

TH72036 868/915MHz FSK/ASK Transmitter

Features

Fully integrated PLL-stabilized VCO	Microcontroller clock output
Frequency range from 850 MHz to 930 MHz	On-chip low voltage detector
Single-ended RF output	High over-all frequency accuracy
FSK through crystal pulling allows modulation	FSK deviation and center frequency
from DC to 40 kbit/s	independently adjustable
High FSK deviation possible for wideband data	Adjustable output power range from
transmission	-11 dBm to +9.5 dBm
ASK achieved by on/off keying of internal	Adjustable current consumption from
power amplifier up to 40 kbit/s	5.5 mA to 13.8 mA
Wide power supply range from 1.95 V to 5.5 V	Conforms to EN 300 220 and similar standards
Very low standby current	10-pin Quad Flat No-Lead Package (QFN)

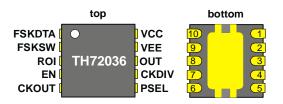
Ordering Information

Part Number	Temperature Code	Package Code	Delivery Form
TH72036	K (-40 °C to 125 °C)	LD (10L QFN 3x3 Dual)	120 pc/tray 5000 pc/T&R

Application Examples

□ General digital data transmission □ Tire Pressure Monitoring Systems (TPMS) □ Remote Keyless Entry (RKE) □ Wireless access control □ Alarm and security systems □ Garage door openers □ Remote Controls □ Home and building automation

Pin Description



General Description

■ Low-power telemetry systems

The TH72036 FSK/ASK transmitter IC is designed for applications in the European 868 MHz industrial-scientific-medical (ISM) band, according to the EN 300 220 telecommunications standard. It can also be used for any other system with carrier frequencies ranging from 850 MHz to 930 MHz (e.g. for applications in the US 902 to 928 MHz ISM band).

The transmitter's carrier frequency f_c is determined by the frequency of the reference crystal f_{ref} . The integrated PLL synthesizer ensures that carrier frequencies, ranging from 850 MHz to 930 MHz, can be achieved. This is done by using a crystal with a reference frequency according to: $f_{ref} = f_c/N$, where N = 32 is the PLL feedback divider ratio.

A clock signal with selectable frequency is provided. It can be used to drive a microcontroller.





Document Content

1	The	eory of Operation	3
	1.1	General	3
	1.2	Block Diagram	3
2	Fu	nctional Description	3
	2.1	Crystal Oscillator	3
	2.2	FSK Modulation	4
	2.3	Crystal Pulling	4
	2.4	ASK Modulation	
	2.5	Output Power Selection	
	2.6	Lock Detection	
	2.7	Low Voltage Detection	
	2.8	Mode Control Logic	
	2.9	Clock Output	
	2.10	Timing Diagrams	
3	Pin	Definition and Description	7
4	Ele	ectrical Characteristics	8
	4.1	Absolute Maximum Ratings	8
	4.2	Normal Operating Conditions	8
	4.3	Crystal Parameters	8
	4.4	DC Characteristics	9
	4.5	AC Characteristics	10
	4.6	AC Characteristics	11
	4.7	Output Power Steps – ASK Mode	11
5	Tes	st Circuit	12
	5.1	Test circuit component list to Fig. 6	12
6	Pa	ckage Description	13
	6.1	Soldering Information	13
	6.2	Recommended PCB Footprints	13
7	Re	liability Information	14
8	ES	D Precautions	14
9	Dis	sclaimer	16



1 Theory of Operation

1.1 General

As depicted in Fig.1, the TH72036 transmitter consists of a fully integrated voltage-controlled oscillator (VCO), a divide-by-32 divider (div32), a phase-frequency detector (PFD) and a charge pump (CP). An internal loop filter determines the dynamic behavior of the PLL and suppresses reference spurious signals. A Colpitts crystal oscillator (XOSC) is used as the reference oscillator of a phase-locked loop (PLL) synthesizer. The VCO's output signal feeds the power amplifier (PA). The RF signal power P_{out} can be adjusted in four steps from $P_{out} = -11$ dBm to +9.5 dBm, either by changing the value of resistor RPS or by varying the voltage V_{PS} at pin PSEL. The open-collector output (OUT) can be used either to directly drive a loop antenna or to be matched to a 500hm load. Bandgap biasing ensures stable operation of the IC at a power supply range of 1.95 V to 5.5 V.

