

SANYO Semiconductors DATA SHEET



Bi-CMOS LSI LV8747T — PWM Constant-Current Control Stepping Motor **Driver and Switching Regulator Controller**

Overview

The LV8747T is a PWM constant-current control stepping motor driver and switching regulator controller IC.

Features

- Two circuits of PWM constant-current control stepping motor driver incorporated
- Two circuits of switching regulator controller incorporated
- Motor driver control power incorporated
- Control of the stepping motor to W1-2 phase excitation possible
- Chopping frequency selectable
- Output short-circuit protection circuit incorporated
- High-precision reference voltage circuit incorporated
- Output-stage push-pull composition enabling high-speed operation
- Timer latch type short-circuit protection circuit incorporated
- Upper and lower regenerative diodes incorporated
- Thermal shutdown circuit incorporated

Specifications

Absolute Maximum Ratings at $Ta = 25^{\circ}C$

Parameter	Symbol	Conditions	Ratings	Unit
Supply voltage	VM max		38	V
Driver output peak current 1	MDI _O peak1	OUT1/OUT2 tw \leq 10ms, duty 20%	1.75	А
Driver output continuous current 1	MDI _O max1	OUT1/OUT2	1.5	А
Driver output peak current 2	MDI _O peak2	OUT3/OUT4 tw \leq 10ms, duty 20%	0.8	А
Driver output continuous current 2	MDI _O max2	OUT3/OUT4	0.5	А
Regulator output current	SWI _O max	OUT5/OUT6 tw $\leq 1 \mu s$	500	mA

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Parameter	Symbol	Conditions	Ratings	Unit
Allowable power dissipation 1	Pd max1	Independent IC	0.4	W
Allowable power dissipation 2	Pd max2	Our recommended four-layer substrate *1, *2	4.85	W
Operating temperature	Topr		-20 to +85	°C
Storage temperature	Tstg		-55 to +150	°C

*1 Specified circuit board : $100 \times 100 \times 1.6 mm^3$: 4-layer glass epoxy printed circuit board

 $^{\ast}2$ For mounting to the backside by soldering, see the precautions.

Allowable Operating Ratings at $Ta=25^{\circ}C$

Parameter	Symbol	Conditions	Ratings	Unit
Supply voltage	VM		10 to 35	V
Logic input voltage	V _{IN}		0 to 5	V
VREF input voltage	VREF		0 to 3	V
Regulator output voltage	VO		VM-5 to VM	V
Regulator output current	Ι _Ο		0 to 200	mA
Error amplifier input voltage	V _O A		0 to 3	V
Timing capacity	СТ		100 to 15000	pF
Timing resistance	RT		5 to 50	kΩ
Triangular wave oscillation frequency	Fosc		10 to 800	kHz

Electrical Characteristics at $Ta = 25^{\circ}C$, VM = 24V, VREF = 1.5V

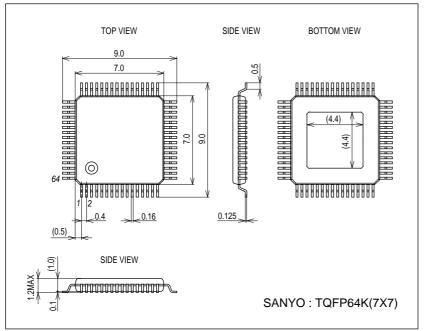
Parameter	Symbol Conditions			Ratings					
ratamotor			min	typ	max	Unit			
General									
VM current drain	IM	PS = "H", no load		6	8	mA			
Thermal shutdown temperature	TSD	Design guarantee		180		°C			
Thermal hysteresis width	ΔTSD	Design guarantee		40		°C			
REG5 output voltage	Vreg5	lreg5 = -1mA	4.5	5.0	5.5	V			
Motor Drivers [Charge pump block]									
Boost voltage	VGH	VM = 24V	28.0	28.7	29.8	V			
Rise time	tONG	$VG = 10\mu F$		50	100	ms			
Oscillation frequency	Fcp	$CHOP = 20k\Omega$	90	120	150	kHz			
Output block (OUT1/OUT2)									
Output on resistance	RonU1	I _O = -1.5A, source side		0.5	0.8	Ω			
	RonD2	I _O = 1.5A, sink side		0.5	0.8	Ω			
Output leak current	IOleak1	V _O = 35V			50	μA			
Diode forward voltage	VD1	ID = -1.5A		1.0	1.3	V			
Output block (OUT3/OUT4)		•							
Output on resistance	RonU2	I _O = -500mA, source side		1.5	1.8	Ω			
	RonD2	I _O = 500mA, sink side		1.1	1.4	Ω			
Output leak current	IOleak2	V _O = 35V			50	μA			
Diode forward voltage	VD2	ID = -500mA		1.0	1.3	V			
Logic input block		•							
Logic pin input current	IINL	V _{IN} = 0.8V	3	8	15	μA			
	IINH	V _{IN} = 5V	30	50	70	μA			
Logic high-level input voltage	V _{IN} H		2.0			V			
Logic low-level input voltage	VINL				0.8	V			
Current control block		•							
VREF input current	IREF	VREF = 1.5V	-0.5			μA			
Chopping frequency	Fchop	CHOP = 20kΩ	45	62.5	75	kHz			
Threshold voltage of current setting	VHH	VREF = 1.5V, I0 = H, I1 = H	0.291	0.300	0.309	V			
comparator	VLH	VREF = 1.5V, I0 = L, I1 = H	0.191	0.200	0.209	V			
	VHL	VREF = 1.5V, I0 = H, I1 = L	0.093	0.100	0.107	V			

