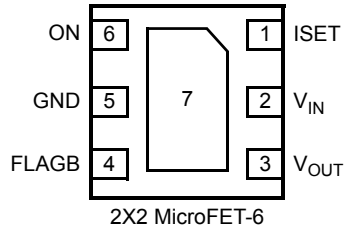
Typical Application Circuit



Functional Block Diagram



Pin Configuration



Pin Description

| Pin | Name | Function |
|------|------------------|--|
| 1 | ISET | Current Limit Set Input: A resistor from ISET to ground sets the current limit for the switch. |
| 2 | V _{IN} | Supply Input: Input to the power switch and the supply voltage For the IC |
| 3 | V _{OUT} | Switch Output: Output of the power switch |
| 4 | FLAGB | Fault Output: Active LO, open drain output which indicates an over current supply under voltage or over temperature state. |
| 5, 7 | GND | Ground |
| 6 | ON | ON Control Input |

Absolute Maximum Ratings

| Parameter | Min | Max | Unit | |
|---|---------------|-------------------|-------|---|
| V _{IN} , V _{OUT} , ON, FLAGB, ISET to GND | -0.3 | 6 | V | |
| Power Dissipation | | 1.2 | W | |
| Operating and Storage Junction Temperature | -65 | 150 | °C | |
| Thermal Resistance, Junction to Ambient | | 86 | °C/W | |
| Electrostatic Discharge Protection | Jedec A114A | HBM | 4000 | V |
| | Jedec C101C | CDM | 2000 | V |
| | Jedec A115 | MM | 400 | V |
| | IEC 61000-4-2 | Air Discharge | 15000 | V |
| | | Contact Discharge | 8000 | V |

Recommended Operating Range

| Parameter | Min | Max | Unit |
|---|-----|-----|------|
| V _{IN} | 1.8 | 5.5 | V |
| Ambient Operating Temperature, T _A | -40 | 85 | °C |

Electrical Characteristics

V_{IN} = 1.8 to 5.5V, T_A = -40 to +85°C unless otherwise noted. Typical values are at V_{IN} = 3.3V and T_A = 25°C.

| Parameter | Symbol | Conditions | Min | Typ | Max | Units | |
|------------------------|-----------------|------------------------|------------------------|-----|-----|-------|----|
| Basic Operation | | | | | | | |
| Operating Voltage | V _{IN} | | 1.8 | | 5.5 | V | |
| Quiescent Current | I _Q | I _{OUT} = 0mA | V _{IN} = 1.8V | | 63 | 100 | μA |
| | | | V _{IN} = 3.3V | | 68 | | |
| | | | V _{IN} = 5.5V | | 77 | 120 | |

Electrical Characteristics Cont.
 $V_{IN} = 1.8$ to $5.5V$, $T_A = -40$ to $+85^\circ C$ unless otherwise noted. Typical values are at $V_{IN} = 3.3V$ and $T_A = 25^\circ C$.

| Parameter | Symbol | Conditions | Min | Typ | Max | Units |
|-----------------------------------|-----------------|--|------|------|------|------------|
| On-Resistance | R_{ON} | $V_{IN} = 3.3V, I_{OUT} = 200mA, T_A = 25^\circ C$ | | 120 | 160 | m Ω |
| | | $V_{IN} = 3.3V, I_{OUT} = 200mA, T_A = 85^\circ C$ | | 135 | 180 | |
| | | $V_{IN} = 3.3V, I_{OUT} = 200mA, T_A = -40^\circ C$ to $+85^\circ C$ | 65 | | 180 | |
| ON Input Logic High Voltage (ON) | V_{IH} | $V_{IN} = 1.8V$ | 0.8 | | | V |
| | | $V_{IN} = 5.5V$ | 1.4 | | | |
| ON Input Logic Low Voltage | V_{IL} | $V_{IN} = 1.8V$ | | | 0.5 | V |
| | | $V_{IN} = 5.5V$ | | | 1 | |
| ON Input Leakage | | $V_{ON} = V_{IN}$ or GND | -1 | | 1 | μA |
| V_{IN} Shutdown Current | | $V_{ON} = 0V, V_{IN} = 5.5V,$ $V_{OUT} =$ short to GND | -2 | | 2 | μA |
| FLAGB Output Logic Low Voltage | | $V_{IN} = 5V, I_{SINK} = 10mA$ | | 0.05 | 0.2 | V |
| | | $V_{IN} = 1.8V, I_{SINK} = 10mA$ | | 0.12 | 0.3 | |
| FLAGB Output High Leakage Current | | $V_{IN} = 5V$, Switch on | | | 1 | μA |
| Reverse Block | | | | | | |
| V_{OUT} Shutdown Current | | $V_{ON} = 0V, V_{OUT} = 5.5V,$ $V_{IN} =$ short to GND | -2 | | 2 | μA |
| Protections | | | | | | |
| Current Limit | I_{LIM} | $V_{IN} = 3.3V, V_{OUT} = 3.0V, R_{SET} = 345\Omega$ | 600 | 800 | 1000 | mA |
| Min. Current Limit | $I_{LIM(min.)}$ | $V_{IN} = 3.3V, V_{OUT} = 3.0V$ | | 150 | | mA |
| Thermal Shutdown | | Shutdown Threshold | | 140 | | $^\circ C$ |
| | | Return from Shutdown | | 130 | | |
| | | Hysteresis | | 10 | | |
| Under Voltage Shutdown | UVLO | V_{IN} Increasing | 1.55 | 1.65 | 1.75 | V |
| Under Voltage Shutdown Hysteresis | | | | 50 | | mV |
| Dynamic | | | | | | |
| Delay On Time | t_{dON} | $R_L = 500\Omega, C_L = 0.1\mu F$ | | 25 | | μs |
| Delay Off Time | t_{dOFF} | $R_L = 500\Omega, C_L = 0.1\mu F$ | | 45 | | μs |
| V_{OUT} Rise Time | t_{RISE} | $R_L = 500\Omega, C_L = 0.1\mu F$ | | 10 | | μs |
| V_{OUT} Fall Time | t_{FALL} | $R_L = 500\Omega, C_L = 0.1\mu F$ | | 110 | | μs |
| Over Current Blanking Time | t_{BLANK} | FPF2163, FPF2164 | 15 | 30 | 60 | ms |
| Auto-Restart Time | t_{RSTRT} | FPF2163 | 225 | 450 | 900 | ms |
| Short Circuit Response Time | | $V_{IN} = V_{OUT} = 3.3V$. Moderate Over-Current Condition. | | 5 | | μs |
| | | $V_{IN} = V_{OUT} = 3.3V$. Hard Short. | | 30 | | ns |

