

## 9-Mbit 4-Word Burst SRAM with DDR-I Architecture

#### **Features**

- 9-Mbit density (256 Kbit x 36)
- · 167-MHz clock for high bandwidth
- · 4-Word Burst for reducing address bus frequency
- Double Data Rate (DDR) interfaces (data transferred at 333 MHz @ 167 MHz)
- Two input clocks (K and K) for precise DDR timing—SRAM uses rising edges only
- Two input clocks for output data (C and C) to minimize clock-skew and flight-time mismatches.
- Separate Port Selects for depth expansion
- . Synchronous internally self-timed writes
- 2.5V core power supply with HSTL inputs and outputs
- Available in 165-ball FBGA package (13 x 15 x 1.4 mm)
- · Variable drive HSTL output buffers
- Expanded HSTL output voltage (1.4V–1.9V)
- JTAG 1149.1 compatible test access port

#### Configuration

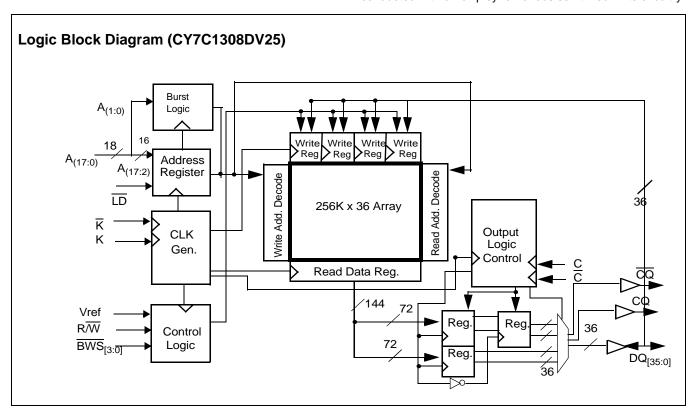
CY7C1308DV25 - 256K x 36

### **Functional Description**

The CY7C1308DV25 is a 2.5V Synchronous Pipelined SRAM equipped with DDR-I (Double Data Rate) architecture. The DDR-I architecture consists of an SRAM core with advanced synchronous peripheral circuitry and a 2-bit burst counter. Addresses for Read and Write are latched on alternate rising edges of the input (K) clock. Write data is registered on the rising edges of both K and K. Read data is driven on the rising edges of C and C if provided, or on the rising edge of K and K if C/C are not provided. Every Read or Write operation is associated with four words that burst sequentially into or out of the device. The burst counter takes in the least two significant bits of the external address and bursts four 36-bit words. Depth expansion is accomplished with Port Selects for each port. Port Selects allow each port to operate independently.

Asynchronous inputs include impedance match (ZQ). Synchronous data outputs (Q, sharing the same physical pins as the data inputs D) are tightly matched to the two output echo clocks CQ/CQ, eliminating the need for separately capturing data from each individual DDR SRAM in the system design. Output data clocks (C/C) are also provided for maximum system clocking and data synchronization flexibility.

All synchronous inputs pass through input registers controlled by the K or K input clocks. All data outputs pass through output registers controlled by the C or C input clocks. Writes are conducted with on-chip synchronous self-timed write circuitry.





### **Selection Guide**

	167 MHz	Unit
Maximum Operating Frequency	167	MHz
Maximum Operating Current	700	mA

### **Pin Configuration**

# 165-ball FBGA (13 x 15 x 1.4 mm) Pinout CY7C1308DV25 (256K x 36)

	1	2	3	4	5	6	7	8	9	10	11
Α	CQ	GND/144M	NC/36M	R/W	BWS <sub>2</sub>	K	BWS <sub>1</sub>	LD	NC/18M	GND/72M	CQ
В	NC	DQ27	DQ18	Α	BWS <sub>3</sub>	K	BWS <sub>0</sub>	Α	NC	NC	DQ8
С	NC	NC	DQ28	VSS	Α	A0	A1	VSS	NC	DQ17	DQ7
D	NC	DQ29	DQ19	VSS	VSS	VSS	VSS	VSS	NC	NC	DQ16
E	NC	NC	DQ20	VDDQ	VSS	VSS	VSS	VDDQ	NC	DQ15	DQ6
F	NC	DQ30	DQ21	VDDQ	VDD	VSS	VDD	VDDQ	NC	NC	DQ5
G	NC	DQ31	DQ22	VDDQ	VDD	VSS	VDD	VDDQ	NC	NC	DQ14
Н	NC	VREF	VDDQ	VDDQ	VDD	VSS	VDD	VDDQ	VDDQ	VREF	ZQ
J	NC	NC	DQ32	VDDQ	VDD	VSS	VDD	VDDQ	NC	DQ13	DQ4
K	NC	NC	DQ23	VDDQ	VDD	VSS	VDD	VDDQ	NC	DQ12	DQ3
L	NC	DQ33	DQ24	VDDQ	VSS	VSS	VSS	VDDQ	NC	NC	DQ2
M	NC	NC	DQ34	VSS	VSS	VSS	VSS	VSS	NC	DQ11	DQ1
N	NC	DQ35	DQ25	VSS	Α	Α	Α	VSS	NC	NC	DQ10
Р	NC	NC	DQ26	Α	Α	С	Α	Α	NC	DQ9	DQ0
R	TDO	TCK	Α	Α	Α	C	Α	Α	Α	TMS	TDI

### **Pin Definitions**

Name	I/O	Description
DQ <sub>[35:0]</sub>	Input/Output- Synchronous	<b>Data Input/Output Signals</b> . Inputs are sampled on the rising edge of K and $\overline{K}$ clocks during valid Write operations. These pins drive out the requested data <u>during</u> a Read operation. Valid data is driven out on the rising edge of both the C and $\overline{C}$ clocks during Read operations or K and $\overline{K}$ when in single clock mode. When Read access is deselected, $Q_{[35:0]}$ are automatically three-stated.
LD	Input- Synchronous	<b>Synchronous Load</b> . This input is brought LOW when a bus cycle sequence is to be defined. This definition includes address and Read/Write direction. All transactions operate on a burst of 4 data (two clock periods of bus activity).
BWS <sub>0</sub> , BWS <sub>1</sub> , BWS <sub>2</sub> , BWS <sub>3</sub>	Input- Synchronous	Byte Write Select 0, 1, 2 and 3 – active LOW. Sampled on the rising edge of the K and K clocks during Write operations. Used to select which byte is written into the device during the current portion of the Write operations. Bytes not written remain unaltered. CY7C1308DV25 – BWS $_0$ controls D $_{[8:0]}$ , BWS $_1$ controls D $_{[17:9]}$ , BWS $_2$ controls D $_{[26:18]}$ and BWS $_3$ controls D $_{[35:27]}$ All the Byte Write Selects are sampled on the same edge as the data. Deselecting a Byte Write Select will cause the corresponding byte of data to be ignored and not written into the device.
A, A0, A1	Input- Synchronous	Address Inputs. These address inputs are multiplexed for both Read and Write operations. A0 and A1 are the inputs to the burst counter. These are incremented in a linear fashion internally. 18 address inputs are needed to access the entire memory array. All the address inputs are ignored when the part is deselected.
R/W	Input- Synchronous	Synchronous Read/Write Input. When LD is LOW, this input designates the access type (Read when R/W is HIGH, Write when R/W is LOW) for loaded address. R/W must meet the set-up and hold times around edge of K.



