



Integrated  
Circuit  
Systems, Inc.

# PRELIMINARY

## ICS8530I-01

LOW SKEW, 1-TO-16

DIFFERENTIAL-TO-3.3V LVPECL FANOUT BUFFER

### GENERAL DESCRIPTION



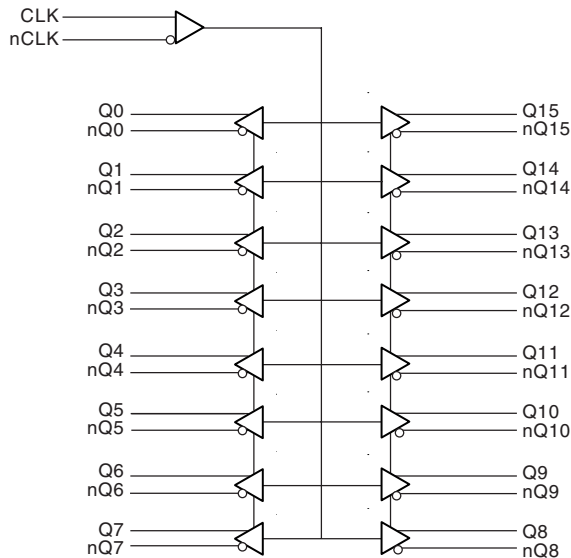
The ICS8530I-01 is a low skew, 1-to-16 Differential-to-3.3V LVPECL Fanout Buffer and a member of the HiPerClockS™ family of High Performance Clock Solutions from ICS. The CLK, nCLK pair can accept most standard differential input levels. The high gain differential amplifier accepts peak-to-peak input voltages as small as 150mV as long as the common mode voltage is within the specified minimum and maximum range.

Guaranteed output and part-to-part skew characteristics make the ICS8530I-01 ideal for those clock distribution applications demanding well defined performance and repeatability.

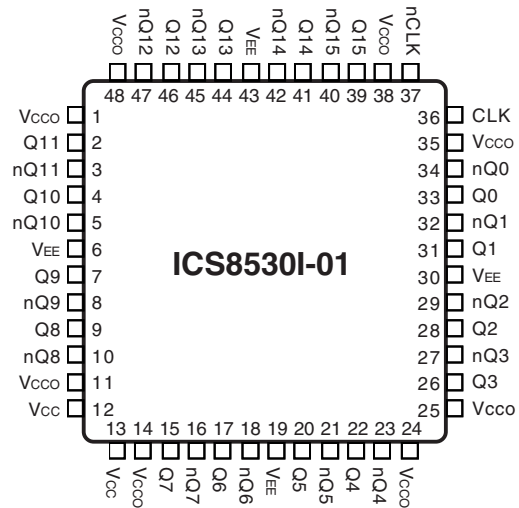
### FEATURES

- (16) differential 3.3V LVPECL outputs
- CLK, nCLK input pair
- CLK, nCLK pair can accept the following differential input levels: LVDS, LVPECL, LVHSTL, SSTL, HCSL
- Maximum output frequency: 500MHz
- Translates any single-ended input signal to 3.3V LVPECL levels with a resistor bias on nCLK input
- Output skew: 50ps (typical)
- Part-to-part skew: 100ps (typical)
- Additive phase jitter, RMS @ 106.25MHz: 0.022ps (typical) @ 25°C
- 3.3V output operating supply
- -40°C to 85°C ambient operating temperature

### BLOCK DIAGRAM



### PIN ASSIGNMENT



**48-Lead TQFP, E-Pad**  
7mm x 7mm x 1.0mm body package  
**Y Package**  
Top View

The Preliminary Information presented herein represents a product in prototyping or pre-production. The noted characteristics are based on initial product characterization. Integrated Circuit Systems, Incorporated (ICS) reserves the right to change any circuitry or specifications without notice.