Note 1: Package power dissipation on 1square inch pad, 2 oz copper board.

Typical Characteristics

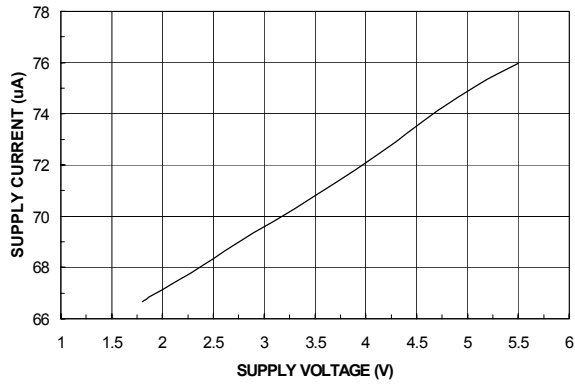


Figure 1. Quiescent Current vs. Input Voltage

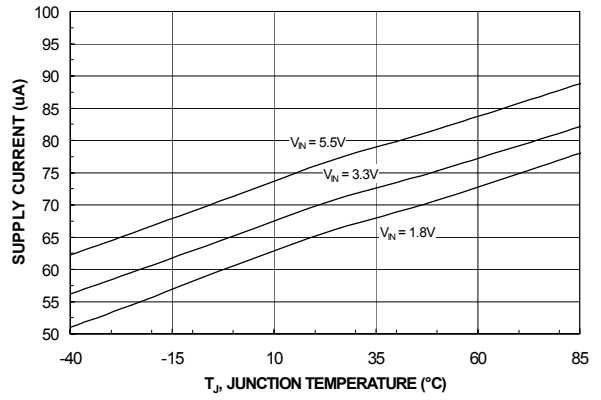


Figure 2. Quiescent Current vs. Temperature

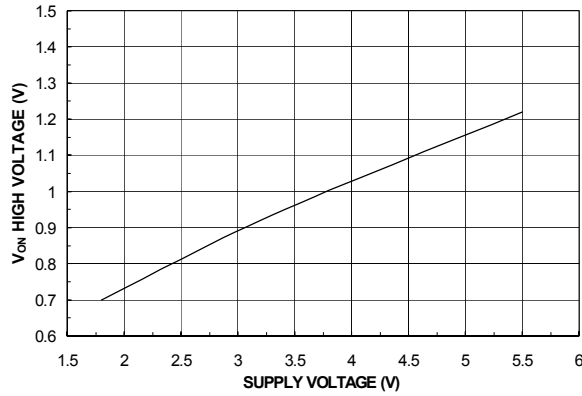


Figure 3. V_{ON} High Voltage vs. Input Voltage

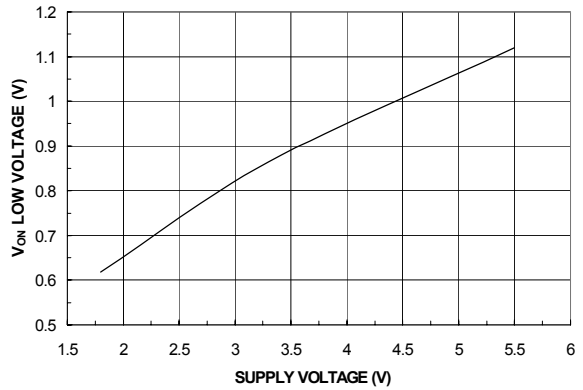


Figure 4. V_{ON} Low Voltage vs. Input Voltage

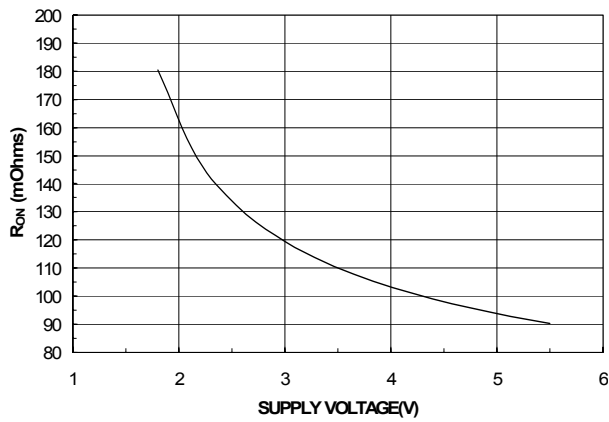


Figure 5. R_{ON} vs. V_{IN}