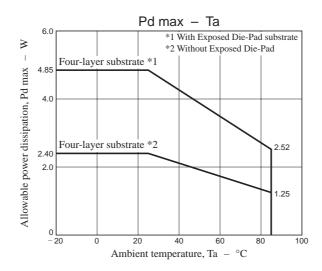
Parameter	Symbol	Conditions		Ratings		Unit
Falameter	Symbol	Conditions	min	typ	max	Unit
Output short-circuit protection circ	uit					
Charge current	IOCP	VOCP = 0V	15	20	25	μA
Threshold voltage	VthOCP		0.8	1.0	1.2	V
Switching regulator Controller [Ref	erence voltage l	block]				
REG25 output voltage	Vreg25	Ireg25 = -1mA	2.475	2.500	2.525	V
Input stability	V _{DL} I	VM = 10 to 35V			10	mV
Load stability	V _{DL} O	Ireg25 = 0 to -3mA			10	mV
Internal regulator block						
REGVM5 output voltage	VregVM5	VregVM5 = 1mA	VM-6.0		VM-5.0	V
Triangular wave oscillator block		•	·			
Oscillation frequency	FOSC	RT = 20kΩ, CT = 620pF	72	80	88	kHz
Frequency fluctuation	FDV	VM = 10 to 35V		1	5	%
Current setting pin voltage	VRT	RT = 20kΩ	0.91	0.98	1.05	V
Protective circuit block			I			
Threshold voltage of comparator	VthFB	FB5, FB6	1.40	1.55	1.70	V
Standby voltage	VstSCP	ISCP = 40μA			100	mV
Source current	ISCP	VSCP = 0V	1.6	2.5	3.4	μA
Threshold voltage	VthSCP		1.65	1.8	1.95	V
Latch voltage	VItSCP	ISCP = 40μA			100	mV
Soft start circuit block						
Source current	ISOFT	VSOFT = 0V	1.3	1.6	1.9	μA
Latch voltage	VItSOFT	ISOFT = 40μA			100	mV
Low-input malfunction preventive of						
Threshold voltage	VUT		8.3	8.7	9.1	V
Hysteresis voltage	VHIS		240	340	440	mV
Error amplifier block		I				
Input offset voltage	V _i O				6	mV
Input offset current	I _i O				30	nA
Input bias current	ljo ljb				100	nA
OPEN open gain	AV			85	100	dB
Common-phase input voltage range	VCM	VM = 10 to 35V		00	3.0	V
Common phase removal ratio	CMRR	10 10 35 0		80	3.0	dB
Max output voltage			4.5			ињ V
	V _O H		4.0	5.0 0.2	0.5	V
Min output voltage	V _O L Isi	FB = 2.5V	300	600	1000	
Output sink current						μA A
Output source current	lso	FB = 2.5V	45	75	105	μA
PWM comparator block	VT400	Putty cyclo = 100%	0.05	1.04	4 07	V
Input threshold voltage (Fosc = 10kHz)	VT100	Duty cycle = 100%	0.95	1.01	1.07	
	VT0	Duty cycle = 0%	0.49	0.52	0.55	V
Input bias current	IBDT	DT6 = 0.4V			1	μA
MAX duty cycle 1	Don1	5ch Internally fixed	95			%
(Fosc = 80kHz) MAX duty cycle 2	Don2	5ch	93			%
(Fosc = 160kHz)		Internally fixed				/0
MAX duty cycle 3	Don3	6ch	56	65	74	%
(Fosc = 10kHz)		VREG25 divided by $17k\Omega$ and $8k\Omega$				
Output block		r		<u>.</u>		
Output ON resistance	RonU3	I _O = -200mA, source side		10	12	Ω
	RonD3	I _O = 200mA, sink side		6	8	Ω
Leak current	ILEAK	V _O = 35V			5	μA