### Pin Definitions (continued)

Name	I/O	Description			
С	Input-Clock	<b>Positive Input Clock for Output Data</b> . C is used in conjunction with $\overline{C}$ to clock out the Read data from the device. C and $\overline{C}$ can be used together to deskew the flight times of various devices on the board back to the controller. See application example for further details.			
C	Input-Clock	Negative Input Clock for Output Data. $\overline{C}$ is used in conjunction with C to clock out the Read data from the device. C and $\overline{C}$ can be used together to deskew the flight times of various devices on the board cack to the controller. See application example for further details.			
К	Input-Clock	<b>Positive Input Clock Input</b> . The rising edge of K is used to capture synchronous inputs to the device and to drive out data through $Q_{[35:0]}$ when in single clock mode. All accesses are initiated on the rising edge of K.			
K	Input-Clock	Negative Input Clock Input. $\overline{K}$ is used to capture synchronous inputs being presented to the device and to drive out data through $Q_{[35:0]}$ when in single clock mode.			
CQ	Echo Clock	CQ is referenced with respect to C. This is a free running clock and is synchroniz to the output clock (C) of the DDR-I. In the single clock mode, CQ is generated with respect to K. The timings for the echo clocks are shown in the AC timing table.			
CQ	Echo Clock	CQ is referenced with respect to C. This is a free running clock and is synchronize to the output clock (C) of the DDR-I. In the single clock mode, CQ is generated with respect to K. The timings for the echo clocks are shown in the AC timing table.			
ZQ	Input	Output Impedance Matching Input. This input is used to tune the device outputs to the system data bus impedance. CQ, CQ and $Q_{[35:0]}$ output impedance are set to $0.2 \times RC$ where RQ is a resistor connected between ZQ and ground. Alternately, this pin can be connected directly to $V_{DDQ}$ , which enables the minimum impedance mode. This pin cannot be connected directly to GND or left unconnected.			
TDO	Output	TDO for JTAG.			
TCK	Input	TCK pin for JTAG.			
TDI	Input	TDI pin for JTAG.			
TMS	Input	TMS pin for JTAG.			
NC	N/A	Not connected to the die. Can be tied to any voltage level.			
NC/18M	N/A	Address expansion for 18M. This is not connected to the die.			
NC/36M	N/A	Address expansion for 36M. This is not connected to the die.			
GND/72M	Input	Address expansion for 72M. This should be tied LOW.			
GND/144M	Input	Address expansion for 144M. This should be tied LOW.			
V <sub>REF</sub>	Input- Reference	Reference Voltage Input. Static input used to set the reference level for HSTL inputs and outputs as well as AC measurement points.			
$V_{DD}$	Power Supply	Power supply inputs to the core of the device.			
V <sub>SS</sub>	Ground	Ground for the device.			
$V_{DDQ}$	Power Supply	Power supply inputs for the outputs of the device.			

#### Introduction

#### **Functional Overview**

The CY7C1308DV25 is a synchronous pipelined Burst SRAM equipped with DDR interface.

Accesses are initiated on the positive input clock (K). All synchronous input timing is referenced from the rising edge of the input clocks (K and K) and all output timing is referenced to the rising edge of output clocks (C and  $\overline{C}$  or K and K when in single clock mode).

All synchronous data inputs (D $_{[35:0]}$ ) pass through input registers controlled by the input clocks (K and K). All synchronous data outputs (Q $_{[35:0]}$ ) pass through output

registers controlled by the rising edge of the output clocks (C and  $\overline{C}$  or K and  $\overline{K}$  when in single clock mode).

<u>All synchronous control</u> (R/ $\overline{W}$ ,  $\overline{LD}$ ,  $\overline{BWS}_0$ ,  $\overline{BWS}_1$ ,  $\overline{BWS}_2$ ,  $\overline{BWS}_3$ ) inputs pass through input registers controlled by the rising edge of the input clocks (K and K).

#### **Read Operations**

The CY7C1308DV25 is organized internally as an array of 256K x 36. Accesses are completed in a burst of four sequential 36-bit data words. Read operations are initiated by asserting R/W HIGH and  $\overline{\text{LD}}$  LOW at the rising edge of the Positive Input Clock (K). The address presented to Address inputs are stored in the Read address register and the least



two significant bits of the address are presented to the burst counter. The burst counter increments the address in a linear fashion. Following the next K clock rise the corresponding 36-bit word of data from this address location is driven onto the Q<sub>[35:0]</sub> using C as the output timing reference. On the subsequent rising edge of C the next 36-bit data word from the address location generated by the burst counter is driven onto the  $\mathsf{Q}_{[35:0]}$  This process continues until all four 36-bit data words have been driven out onto  $Q_{[35:0]}$ . The requested data will be valid 3 ns from the rising edge of the output clock (C or C, 167-MHz device). In order to maintain the internal logic, each Read access must be allowed to complete. Each Read access consists of four 36-bit data words and takes two clock cycles to complete. Therefore, Read accesses to the device can not be initiated on two consecutive K clock rises. The internal logic of the device will ignore the second Read request. Read accesses can be initiated on every other K clock rise. Doing so will pipeline the data flow such that data is transferred out of the device on every rising edge of the output clocks (C and C or K and K when in single clock mode).

When the read port is deselected, the CY7C1308DV25 will first complete the pending read transactions. Synchronous internal circuitry will automatically three-state the outputs following the next rising edge of the positive output clock (C). This will allow for a seamless transition between devices without the insertion of wait states in a depth expanded memory.

#### **Write Operations**

Write operations are initiated by asserting R/W LOW and LD LOW at the rising edge of the positive input clock (K). The address presented to Address inputs are stored in the Write address register and the least two significant bits of the address are presented to the burst counter. The burst counter increments the address in a linear fashion. On the following K clock rise the data presented to  $D_{[35:0]}$  is latched and stored into the 36-bit Write Data register provided BWS[3:0] are asserted active. On the subsequent rising edge of the Negative Input Clock ( $\overline{K}$ ) the information presented to  $D_{[35:0]}$ is also stored into the Write Data Register provided BWS[3:0] are asserted active. This process continues for one more cycle until four 36-bit words (a total of 144 bits) of data are stored in the SRAM. The 144 bits of data are then written into the memory array at the specified location. Therefore, Write accesses to the device can not be initiated on two consecutive K clock rises. The internal logic of the device will ignore the second Write request. Write accesses can be initiated on every other rising edge of the positive input clock (K). Doing so will pipeline the data flow such that 36-bits of data can be transferred into the device on every rising edge of the input clocks (K and K).

When deselected, the Write port will ignore all inputs after the pending Write operations have been completed.