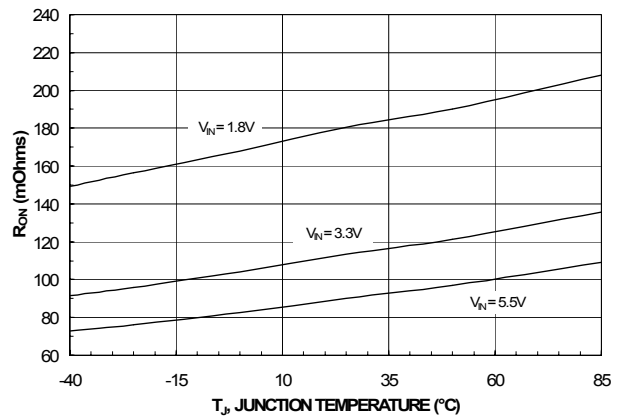


Figure 6. R_{ON} vs. Temperature

Typical Characteristics

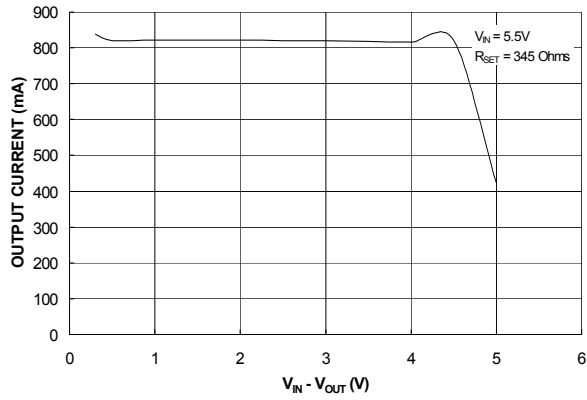


Figure 7. Current Limit vs. Output Voltage

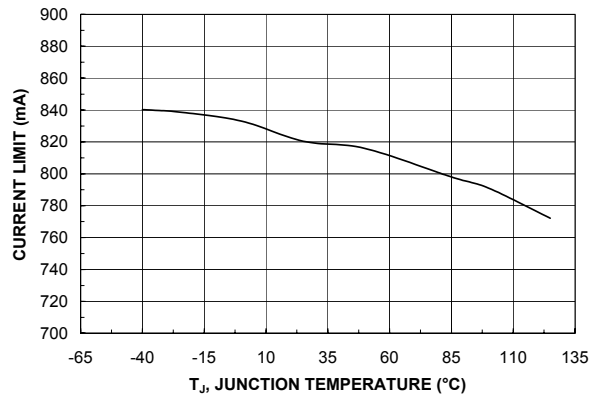


Figure 8. Current Limit vs. Temperature

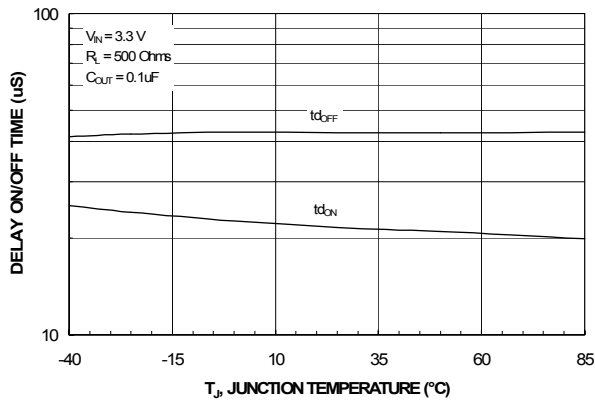


Figure 9. t_{dON} / t_{dOFF} vs. Temperature

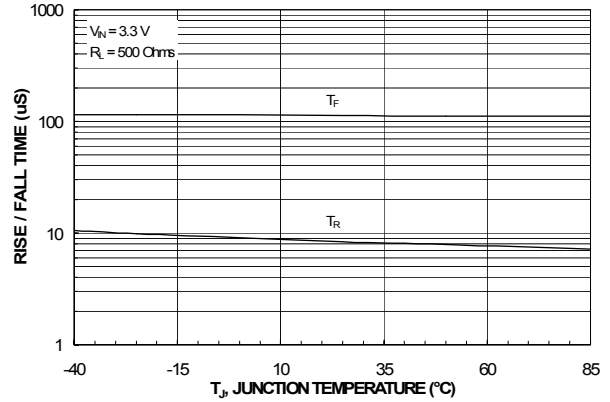


Figure 10. T_{RISE} / T_{FALL} vs. Temperature

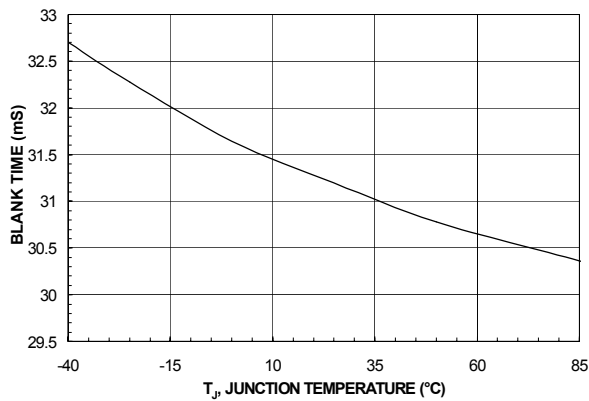


Figure 11. T_{BLANK} vs. Temperature