Package Dimensions

unit : mm (typ)

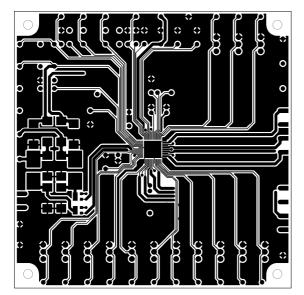
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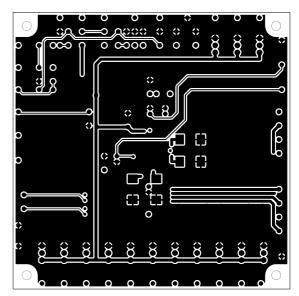


Substrate Specifications (Substrate recommended for operation of LV8747T)

Size	: $100mm \times 100mm \times 1.6mm$ (four-layer substrate [2S2P])
Material	: Glass epoxy
Copper wiring density	: L1 = 85% / L4 = 90%



L1 : Copper wiring pattern diagram



L4 : Copper wiring pattern diagram

Cautions

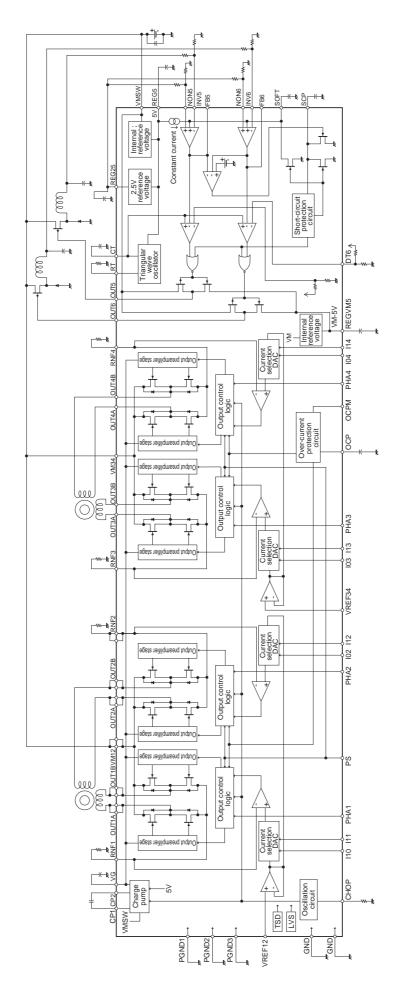
- 1) The data for the case with the Exposed Die-Pad substrate mounted shows the values when 80% or more of the Exposed Die-Pad is wet.
- 2) For the set design, employ the derating design with sufficient margin.
 - Stresses to be derated include the voltage, current, junction temperature, power loss, and mechanical stresses such as vibration, impact, and tension.
 - Accordingly, the design must ensure these stresses to be as low or small as possible.
 - The guideline for ordinary derating is shown below :
 - (1)Maximum value 80% or less for the voltage rating
 - (2)Maximum value 80% or less for the current rating
 - (3)Maximum value 80% or less for the temperature rating
- 3) After the set design, be sure to verify the design with the actual product.

Confirm the solder joint state and verify also the reliability of solder joint for the Exposed Die-Pad, etc. Any void or deterioration, if observed in the solder joint of these parts, causes deteriorated thermal conduction, possibly resulting in thermal destruction of IC.

Pin Assignment

		64	63	62	61	60	59	58	57	56	55	54	53	52	51	50	49		-
		14	104	PS	VREF34	ОСР	OCPM	OUT5	OUT6	REGVM5	GND	NON5	INV5	FB5	NON 6	INV6	FB6		
1	GND				VRE	-	õ	0	0	REG	-	Z	_		Z			DT6	48
2	PHA4																	RT	47
3	OUT4	В																СТ	46
4	RNF4																RI	EG25	45
5	OUT4	A															F	REG5	44
6	VM34																	SCP	43
7	OUT3	В															5	SOFT	42
8	RNF3									LV8747T							V	MSW	41
9	OUT3	A								F							VR	EF12	40
10	PGND)3															С	HOP	39
11	103																	CP1	38
12	113								Тор	View								CP2	37
13	PHA3																	VG	36
14	102																	101	35
15	112																	111	34
16	PHA2	2	В	В			A	A			В	В			A	A	۲ ا	PHA1	33
		PGND2	OUT2B	OUT2B	RNF2	RNF2	OUT2A	OUT2A	VM12	VM12	OUT1B	OUT1B	RNF1	RNF1	OUT1A	OUT1A	PGND1		
	L	17	18	19	20	21	22	23	24	25	26	27	28	29	30	31	32		1