#### **Byte Write Operations**

Byte Write operations are supported by the CY7C1308DV25. A Write operation is initiated as described in the Write Operation section above. The bytes that are written are determined by BWS<sub>[3:0]</sub> which are sampled with each set of 36-bit data word. Asserting the appropriate Byte Write Select input

during the data portion of a write will allow the data being presented to be latched and written into the device. Deasserting the Byte Write Select input during the data portion of a Write will allow the data stored in the device for that byte to remain unaltered. This feature can be used to simplify Read/Modify/Write operations to a Byte Write operation.

#### **Single Clock Mode**

The CY7C1308DV25 can be used with a single clock that controls both the input and output registers. In this mode the device will recognize only a single pair of input clocks (K and  $\overline{K}$ ) that control both the input and output registers. This operation is identical to the operation if the device had zero skew between the  $K/\overline{K}$  and  $C/\overline{C}$  clocks. All timing parameters remain the same in this mode. To use this mode of operation, the user must tie C and  $\overline{C}$  HIGH at power-on. This function is a strap option and not alterable during device operation.

#### **DDR Operation**

The CY7C1308DV25 enables high-performance operation through high clock frequencies (achieved through pipelining) and double data rate mode of operation. At slower frequencies, the CY7C1308DV25 requires a single No Operation (NOP) cycle when transitioning from a Read to a Write cycle. At higher frequencies, a second NOP cycle may be required to prevent bus contention.

If a Read occurs after a Write cycle, address and data for the Write are stored in registers. The Write information must be stored because the SRAM can not perform the last word Write to the array without conflicting with the Read. The data stays in this register until the next Write cycle occurs. On the first Write cycle after the Read(s), the stored data from the earlier Write will be written into the SRAM array. This is called a Posted Write.

#### **Depth Expansion**

Depth expansion requires replicating the LD control signal for each bank. All other control signals can be common between banks as appropriate.

#### **Echo Clocks**

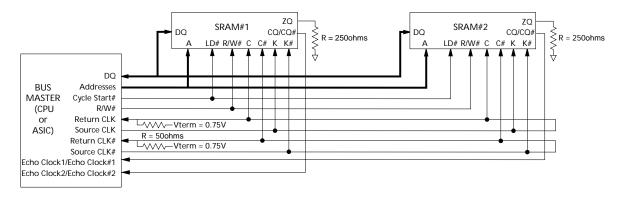
Echo clocks are provided on the DDR-I to simplify data capture on high-speed systems. Two echo clocks are gene<u>rated</u> by the DDR-I. CQ is referenced with respect to C and  $\overline{\text{CQ}}$  is referenced with respect to  $\overline{\text{C}}$ . These are free-running clocks and are synchronized to the output clock of the DDR-I. In the <u>single</u> clock mode, CQ is generated with respect to K and  $\overline{\text{CQ}}$  is generated with respect to K. The timings for the echo clocks are shown in the AC Timing table.

#### **Programmable Impedance**

An external resistor, RQ must be connected between the ZQ pin on the SRAM and  $V_{SS}$  to allow the SRAM to adjust its output driver impedance. The value of RQ must be 5X the value of the intended line impedance driven by the SRAM, The allowable range of RQ to guarantee impedance matching with a tolerance of  $\pm 15\%$  is between  $175\Omega$  and  $350\Omega$ , with  $V_{DDQ}$ =1.5V. The output impedance is adjusted every 1024 cycles to adjust for drifts in supply voltage and temperature.



### Application Example[1]



### **Truth Table**<sup>[2, 3, 4, 5, 6, 7]</sup>

Operation	K	LD	R/W	DQ	DQ	DQ	DQ
Write Cycle: Load address; wait one cycle; input write data on 2 consecutive K and K rising edges.	L-H	L	L <sup>[8]</sup>	D(A1)at K(t+1)↑	<u>D</u> (A2) at K(t+1)↑	D(A3) at K(t+2) ↑	<u>D</u> (A4) at K(t+2) ↑
Read Cycle: Load address; wait one cycle; read data on 2 consecutive C and C rising edges.		L	H <sup>[9]</sup>	Q(A1) at C(t+1)↑	Q(A2) at C(t+1) ↑	Q(A3) at C(t+2)↑	<u>Q</u> (A4) at C(t+2) ↑
NOP: No Operation	L-H	Н	Х	High-Z	High-Z	High-Z)	High-Z
Standby: Clock Stopped	Stopped	Χ	Х	Previous State	Previous State	Previous State	Previous State

#### **Linear Burst Address Table**

First Address (External)	Second Address (Internal)	Third Address (Internal)	Fourth Address (Internal)
XX00	XX01	XX10	XX11
XX01	XX10	XX11	XX00
XX10	XX11	XX00	XX01
XX11	XX00	XX01	XX10

- 1. The above application shows 2 DDR-I being used.
- 2. X = "Don't Care", H = Logic HIGH, L = Logic LOW, ↑represents rising edge.
- 3. Device will power-up deselected and the outputs in a three-state condition.
- 4. "A1" represents address location latched by the devices when transaction was initiated. A2, A3 and A4 represents the internal address sequence in the burst.

- 5. "t" represents the cycle at which a Read/Write operation is started. t+1 and t+2 are the first and second clock cycles succeeding the "t" clock cycle.

  6. Data inputs are registered at K and K rising edges. Data outputs are delivered on C and C rising edges, except when in single clock mode.

  7. It is recommended that K = K and C = C when clock is stopped. This is not essential, but permits most rapid restart by overcoming transmission line charging
- 8. This signal was HIGH on previous K clock rise. Initiating consecutive Write operations on consecutive K clock rises is not permitted. The device will ignore the
- This signal was LOW on previous K clock rise. Initiating consecutive Read operations on consecutive K clock rises is not permitted. The device will ignore the second Read request.



## Write Cycle Descriptions<sup>[2,10]</sup>

BWS <sub>0</sub>	BWS <sub>1</sub>	BWS <sub>2</sub>	BWS <sub>3</sub>	K	K	Comments	
L	L	L	L	L-H	1	During the Data portion of a Write sequence, all four bytes ( $D_{[35:0]}$ ) are written into the device.	
L	L	L	L	-	Ļ	During the Data portion of a Write sequence, all four bytes ( $D_{[35:0]}$ ) are written into the device.	
L	Н	Η	Ι	L-H	1	During the Data portion of a Write sequence, only the lower byte $(D_{[8:0]})$ is written into the device. $D_{[35:9]}$ will remain unaltered.	
L	Н	Η	Η	-	L-H	During the Data portion of a Write sequence, only the lower byte $(D_{[8:0]})$ is written into the device. $D_{[35:9]}$ will remain unaltered.	
Н	L	Η	Η	L-H	•	During the Data portion of a Write sequence, only the byte $(D_{[17:9]})$ is written into the device. $D_{[8:0]}$ and $D_{[35:18]}$ will remain unaltered.	
Н	L	Η	Н	-	L-H	During the Data portion of a Write sequence, only the byte $(D_{[17:9]})$ is written into the device. $D_{[8:0]}$ and $D_{[35:18]}$ will remain unaltered.	
Н	Н	L	Η	L-H	1	During the Data portion of a Write sequence, only the byte $(D_{[26:18]})$ is written into the device. $D_{[17:0]}$ and $D_{[35:27]}$ will remain unaltered.	
Н	Н	L	Ι	-	Ļ	During the Data portion of a Write sequence, only the byte $(D_{[26:18]})$ is written into the device. $D_{[17:0]}$ and $D_{[35:27]}$ will remain unaltered.	
Н	Н	Η	L	L-H		During the Data portion of a Write sequence, only the byte $(D_{[35:27]})$ is written into the device. $D_{[26:0]}$ will remain unaltered.	
Н	Н	Η	L	-	L-H	During the Data portion of a Write sequence, only the byte $(D_{[35:27]})$ is written int he device. $D_{[26:0]}$ will remain unaltered.	
Н	Н	Η	Η	L-H	-	No data is written into the device during this portion of a Write operation.	
Н	Н	Н	Н	-	L-H	No data is written into the device during this portion of a Write operation.	

