

CASCADABLE BROADBAND GaAs MMIC AMPLIFIER DC TO 10GHz

Gain Stage or Driver Amplifiers for

MWRadio/Optical Designs (PTP/PMP/

LMDS/UNII/VSAT/WLAN/Cellular/DWDM)

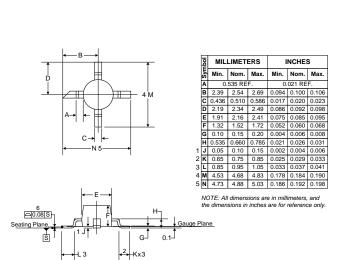
RoHS Compliant & Pb-Free Product

Typical Applications

- Narrow and Broadband Commercial and Military Radio Designs
- Linear and Saturated Amplifiers

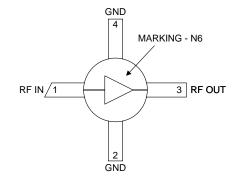
Product Description

The NLB-310 cascadable broadband InGaP/GaAs MMIC amplifier is a low-cost, high-performance solution for general purpose RF and microwave amplification needs. This 50 Ω gain block is based on a reliable HBT proprietary MMIC design, providing unsurpassed performance for small-signal applications. Designed with an external bias resistor, the NLB-310 provides flexibility and stability. The NLB-310 is packaged in a low-cost, surface-mount plastic package, providing ease of assembly for high-volume tape-and-reel requirements.



Optimum Technology Matching® Applied

🗌 Si BJT	🗌 GaAs HBT	GaAs MESFET
Si Bi-CMOS	SiGe HBT	Si CMOS
InGaP/HBT	🗌 GaN HEMT	SiGe Bi-CMOS



Functional Block Diagram

Package Style: Micro-X, 4-Pin, Plastic

Features

- Reliable, Low-Cost HBT Design
- 12.7dB Gain, +12.6dBm P1dB@2GHz
- High P1dB of +14.9dBm @ 6.0GHz and +13.1dBm @ 10.0GHz
- Single Power Supply Operation
- 50 Ω I/O Matched for High Freq. Use

Ordering Information

NLB-310	Cascadable Broadband G 10GHz	GaAs MMIC Amplifier DC to
NLB-310-T1	Tape & Reel, 1000 Pieces	5
NLB-310-E	Fully Assembled Evaluati	on Board
NBB-X-K1		aP Amp Designer's Tool Kit
RF Micro Devices	, Inc.	Tel (336) 664 1233
7628 Thorndike R	oad	Fax (336) 664 0454
Greensboro, NC 2	7409, USA	http://www.rfmd.com

Absolute Maximum Ratings

	0			
Parameter	Rating	Unit		
RF Input Power	+20	dBm		
Power Dissipation	300	mW		
Device Current	70	mA		
Channel Temperature	200	°C		
Operating Temperature	-45 to +85	°C		
Storage Temperature	-65 to +150	°C		
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RF Micro Devices believes the furnished information is correct and accurate at the time of this printing. RoHS marking based on EU Directive2002/95/EC (at time of this printing). However, RF Micro Devices reserves the right to make changes to its products without notice. RF Micro Devices does not assume responsibility for the use of the described product(s).

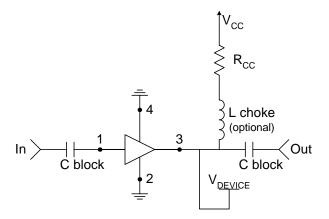
Exceeding any one or a combination of these limits may cause permanent damage.