Block Diagram

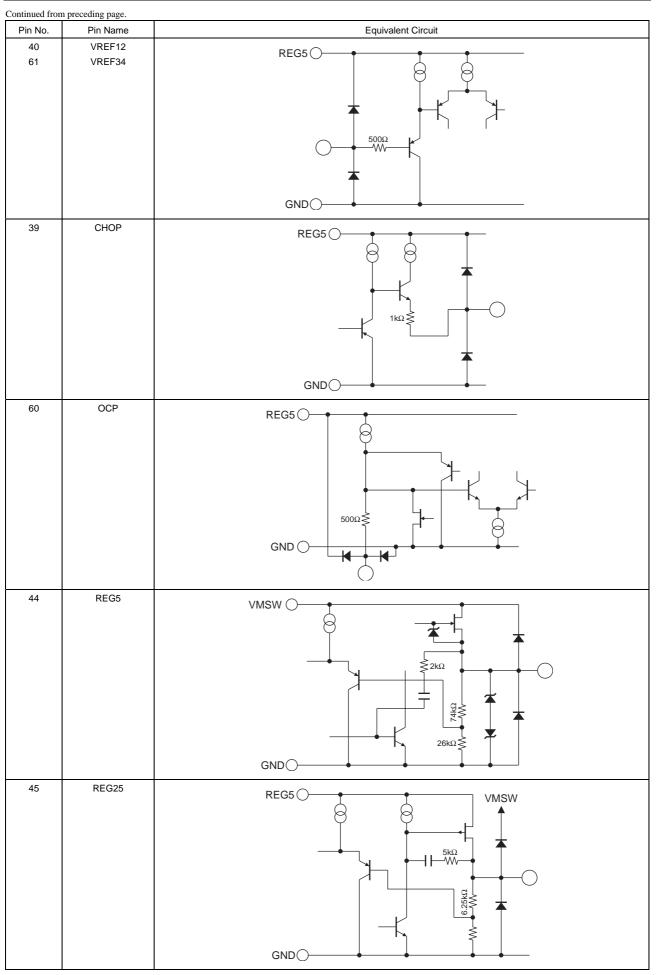


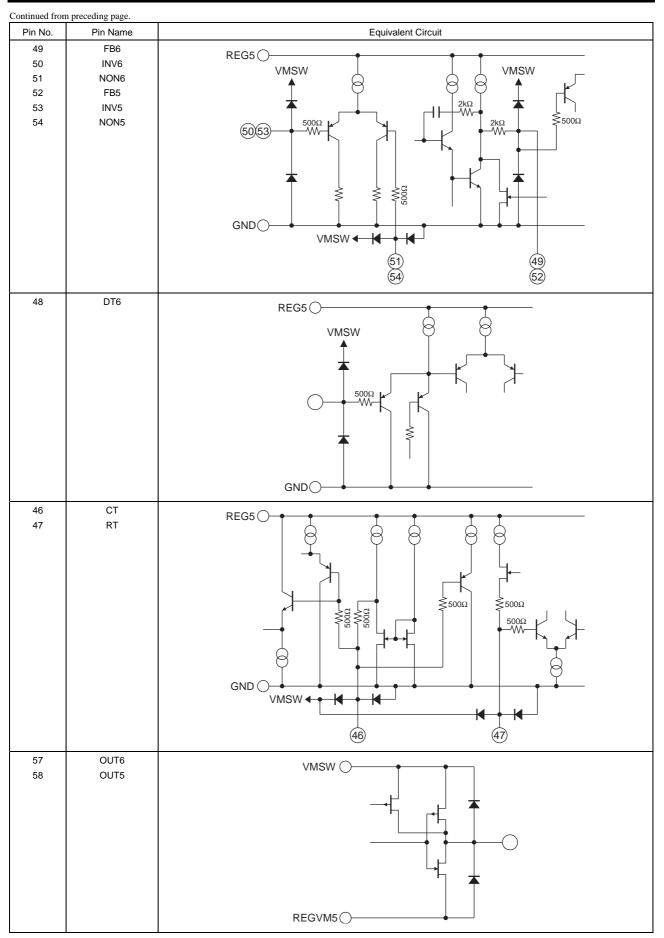
Pin Functions

Pin No	Pin	Description
24	VM12	Driver 1/2ch Pin to connect to power supply
25		
30	OUT1A	Driver 1ch OUTA output pin
31		
26	OUT1B	Driver 1ch OUTB output pin
27	1	
28	RNF1	Driver 1ch Current sense resistor connection pin
29	-	
22	OUT2A	Driver 2ch OUTA output pin
23		
18	OUT2B	Driver 2ch OUTB output pin
19	00125	
	DNIEG	
20	RNF2	Driver 2ch Current sense resistor connection pin
21		
35	101	Driver 1ch Output current setting input pin
34	111	
33	PHA1	Driver 1ch Output phase shift input pin
14	102	Driver 2ch Output current setting input pin
15	l12	
16	PHA2	Driver 2ch Output phase shift input pin
40	VREF12	Driver 1/2ch Output current setting reference voltage input pin
32	PGND1	Driver output Power GND
17	PGND2	Driver output Power GND
6	VM34	Driver 3/4ch Power connection pin
9	OUT3A	Driver 3ch OUTA output pin
7	OUT3B	Driver 3ch OUTB output pin
8	RNF3	
		Driver 3ch Current sense resistor connection pin
5	OUT4A	Driver 4ch OUTA output pin
3	OUT4B	Driver 4ch OUTB output pin
4	RNF4	Driver 4ch Current sense resistor connection pin
11	103	Driver 3ch Output current setting input pin
12	l13	
13	PHA3	Driver 3ch Output phase shift input pin
63	104	Driver 4ch Output current setting input pin
64	114	
2	PHA4	Driver 4ch Output phase shift input pin
61	VREF34	Driver 3/4ch Output current setting reference voltage input pin
10	PGND3	Driver output Power GND
60	OCP	Pin to connect to the output short-circuit state detection time setting capacitor
59	OCPM	Over-current mode changeover pin
39	СНОР	Pin to connect to the resistor to set the chopping frequency
62	PS	Driver Power save input pin
36	VG	
		Charge pump capacitor connection pin
38	CP1	Charge pump capacitor connection pin
37	CP2	Charge pump capacitor connection pin