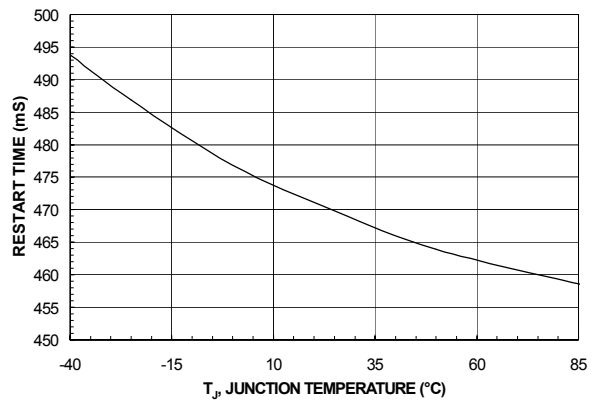


Figure 12. $T_{RESTART}$ vs. Temperature

Typical Characteristics

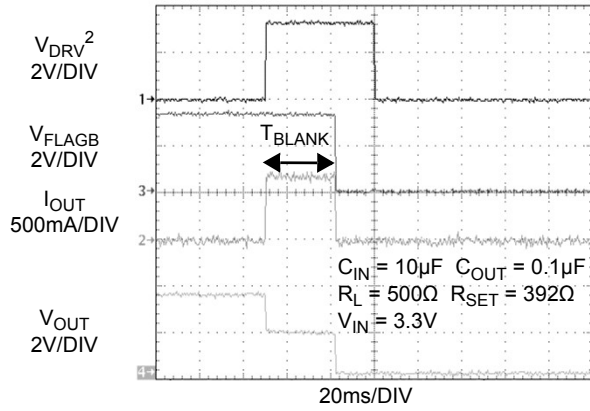


Figure 13. T_{BLANK} Response

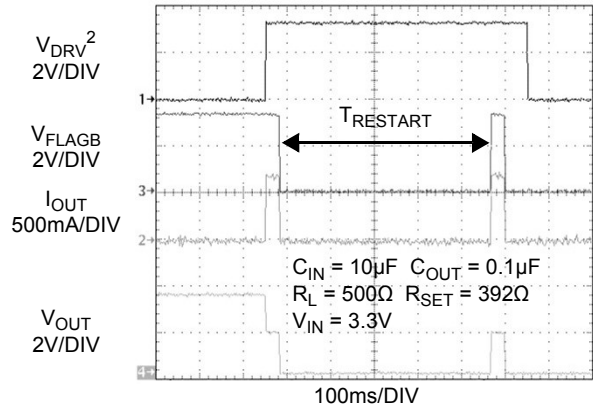


Figure 14. $T_{RESTART}$ Response

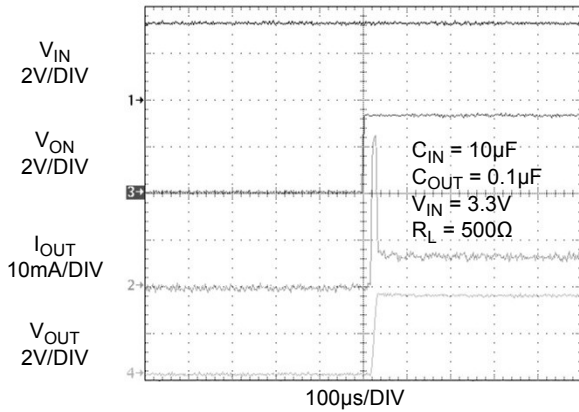


Figure 15. t_{dON} Response

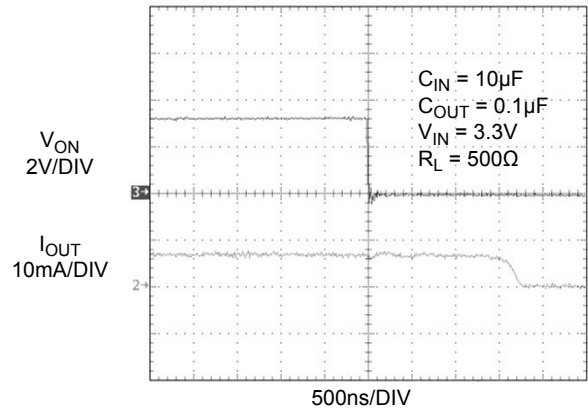


Figure 16. t_{dOFF} Response

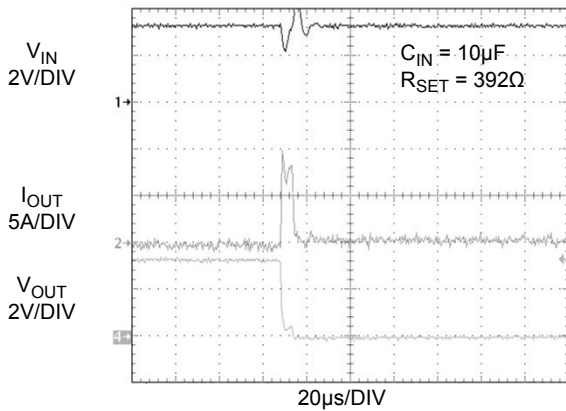


Figure 17. Short Circuit Response Time
(Output shorted to GND)