#### Note:

<sup>10.</sup> Assumes a Write cycle was initiated per the Write Port Cycle Description Truth Table. BWS<sub>0</sub>, BWS<sub>1</sub>, BWS<sub>2</sub>, BWS<sub>3</sub> can be altered on different portions of a Write cycle, as long as the set-up and hold requirements are achieved.



#### IEEE 1149.1 Serial Boundary Scan (JTAG)

These SRAMs incorporate a serial boundary scan test access port (TAP) in the FBGA package. This part is fully compliant with IEEE Standard #1149.1-1900. The TAP operates using JEDEC standard 2.5V I/O logic levels.

#### **Disabling the JTAG Feature**

It is possible to operate the SRAM without using the JTAG feature. To disable the TAP controller, TCK must be tied LOW (V\_SS) to prevent clocking of the device. TDI and TMS are internally pulled up and may be unconnected. They may alternately be connected to  $V_{DD}$  through a pull-up resistor. TDO should be left unconnected. Upon power-up, the device will come up in a reset state which will not interfere with the operation of the device.

#### Test Access Port—Test Clock

The test clock is used only with the TAP controller. All inputs are captured on the rising edge of TCK. All outputs are driven from the falling edge of TCK.

#### **Test Mode Select**

The TMS input is used to give commands to the TAP controller and is sampled on the rising edge of TCK. It is allowable to leave this pin unconnected if the TAP is not used. The pin is pulled up internally, resulting in a logic HIGH level.

#### Test Data-In (TDI)

The TDI pin is used to serially input information into the registers and can be connected to the input of any of the registers. The register between TDI and TDO is chosen by the instruction that is loaded into the TAP instruction register. For information on loading the instruction register, see the TAP Controller State Diagram. TDI is internally pulled up and can be unconnected if the TAP is unused in an application. TDI is connected to the most significant bit (MSB) on any register.

#### Test Data-Out (TDO)

The TDO output pin is used to serially clock data-out from the registers. The output is active depending upon the current state of the TAP state machine (see Instruction codes). The output changes on the falling edge of TCK. TDO is connected to the least significant bit (LSB) of any register.

#### Performing a TAP Reset

A Reset is performed by forcing TMS HIGH ( $V_{DD}$ ) for five rising edges of TCK. This RESET does not affect the operation of the SRAM and may be performed while the SRAM is operating. At power-up, the TAP is reset internally to ensure that TDO comes up in a high-Z state.

#### **TAP Registers**

Registers are connected between the TDI and TDO pins and allow data to be scanned into and out of the SRAM test circuitry. Only one register can be selected at a time through the instruction registers. Data is serially loaded into the TDI pin on the rising edge of TCK. Data is output on the TDO pin on the falling edge of TCK.

#### Instruction Register

Three-bit instructions can be serially loaded into the instruction register. This register is loaded when it is placed between the

TDI and TDO pins as shown in TAP Controller Block Diagram. Upon power-up, the instruction register is loaded with the IDCODE instruction. It is also loaded with the IDCODE instruction if the controller is placed in a reset state as described in the previous section.

When the TAP controller is in the Capture IR state, the two least significant bits are loaded with a binary "01" pattern to allow for fault isolation of the board level serial test path.

#### Bypass Register

To save time when serially shifting data through registers, it is sometimes advantageous to skip certain chips. The bypass register is a single-bit register that can be placed between TDI and TDO pins. This allows data to be shifted through the SRAM with minimal delay. The bypass register is set LOW  $(V_{SS})$  when the BYPASS instruction is executed.

#### Boundary Scan Register

The boundary scan register is connected to all of the input and output pins on the SRAM. Several no connect (NC) pins are also included in the scan register to reserve pins for higher density devices.

The boundary scan register is loaded with the contents of the RAM Input and Output ring when the TAP controller is in the Capture-DR state and is then placed between the TDI and TDO pins when the controller is moved to the Shift-DR state. The EXTEST, SAMPLE/PRELOAD and SAMPLE Z instructions can be used to capture the contents of the Input and Output ring.

The Boundary Scan Order tables show the order in which the bits are connected. Each bit corresponds to one of the bumps on the SRAM package. The MSB of the register is connected to TDI, and the LSB is connected to TDO.

#### Identification (ID) Register

The ID register is loaded with a vendor-specific, 32-bit code during the Capture-DR state when the IDCODE command is loaded in the instruction register. The IDCODE is hardwired into the SRAM and can be shifted out when the TAP controller is in the Shift-DR state. The ID register has a vendor code and other information described in the Identification Register Definitions table.

#### **TAP Instruction Set**

Eight different instructions are possible with the three-bit instruction register. All combinations are listed in the Instruction Code table. Three of these instructions are listed as RESERVED and should not be used. The other five instructions are described in detail below.

Instructions are loaded into the TAP controller during the Shift-IR state when the instruction register is placed between TDI and TDO. During this state, instructions are shifted through the instruction register through the TDI and TDO pins. To execute the instruction once it is shifted in, the TAP controller needs to be moved into the Update-IR state.

#### **IDCODE**

The IDCODE instruction causes a vendor-specific, 32-bit code to be loaded into the instruction register. It also places the instruction register between the TDI and TDO pins and allows the IDCODE to be shifted out of the device when the TAP controller enters the Shift-DR state. The IDCODE instruction



is loaded into the instruction register upon power-up or whenever the TAP controller is given a test logic reset state.

#### SAMPLE Z

The SAMPLE Z instruction causes the boundary scan register to be connected between the TDI and TDO pins when the TAP controller is in a Shift-DR state. The SAMPLE Z command puts the output bus into a High-Z state until the next command is given during the "Update IR" state.

#### SAMPLE/PRELOAD

SAMPLE/PRELOAD is a 1149.1 mandatory instruction. When the SAMPLE/PRELOAD instructions are loaded into the instruction register and the TAP controller is in the Capture-DR state, a snapshot of data on the inputs and output pins is captured in the boundary scan register.