Parameter	Specification			Unit	Condition
Parameter	Min.	Тур.	Max.	Unit	Condition
Overall					V _D =+4.6V, I _{CC} =50mA, Z ₀ =50Ω, T _A =+25°C
Small Signal Power Gain, S21	12.0	12.7		dB	f=0.1GHz to 1.0GHz
-		10.7		dB	f=1.0GHz to 4.0GHz
		10.0		dB	f=4.0GHz to 6.0GHz
	8.5	9.7		dB	f=6.0GHz to 10.0GHz
		9.6		dB	f=10.0GHz to 12.0GHz
Gain Flatness, GF		±0.3		dB	f=5.0GHz to 10.0GHz
Input VSWR		1.6:1			f=0.1GHz to 4.0GHz
		1.75:1			f=4.0GHz to 7.0GHz
		1.6:1			f=7.0GHz to 11.0GHz
Output VSWR		1.5:1			f=0.1GHz to 4.0GHz
		1.8:1			f=4.0GHz to 7.0GHz
		1.6:1			f=7.0GHz to 11.0GHz
Output Power @					
-1dB Compression, P1dB		12.6		dBm	f=2.0GHz
		14.9		dBm	f=6.0GHz
		13.1		dBm	f=10.0GHz
Noise Figure, NF		5.0		dB	f=3.0GHz
Third Order Intercept, IP3		+28.9		dBm	f=2.0GHz
		+27.9			f=6.0GHz
Reverse Isolation, S12		-17		dB	f=0.1GHz to 20.0GHz
Device Voltage, V _D	4.4	4.6	4.8	V	
Gain Temperature Coefficient, δG _T /δT		-0.0015		dB/°C	
MTTF versus Temperature					
@ I _{CC} =50mA					
Case Temperature		85		°C	
Junction Temperature		125		°C	
MTTF		>1,000,000		hours	
Thermal Resistance					
θ _{JC}		174		°C/W	$\frac{J_T - T_{CASE}}{V_D \cdot I_{CC}} = \theta_{JC}(°C/Watt)$

Pin	Function	Description	Interface Schematic
1	RF IN	RF input pin. This pin is NOT internally DC blocked. A DC blocking capacitor, suitable for the frequency of operation, should be used in most applications. DC coupling of the input is not allowed, because this will override the internal feedback loop and cause temperature instability.	
2	GND	Ground connection. For best performance, keep traces physically short and connect immediately to ground plane.	
3	RF OUT	RF output and bias pin. Biasing is accomplished with an external series resistor and choke inductor to V _{CC} . The resistor is selected to set the DC current into this pin to a desired level. The resistor value is determined by the following equation: $R = \frac{(V_{CC} - V_{DEVICE})}{I_{CC}}$ Care should also be taken in the resistor selection to ensure that the current into the part never exceeds maximum datasheet operating current over the planned operating temperature. This means that a resistor between the supply and this pin is always required, even if a supply near 5.0V is available, to provide DC feedback to prevent thermal runaway. Because DC is present on this pin, a DC blocking capacitor, suitable for the frequency of operation, should be used in most applications. The supply side of the bias network should also be well bypassed.	RF IN O
4	GND	Same as pin 2.	

Typical Bias Configuration

Application notes related to biasing circuit, device footprint, and thermal considerations are available on request.

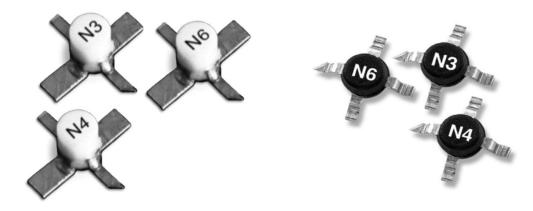


Recommended Bias Resistor Values					
Supply Voltage, V _{CC} (V)	8	10	12	15	20
Bias Resistor, R_{CC} (Ω)	60	100	140	200	300

Extended Frequency InGaP Amplifier Designer's Tool Kit NBB-X-K1

This tool kit was created to assist in the design-in of the RFMD NBB- and NLB-series InGap HBT gain block amplifiers. Each tool kit contains the following.