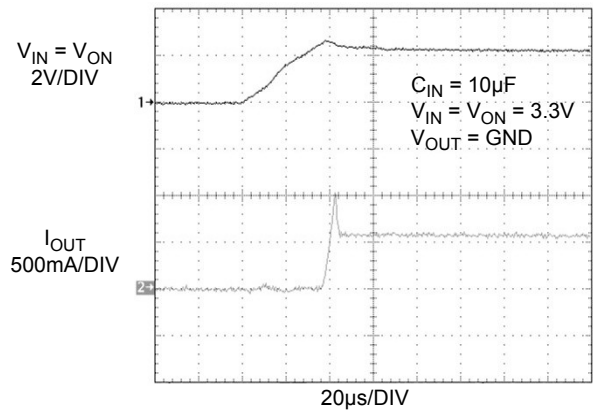


Figure 18. Current Limit Response Time
(Switch is powered into a short)

Typical Characteristics

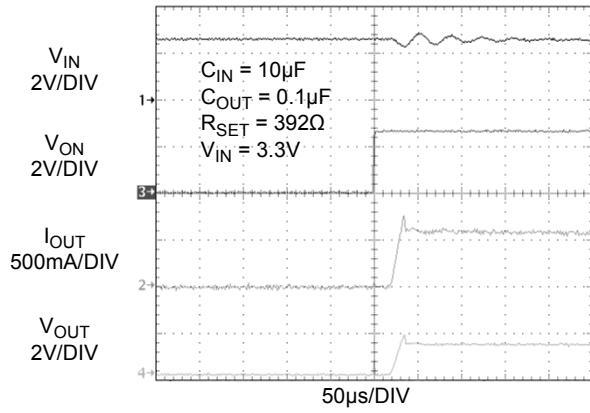


Figure 19. Current Limit Response Time
(Output is loaded by 2.2Ω, $C_{OUT} = 0.1\mu F$)

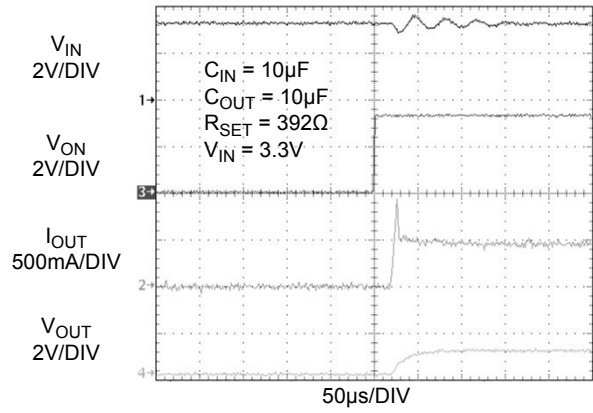


Figure 20. Current Limit Response Time
(Output is loaded by 2.2Ω, $C_{OUT} = 10\mu F$)

Note 2: V_{DRV} signal forces the device to go into overcurrent condition by loading.

Description of Operation

The FPF2163, FPF2164, and FPF2165 are current limited switches that protect systems and loads which can be damaged or disrupted by the application of high currents. The core of each device is a 0.12Ω P-channel MOSFET and a controller capable of functioning over a wide input operating range of 1.8-5.5V. The controller protects against system malfunctions through current limiting, undervoltage lockout and thermal shutdown. The current limit is adjustable from 150mA to 1.5A through the selection of an external resistor.

On/Off Control

The ON pin controls the state of the switch. When ON is high, the switch is in the on state. Activating ON continuously holds the switch in the on state so long as there is no fault. For all versions, an undervoltage on V_{IN} or a junction temperature in excess of 140°C overrides the ON control to turn off the switch. In addition, excessive currents will cause the switch to turn off in the FPF2163 and FPF2164. The FPF2163 has an Auto-Restart feature which will automatically turn the switch on again after 450ms. For the FPF2164, the ON pin must be toggled to turn-on the switch again. The FPF2165 does not turn off in response to an over current condition but instead remains operating in a constant current mode so long as ON is active and the thermal shutdown or undervoltage lockout have not activated.

The ON pin control voltage and V_{IN} pin have independent recommended operating ranges. The ON pin voltage can be driven by a voltage level higher than the input voltage.

Fault Reporting

Upon the detection of an over-current, an input undervoltage, or an over-temperature condition, the FLAGB signals the fault mode by activating LO. For the FPF2163 and FPF2164, the FLAGB goes LO at the end of the blanking time while FLAGB goes LO immediately for the FPF2165. FLAGB remains LO through the Auto-Restart Time for the FPF2165. For the FPF2164, FLAGB is latched LO and ON must be toggled to release it. With the FPF2165, FLAGB is LO during the faults and immediately returns HI at the end of the fault condition. FLAGB is an open-drain MOSFET which requires a pull-up resistor between V_{IN} and FLAGB. During shutdown, the pull-down on FLAGB is disabled to reduce current draw from the supply.