The user must be aware that the TAP controller clock can only operate at a frequency up to 10 MHz, while the SRAM clock operates more than an order of magnitude faster. Because there is a large difference in the clock frequencies, it is possible that during the Capture-DR state, an input or output will undergo a transition. The TAP may then try to capture a signal while in transition (metastable state). This will not harm the device, but there is no guarantee as to the value that will be captured. Repeatable results may not be possible.

To guarantee that the boundary scan register will capture the correct value of a signal, the SRAM signal must be stabilized long enough to meet the TAP controller's capture set-up plus hold times ( $t_{CS}$  and  $t_{CH}$ ). The SRAM clock input might not be captured correctly if there is no way in a design to stop (or slow) the clock during a SAMPLE/PRELOAD instruction. If this is an issue, it is still possible to capture <u>all other signals and simply ignore the value of the CK and CK captured in the boundary scan register.</u>

Once the data is captured, it is possible to shift out the data by putting the TAP into the Shift-DR state. This places the boundary scan register between the TDI and TDO pins.

PRELOAD allows an initial data pattern to be placed at the latched parallel outputs of the boundary scan register cells prior to the selection of another boundary scan test operation.

The shifting of data for the SAMPLE and PRELOAD phases can occur concurrently when required—that is, while data captured is shifted out, the preloaded data can be shifted in.

#### **BYPASS**

When the BYPASS instruction is loaded in the instruction register and the TAP is placed in a Shift-DR state, the bypass register is placed between the TDI and TDO pins. The advantage of the BYPASS instruction is that it shortens the boundary scan path when multiple devices are connected together on a board.

#### **EXTEST**

The EXTEST instruction enables the preloaded data to be driven out through the system output pins. This instruction also selects the boundary scan register to be connected for serial access between the TDI and TDO in the shift-DR controller state.

#### EXTEST Output Bus Tri-state

IEEE Standard 1149.1 mandates that the TAP controller be able to put the output bus into a tri-state mode.

The boundary scan register has a special bit located at bit #47. When this scan cell, called the "extest output bus tri-state", is latched into the preload register during the "Update-DR" state in the TAP controller, it will directly control the state of the output (Q-bus) pins, when the EXTEST is entered as the current instruction. When HIGH, it will enable the output buffers to drive the output bus. When LOW, this bit will place the output bus into a High-Z condition.

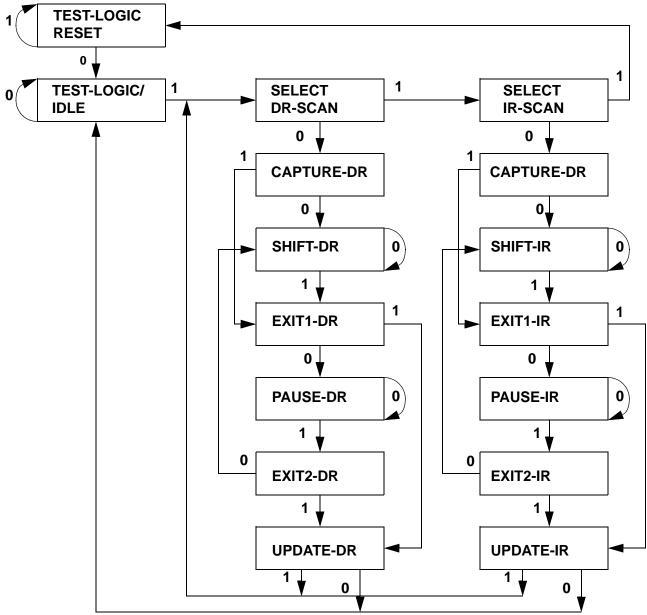
This bit can be set by entering the SAMPLE/PRELOAD or EXTEST command, and then shifting the desired bit into that cell, during the "Shift-DR" state. During "Update-DR", the value loaded into that shift-register cell will latch into the preload register. When the EXTEST instruction is entered, this bit will directly control the output Q-bus pins. Note that this bit is pre-set HIGH to enable the output when the device is powered-up, and also when the TAP controller is in the "Test-Logic-Reset" state.

#### Reserved

These instructions are not implemented but are reserved for future use. Do not use these instructions.



### TAP Controller State Diagram<sup>[11]</sup>

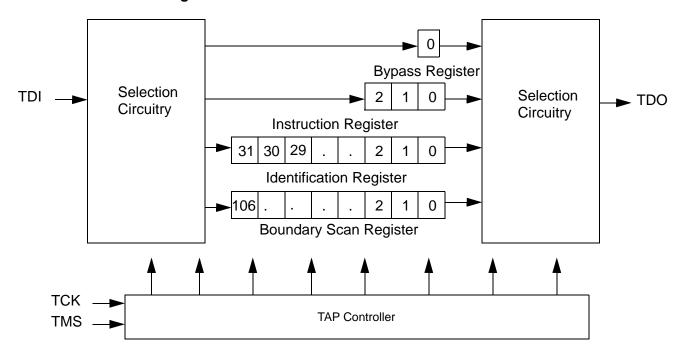


Note:

11. The 0/1 next to each state represents the value at TMS at the rising edge of TCK.



### **TAP Controller Block Diagram**



TAP Electrical Characteristics Over the Operating Range [12, 15, 17]

Parameter	Description	Test Conditions	Min.	Max.	Unit
V <sub>OH1</sub>	Output HIGH Voltage	$I_{OH} = -2.0 \text{ mA}$	1.7		V
V <sub>OH2</sub>	Output HIGH Voltage	$I_{OH} = -100 \mu A$	2.1		V
V <sub>OL1</sub>	Output LOW Voltage	I <sub>OL</sub> = 2.0 mA		0.7	V
V <sub>OL2</sub>	Output LOW Voltage	I <sub>OL</sub> = 100 μA		0.2	V
V <sub>IH</sub>	Input HIGH Voltage		1.7	V <sub>DD</sub> + 0.3	V
V <sub>IL</sub>	Input LOW Voltage		-0.3	0.7	V
I <sub>X</sub>	Input and Output Load Current	$GND \leq V_I \leq V_DDQ$	-5	5	μΑ

TAP AC Switching Characteristics Over the Operating Range<sup>[13, 14]</sup>

Parameter	Description	Min.	Max.	Unit
t <sub>TCYC</sub>	TCK Clock Cycle Time	50		ns
t <sub>TF</sub>	TCK Clock Frequency		20	MHz
t <sub>TH</sub>	TCK Clock HIGH	20		ns
t <sub>TL</sub>	TCK Clock LOW	20		ns
Set-up Times		<u> </u>		
t <sub>TMSS</sub>	TMS Set-up to TCK Clock Rise	10		ns
t <sub>TDIS</sub>	TDI Set-up to TCK Clock Rise	10		ns
t <sub>CS</sub>	Capture Set-up to TCK Rise	10		ns
Hold Times		<u> </u>		
t <sub>TMSH</sub>	TMS Hold after TCK Clock Rise	10		ns
t <sub>TDIH</sub>	TDI Hold after Clock Rise	10		ns
t <sub>CH</sub>	Capture Hold after Clock Rise	10		ns

<sup>12.</sup> These characteristics pertain to the TAP inputs (TMS, TCK, TDI and TDO). Parallel load levels are specified in the Electrical Characteristics table.