**TABLE 1. PIN DESCRIPTIONS**

Number	Name	Type		Description
1, 11, 14, 24, 25, 35, 38, 48	V <sub>CCO</sub>	Power		Output supply pins.
2, 3	Q11, nQ11	Output		Differential output pair. LVPECL interface levels.
4, 5	Q10, nQ10	Output		Differential output pair. LVPECL interface levels.
6, 19, 30, 43	V <sub>EE</sub>	Power		Negative supply pins.
7, 8	Q9, nQ9	Output		Differential output pair. LVPECL interface levels.
9, 10	Q8, nQ8	Output		Differential output pair. LVPECL interface levels.
12, 13	V <sub>CC</sub>	Power		Core supply pins.
15, 16	Q7, nQ7	Output		Differential output pair. LVPECL interface levels.
17, 18	Q6, nQ6	Output		Differential output pair. LVPECL interface levels.
20, 21	Q5, nQ5	Output		Differential output pair. LVPECL interface levels..
22, 23	Q4, nQ4	Output		Differential output pair. LVPECL interface levels.
26, 27	Q3, nQ3	Output		Differential output pair. LVPECL interface levels.
28, 29	Q2, nQ2	Output		Differential output pair. LVPECL interface levels.
36	CLK	Input	Pulldown	Non-inverting differential clock input.
37	nCLK	Input	Pullup	Inverting differential clock input.
39, 40	Q15, nQ15	Output		Differential output pair. LVPECL interface levels.
41, 42	Q14, nQ14	Output		Differential output pair. LVPECL interface levels.
44, 45	Q13, nQ13	Output		Differential output pair. LVPECL interface levels.
46, 47	Q12, nQ12	Output		Differential output pair. LVPECL interface levels.

NOTE: *Pullup* and *Pulldown* refer to internal input resistors. See Table 2, Pin Characteristics, for typical values.

**TABLE 2. PIN CHARACTERISTICS**

Symbol	Parameter	Test Conditions	Minimum	Typical	Maximum	Units
C <sub>IN</sub>	Input Capacitance			4		pF
R <sub>PULLUP</sub>	Input Pullup Resistor			51		kΩ
R <sub>PULLDOWN</sub>	Input Pulldown Resistor			51		kΩ

**TABLE 3. FUNCTION TABLE**

Inputs		Outputs		Input to Output Mode	Polarity
CLK	nCLK	Q0:Q15	nQ0:nQ15		
0	1	LOW	HIGH	Differential to Differential	Non Inverting
1	0	HIGH	LOW	Differential to Differential	Non Inverting
0	Biased; NOTE 1	LOW	HIGH	Single Ended to Differential	Non Inverting
1	Biased; NOTE 1	HIGH	LOW	Single Ended to Differential	Non Inverting
Biased; NOTE 1	0	HIGH	LOW	Single Ended to Differential	Inverting
Biased; NOTE 1	1	LOW	HIGH	Single Ended to Differential	Inverting

NOTE 1: Please refer to the Application Information section, "Wiring the Differential Input to Accept Single Ended Levels".



**ABSOLUTE MAXIMUM RATINGS**

Supply Voltage, $V_{CC}$	4.6V
Inputs, $V_i$	-0.5V to $V_{CC} + 0.5V$
Outputs, $I_o$	
Continuous Current	50mA
Surge Current	100mA
Package Thermal Impedance, $\theta_{JA}$	27.6°C/W (0 lfpm)
Storage Temperature, $T_{STG}$	-65°C to 150°C

NOTE: Stresses beyond those listed under Absolute Maximum Ratings may cause permanent damage to the device. These ratings are stress specifications only. Functional operation of product at these conditions or any conditions beyond those listed in the *DC Characteristics* or *AC Characteristics* is not implied. Exposure to absolute maximum rating conditions for extended periods may affect product reliability.

**TABLE 4A. POWER SUPPLY DC CHARACTERISTICS,  $V_{CC} = V_{CCO} = 3.3V \pm 5\%$ ,  $T_A = -40^\circ C$  TO  $85^\circ C$**

Symbol	Parameter	Test Conditions	Minimum	Typical	Maximum	Units
$V_{CC}$	Core Supply Voltage		3.135	3.3	3.465	V
$V_{CCO}$	Output Supply Voltage		3.135	3.3	3.465	V
$I_{EE}$	Power Supply Current			115		mA

**TABLE 4B. DIFFERENTIAL DC CHARACTERISTICS,  $V_{CC} = V_{CCO} = 3.3V \pm 5\%$ ,  $T_A = -40^\circ C$  TO  $85^\circ C$**