Current Limiting

The current limit ensures that the current through the switch doesn't exceed a maximum value while not limiting at less than a minimum value. The current at which the parts will limit is adjustable through the selection of an external resistor connected to ISET. Information for selecting the resistor is found in the Application Info section. The FPF2163 and FPF2164 have a blanking time of 30ms, nominally, during which the switch will act as a constant current source. At the end of the blanking time, the switch will be turned-off. The FPF2165 has no current limit blanking period so it will remain in a constant current state until the ON pin is deactivated or the thermal shutdown turns-off the switch.

For preventing the switch from large power dissipation during heavy load a short circuit detection feature is introduced. Short circuit condition is detected by observing the output voltage. The switch is put into short circuit current limiting mode if the

switch is loaded with a heavy load. When the output voltage drops below V_{SCTH} , short circuit detection threshold voltage, the current limit value re-conditioned and short circuit current limit value is decreased to 62.5% of the current limit value. This keeps the power dissipation of the part below a certain limit even at dead short conditions at 5.5V input voltage. The V_{SCTH} value is set to be 1V. At around 1.1V of output voltage the switch is removed from short circuit current limiting mode and the current limit is set to the current limit value.

Undervoltage Lockout

The undervoltage lockout turns-off the switch if the input voltage drops below the undervoltage lockout threshold. With the ON pin active the input voltage rising above the undervoltage lockout threshold will cause a controlled turn-on of the switch which limits current over-shoots.

Reverse Current Blocking

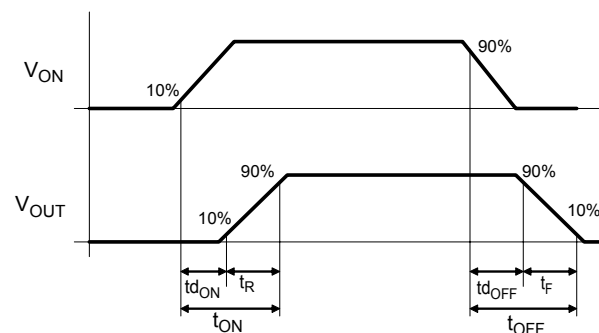
The entire FPF2163/65 family has a Reverse Current Blocking feature that protects input source against current flow from output to input. For a standard USB power design, this is an important feature which protects the USB host from being damaged due to reverse current flow on V_{BUS} . The reverse current blocking feature is active when the load switch is turned off.

If ON pin is LO and output voltage become greater than input voltage, no current can flow from the output to the input. The FLAGB operation is independent of the Reverse Current blocking feature and will not report a fault condition if this feature is activated.

Thermal Shutdown

The thermal shutdown protects the die from internally or externally generated excessive temperatures. During an over-temperature condition the FLAGB is activated and the switch is turned-off. The switch automatically turns-on again if temperature of the die drops below the threshold temperature.

Timing Diagram

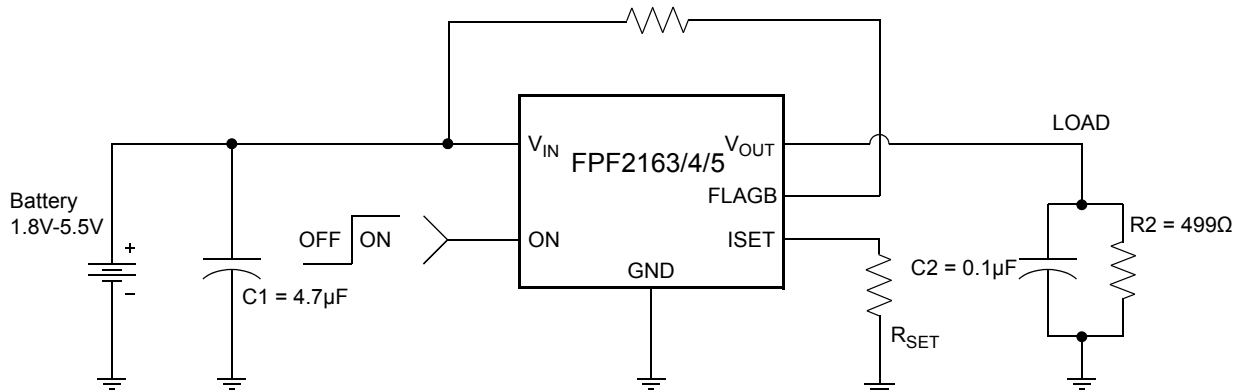


where:

- $t_{d_{ON}}$ = Delay On Time
- t_r = V_{OUT} Rise Time
- t_{ON} = Turn On Time
- $t_{d_{OFF}}$ = Delay Off Time
- t_f = V_{OUT} Fall Time
- t_{OFF} = Turn Off Time