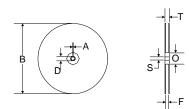
- 5 each NBB-300, NBB-310 and NBB-400 Ceramic Micro-X Amplifiers
- 5 each NLB-300, NLB-310 and NLB-400 Plastic Micro-X Amplifiers
- 2 Broadband Evaluation Boards and High Frequency SMA Connectors
- Broadband Bias Instructions and Specification Summary Index for ease of operation



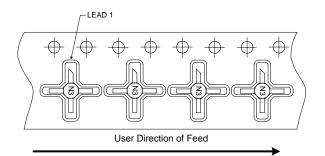
NLB-310

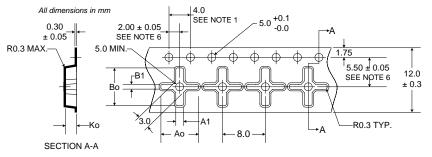
Tape and Reel Dimensions

All Dimensions in Millimeters



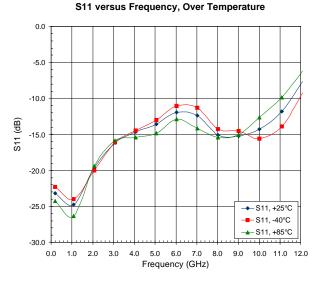
14.732 mm (7") REEL			Plastic, Micro-X		
	ITEMS SYMBOI		SIZE (mm)	SIZE (inches)	
	Diameter	В	178 +0.25/-4.0	7.0 +0.079/-0.158	
FLANGE	Thickness	Т	18.4 MAX	0.724 MAX	
	Space Between Flange	F	12.8 +2.0	0.50 +0.08	
	Outer Diameter	0	76.2 REF	3.0 REF	
HUB	Spindle Hole Diameter	S	13.716 +0.5/-0.2	0.540 +0.020/-0.008	
	Key Slit Width	A	1.5 MIN	0.059 MIN	
	Key Slit Diameter	D	20.2 MIN	0.795 MIN	



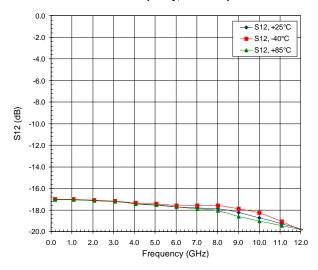


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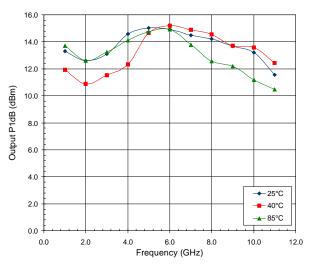
NOTES:	Ao = 7.0 MM
1. 10 sprocket hole pitch cumulative tolerance ±0.2.	A1 = 1.8 MM
2. Camber not to exceed 1 mm in 100 mm.	
3. Material; PS+C.	Bo = 7.0 MM
4. Ao and Bo measured on a plane 0.3 mm above the bottom of the pocket.	B1 = 1.3 MM
5. Ko measured from a plane on the inside bottom of the pocket to the surface of the carrier.	Ko = 2.1 MM
6. Pocket position relative to sprocket hole measured as true position of pocket, not pocket hole.	

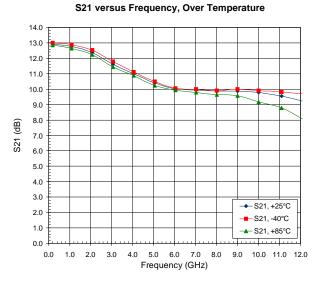


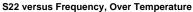
S12 versus Frequency, Over Temperature

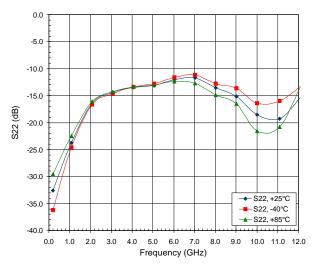


Output P1dB versus Frequency Across Temperature

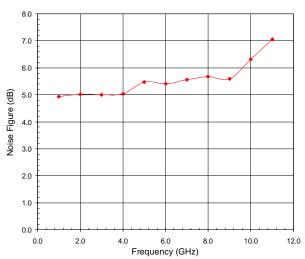














Note: The s-parameter gain results shown include device performance as well as evaluation board and connector loss variations. The insertion losses of the evaluation board and connectors are as follows:

1GHz to 4GHz=-0.06dB 5GHz to 9GHz=-0.22dB 10GHz to 14GHz=-0.50dB 15GHz to 20GHz=-1.08dB