# TOSHIBA Bi-CMOS Digital Integrated Circuit Silicon Monolithic 

## TB2905HQ

## Class KB High-Efficiency, Low-Frequency Power Amplifier IC Maximum Power: $47 \mathrm{~W} \times 4$ Channels

The TB2905HQ is a high-efficiency class KB power amplifier IC developed for car audio applications that incorporates four BTL amplifier channels.

It employs a pure complementary DM OS output stage consisting of P -ch upper and N -ch lower sections, offering a maximum output power (POUT) of 47 W .

Class KB (keyed BTL) amplifiers exhibit less than half the heat generation of comparable class AB solutions under normal operating conditions. Therefore, it is possible to design a smaller heatsink and maintain lower internal temperature in the car


Weight: 7.7 g (typ.) audio sets.

Additionally, the TB2905HQ has many built-in functions for car audio, such as standby switching, muting, protective circuits, and self diagnosis.

## Features

- High output power

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: Pout MAX (1) = 47 W (typ.)
    ( \(\mathrm{VCC}=14.4 \mathrm{~V}, \mathrm{f}=1 \mathrm{kHz}, \mathrm{J}\) EITA max, \(\mathrm{RL}=4 \Omega\) )
    : Pout MAX (2) = 43 W (typ.)
        \((\mathrm{VCC}=13.7 \mathrm{~V}, \mathrm{f}=1 \mathrm{kHz}, \mathrm{J}\) EITA max, \(\mathrm{RL}=4 \Omega\) )
    : Pout Pout MAX (3) = 80 W (typ.)
        \(\left(\mathrm{VCC}=14.4 \mathrm{~V}, \mathrm{f}=1 \mathrm{kHz}, \mathrm{J}\right.\) EITA max, \(\mathrm{R}_{\mathrm{L}}=2 \Omega\) )
    : Pout (1) = 29 W (typ.)
        (VCC = 14.4 V, f = 1 kHz, THD \(=10 \%, \mathrm{RL}=4 \Omega\) )
    : Pout (2) = 25 W (typ.)
        \((\mathrm{VCC}=13.2 \mathrm{~V}, \mathrm{f}=1 \mathrm{kHz}, \mathrm{THD}=10 \%, \mathrm{RL}=4 \Omega)\)
    : Pout (3) = 55 W (typ.)
        ( \(\mathrm{VCC}=14.4 \mathrm{~V}, \mathrm{f}=1 \mathrm{kHz}, \mathrm{THD}=10 \%, \mathrm{RL}=2 \Omega\) )
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- Low distortion ratio
: THD = 0.03\% (typ.) $(\mathrm{VCC}=13.2 \mathrm{~V}, \mathrm{f}=1 \mathrm{kHz}, \mathrm{GV}=26 \mathrm{~dB}, \mathrm{POUT}=2 \mathrm{~W}, \mathrm{RL}=4 \Omega)$
- Low noise
: $\mathrm{V}_{\mathrm{NO}}=120 \mu \mathrm{Vrms}$ (typ.) $(\mathrm{VCC}=13.2 \mathrm{~V}, \mathrm{GV}=26 \mathrm{~dB}, \mathrm{Rg}=0 \Omega, \mathrm{BW}=20 \mathrm{~Hz} \sim 20 \mathrm{kHz}, \mathrm{RL}=4 \Omega$ )
- Built-in standby (pin 4) and muting (pin22) functions
- Built-in offset/clipping detection (pin 25)
- Protective circuits: Thermal shutdown, overvoltage, out to GND, out to VCC, out to out short
- Operating supply voltage: VCC (opr) $=9$ to $18 \mathrm{~V}\left(\mathrm{RL}_{\mathrm{L}}=4 \Omega\right)$

Note 1: Some pins of this product are sensitive to electrostatic discharge. When handling this product, ensure that the environment is protected against electrostatic discharge.

Note 2: Install the product correctly. Otherwise, the product or connected equipment may get damaged or degrade.
Note 3: These protective features are intended to temporarily prevent an output short circuit or other abnormal conditions from occurring. Toshiba does not guarantee that they prevent the IC from being damaged.
If the product is operating outside any of the guaranteed operating ranges, these protective features may not operate and an output short circuit may result in the IC being damaged.

## Block Diagram



Some of the functional blocks, circuits, or constants in the block diagram may be omitted or simplified for explanatory purpose.

## Operational Description (Each description applies to a single channel)

## 1. Voltage Gain

The TB2905HQ has no NF (negative feedback) pins. Therefore, the voltage gain (Gv) is determined within the IC.