41	VMSW	Power connection pin
44	REG5	Internal regulator output pin
56	REGVM5	Internal regulator output pin
45	REG25	Regulator Reference voltage output pin
46	СТ	Regulator Timing capacity external pin
47	RT	Regulator Timing resistor external pin
42	SOFT	Regulator Soft start setting pin
43	SCP	Regulator Timer and latch setting pin
-		

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Pin No	Pin	Description
53	INV5	Regulator Error amplifier 5 input – pin
52	FB5	Regulator Error amplifier 5 output pin
58	OUT5	Regulator Output 5
51	NON6	Regulator Error amplifier 6 input + pin
50	INV6	Regulator Error amplifier 6 input – pin
49	FB6	Regulator Error amplifier 6 output pin
57	OUT6	Regulator Output 6
48	DT6	Regulator Output 6 MAX DUTY setting pin
55	GND	GROUND
1	GND	GROUND

Equivale	nt Circuits	
Pin No.	Pin Name	Equivalent Circuit
2	PHA4	REG5 () + +
11	103	
12	l13	
13	PHA3	
14	102	
15	112	
16	PHA2	
33 34	PHA1	
34 35	l11 l01	
59	OCPM	
62	PS	
63	104	
64	114	
36	VG	VMSW 22
37	CP2	$(38) \qquad \qquad$
38	CP1	
		GND
3	OUT4B	(6)
4	RNF4 OUT4A	Ť
5 6	VM34	
8 7	OUT3B	$REG5\bigcirc \qquad \qquad$
8	RNF3	
9	OUT3A	59-37
10	PGND3	
4-	DONIDO	
17	PGND2	(24)
18	OUT2B	
19	OUT2B	Ϋ́
20 21	RNF2 RNF2	
21	OUT2A	
22	OUT2A	╷ →╷ 주 ┍┝──
23	VM12	
25	VM12	
26	OUT1B	
27	OUT1B	
28	RNF1	
29	RNF1	
30	OUT1A	
31	OUT1A	GND () (20)28
32	PGND1	(17)32 (21)29
		Continued on next page





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Pin No.	Pin Name	Equivalent Circuit
56	REGVM5	VMSW O
42	SOFT	REG5
43	SCP	REG5 GND VMSW

Stepping Motor Driver OUT1/OUT2(OUT3/OUT4)

(1) Output control logic

Parallel in	put (Note)	Ou	tput	Current direction	
PS	PHA	OUTA	OUTB	Current direction	
Low	*	Off	Off	Standby	
High	Low	Low	High	OUTB→OUTA	
High	High	High	Low	OUTA→OUTB	

(Note) : Enter either "H" or "L" externally for the logic input pin. Never use the input pin in the OPEN state.