Application Information

Typical Application



Setting Current Limit

The FPF2163, FPF2164, and FPF2165 have a current limit which is set with an external resistor connected between ISET and GND. This resistor is selected by using the following equation,

$$R_{SET} = \frac{275.6}{I_{LIM}} \quad (1)$$

R_{SET} is in Ohms and that of I_{LIM} is Amps

The table below can also be used to select R_{SET} . A typical application would be the 500mA current that is required by a single USB port. Using the table below an appropriate selection for the R_{SET} resistor would be 394Ω. This will ensure that the port load could draw 525mA, but not more than 875mA. Likewise for a dual port system, an R_{SET} of 185Ω would always deliver at least 1125mA and never more than 1875mA

Current Limit Various R_{SET} Values

| R_{SET} [Ω] | Min. Current Limit [mA] | Typ. Current Limit [mA] | Max. Current Limit [mA] |
|------------------|-------------------------------|-------------------------------|-------------------------------|
| 185 | 1125 | 1500 | 1875 |
| 220 | 938 | 1250 | 1562 |
| 275 | 750 | 1000 | 1250 |
| 306 | 675 | 900 | 1125 |
| 345 | 600 | 800 | 1000 |
| 394 | 525 | 700 | 875 |
| 460 | 450 | 600 | 750 |
| 550 | 375 | 500 | 625 |
| 610 | 338 | 450 | 563 |
| 690 | 300 | 400 | 500 |
| 790 | 263 | 350 | 438 |
| 920 | 225 | 300 | 375 |
| 1100 | 188 | 250 | 313 |
| 1380 | 150 | 200 | 250 |
| 1830 | 113 | 150 | 188 |

Input Capacitor

To limit the voltage drop on the input supply caused by transient in-rush currents when the switch is turned on into a discharged load capacitor or a short-circuit, a capacitor needs to be placed between V_{IN} and GND. A 4.7µF ceramic capacitor, C_{IN} , must be placed close to the V_{IN} pin. A higher value of C_{IN} can be used to further reduce the voltage drop experienced as the switch is turned on into a large capacitive load.

Output Capacitor

A 0.1µF capacitor C_{OUT} , should be placed between V_{OUT} and GND. This capacitor will prevent parasitic board inductances from forcing V_{OUT} below GND when the switch turns-off. For the FPF2163 and FPF2164, the total output capacitance needs to be kept below a maximum value, $C_{OUT(max)}$, to prevent the part from registering an over-current condition and turning-off the switch. The maximum output capacitance can be determined from the following formula,

$$C_{OUT(max)} = \frac{I_{LIM(max)} \times t_{BLANK(min)}}{V_{IN}} \quad (2)$$

Power Dissipation

During normal operation as a switch, the power dissipated in the part will depend upon the level at which the current limit is set. The maximum allowed setting for the current limit is 1.5A and this will result in a power dissipation of,

$$P = (I_{LIM})^2 \times R_{DS} = (1.5)^2 \times 0.12 = 270mW \quad (3)$$

If the part goes into current limit the maximum power dissipation will occur when the output is shorted to ground. For the FPF2163, the power dissipation will scale by the Auto-Restart Time, t_{RSTRT} , and the Over Current Blanking Time, t_{BLANK} , so that the maximum power dissipated is,

$$P(max) = \frac{t_{BLANK}}{t_{BLANK} + t_{RSTRT}} \times V_{IN(max)} \times I_{LIM(max)}$$

$$= \frac{30}{30 + 450} \times 5.5 \times 1.5 = 515.6mW \quad (4)$$

This is more power than the package can dissipate, but the thermal shutdown of the part will activate to protect the part from damage due to excessive heating. When using the FPF2164, attention must be given to the manual resetting of the part. The junction temperature will only be able to increase to the thermal shutdown threshold. Once this temperature has been reached, toggling ON will not turn-on the switch until the junction temperature drops. For the FPF2165, a short on the output will cause the part to operate in a constant current state dissipating a worst case power of,

$$\begin{aligned} P(\text{max}) &= V_{\text{IN}}(\text{max}) \times I_{\text{LIM}}(\text{max}) & (5) \\ &= 5.5 \times 1.5 = 8.25\text{W} \end{aligned}$$