Symbol	Parameter	Test Conditions	Minimum	Typical	Maximum	Units
$I_{IH}$	Input High Current	CLK	$V_{CC} = V_{IN} = 3.465V$		150	$\mu A$
		nCLK	$V_{CC} = V_{IN} = 3.465V$		5	$\mu A$
$I_{IL}$	Input Low Current	CLK	$V_{CC} = 3.465V, V_{IN} = 0V$	-5		$\mu A$
		nCLK	$V_{CC} = 3.465V, V_{IN} = 0V$	-150		$\mu A$
$V_{PP}$	Peak-to-Peak Input Voltage		0.15		1.3	V
$V_{CMR}$	Common Mode Input Voltage; NOTE 1, 2		$V_{EE} + 0.5$		$V_{CC} - 0.85$	V

NOTE 1: For single ended applications, the maximum input voltage for CLK, nCLK is  $V_{CC} + 0.3V$ .

NOTE 2: Common mode voltage is defined as  $V_{IH}$ .

**TABLE 4C. LVPECL DC CHARACTERISTICS,  $V_{CC} = V_{CCO} = 3.3V \pm 5\%$ ,  $T_A = -40^\circ C$  TO  $85^\circ C$**

Symbol	Parameter	Test Conditions	Minimum	Typical	Maximum	Units
$V_{OH}$	Output High Voltage; NOTE 1		$V_{CCO} - 1.4$		$V_{CCO} - 0.9$	V
$V_{OL}$	Output Low Voltage; NOTE 1		$V_{CCO} - 2.0$		$V_{CCO} - 1.7$	V
$V_{SWING}$	Peak-to-Peak Output Voltage Swing		0.6		1.0	V

NOTE 1: Outputs terminated with  $50\Omega$  to  $V_{CCO} - 2V$ .



**TABLE 5. AC CHARACTERISTICS,  $V_{CC} = V_{CC0} = 3.3V \pm 5\%$ ,  $T_A = -40^\circ C$  TO  $85^\circ C$**

Symbol	Parameter	Test Conditions	Minimum	Typical	Maximum	Units
$f_{MAX}$	Output Frequency				500	MHz
$t_{PD}$	Propagation Delay; NOTE 1			1.35		ns
$tsk(o)$	Output Skew; NOTE 2, 4			50		ps
$tsk(pp)$	Part-to-Part Skew; NOTE 3, 4			100		ps
$f_{jit}$	Buffer Additive Phase Jitter, RMS; refer to Additive Phase Jitter Section	106.25MHz, 25°C Integration Range: 12KHz to 20MHz		0.022		ps
		106.25MHz, 85°C Integration Range: 12KHz to 20MHz		0.026		ps
		212.5MHz, 25°C Integration Range: 12KHz to 20MHz		0.033		ps
		212.5MHz, 85°C Integration Range: 12KHz to 20MHz		0.034		ps
$t_R$	Output Rise Time	20% to 80% @ 50MHz	300		700	ps
$t_F$	Output Fall Time	20% to 80% @ 50MHz	300		700	ps
odc	Output Duty Cycle			50		%

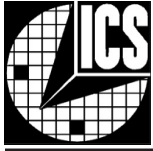
All parameters measured at 250MHz unless noted otherwise.

NOTE 1: Measured from the differential input crossing point to the differential output crossing point.

NOTE 2: Defined as skew between outputs at the same supply voltage and with equal load conditions. Measured at the output differential cross points.

NOTE 3: Defined as skew between outputs on different devices operating at the same supply voltages and with equal load conditions. Using the same type of inputs on each device, the outputs are measured at the differential cross points.