1.2 Block Diagram

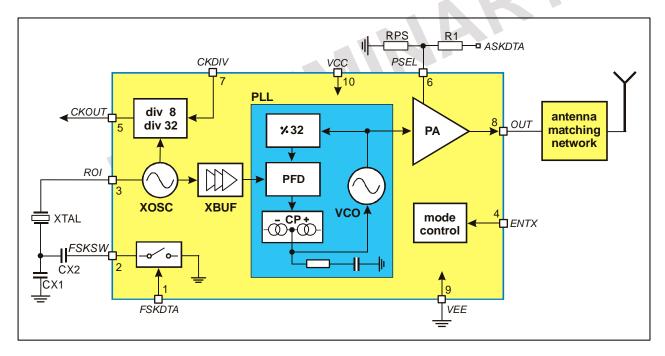


Fig. 1: Block diagram with external components

2 Functional Description

2.1 Crystal Oscillator

A Colpitts crystal oscillator with integrated functional capacitors is used as the reference oscillator for the PLL synthesizer. The equivalent input capacitance CRO offered by the crystal oscillator input pin ROI is about 18pF. The crystal oscillator is provided with an amplitude control loop in order to have a very stable frequency over the specified supply voltage and temperature range in combination with a short start-up time.

3901072036 Page 3 of 16 Data Sheet Rev. 005 June/07



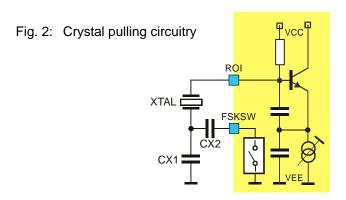
2.2 FSK Modulation

FSK modulation can be achieved by pulling the oscillator frequency. A CMOScompatible data stream applied at the pin FSKDTA digitally modulates the XOSC via an integrated NMOS switch. Two external pulling capacitors CX1 and CX2 allow the FSK deviation Δf and the center frequency f_c to be adjusted independently. At FSKDTA = 0, CX2 is connected in parallel to CX1 leading to the lowfrequency component of the FSK spectrum (f_{min}); while at FSKDTA = 1, CX2 is deactivated and the XOSC is set to its high frequency f_{max}. An external reference signal can be directly ACcoupled to the reference oscillator input pin ROI. Then the transmitter is used without a crystal. Now the reference signal sets the carrier frequency and may also contain the FSK (or FM) modulation.

2.3 Crystal Pulling

A crystal is tuned by the manufacturer to the required oscillation frequency f_0 at a given load capacitance CL and within the specified calibration tolerance. The only way to pull the oscillation frequency is to vary the effective load capacitance CL_{eff} seen by the crystal.

Figure 3 shows the oscillation frequency of a crystal as a function of the effective load capacitance. This capacitance changes in accordance with the logic level of FSKDTA around the specified load capacitance. The figure illustrates the relationship between the external pulling capacitors and the frequency deviation. It can also be seen that the pulling sensitivity increases with the reduction of CL. Therefore, applications with a high frequency deviation require a low load capacitance. For narrow band FSK applications, a higher load capacitance could be chosen in order to reduce the frequency drift caused by the tolerances of the chip and the external pulling capacitors.



FSKDTA	Description
0	f_{min} = f_c - Δf (FSK switch is closed)
1	$f_{max} = f_c + \Delta f$ (FSK switch is open)

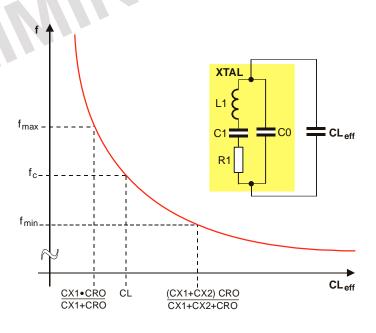


Fig. 3: Crystal pulling characteristic

For ASK applications CX2 can be omitted. Then CX1 has to be adjusted for center frequency.