(2) Constant-current setting

I0 (Note)	I1 (Note)	Output current
High	High	I _O = (VREF/5) /RNF
Low	High	I _O = ((VREF/5) /RNF) × 2/3
High	Low	I _O = ((VREF/5) /RNF) × 1/3
Low	Low	IO = 0

(Note) : Enter either "H" or "L" externally for the logic input pin. Never use the input pin in the OPEN state.

Set current calculation method

The constant-current control setting of STM driver is determined as follows from the setting of VREF voltage, and I0 and I1, and resistor (RNF) connected between RNF and GND :

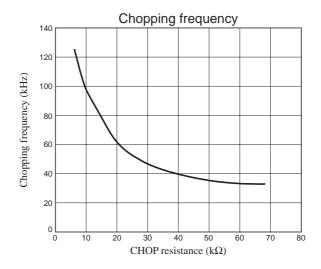
Iconst [A] = ((VREF [V] /5) /RNF $[\Omega]$) × attenuation factor

(Example) For VREF = 1.5V, I0 = I1 = "H" and RNF = 1 Ω ; Iconst = 1.5V/5/1 $\Omega \times 1 = 0.3A$

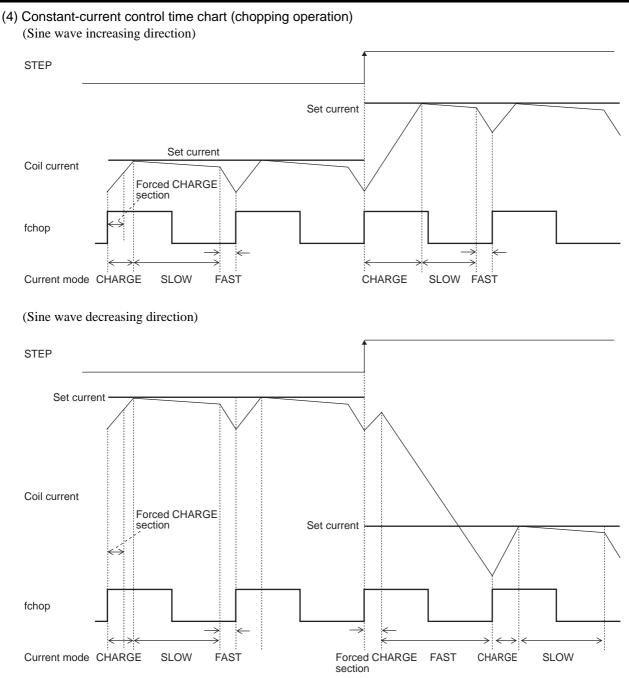
(3) Setting the chopping frequency

For constant-current control, chopping operation is made with the frequency determined by the external resistor (connected to the CHOP pin).

The chopping frequency to be set with the resistance connected to the CHOP pin (pin 39) is as shown below.



The recommended chopping frequency ranges from 30kHz to 120kHz.



In each current mode, the operation sequence is as described below :

- At rise of chopping frequency, the CHARGTE mode begins.(The section in which the CHARGE mode is forced regardless of the magnitude of the coil current (ICOIL) and set current (IREF) exists for 1/16 of one chopping cycle.)
- The coil current (ICOIL) and set current (IREF) are compared in this forced CHARGE section.
 - When (ICOIL<IREF) state exists in the forced CHARGE section;

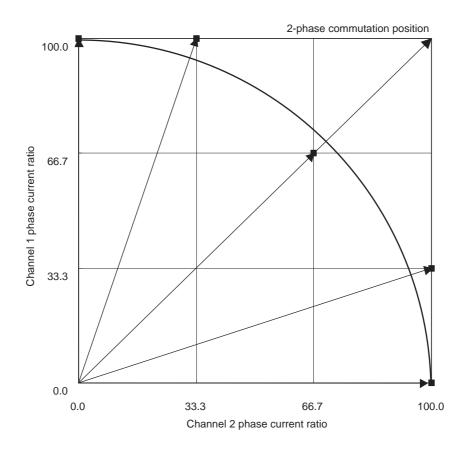
CHARGE mode up to ICOIL \geq IREF, then followed by changeover to the SLOW DECAY mode, and finally by the FAST DECAY mode for the 1/16 portion of one chopping cycle.

When (ICOIL<IREF) state does not exist in the forced CHARGE section;

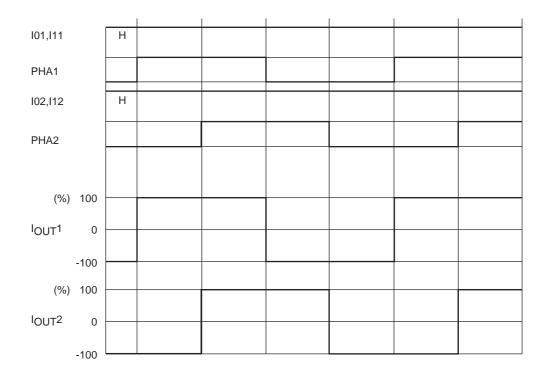
The FAST DECAY mode begins. The coil current is attenuated in the FAST DECAY mode till one cycle of chopping is over.