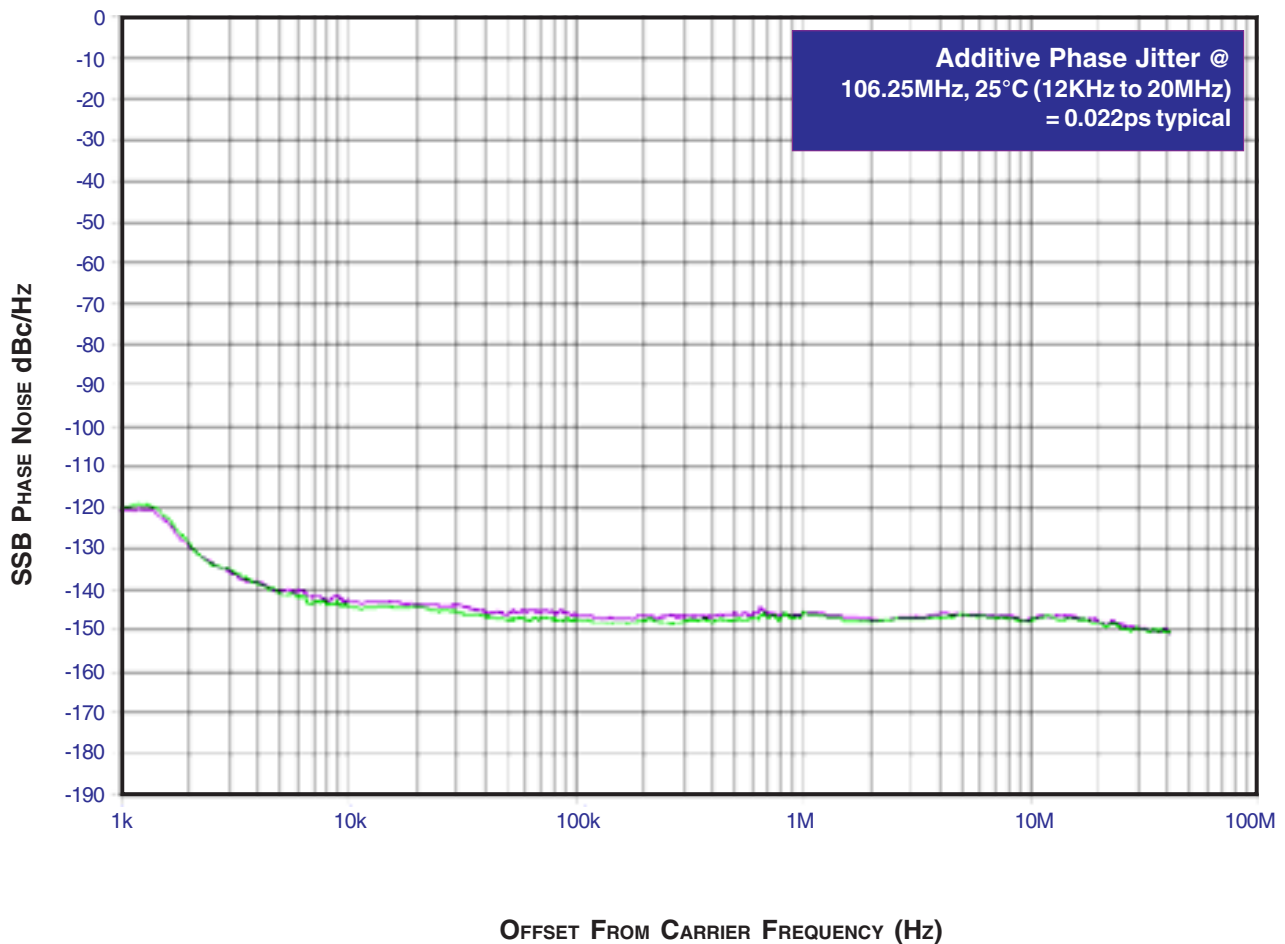
NOTE 4: This parameter is defined in accordance with JEDEC Standard 65.



ADDITIVE PHASE JITTER

The spectral purity in a band at a specific offset from the fundamental compared to the power of the fundamental is called the **dBc Phase Noise**. This value is normally expressed using a Phase noise plot and is most often the specified plot in many applications. Phase noise is defined as the ratio of the noise power present in a 1Hz band at a specified offset from the fundamental frequency to the power value of the fundamental. This ratio is expressed in decibels (dBm) or a ratio of the power in

the 1Hz band to the power in the fundamental. When the required offset is specified, the phase noise is called a **dBc** value, which simply means dBm at a specified offset from the fundamental. By investigating jitter in the frequency domain, we get a better understanding of its effects on the desired application over the entire time record of the signal. It is mathematically possible to calculate an expected bit error rate given a phase noise plot.



