## RoHS Compliant \& Pb-Free Product

## Typical Applications

- UHF Digital and Analog Receivers
- Digital Communication Systems
- Spread-Spectrum Communication Systems
- Commercial and Consumer Systems
- 433MHz and 915 MHz ISM Band Receivers
- General Purpose Frequency Conversion


## Product Description

The RF2418 is a monolithic integrated UHF receiver front-end. The IC contains all of the required components to implement the RF functions of the receiver except for the passive filtering and LO generation. It contains an LNA (low-noise amplifier), a second RF amplifier, a dualgate GaAs FET mixer, and an IF output buffer amplifier which will drive a $50 \Omega$ load. In addition, the IF buffer amplifier may be disabled and a high impedance output is provided for easy matching to IF filters with high impedances. The output of the LNA is made available as an output to permit the insertion of a bandpass filter between the LNA and the RF/Mixer section. The LNA section may be disabled by removing the VDD1 connection to the IC.

Optimum Technology Matching ${ }^{\circledR}$ Applied

| $\square$ Si BJT | $\square$ GaAs HBT | $\square$ GaAs MESFET |
| :--- | :--- | :--- |
| $\square$ Si Bi-CMOS | $\square$ SiGe HBT | $\square$ Si CMOS |
| $\square$ InGaP/HBT | $\square$ GaN HEMT | $\square$ SiGe Bi-CMOS |



Functional Block Diagram


Package Style: SOIC-14

## Features

- Single 3V to 6.5V Power Supply
- High Dynamic Range
- Low Current Drain
- High LO Isolation
- LNA Power Down Mode for Large Signals


## Ordering Information

| RF2418 | Low Current LNA/Mixer |
| :--- | :--- |
| RF2418 PCBA | Fully Assembled Evaluation Board |

## Absolute Maximum Ratings

| Parameter | Rating | Unit |
| :--- | :---: | :---: |
| Supply Voltage | -0.5 to 7 | $\mathrm{~V}_{\mathrm{DC}}$ |
| Input LO and RF Levels | +6 | $\mathrm{dBm}^{\circ} \mathrm{Co}$ |
| Ambient Operating Temperature | -40 to +85 | ${ }^{\circ} \mathrm{C}$ |
| Storage Temperature | -40 to +150 | ${ }^{\circ} \mathrm{C}$ |



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| Parameter | Specification |  |  | Unit | Condition |
| :---: | :---: | :---: | :---: | :---: | :---: |
|  | Min. | Typ. | Max. |  |  |
| Overall <br> RF Frequency Range Cascade Power Gain Cascade $\mathrm{IP}_{3}$ Cascade Noise Figure |  | $\begin{gathered} 400 \text { to } 1100 \\ 23 \\ -13 \\ 2.4 \end{gathered}$ |  | $\begin{gathered} \mathrm{MHz} \\ \mathrm{~dB} \\ \mathrm{dBm} \\ \mathrm{~dB} \end{gathered}$ | $\begin{aligned} & \mathrm{T}=25^{\circ} \mathrm{C}, \mathrm{~V}_{\mathrm{CC}}=5 \mathrm{~V}, \mathrm{RF}=850 \mathrm{MHz}, \\ & \mathrm{LO}=921 \mathrm{MHz} \end{aligned}$ <br> High impedance output <br> Referenced to the input <br> Single sideband, includes image filter with 1.0 dB insertion loss |
| First Section (LNA) <br> Noise Figure Input VSWR Input IP3 <br> Gain <br> Reverse Isolation Output VSWR | $\begin{gathered} +3.0 \\ 13 \end{gathered}$ | $\begin{gathered} 1.8 \\ 1.5: 1 \\ +4.0 \\ 14 \\ 40 \\ 1.5: 1 \end{gathered}$ | 2.0 | dB <br> dBm <br> dB <br> dB | With external series matching inductor |
| Second Section (RF Amp, Mixer, IF1) <br> Noise Figure <br> Input VSWR <br> Input IP3 <br> Conversion Power Gain <br> Output Impedance | 7 | $\begin{gathered} 9.5 \\ 1.5: 1 \\ +1 \\ 9 \\ 4000 \\| 10 \mathrm{pF} \end{gathered}$ |  | $\begin{gathered} \mathrm{dB} \\ \mathrm{dBm} \\ \mathrm{~dB} \\ \Omega \\ \hline \end{gathered}$ | High impedance output <br> Single Sideband <br> With external series matching inductor <br> Open Collector |
| Second Section (RF Amp, Mixer, IF2) <br> Noise Figure <br> Input VSWR <br> Input IP3 <br> Conversion Gain <br> Output Impedance | $\begin{gathered} -0.5 \\ 5 \end{gathered}$ | $\begin{gathered} 10 \\ 1.5: 1 \\ 0 \\ 6 \\ 30 \\ \hline \end{gathered}$ |  | $\begin{gathered} \mathrm{dB} \\ \\ \mathrm{dBm} \\ \mathrm{~dB} \\ \Omega \\ \hline \end{gathered}$ | Buffered output, $50 \Omega$ load <br> Single Sideband <br> With external series matching inductor |
| LO Input <br> LO Frequency <br> LO Level <br> LO to RF Rejection LO to IF Rejection LO Input VSWR |  | $\begin{gathered} 300 \text { to } 1200 \\ -6 \text { to }+6 \\ 15 \\ 40 \\ 1.3: 1 \end{gathered}$ |  | MHz dBm dB dB | With pin 5 connected to ground. In order to achieve a low VSWR match at this input, an $82 \Omega$ resistor to ground is placed in parallel with this port. |
| Power Supply <br> Voltage <br> Current Consumption | $\begin{gathered} 3.0 \\ 12 \\ 6 \end{gathered}$ | $\begin{gathered} 14 \\ 20 \\ 9 \end{gathered}$ | $\begin{aligned} & 6.5 \\ & 26 \\ & 20 \end{aligned}$ | $\begin{gathered} \mathrm{V} \\ \mathrm{~mA} \\ \mathrm{~mA} \\ \mathrm{~mA} \end{gathered}$ | $\mathrm{V}_{\mathrm{CC}}=5.0 \mathrm{~V}$, LNA On, Mixer On, Buffer Off $\mathrm{V}_{\mathrm{CC}}=5.0 \mathrm{~V}$, LNA On, Mixer On, Buffer On $\mathrm{V}_{\mathrm{CC}}=5.0 \mathrm{~V}$, LNA Off, Mixer On, Buffer Off |