2.4 ASK Modulation

The TH72036 can be ASK-modulated by applying data directly at pin PSEL. This turns the PA on and off which leads to an ASK signal at the output.

2.5 Output Power Selection

The transmitter is provided with an output power selection feature. There are four predefined output power steps and one off-step accessible via the power selection pin PSEL. A digital power step adjustment was chosen because of its high accuracy and stability. The number of steps and the step sizes as well as the corresponding power levels are selected to cover a wide spectrum of different applications.

The implementation of the output power control logic is shown in figure 4. There are two matched current sources with an amount of about 8 µA. One current source is directly applied to the PSEL pin. The other current source is used for the generation of reference voltages with a resistor ladder. These reference voltages are defining the thresholds between the power steps. The four comparators deliver thermometer-coded control signals depending on the voltage level at the pin PSEL. In order to have a certain amount of ripple tolerance in a noisy environment the comparators are provided with a little hysteresis of about 20 mV. With these control signals, weighted current sources of the power amplifier are switched on or off to set the desired output power level (Digitally Controlled Current Source). The LOCK, ASK signal and the output of the low voltage detector are gating this current source.

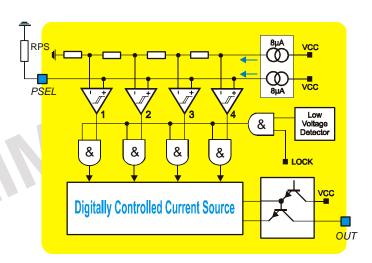


Fig. 4: Block diagram of output power control circuitry

There are two ways to select the desired output power step. First by applying a DC voltage at the pin PSEL, then this voltage directly selects the desired output power step. This kind of power selection can be used if the transmission power must be changed during operation. For a fixed-power application a resistor can be used which is connected from the PSEL pin to ground. The voltage drop across this resistor selects the desired output power level. For fixed-power applications at the highest power step this resistor can be omitted. The pin PSEL is in a high impedance state during the "TX standby" mode.

2.6 Lock Detection

The lock detection circuitry turns on the power amplifier only after PLL lock. This prevents from unwanted emission of the transmitter if the PLL is unlocked.

2.7 Low Voltage Detection

The supply voltage is sensed by a low voltage detect circuitry. The power amplifier is turned off if the supply voltage drops below a value of about 1.85 V. This is done in order to prevent unwanted emission of the transmitter if the supply voltage is too low.

3901072036 Page 5 of 16 Data Sheet Rev. 005 June/07



2.8 Mode Control Logic

The mode control logic allows two different modes of operation as listed in the following table. The mode control pin EN is pulled-down internally. This guarantees that the whole circuit is shut down if this pin is left floating.

EN	Mode	Description
0	TX standby	TX disabled
1	TX active CKOUT active	TX / CKOUT enabled

2.9 Clock Output

The clock output CKOUT is CMOS-compatible and can be used to drive a microcontroller. The frequency of the clock can be selected by the clock divider control signal CKDIV, according to the following table. A capacitor at pin CKOUT can be used to control the clock voltage swing and the RF spurious emission.

CKDIV	Clock divider ratio	Clock frequency / f _c =868.3MHz
0	8	3.39MHz
1	32	848kHz

2.10 Timing Diagrams

After enabling the transmitter by the EN signal, the power amplifier remains inactive for the time t_{on} , the transmitter start-up time. The crystal oscillator starts oscillation and the PLL locks to the desired output frequency within the time duration t_{on} . After successful PLL lock, the LOCK signal turns on the power amplifier, and then the RF carrier can be FSK or ASK modulated.