As with most timing specifications, phase noise measurements have issues. The primary issue relates to the limitations of the equipment. Often the noise floor of the equipment is higher than the noise floor of the device. This is illustrated above. The de-

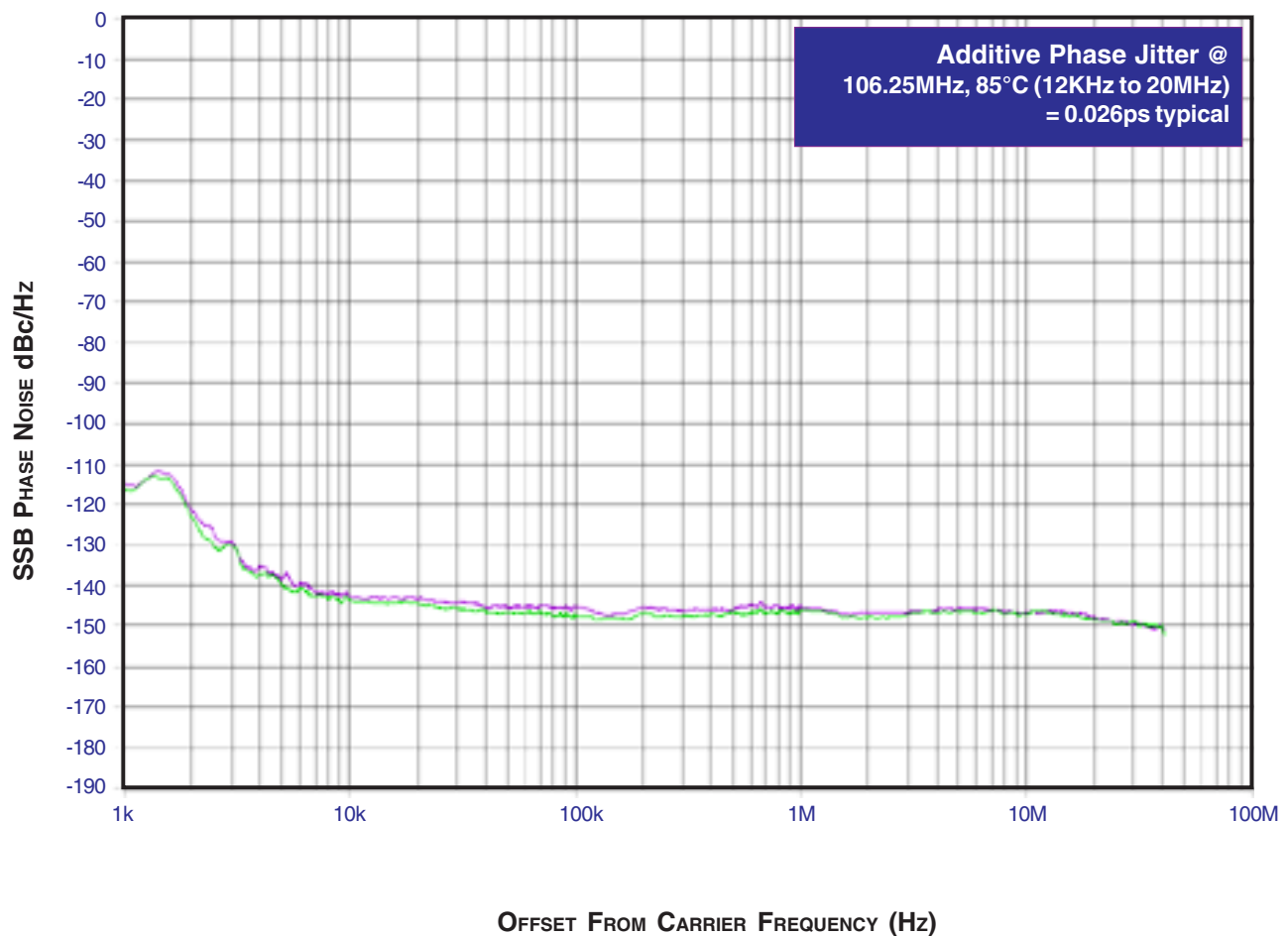
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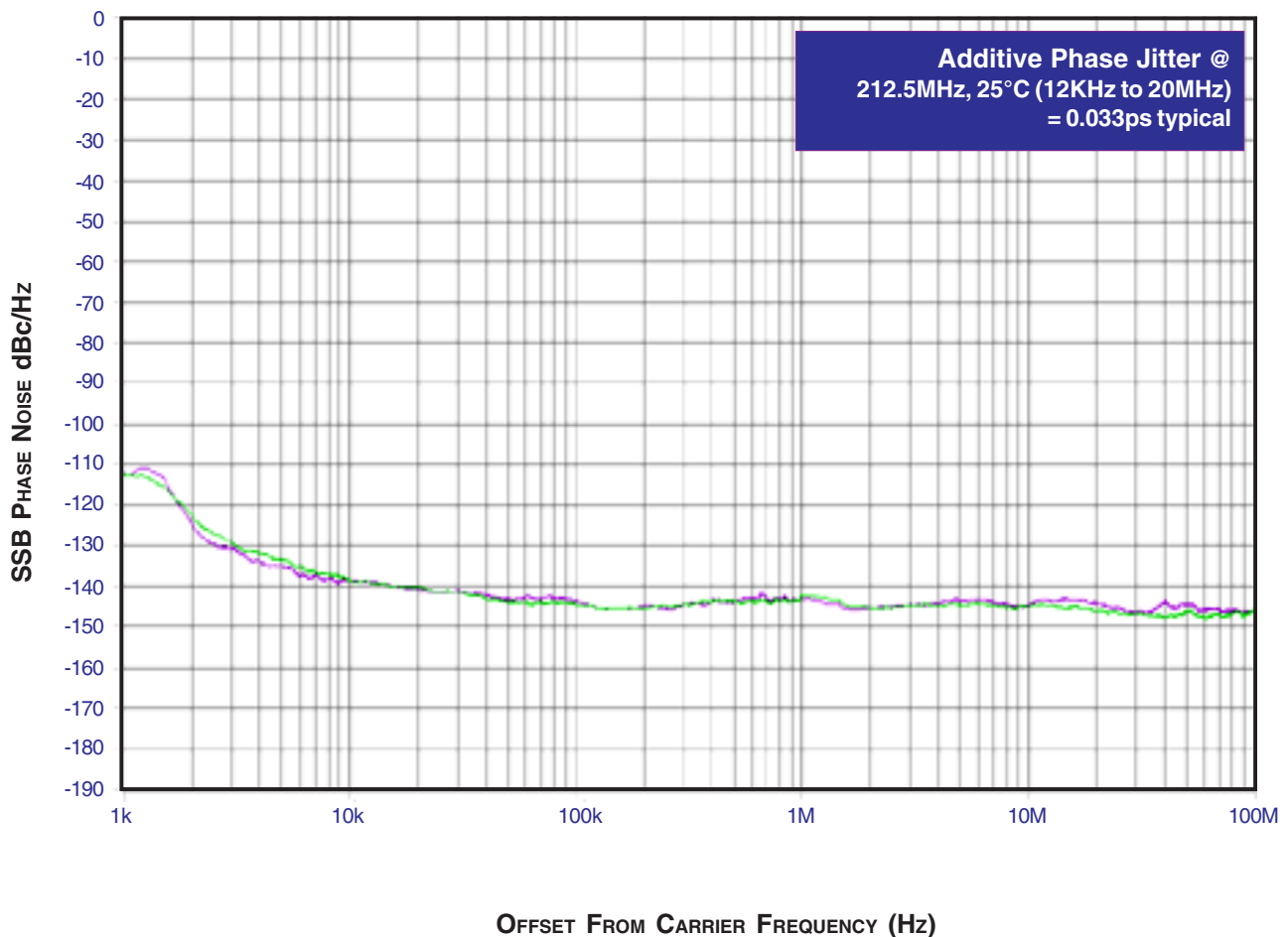
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