<sup>13.</sup> t<sub>CS</sub> and t<sub>CH</sub> refer to the set-up and hold time requirements of latching data from the boundary scan register.

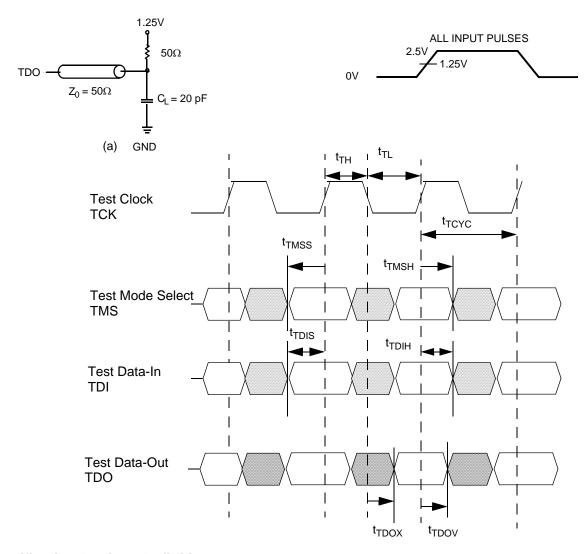
14. Test conditions are specified using the load in TAP AC test conditions. t<sub>R</sub>/t<sub>F</sub> = 1 ns.



## TAP AC Switching Characteristics Over the Operating Range $^{[13,\ 14]}$

Parameter	Description	Min.	Max.	Unit
Output Times				
t <sub>TDOV</sub>	TCK Clock LOW to TDO Valid		20	ns
t <sub>TDOX</sub>	TCK Clock LOW to TDO Invalid	0		ns

### TAP Timing and Test Conditions<sup>[14]</sup>



### **Identification Register Definitions**

	Value	
Instruction Field	CY7C1308DV25	Description
Revision Number (31:29)	000	Version number.
Cypress Device ID (28:12)	01011111011100110	Defines the type of SRAM.
Cypress JEDEC ID (11:1)	00000110100	Allows unique identification of SRAM vendor.
ID Register Presence (0)	1	Indicate the presence of an ID register.



### **Scan Register Sizes**

Register Name	Bit Size
Instruction	3
Bypass	1
ID	32
Boundary Scan	107

### **Instruction Codes**

Instruction	Code	Description
EXTEST	000	Captures the Input/Output ring contents.
IDCODE	001	Loads the ID register with the vendor ID code and places the register between TDI and TDO. This operation does not affect SRAM operation.
SAMPLE Z	010	Captures the Input/Output contents. Places the boundary scan register between TDI and TDO. Forces all SRAM output drivers to a High-Z state.
RESERVED	011	Do Not Use: This instruction is reserved for future use.
SAMPLE/PRELOAD	100	Captures the Input/Output ring contents. Places the boundary scan register between TDI and TDO. Does not affect the SRAM operation.
RESERVED	101	Do Not Use: This instruction is reserved for future use.
RESERVED	110	Do Not Use: This instruction is reserved for future use.
BYPASS	111	Places the bypass register between TDI and TDO. This operation does not affect SRAM operation.



### **Boundary Scan Order**

Bouridary C	ocali Oluei
Bit #	Bump ID
0	6R
1	6P
2	6N
3	7P
4	7N
5	7R
6	8R
7	8P
8	9R
9	11P
10	10P
11	10N
12	9P
13	10M
14	11N
15	9M
16	9N
17	11L
18	11M
19	9L
20	10L
21	11K
22	10K
23	9J
24	9K
25	10J
26	11J

Bit #	Bump ID
27	11H
28	10G
29	9G
30	11F
31	11G
32	9F
33	10F
34	11E
35	10E
36	10D
37	9E
38	10C
39	11D
40	9C
41	9D
42	11B
43	11C
44	9B
45	10B
46	11A
47	Internal
48	9A
49	8B
50	7C
51	6C
52	8A
53	7A

Bit #	Bump ID		
54	7B		
55	6B		
56	6A		
57	5B		
58	5A		
59	4A		
60	5C		
61	4B		
62	3A		
63	1H		
64	1A		
65	2B		
66	3B		
67	1C		
68	1B		
69	3D		
70	3C		
71	1D		
72	2C		
73	3E		
74	2D		
75	2E		
76	1E		
77	2F		
78	3F		
79	1G		
80	1F		

Bit #	Bump ID
81	3G
82	2G
83	1J
84	2J
85	3K
86	3J
87	2K
88	1K
89	2L
90	3L
91	1M
92	1L
93	3N
94	3M
95	1N
96	2M
97	3P
98	2N
99	2P
100	1P
101	3R
102	4R
103	4P
104	5P
105	5N
106	5R



### **Maximum Ratings**

(Above which the useful life may be impaired.) Storage Temperature ......-65°C to + 150°C Ambient Temperature with Power Applied.....-55°C to + 125°C Supply Voltage on  $V_{DD}$  Relative to GND......-0.5V to + 3.6V Supply Voltage on V<sub>DDO</sub> Relative to GND .....-0.5V to + V<sub>DD</sub> DC Applied to Outputs in High-Z.....-0.5V to V<sub>DDQ</sub> + 0.5V DC Input Voltage<sup>[15]</sup>.....-0.5V to  $V_{DD}$  + 0.5V

Current into Outputs (LOW)	20 mA
Static Discharge Voltage(per MIL-STD-883, Method 3015)	>2 001V
Latch-up Current	> 200 mA

### **Operating Range**

Range	Ambient Temperature (T <sub>A</sub> )	<b>V</b> <sub>DD</sub> <sup>[16]</sup>	<b>V</b> DDQ <sup>[16]</sup>
Com'l	0°C to + 70°C	2.5 ± 0.1V	1.4V to 1.9V
Ind'I	-40°C to + 85°C		