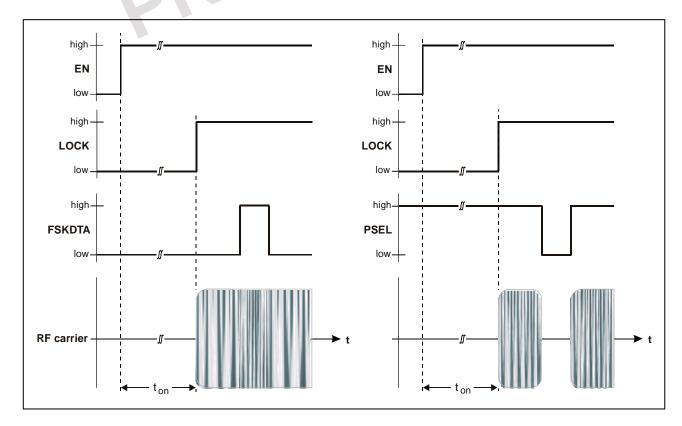


Fig. 5: Timing diagrams for FSK and ASK modulation

3901072036 Page 6 of 16 Data Sheet Rev. 005 June/07



3 Pin Definition and Description

Pin No.	Name	I/O Type	Functional Schematic	Description
1	FSKDTA	input	FSKDTA 1.5kΩ 1	FSK data input, CMOS compatible with in- ternal pull-up circuit TX standby: no pull-up TX active: pull-up
2	FSKSW	analog I/O	T VEE T VCC FSKSW 2 VEE T	XOSC FSK pulling pin, MOS switch
3	ROI	analog I/O	36p VCE TO	XOSC connection to XTAL, Colpitts type crystal oscillator
4	EN	input	EN 1.5kΩ VCC VCC VCC VEE	mode control input, CMOS-compatible with in- ternal pull-down circuit
5	CKOUT	output	CKOUT 400Ω VEE	clock output, CMOS-compatible
6	PSEL	analog I/O	PSEL 1.5kΩ BµA	power select input, high- impedance comparator logic TX standby: I _{PSEL} = 0 TX active: I _{PSEL} = 8µA
7	CKDIV	input	CKDIV 1.5kΩ 0: ENTX=0	clock divider control input, CMOS compatible with in- ternal pull-down circuit TX standby: no pull-down TX active: pull-down
8	OUT	output	OUT VEE VCC	power amplifier output, open collector
9	VEE	ground		negative power supply
10	VCC	supply		positive power supply



4 Electrical Characteristics

4.1 Absolute Maximum Ratings

Parameter	Symbol	Condition	Min	Max	Unit
Supply voltage	V _{cc}		0	7.0	V
Input voltage	V_{IN}		-0.3	V _{CC} +0.3	V
Storage temperature	T _{STG}		-65	150	°C
Junction temperature	TJ			150	°C
Thermal Resistance	R_{thJA}			49	K/W
Power dissipation	P _{diss}			0.12	W
Electrostatic discharge	V _{ESD}	human body model (HBM) according to CDF-AEC- Q100-002	±2.0		kV

4.2 Normal Operating Conditions

Parameter	Symbol	Condition	Min	Max	Unit
Supply voltage	V _{cc}		1.95	5.5	V
Operating temperature	T _A		-40	125	°C
Input low voltage CMOS	V_{IL}	EN, FSKDTA		0.3*V _{CC}	V
Input high voltage CMOS	V_{IH}	EN, FSKDTA	0.7*V _{CC}		V
XOSC frequency	f _{ref}	set by the crystal	26.6	29	MHz
VCO frequency	f _c	$f_c = 32 \bullet f_{ref}$	850	930	MHz
Clock frequency	f _{CLK}	CKDIV=0, f _{CLK} = f _{ref} / 8	3.3	3.6	MHz
		CKDIV=1, f _{CLK} = f _{ref} / 32	831	906	kHz
FSK deviation	Δf	depending on CX1, CX2 and crystal parameters	±2.5	±40	kHz
FSK Data rate	R	NRZ		40	kbit/s
ASK Data rate	R	NRZ		40	kbit/s

4.3 Crystal Parameters

Parameter	Symbol	Condition	Min	Max	Unit
Crystal frequency	f ₀	fundamental mode, AT	26.6	29	MHz
Load capacitance	CL		10	15	pF
Static capacitance	C ₀			7	pF
Series resistance	R ₁			50	Ω
Spurious response	a _{spur}	only required for FSK		-10	dB

3901072036 Page 8 of 16 Data Sheet Rev. 005 June/07



4.4 DC Characteristics

all parameters under normal operating conditions, unless otherwise stated; typical values at T_A = 23 $^{\circ}C$ and V_{CC} = 3 V