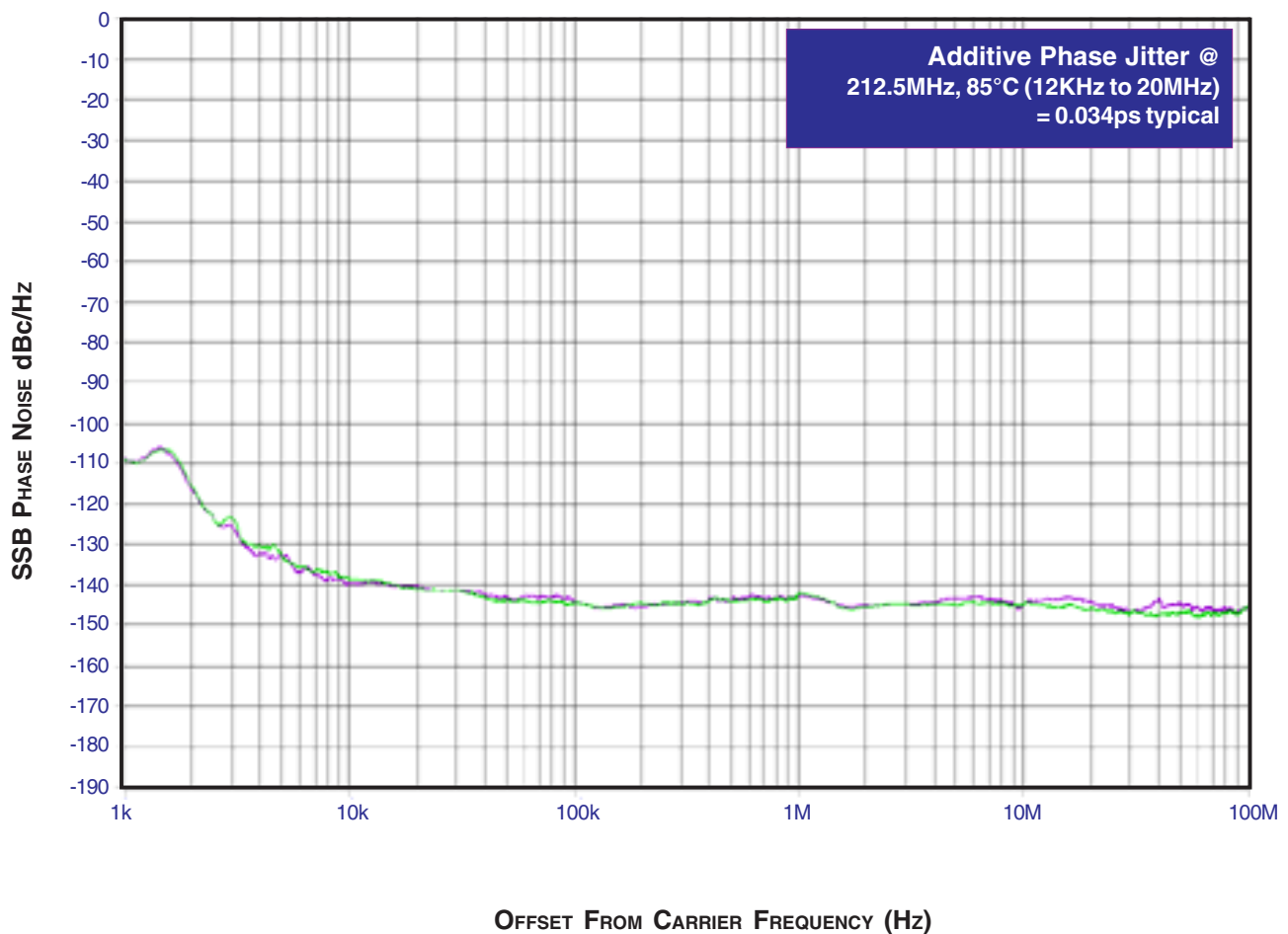
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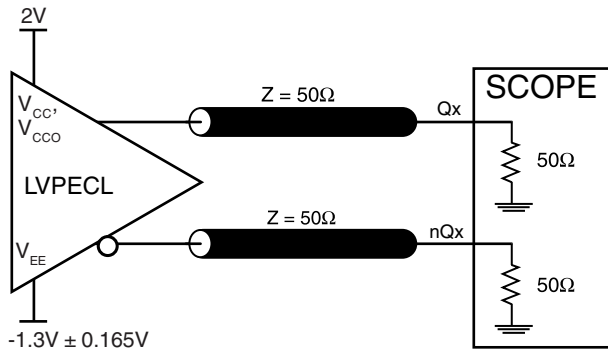
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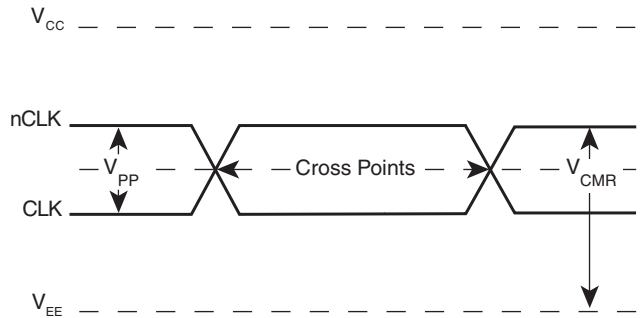