Figure 1 Amplifier Configuration

Voltage gain of amplifier $1: \mathrm{G}_{\mathrm{V} 1}=0 \mathrm{~dB}$ Voltage gain of amplifiers 2A and 2B: Gv2 $=20 \mathrm{~dB}$
Voltage gain obtained by BTL connection: $\mathrm{GV}(\mathrm{BTL})=6 \mathrm{~dB}$
Therefore, the total voltage gain is determined by the following expression:

$$
\mathrm{Gv}=\mathrm{G} v 1+\mathrm{Gv} 2+\mathrm{Gv}(\mathrm{BTL})=0+20+6=26 \mathrm{~dB}
$$

Although this configuration without an NF pin does not allow the user to adjust voltage gain, it eliminates the need for an NF capacitor, resulting in lower total application cost and smaller mounting space.

## 2. Standby Switching Function (Pin 4)

The TB2905HQ can be powered up or down by controlling the state of pin 4 (standby pin).
The threshold voltage for pin 4 is approximately 3 VBE (typ.). The supply current in standby state is approximately 2 $\mu \mathrm{A}$ (typ.).

Pin 4 control voltage: VSB

| Stand-by | Power | $\mathrm{V}_{\text {SB }}(\mathrm{V})$ |
| :---: | :---: | :---: |
| ON | OFF | $0 \sim 0.5$ |
| OFF | ON | $2.5 \sim 6 \mathrm{~V}$ |



Figure 2 Driving pin 4 high powers up the TB2905HQ

When changing the time constant for pin 4 , check the pop noise produced.

## <Advantages of standby switching >

(1) The microcontroller can directly turn $\mathrm{V}_{\mathrm{CC}}$ on or off without using a switching relay.
(2) Since the control current is microscopic, a switching relay with small current capacity is satisfactory for switching.




From microcontroller


- Standby switching -

Figure 3 Standby switching

## 3. Muting Function (Pin 22)

Driving pin 22 low triggers audio muting.
The time constant for muting is determined from R1 and C4. Select the constants considering the pop noise that is produced when powering the TB2905HQ on/off or turning muting on/off. (See Figures 4 and 5.) Pin 22 is designed to be controlled at 3.3 V .
The pin functions as a current source switch for the internal muting circuit and is designed so that its discharging current is $200 \mu \mathrm{~A}$. The value of the external pull-up resistor is based on this current.
Example: When changing the control voltage from 3.3 V to $5 \mathrm{~V}, 5 \mathrm{~V} / 3.3 \mathrm{~V} \times 47 \mathrm{k} \Omega=71 \mathrm{k} \Omega$
The TB2905HQ internally triggers muting when the voltage is dropped, taking in a current of $200 \mu \mathrm{~A}$. It cannot take in the current if the pull-up resistance is too low. The series resistance (R1) for pin 22 must, therefore, be at least $47 \mathrm{k} \Omega$.


Figure 4 Muting function


Figure 5 Muting attenuation - VMUTE (V)

## 4. Offset Detection Function

Pin 25 can be used to detect an offset voltage that may appear on an output pin due to input capacitor leakage or other reasons.


Figure 6 Example application and detection mechanism


## 5. Output Clipping Detection Function (Pin 25)

Pin 25 has open-collector output (active low) structure, as shown in Figure 7.
If the output waveform is clipped, the clipping detection circuit in the IC activates and turns the Q1 NPN transistor on.
The microcontroller can use this signal to control the volume and tone control circuits, thus improving sound quality.

Pin 25 should be left open when this function is not used.
(Example application)


Pin 25: Open-collector output (active low)
(A) Output AC waveform
(B) Clipping detection circuit


Figure 8 Clipping detection mechanism

## 6. Pop Noise Suppression

The TB2905HQ uses AC-GND as a common NF pin for all amplifiers, thus requiring that the ratio of input capacitance C 1 to $\mathrm{AC}-\mathrm{GND}$ capacitance C 6 be 1:4.

Powering up the IC initiates the charging of C1 and C6. If the IC is turned off before the charging of C1 and C6 completes, the input DC balance becomes unbalanced, causing a pop noise to be produced.
To suppress the noise, it is recommended that a longer charging time be used for C2 as well as for C1 and C6. Note that the time which audio output takes to start will be longer, since C2 determines the muting time (the time from when the power is turned on to when audio output starts).
The pop noise which is generated when the muting function is turned on/off will vary according to the time constant for C4.

The greater the capacitance, the lower the pop noise. Note that the time from when the muting control signal is applied to C4 to when the muting function is turned on/off will be longer.