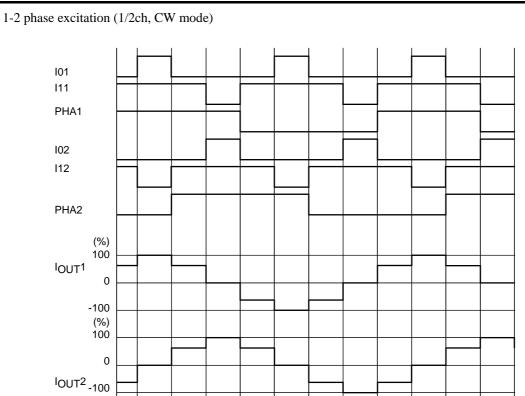
Above operations are repeated. Normally, the SLOW (+FAST) DECAY mode continues in the sine wave increasing direction, then entering the FAST DECAY mode till the current is attenuated to the set level and followed by the SLOW DECAY mode.

(5) Output current vector locus (one step is normalized to 90 degrees)



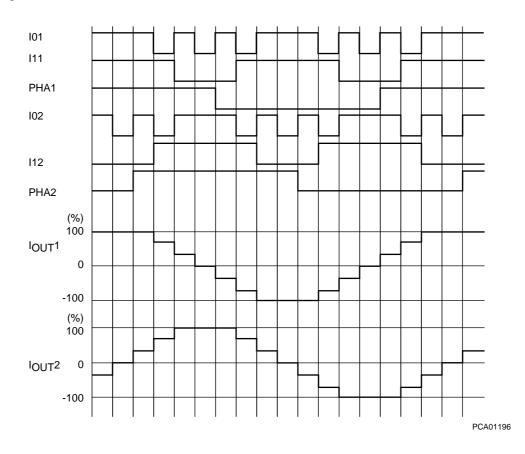
(6) Typical current waveform in each excitation mode Two-phase excitation (1/2ch, CW mode)





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W1-2 phase excitation (1/2ch, CW mode)



Output short-circuit protection circuit

To protect IC from damage due to short-circuit of the output caused by lightening or ground fault, the output short-circuit protection circuit to put the output in the standby mode is incorporated.

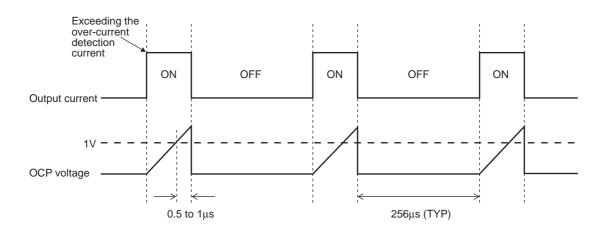
(1) Output short-circuit protection operation changeover function

Changeover to the output short-circuit protection of IC is made by the setting of OCPM pin.

OCPM	State
"Low"	Auto reset method
"High"	Latch method

(Auto reset method)

When the output current is below the output short-circuit protection current, the output is controlled by the input signal. When the output current exceeds the detection current, the switching waveform as shown below appears instead.



When detecting the output short-circuit state, the short-circuit detection circuit is activated.

When the short-circuit detection circuit operation exceeds the timer latch time described later, the output is changed over to the standby mode and reset to the ON mode again in $256\mu s$ (TYP). In this event, if the over-current mode still continues, the above switching mode is repeated till the over-current mode is canceled.

(Latch method)

Similarly to the case of automatic reset method, the short-circuit detection circuit is activated when it detects the output short-circuit state.

When the short-circuit detection circuit operation exceeds the timer latch time described later, the output is changed over to the standby mode.