Parameter	Symbol	Condition	Min	Тур	Max	Unit
Operating Currents					<u> </u>	
Standby current	I _{SBY}	EN=0, T _A =85°C		0.2	200	nA
Starraby sarroin	1281	EN=0, T _A =125°C			4	μA
Supply current in power step 0	I _{CC0}	EN=1	2.5	4.3	6.5	mA
Supply current in power step 1	I _{CC1}	EN=1	3.5	5.5	8.1	mA
Supply current in power step 2	I _{CC2}	EN=1	4.5	6.8	9.6	mA
Supply current in power step 3	I _{CC3}	EN=1	6.2	9.0	12.2	mA
Supply current in power step 4	I _{CC4}	EN=1	9.4	13.8	18.1	mA
Digital Pin Characteristics					•	
Input low voltage CMOS	V _{IL}	EN, FSKDTA	-0.3		0.3*V _{cc}	V
Input high voltage CMOS	V _{IH}	EN, FSKDTA	0.7*V _{CC}		V _{CC} +0.3	V
Pull down current, EN	I _{PDEN}	EN=1	0.2	4.0	40	μA
Low level input current, EN	I _{INLEN}	EN=0			0.02	μA
High level input current, FSKDTA	I _{INHDTA}	FSKDTA=1			0.02	μA
Pull up current FSKDTA active mode	I _{PUDTAa}	FSKDTA=0, EN=1	0.1	1.5	12	μΑ
Pull up current FSK standby mode	I _{PUDTAs}	FSKDTA=0, EN=0			0.02	μΑ
Low level input current CKDIV	I _{INLCKDIV}	CKDIV=0			0.02	μA
Pull-down current CKDIV active mode	I _{PDCKDIVa}	CKDIV=1, EN=1	0.1	1.5	12	μΑ
Pull-down current CKDIV standby mode	I _{PDCKDIVs}	CKDIV=1, EN=0			0.02	μΑ
FSK Switch Resistance						
MOS switch On resistance	R _{ON}	FSKDTA=0, EN=1		20	70	Ω
MOS switch Off resistance	R _{OFF}	FSKDTA=1, EN=1	1			MΩ
Power Select Characteristics					·	
Power select current	I _{PSEL}	EN=1	7.0	8.6	9.9	μA
Power select voltage step 0	V _{PS0}	EN=1			0.035	V
Power select voltage step 1	V_{PS1}	EN=1	0.14		0.24	V
Power select voltage step 2	V_{PS2}	EN=1	0.37		0.60	V
Power select voltage step 3	V_{PS3}	EN=1	0.78		1.29	V
Power select voltage step 4	V_{PS4}	EN=1	1.55			V
Low Voltage Detection Charac	cteristic					
Low voltage detect threshold	V_{LVD}	EN=1	1.75	1.85	1.95	V



4.5 AC Characteristics

all parameters under normal operating conditions, unless otherwise stated; typical values at T_A = 23 °C and V_{CC} = 3 V; test circuit shown in Fig. 6, f_c = 868.3MHz

Parameter	Symbol	Condition	Min	Тур	Max	Unit
CW Spectrum Characteristics						
Output power in step 0 (Isolation in off-state)	P _{off}	EN=1			-70	dBm
Output power in step 1	P ₁	EN=1	-13	-12	-11 ¹⁾	dBm
Output power in step 2	P ₂	EN=1	-4	-3	-2 ¹⁾	dBm
Output power in step 3	P ₃	EN=1	1	2.5	3.5 ¹⁾	dBm
Output power in step 4	P ₄	EN=1	4	7.5	9.5 ¹⁾	dBm
Phase noise	L(f _m)	@ 200kHz offset		-87	-82	dBc/Hz
Spurious emissions according to EN 300 220-1 (2000.09) table 13	P _{spur}	47MHz< f <74MHz 87.5MHz< f <118MHz 174MHz< f <230MHz 470MHz< f <862MHz B=100kHz	JP		-54	dBm
		f < 1GHz, B=100kHz			-36	dBm
		f > 1GHz, B=1MHz			-30	dBm
Clock output Characteristics						
Output low voltage CMOS	V _{OL}	depending on capaci-			0.3*V _{CC}	V
Output high voltage CMOS	V _{OH}	tor CCK and CKDIV	0.7*V _{CC}			V
Start-up Parameters						
Start-up time	t _{on}	from standby to transmit mode		0.6	1	ms
Frequency Stability	•					
Frequency stability vs. supply voltage	df _{VCC}				±3	ppm
Frequency stability vs. temperature	df _{TA}	crystal at constant temperature			±10	ppm
Frequency stability vs. variation range of C _{RO}	df _{CRO}				±20	ppm