### Electrical Characteristics Over the Operating Range [17]

#### **DC Electrical Characteristics**

Parameter	Description	Test Conditions	Min.	Тур.	Max.	Unit
$V_{DD}$	Power Supply Voltage		2.4	2.5	2.6	V
$V_{DDQ}$	I/O Supply Voltage		1.4	1.5	1.9	V
V <sub>OH</sub>	Output HIGH Voltage	Note 19	V <sub>DDQ</sub> /2 – 0.12		$V_{DDQ}/2 + 0.12$	V
V <sub>OL</sub>	Output LOW Voltage	Note 20	V <sub>DDQ</sub> /2 – 0.12		$V_{DDQ}/2 + 0.12$	V
V <sub>OH(LOW)</sub>	Output HIGH Voltage	$I_{OH} = -0.1$ mA, Nominal Impedance	V <sub>DDQ</sub> - 0.2		$V_{DDQ}$	V
V <sub>OL(LOW)</sub>	Output LOW Voltage	I <sub>OL</sub> = 0.1 mA, Nominal Impedance	V <sub>SS</sub>		0.2	V
V <sub>IH</sub>	Input HIGH Voltage <sup>[15]</sup>		V <sub>REF</sub> + 0.1		V <sub>DDQ</sub> + 0.3	V
V <sub>IL</sub>	Input LOW Voltage <sup>[15,18]</sup>		-0.3		V <sub>REF</sub> - 0.1	V
I <sub>X</sub>	Input Leakage Current	$GND \le V_I \le V_{DDQ}$	<b>-</b> 5		5	μΑ
l <sub>OZ</sub>	Output Leakage Current	$GND \le V_I \le V_{DDQ}$ , Output Disabled	<b>-</b> 5		5	μΑ
$V_{REF}$	Input Reference Voltage <sup>[21]</sup>	Typical Value = 0.75V	0.68	0.75	0.95	V
I <sub>DD</sub>	V <sub>DD</sub> Operating Supply	$V_{DD} = Max., I_{OUT} = 0 \text{ mA},$ $f = f_{MAX} = 1/t_{CYC}$			700	mA
I <sub>SB1</sub>	Automatic Power-Down	$\begin{aligned} &\text{Max. V}_{DD},  \text{Both Ports} \\ &\text{Deselected, V}_{IN} \geq V_{IH}  \text{or V}_{IN} \leq V_{IL} \\ &\text{f = f}_{MAX} = 1/t_{CYC,}  \text{Inputs Static} \end{aligned}$			450	mA

#### **AC Input Requirements**

Parameter	Description	Test Conditions	Min.	Тур.	Max.	Unit
$V_{IH}$	Input HIGH Voltage		V <sub>REF</sub> + 0.2	_	_	V
$V_{IL}$	Input LOW Voltage		_	1	V <sub>REF</sub> – 0.2	V

#### Thermal Resistance<sup>[22]</sup>

Parameter	Description	Test Conditions	165 FBGA Package	Unit
$\Theta_{JA}$	Thermal Resistance (Junction to Ambient)		16.7	°C/W
Θ <sub>JC</sub>	i i neimai kesisiance Gunciion io Casei	methods and procedures for measuring thermal impedance, per EIA/JESD51.	2.5	°C/W

- 15. Overshoot:  $V_{IH}(AC) < V_{DDQ} + 0.85V$  (Pulse width less than  $t_{CYC/2}$ ). Undershoot:  $V_{IL}(AC) > -1.5V$  (Pulse width less than  $t_{CYC/2}$ ). 16. Power-up: Assumes a linear ramp from 0V to  $V_{DD}(min.)$  within 200 ms. During this time  $V_{IH} < V_{DD}$  and  $V_{DDQ} \le V_{DD}$ .
- 17. All voltage referenced to ground.

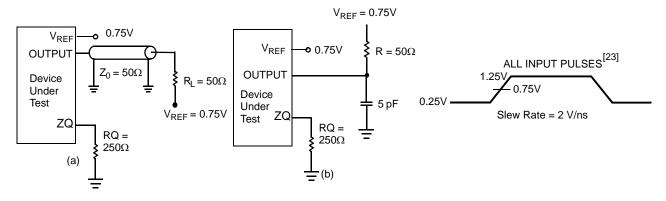
- 17. All voltage referenced to ground.
  18. This spec is for all inputs except C and C Clock. For C and C Clock, V<sub>IL</sub>(Max.) = V<sub>REF</sub> 0.2V.
  19. Output are impedance controlled. I<sub>OH</sub> = -(V<sub>DDQ</sub>/2)/(RQ/5) for values of 175Ω <= RQ <= 350Ω.</li>
  20. Output are impedance controlled. I<sub>OL</sub> = (V<sub>DDQ</sub>/2)/(RQ/5) for values of 175Ω <= RQ <= 350Ω.</li>
  21. V<sub>REF</sub> (Min.) = 0.68V or 0.46V<sub>DDQ</sub>, whichever is larger, V<sub>REF</sub> (Max.) = 0.95V or 0.54V<sub>DDQ</sub>, whichever is smaller.
  22. Tested initially and after any design or process change that may affect these parameters.



### Capacitance<sup>[22]</sup>

Parameter	Description	Test Conditions	Max.	Unit
C <sub>IN</sub>	Input Capacitance	$T_A = 25^{\circ}C, f = 1 \text{ MHz},$	5	pF
C <sub>CLK</sub>	Clock Input Capacitance	V <sub>DD</sub> = 2.5V V <sub>DDQ</sub> = 1.5V	6	pF
Co	Output Capacitance	1 DDQ = 1.0 V	7	pF

#### **AC Test Loads and Waveforms**



### Switching Characteristics Over the Operating Range [23]

Cypress	Consortium			167 MHz	
Parameter	Parameter	Description	Min.	Max.	Unit
t <sub>Power</sub> <sup>[24]</sup>		V <sub>CC</sub> (typical) to the First Access Read or Write	10		μS
Cycle Time	•				
t <sub>CYC</sub>	t <sub>KHKH</sub>	K Clock and C Clock Cycle Time	6.0		ns
t <sub>KH</sub>	t <sub>KHKL</sub>	Input Clock (K/K and C/C) HIGH	2.4		ns
t <sub>KL</sub>	t <sub>KLKH</sub>	Input Clock (K/K and C/C) LOW	2.4		ns
<sup>t</sup> ĸн <del>к</del> н	t <sub>KHK</sub> H	K/K Clock Rise to K/K Clock Rise and C/C to C/C Rise (rising edge to rising edge)	2.8	3.2	ns
t <sub>KHCH</sub>	t <sub>KHCH</sub>	K/K Clock Rise to C/C Clock Rise (rising edge to rising edge)	0.0	2.0	ns
Set-up Time	es				
t <sub>SA</sub>	t <sub>SA</sub>	Address Set-up to Clock (K and K) Rise	0.7		ns
t <sub>SC</sub>	t <sub>SC</sub>	Control Set-up to Clock (K and K) Rise (RPS, WPS, BWS <sub>0</sub> , BWS <sub>1</sub> )	0.7		ns
t <sub>SD</sub>	t <sub>SD</sub>	$D_{[35:0]}$ Set-up to Clock (K and $\overline{K}$ ) Rise	0.7		ns
<b>Hold Times</b>				•	•
t <sub>HA</sub>	t <sub>HA</sub>	Address Hold after Clock (K and $\overline{K}$ ) Rise	0.7		ns
t <sub>HC</sub>	t <sub>HC</sub>	$\frac{Control\ Signals\ Hold\ after\ Clock\ (K\ and\ \overline{K})\ Rise\ (\overline{RPS},\overline{WPS},\overline{BWS}_0,}{BWS_1)}$	0.7		ns
t <sub>HD</sub>	t <sub>HD</sub>	D <sub>[35:0]</sub> Hold after Clock (K and $\overline{K}$ ) Rise	0.7		ns