## 7. External Component Constants

| Component Name | Recommended Value | Purpose | Effect |  | Remarks |
| :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  | Smaller than Recommended Value | Larger than Recommended Value |  |
| C1 | $0.22 \mu \mathrm{~F}$ | To eliminate DC | Cutoff frequency is increased | Cutoff frequency is reduced | Affects the pop noise generated when VCC is turned on |
| C2 | $10 \mu \mathrm{~F}$ | To reduce ripple | Powering on/off is faster | Powering on/off is slower |  |
| C3 | $0.1 \mu \mathrm{~F}$ | To provide sufficient oscillation margin | Reduces noise and provides sufficient oscillation margin |  |  |
| C4 | $1 \mu \mathrm{~F}$ | To reduce pop noise | High pop noise Duration until muting function is turned on/off is short | Low pop noise Duration until muting function is turned on/off is long |  |
| C5 | 3900 F | Ripple filter | Power supply humming and ripple filtering |  |  |
| C6 | $1 \mu \mathrm{~F}$ | NF for all outputs | Pop noise is suppressed when $\mathrm{C} 1: \mathrm{C} 6=1: 4$ |  | Affects the pop noise generated when VCC is turned on |

## 8. Preventive measure against oscillation

For preventing the oscillation, check that the application circuit and actual load makes no abnormal oscillation under all the necessary condition.
Especially, perform the temperature test to check the oscillation margin since the oscillation margin is varied according to the causes described below,

1) Layout of printed board
2) Type of Speaker
3) Value and kind of the capacitor between the output(+) and output(-)
4) Value and kind of the CR filter or the capacitor between each output and GND.

Maximum Ratings $\left(\mathbf{T a}=\mathbf{2 5}{ }^{\circ} \mathrm{C}\right.$ )

| Characteristics | Symbol | Rating | Unit |
| :--- | :---: | :---: | :---: |
| Peak supply voltage (0.2 s) | $\mathrm{V}_{\mathrm{CC}}$ (surge) | 50 | V |
| DC supply voltage | $\mathrm{V}_{\mathrm{CC}}(\mathrm{DC})$ | 25 | V |
| Operating supply voltage | $\mathrm{V}_{\mathrm{CC}}$ (opr) | 18 | V |
| Output current (peak) | $\mathrm{I}_{\mathrm{O}}$ (peak) | 9 | A |
| Power dissipation | $\mathrm{P}_{\mathrm{D}}($ Note 5$)$ | 125 | W |
| Operating temperature | $\mathrm{T}_{\mathrm{opr}}$ | $-40 \sim 85$ | ${ }^{\circ} \mathrm{C}$ |
| Storage temperature | $\mathrm{T}_{\mathrm{stg}}$ | $-55 \sim 150$ | ${ }^{\circ} \mathrm{C}$ |

Note 5: Package thermal resistance $\left(\theta \mathrm{j} T=1^{\circ} \mathrm{C} / \mathrm{W}\right)\left(\mathrm{Ta}=25^{\circ} \mathrm{C}\right.$, with infinite heat sink)

The absolute maximum ratings of a semiconductor device are a set of specified parameter values which must not be exceeded during operation, even for an instant.
Exposure to conditions beyond those listed above may cause permanent damage to the device or affect device reliability, which could increase potential risks of personal injury due to IC blowup and/or burning.
The equipment manufacturer should design so that no maximum rating value is exceeded with respect to current, voltage, power dissipation, temperature, etc.
Ensuring that the parameter values remain within these specified ranges during device operation will help to ensure that the integrity of the device is not compromised.

Electrical Characteristics
(unless otherwise specified, $\mathrm{V}_{\mathrm{CC}}=13.2 \mathrm{~V}, \mathrm{f}=1 \mathrm{kHz}, \mathrm{R}_{\mathrm{L}}=4 \Omega, \mathrm{Ta}=25^{\circ} \mathrm{C}$ )