¹⁾ output matching network tuned for 5V supply



4.6 AC Characteristics

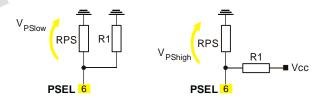
Power step	0	1	2	3	4
RPS / kΩ < 3		22	56	120	not connected

4.7 Output Power Steps - ASK Mode

typical values at T_A = 23 °C and V_{CC} = 3 V; test circuit shown in Fig. 6

Power step 1		2	3	4	
RPS / kΩ	2.4	2.8	3.5	not connected	
R1 / kΩ	36	14	7	0	

 V_{PSlow} = voltage across RPS if ASK_DTA at 0V V_{PShigh} = voltage across RPS if ASK_DTA at Vcc



If the transmitter is operated at any supply voltage V_{cc} , the values for R1 and RPS can be calculated as follows:

$$R_{1} = \frac{V_{CC} \cdot V_{PSlow}}{I_{PSEL} \cdot V_{PShigh}} \qquad \qquad R_{PS} = R_{1} \frac{V_{PShigh}}{V_{CC} - V_{PShigh}}$$



5 Test Circuit

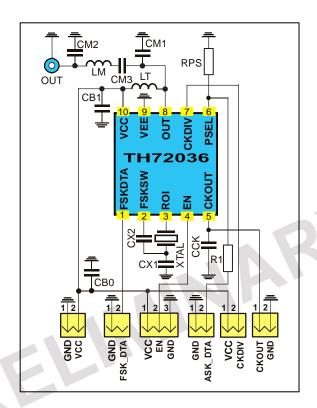


Fig. 6: Test circuit for FSK and ASK with 50 Ω matching network

5.1 Test circuit component list to Fig. 6

Part	Size	Value @ 868.3 MHz	Value @ 915 MHz	Tolerance	Description		
CM1	0805	1.8 pF	2.2 pF	±5%	impedance matching capacitor		
CM2	0805	5.6 pF	5.6 pF	±5%	impedance matching capacitor		
CM3	0805	68 pF	68 pF	±5%	impedance matching capacitor		
LM	0805	12 nH	10 nH	±5%	impedance matching inductor, note 2		
LT	0805	15 nH	10 nH	±5%	output tank inductor, note 2		
CX1_FSK	0805	22 pF	22 pF	±5%	XOSC FSK capacitor ($\Delta f = \pm 20 \text{ kHz}$), note 1		
CX1_ASK	0805	27 pF	27 pF	±5%	XOSC ASK capacitor, trimmed to f _C , note 1		
CX2	0805	12 pF	12 pF	±5%	XOSC capacitor ($\Delta f = \pm 20 \text{ kHz}$), note 1 only needed for FSK		
CCK	0805	15 pF / 180 pF		±5%	clock spur suppression capacitor, CKDIV 0 / 1		
RPS	0805	see section 4.6		±5%	FSK or CW mode power-select resistor		
R1		see section 4.7			ASK power-select resistor, not used at FSK		
CB0	0805	220 nF		±20%	de-coupling capacitor		
CB1	0805	330 pF		±10%	de-coupling capacitor		
XTAL	SMD 6x3.5	27.13438 MHz	28.59375 MHz	±30ppm calibr. ±30ppm temp.	fundamental wave crystal, CL = 12 pF, C0, max = 7 pF, R1 = 40 Ω		

Note 1: value depending on crystal parameters

Note 2: for high-power applications high-Q wire-wound inductors should be used



6 Package Description

(e)

The device TH72036 is RoHS compliant.