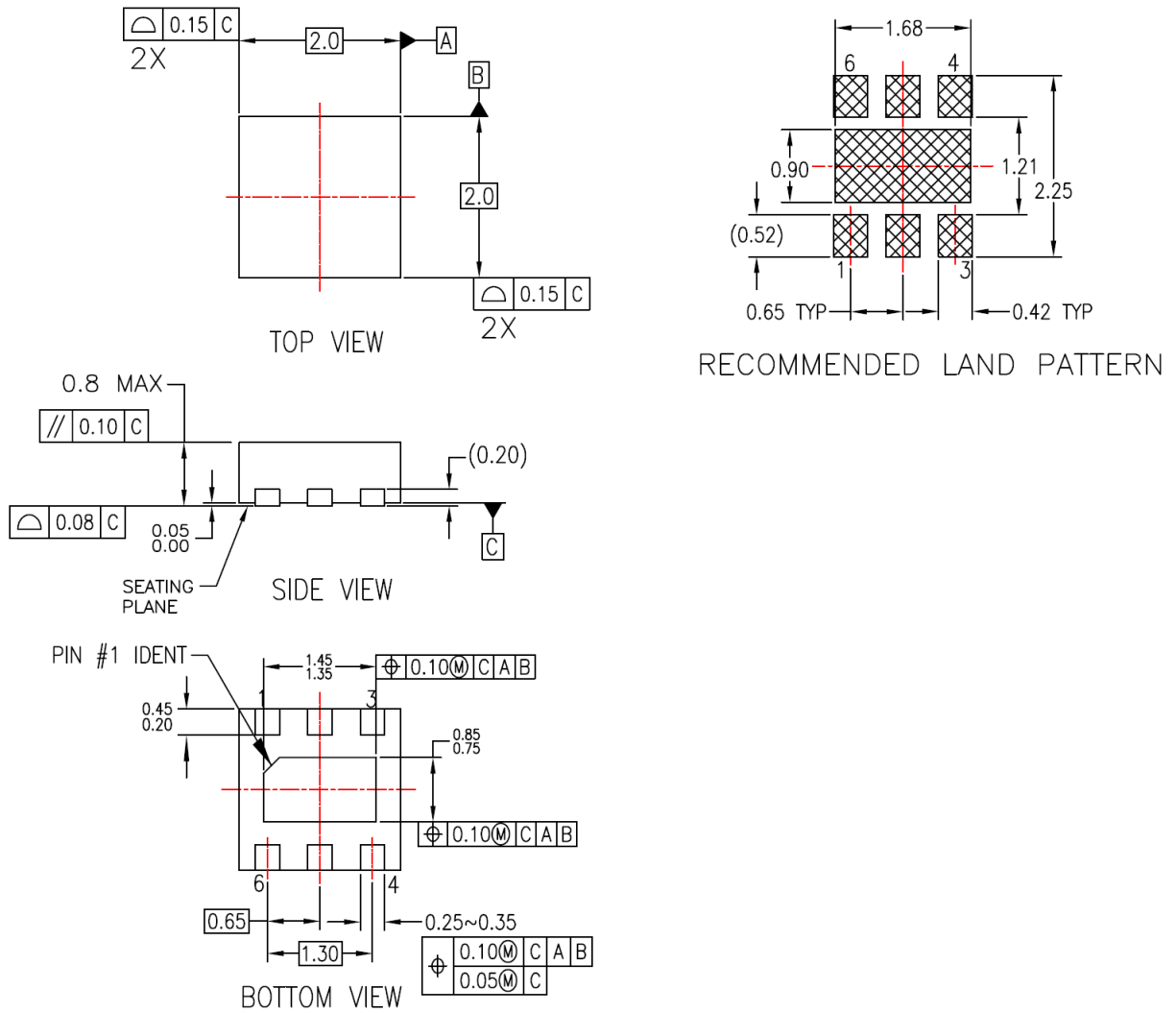
This large amount of power will activate the thermal shutdown and the part will cycle in and out of thermal shutdown so long as the ON pin is active and the short is present.

Board Layout

For best performance, all traces should be as short as possible. To be most effective, the input and output capacitors should be placed close to the device to minimize the effects that parasitic trace inductances may have on normal and short-circuit operation. Using wide traces for V_{IN} , V_{OUT} and GND will help minimize parasitic electrical effects along with minimizing the case to ambient thermal impedance.

The middle pad (pin 7) should be connected to the GND plate of PCB for improving thermal performance of the load switch. An improper layout could result higher junction temperature and triggering the thermal shutdown protection feature. This concern applies when the switch is set at higher current limit value and an overcurrent condition occurs. In this case power dissipation of the switch ($P_{\text{D}} = (V_{\text{IN}} - V_{\text{OUT}}) \times I_{\text{LIM}}(\text{max})$) could exceed the maximum absolute power dissipation of 1.2W.

Dimensional Outline and Pad Layout



NOTES:







- A. NON-CONFORMS TO JEDEC REGISTRATION,
- B. DIMENSIONS ARE IN MILLIMETERS.
- C. DIMENSIONS AND TOLERANCES PER ASME Y14.5M, 1994

MLP06KrevA



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| CROSSVOLT™ | Green FPS™ e-Series™ | Quiet Series™ | TinyBuck™ |
| CTL™ | GTO™ | RapidConfigure™ | TinyLogic® |
| Current Transfer Logic™ | IntelliMAX™ |  ™ | TINYOPTO™ |
| EcoSPARK® | ISOPLANAR™ | Saving our world, 1mW/W/kW at a time™ | TinyPower™ |
| EfficientMax™ | MegaBuck™ | SmartMax™ | TinyPWM™ |
| EZSWITCH™ * | MICROCOUPLER™ | SMART START™ | TinyWire™ |
|  ™ | MicroFET™ | SPM® | µSerDes™ |
|  ™ | MillerDrive™ | STEALTH™ |  ™ |
| Fairchild® | MotionMax™ | SuperFET™ | UHC® |
| Fairchild Semiconductor® | Motion-SPM™ | SuperSOT™_3 | Ultra FRFET™ |
| FACT Quiet Series™ | OPTOLOGIC® | SuperSOT™_6 | UniFET™ |
| FACT® | OPTOPLANAR® | SuperSOT™_8 | VCX™ |
| FAST® |  ™ | SupreMOS™ | VisualMax™ |
| FastvCore™ | PDP SPM™ | SyncFET™ | XS™ |
| FlashWriter® * | Power-SPM™ |  ™ | |
| FPS™ | PowerTrench® | The Power Franchise® | |
| F-PFS™ | PowerXS™ | | |

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Definition of Terms

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Rev. I37