| Characteristics | Symbol | Test Circuit | Test Conditions | Min | Typ. | Max | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Quiescent supply current | ICCQ | - | $\mathrm{V}_{\text {IN }}=0$ | - | 200 | 300 | mA |
| Output power | Pout MAX (1) | - | $\mathrm{V}_{\mathrm{CC}}=14.4 \mathrm{~V}$, max POWER | - | 47 | - | W |
|  | Pout MAX (2) | - | $\mathrm{V}_{\mathrm{Cc}}=13.7 \mathrm{~V}$, max POWER | - | 43 | - |  |
|  | Pout (1) | - | $\mathrm{V}_{\mathrm{CC}}=14.4 \mathrm{~V}, \mathrm{THD}=10 \%$ | - | 29 | - |  |
|  | Pout (2) | - | THD $=10 \%$ | 23 | 25 | - |  |
| Output power ( $\left.\mathrm{R}_{\mathrm{L}}=2 \Omega\right)$ | Pout MAX (1) | - | $\mathrm{V}_{\mathrm{CC}}=14.4 \mathrm{~V}$, max POWER | - | 80 | - | W |
|  | Pout MAX (2) | - | $\mathrm{V}_{\mathrm{CC}}=13.7 \mathrm{~V}$, max POWER | - | 77 | - |  |
|  | Pout (1) | - | $\mathrm{V}_{C C}=14.4 \mathrm{~V}, \mathrm{THD}=10 \%$ | - | 55 | - |  |
|  | Pout (2) | - | THD = 10\% | 42 | 45 | - |  |
| Total harmonics distortion | THD | - | POUT $=2 \mathrm{~W}$ | - | 0.03 | 0.2 | \% |
| Voltage gain | Gv | - | $\mathrm{V}_{\text {OUT }}=0.775 \mathrm{Vrms}$ | 24 | 26 | 28 | dB |
| Interchannel voltage gain | $\triangle \mathrm{Gv}$ | - | $\mathrm{V}_{\text {OUT }}=0.775 \mathrm{Vrms}$ | -1.0 | 0 | 1.0 | dB |
| Output noise voltage | $\mathrm{V}_{\mathrm{NO}}$ (1) | - | $\mathrm{Rg}=0 \Omega$, DIN45405 | - | 130 | - | $\mu \mathrm{Vrms}$ |
|  | $\mathrm{V}_{\mathrm{NO}}(2)$ | - | $\mathrm{Rg}=0 \Omega$, $\mathrm{BW}=20 \mathrm{~Hz} \sim 20 \mathrm{kHz}$ | - | 120 | 270 |  |
| Ripple rejection retio | R.R. | - | $\begin{aligned} & \mathrm{f}_{\text {rip }}=100 \mathrm{~Hz}, \mathrm{R}_{\mathrm{g}}=620 \Omega \\ & \mathrm{~V}_{\text {rip }}=0.775 \mathrm{Vrms} \end{aligned}$ | 50 | 60 | - | dB |
| Crosstalk | C.T. | - | $\begin{aligned} & \mathrm{R}_{\mathrm{g}}=620 \Omega \\ & \mathrm{~V}_{\text {OUT }}=0.775 \mathrm{Vrms} \\ & \hline \end{aligned}$ | - | 70 | - | dB |
| Output offset voltage | V ${ }_{\text {OFFSET }}$ | - | - | -150 | 0 | 150 | mV |
| Input resistance | $\mathrm{R}_{\mathrm{IN}}$ | - | - | - | 90 | - | $\mathrm{k} \Omega$ |
| Standby current | ISB | - | Standby state | - | 2 | 10 | $\mu \mathrm{A}$ |
| Standby control voltage | $\mathrm{V}_{\text {SB }} \mathrm{H}$ | - | POWER: ON, clipping detection on pin 25 | 7.5 | - | $\mathrm{V}_{\mathrm{CC}}$ | V |
|  | $\mathrm{V}_{S B} \mathrm{M}$ | - | POWER: ON, offset detection on pin 25 | 2.5 | - | 6.0 |  |
|  | VSB L | - | POWER: OFF | 0 | - | 0.5 |  |
| Muting control voltage | $\mathrm{V}_{\mathrm{M}} \mathrm{H}$ | - | MUTE: OFF | 2.5 | - | 6.0 | V |
|  | $\mathrm{V}_{\mathrm{M}} \mathrm{L}$ | - | MUTE: ON, $\mathrm{R}_{1}=47 \mathrm{k} \Omega$ | 0 | - | 0.5 |  |
| Muting attenuation | ATT M | - | MUTE: ON <br> VOUT $=7.75 \mathrm{Vrms} \rightarrow$ Mute: OFF | 80 | 90 | - | dB |
| Offset detection threshold voltage | $V_{\text {off-set }}$ | - | Rpull-up $=47 \mathrm{k} \Omega,+\mathrm{V}=5.0 \mathrm{~V}$ Based on normal output DC voltage | $\pm 1.0$ | $\pm 1.5$ | $\pm 2.0$ | V |

## Test Circuit



Components in the testing circuit are only used to determine the device's characteristics.
Toshiba does not guarantee that those components prevent the application equipment from malfunctioning or failing.
T.H.D - Pout (OUT)

T.H.D - POUT (OUT)

T.H.D - POUT (OUT)


T.H.D - Pout (OUT)

T.H.D - POUT (OUT)

T.H.D - POUT (OUT)


T.H.D - f

C.T. - f (OUT1)

C.T. - f (OUT3)

T.H.D - f

C.T. - f (OUT2)

C.T. -f (OUT4)









## Package Dimensions

HZIP25-P-1.00F
Unit: mm


Weight: 7.7 g (typ.)

About solderability, following conditions were confirmed

- Solderability
(1) Use of $\mathrm{Sn}-63 \mathrm{~Pb}$ solder Bath
- solder bath temperature $=230^{\circ} \mathrm{C}$
- dipping time $=5$ seconds
- the number of times = once
- use of R-type flux
(2) Use of $\mathrm{Sn}-3.0 \mathrm{Ag}-0.5 \mathrm{Cu}$ solder Bath
- solder bath temperature $=245^{\circ} \mathrm{C}$
- dipping time $=5$ seconds
- the number of times = once
- use of R-type flux


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