In this method, latch is released by setting PS = "L"

(2) OCP pin constant setting method (timer latch setting)

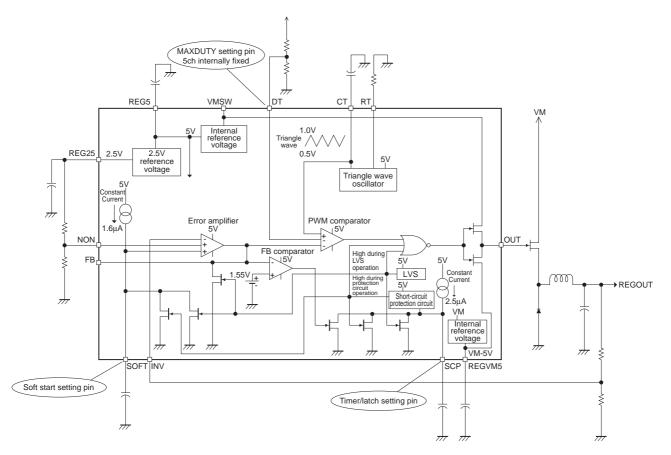
Connect C between the OCP pin and GND, and the time up to the output OFF can be set in case of output short-circuit. The C value can be determined as follows :

Timer latch : Tocp

$$\begin{split} &\text{Tocp} \approx C \times V/I \text{ [s]} \\ &\text{V}: \text{Threshold voltage TYP 1V} \\ &\text{I}: \text{OCP charge current TYP 20} \mu\text{A} \\ &\text{(C: Recommended constant value 100pF to 200pF)} \end{split}$$

Switching Regulator Controller

(1) Regulator block diagram



(2) Timing chart

Short-circuit protection comparator reference voltage Oscillator triangular wave output (CT) Max_Duty setting voltage (DT) Error amplifier output (FB)

Output (OUT)

Triangular wave conversion output

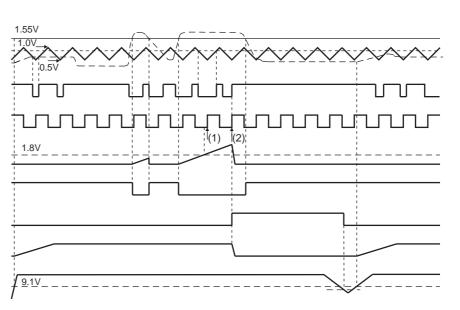
SCP pin waveform

Short-circuit protection comparator output

Latch output

SOFT pin waveform

VMSW supply voltage



(3) SOFT pin constant setting method (Soft start setting)

The switching regulator can be set to soft-start by connecting C between the SOFT pin and GND. Determine the C value as follows :

Soft start time : Tsoft	Tsoft $\approx C \times V/I$ [s]
	V : Error amplifier input + pin voltage (NON5/NON6)
	I : SOFT charge current TYP 1.6µA

(4). SCP pin constant setting method (Timer latch setting)

The time up to the output OFF in case of regulator output short-circuit can be set by connecting C between the SCP pin and GND.

Determine the C value as follows :

Timer latch : TscpTscp $\approx C \times V/I$ [s]V : Threshold voltage TYP 1.8VI : SCP charge current TYP 2.5 μ A

(5) RT pin constant setting method (Capacitor charge/discharge current setting)

The CT pin capacitor charge/discharge current can be set for triangular wave generation by connecting R between the RT pin and GND.

Determine the R value as follows :

Charge/discharge current : Irt $Irt \approx V/R$ [A] V : R pin voltage TYP 0.98V

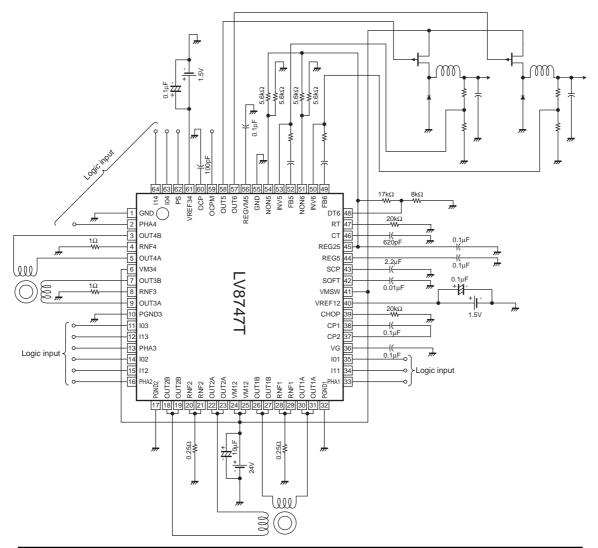
(6) CT pin constant setting method (Triangular wave oscillation frequency setting)

The triangular wave oscillation can be set (together with the setting of charge/discharge current setting of RT pin) by connecting C between the CT pin and GND.

Determine the C value as follows :

Triangular wave oscillation frequency : Fosc	Fosc $\approx 1/\{2 \times C \times V/I\}$ [Hz]
	V : Triangle wave amplitude TYP 0.5V (Fosc = 10kHz)
	*Note that the amplitude increases with the frequency.
	I : Capacitor charge/discharge current. See the RT pin constant
	setting method of (5).

Application Circuit



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