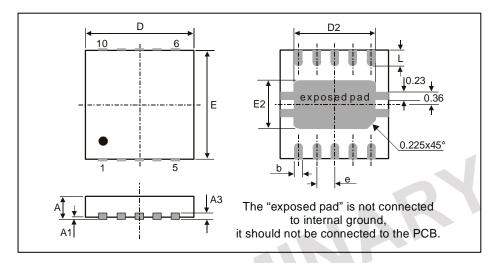


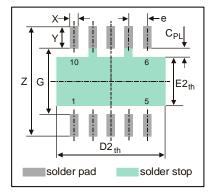
Fig. 7: 10L QFN 3x3 Dual

all Dimensions in mm										
	D	Е	D2	E2	Α	A1	А3	L	е	b
min	2.85	2.85	2.23	1.49	0.80	0	0.20	0.3	0.50	0.18
max	3.15	3.15	2.48	1.74	1.00	0.05	0.20	0.5		0.30
all D	all Dimensions in inch									
min	0.112	0.112	0.0878	0.051	0.0315	0	0.0079	0.0118	0.0197	0.0071
max	0.124	0.124	0.0976	0.055	0.0393	0.002		0.0197		0.0118

6.1 Soldering Information

 The device TH72036 is qualified for MSL3 with soldering peak temperature 260 deg C according to JEDEC J-STD-20

6.2 Recommended PCB Footprints



all D	all Dimensions in mm									
	Z	G	D2 _{th}	E2 _{th}	X	Υ	C _{PL}	е		
min	3.55	1.9	3.2	1.3	0.25	0.7	0.3	0.5		
max	3.90	2.3	3.6	1.7	0.30	1.0	0.5	0.5		
all D	all Dimensions in inch									
min	0.1398	0.0748	0.1260	0.0512	0.0098	0.0276	0.0591	0.0197		
max	0.1535	0.0906	0.1417	0.0669	0.0118	0.0394	0.0197	0.0197		

Fig. 8: PCB land pattern style

3901072036 Page 13 of 16 Data Sheet Rev. 005 June/07

TH72036 868/915MHz FSK/ASK Transmitter

7 Reliability Information

This Melexis device is classified and qualified regarding soldering technology, solderability and moisture sensitivity level, as defined in this specification, according to following test methods:

Reflow Soldering SMD's (Surface Mount Devices)

- IPC/JEDEC J-STD-020
 - "Moisture/Reflow Sensitivity Classification for Nonhermetic Solid State Surface Mount Devices (classification reflow profiles according to table 5-2)"
- EIA/JEDEC JESD22-A113
 - "Preconditioning of Nonhermetic Surface Mount Devices Prior to Reliability Testing (reflow profiles according to table 2)"

Wave Soldering SMD's (Surface Mount Devices) and THD's (Through Hole Devices)

- EN60749-20
 - "Resistance of plastic- encapsulated SMD's to combined effect of moisture and soldering heat"
- EIA/JEDEC JESD22-B106 and EN60749-15
 - "Resistance to soldering temperature for through-hole mounted devices"

Iron Soldering THD's (Through Hole Devices)

EN60749-15

"Resistance to soldering temperature for through-hole mounted devices"

Solderability SMD's (Surface Mount Devices) and THD's (Through Hole Devices)

 EIA/JEDEC JESD22-B102 and EN60749-21 "Solderability"

For all soldering technologies deviating from above mentioned standard conditions (regarding peak temperature, temperature gradient, temperature profile etc) additional classification and qualification tests have to be agreed upon with Melexis.

The application of Wave Soldering for SMD's is allowed only after consulting Melexis regarding assurance of adhesive strength between device and board.

Melexis is contributing to global environmental conservation by promoting **lead free** solutions. For more information on qualification of **RoHS** compliant products (RoHS = European directive on the Restriction Of the Use of Certain Hazardous Substances) please visit the quality page on our website:

http://www.melexis.com/quality_leadfree.aspx

8 ESD Precautions

Electronic semiconductor products are sensitive to Electro Static Discharge (ESD).

Always observe Electro Static Discharge control procedures whenever handling semiconductor products.

3901072036 Page 14 of 16 Data Sheet Rev. 005 June/07



Your Notes



TH72036 868/915MHz FSK/ASK Transmitter

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