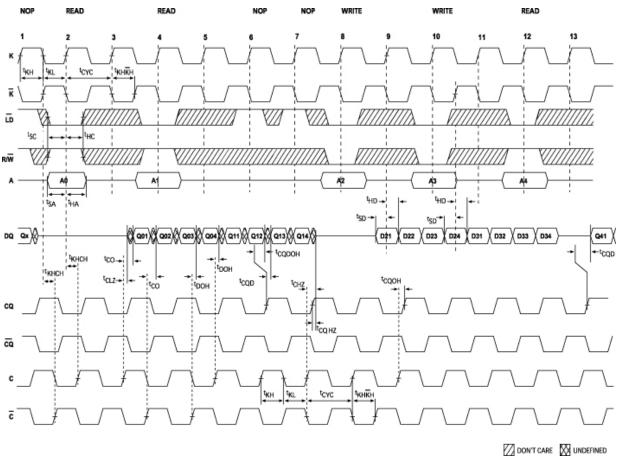
 <sup>23.</sup> Unless otherwise noted, test conditions assume signal transition time of 2V/ns, timing reference levels of 0.75V, Vref = 0.75V, RQ = 250Ω, V<sub>DDQ</sub> = 1.5V, input pulse levels of 0.25V to 1.25V, and output loading of the specified l<sub>QL</sub>/l<sub>QH</sub> and load capacitance shown in (a) of AC test loads.
 24. This part has a voltage regulator that steps down the voltage internally; t<sub>Power</sub> is the time power needs to be supplied above V<sub>DD</sub> minimum initially before a Read or Write operation can be initiated.



### Switching Characteristics Over the Operating Range (continued)<sup>[23]</sup>

Cypress	Consortium		167 MHz		
Parameter		Description	Min.	Max.	Unit
Output Tim	es				
t <sub>co</sub>	t <sub>CHQV</sub>	C/C Clock Rise (or K/K in single clock mode) to Data Valid		3.0	ns
t <sub>DOH</sub>	t <sub>CHQX</sub>	Data Output Hold after Output C/C Clock Rise (Active to Active)	0.8		ns
t <sub>CHZ</sub>	t <sub>CHZ</sub>	Clock (C and $\overline{C}$ ) Rise to High-Z (Active to High-Z) <sup>[25, 26]</sup>		3.0	ns
t <sub>CLZ</sub>	t <sub>CLZ</sub>	Clock (C and $\overline{C}$ ) Rise to Low-Z <sup>[25, 26]</sup>	0.8		ns
tccqo	t <sub>CHCQV</sub>	C/C Clock Rise to Echo Clock Valid	0.8	3.2	ns
t <sub>CQD</sub>	t <sub>CQHQV</sub>	Echo Clock High to Data Valid		0.40	ns
t <sub>CQDOH</sub>	t <sub>CQHQX</sub>	Echo Clock High to Data Invalid	-0.40		ns
t <sub>CQHZ</sub>	t <sub>CHZ</sub>	Clock (CQ and CQ) Rise to High-Z (Active to High-Z)[25, 26]		0.40	ns
t <sub>CQLZ</sub>	t <sub>CLZ</sub>	Clock (CQ and CQ) Rise to Low-Z[25, 26]	-0.40		ns

### Switching Waveforms<sup>[27, 28, 29]</sup>



#### Notes:

- 25. t<sub>CHZ</sub>, t<sub>CLZ</sub>, are specified with a load capacitance of 5 pF as in part (b) of AC Test Loads. Transition is measured ± 100 mV from steady-state voltage.
- 26. At any given voltage and temperature t<sub>CHZ</sub> is less than t<sub>CLZ</sub> and, t<sub>CHZ</sub> less than t<sub>CC</sub>.

  27. Q01 refers to output from address A0. Q02 refers to output from the next internal burst address following A0, i.e., A0+1.
- 28. Outputs are disabled (High-Z) one clock cycle after a NOP.
- 29. In this example, if address A4 = A3, then data Q41 = D31, Q42 = D32, Q43 = D33, and Q44 = D34. Write data is forwarded immediately as Read results. This note applies to the whole diagram.



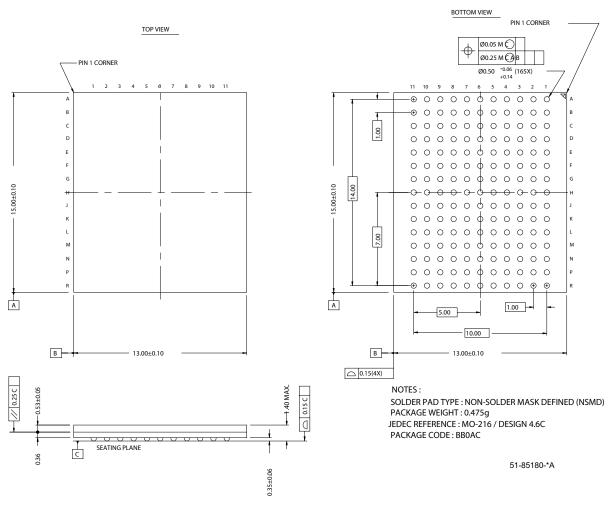
#### **Ordering Information**

"Not all of the speed, package and temperature ranges are available. Please contact your local sales representative or visit <a href="https://www.cypress.com">www.cypress.com</a> for actual products offered".

Speed (MHz)	Ordering Code	Package Diagram	Package Type	Operating Range
167	CY7C1308DV25-167BZC	51-85180	165-ball Fine Pitch Ball Grid Array (13 x 15 x 1.4 mm)	Commercial
	CY7C1308DV25-167BZXC		165-ball Fine Pitch Ball Grid Array (13 x 15 x 1.4 mm) Lead free	
	CY7C1308DV25-167BZI		165-ball Fine Pitch Ball Grid Array (13 x 15 x 1.4 mm)	Industrial
	CY7C1308DV25-167BZXI		165-ball Fine Pitch Ball Grid Array (13 x 15 x 1.4 mm) Lead free	

#### **Package Diagram**

#### 165-ball FBGA (13 x 15 x 1.4 mm) (51-85180)



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## **Document History Page**

Rev.	ECN No.	Issue Date	Orig. of Change	Description of Change
**	253050	See ECN	SYT	New Data Sheet
*A	436882	See ECN	NXR	Converted from Preliminary to Final. Removed 133 MHz & 100 MHz from product offering. Included Industrial Operating Range. Changed C/C Description in the Features Section & Pin Description Table. Changed $t_{TCYC}$ from 100 ns to 50 ns, changed $t_{TF}$ from 10 MHz to 20 MHz a changed $t_{TH}$ and $t_{TL}$ from 40 ns to 20 ns in TAP AC Switching Characteristics table Modified the ZQ pin definition as follows: Alternately, this pin can be connected directly to $V_{DDQ}$ , which enables the minimum impedance mode Included Maximum Ratings for Supply Voltage on $V_{DDQ}$ Relative to GND Changed the Maximum Ratings for DC Input Voltage from $V_{DDQ}$ to $V_{DD}$ . Modified the Description of $I_X$ from Input Load current to Input Leakage Curre on page # 14. Modified test condition in note# 16 from $V_{DDQ} < V_{DD}$ to $V_{DDQ} \le V_{DD}$ Replaced the Package Name Column with Package Diagram.