| Pin | Function | Description | Interface Schematic |
| :---: | :---: | :--- | :--- |
| $\mathbf{1}$ | LNA IN | A series 10nH matching inductor is necessary to achieve specified gain <br> and noise figure at 900 MHz. This pin is NOT internally DC-blocked. An <br> external blocking capacitor must be provided if the pin is connected to a <br> device with DC present. A DC path to ground (i.e. an inductor or resis- <br> tor to ground) is, however, acceptable at this pin. If a blocking capacitor <br> is required, a value of 22pF is recommended. |  |
| $\mathbf{2}$ | GND | Ground connection. Keep traces physically short and connect immedi- <br> ately to ground plane for best performance. |  |
| $\mathbf{3}$ | VDD1 | Supply Voltage for the LNA only. A 22pF external bypass capacitor is <br> required and an additional 0.01 $\mu$ is required if no other low frequency <br> bypass capacitors are near by. The trace length between the pin and <br> the bypass capacitors should be minimized. The ground side of the <br> bypass capacitors should connect immediately to ground plane. |  |
| $\mathbf{4}$ | VDD2 | For large input signals, VDD1 may be disconnected, resulting in the <br> LNA's gain changing from +11dB to -26dB and current drain decreas- <br> ing by 4mA. If the LNA is never required for use, then this pin can be <br> left unconnected or grounded, and Pin 11 is used as the first input. | Power supply for the IF buffer amplifier. If the high impedance mixer <br> output is being used, then this pin is not connected. |
| $\mathbf{5}$ | IF BYP | If this pin is connected to ground, an internal 10pF capacitor is con- <br> nected in parallel with the mixer output. This capacitor functions as an <br> LO trap, which reduces the amount of LO to IF bleed-through and pre- <br> vents high LO voltages at the mixer output from degrading the mixer's <br> dynamic range. At higher IF frequencies, this capacitance, along with <br> parasitic layout capacitance, should be parallel resonated out by the <br> choice of the bias inductor value at pin 7. If the internal capacitor is not <br> connected to ground, the buffer amplifier could become unstable. A <br> $\sim 10 p F ~ c a p a c i t o r ~ s h o u l d ~ b e ~ a d d e d ~ a t ~ t h e ~ o u t p u t ~ t o ~ m a i n t a i n ~ t h e ~ b u f f e r s ~$ |  |
| stability, but the gain will not be significantly affected. |  |  |  |

## RF2418

| Pin | Function | Description | Interface Schematic |
| :---: | :---: | :---: | :---: |
| 10 | GND | Same as pin 2. |  |
| 11 | RF IN | Mixer RF Input port. For a $50 \Omega$ match at 900 MHz use a 15 nH series inductor. This pin is NOT internally DC-blocked. An external blocking capacitor must be provided if the pin is connected to a device with DC present. A DC path to ground (i.e. an inductor or resistor to ground) is, however, acceptable at this pin. If a blocking capacitor is required, a value of 22 pF is recommended. To minimize the mixer's noise figure, it is recommended to have a RF bandpass filter before this input. This will prevent the noise at the image frequency from being converted to the IF. |  |
| 12 | GND | Same as pin 2. |  |
| 13 | GND | Same as pin 2. |  |
| 14 | LNA OUT | $50 \Omega$ output. Internally DC-blocked. |  |

## Application Schematic High Impedance Output Configuration 850 MHz



[^0]
## Application Schematic Buffered Output Configuration 850 MHz



## RF2418

## Evaluation Board Schematic $R F=850 \mathrm{MHz}, I F=71 \mathrm{MHz}$

(Download Bill of Materials from www.rfmd.com.)


## Evaluation Board Layout Board Size 1.52" x 1.52"

Board Thickness 0.031", Board Material FR-4


## RF2418

High Impedance Mixer Gain versus Voltage, RF=850MHz


High Impedance Mixer Input IP3 versus Voltage,


Buffered LNA Gain versus Voltage,


High Impedance Casc. Gain versus Voltage,


High Impedance Casc. Input IP3 versus Voltage,


Buffered Mixer Gain versus Voltage, RF=850MHz




Buffered LNA Noise Figure versus Voltage, RF=850MHz Part to Part Variation


Buffered LNA Input versus Voltage,


Buffered Casc. Input IP3 versus Voltage,


Buffered Mixer Noise Figure versus Voltage,



[^0]:    L1 and C1 are picked to match the mixer's output impedance ( $4 \mathrm{k} \Omega \mathrm{II} 10 \mathrm{pF}$ ) to the IF filter's impedance, at the IF frequency. C1 also serves as a DC block, in case the IF filter is not an open circuit at DC.