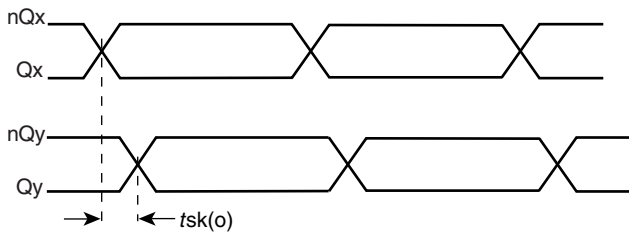
**PARAMETER MEASUREMENT INFORMATION**



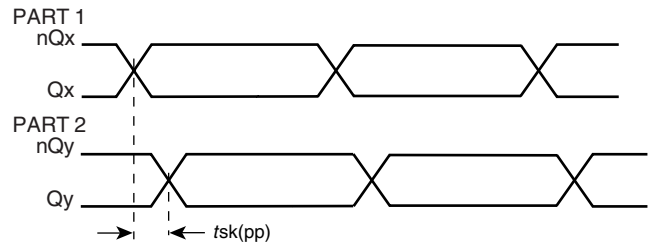
**3.3V OUTPUT LOAD AC TEST CIRCUIT**



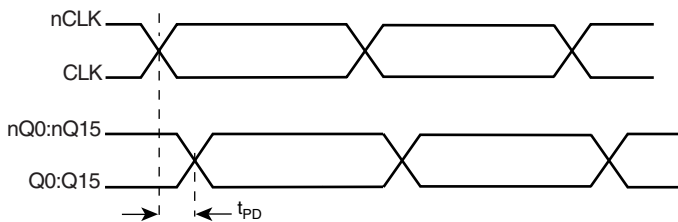
**DIFFERENTIAL INPUT LEVEL**



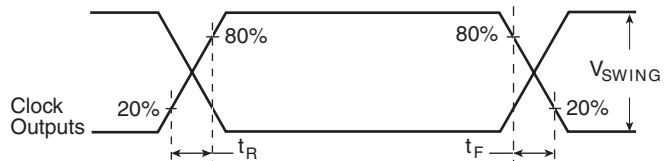
**OUTPUT SKEW**



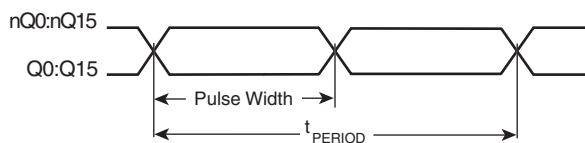
**PART-TO-PART SKEW**



**PROPAGATION DELAY**



**OUTPUT RISE/FALL TIME**



**OUTPUT DUTY CYCLE/PULSE WIDTH/PERIOD**

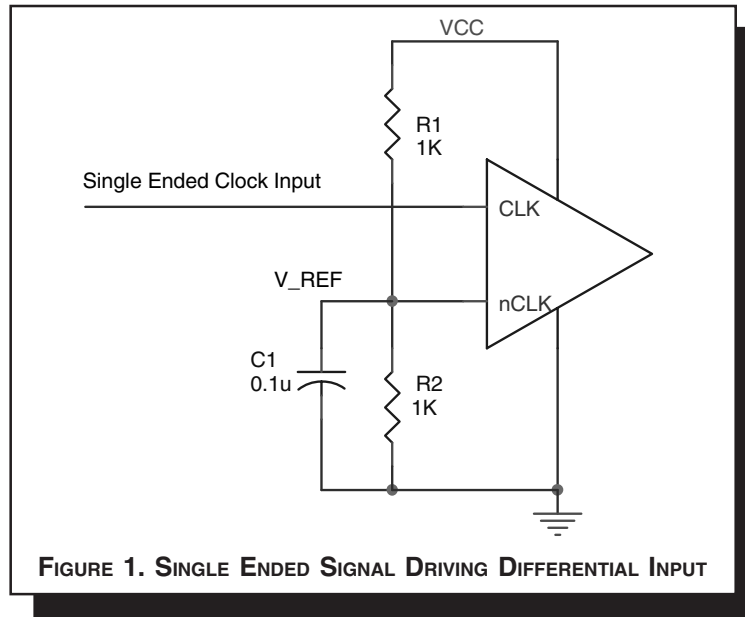
$$odc = \frac{t_{PW}}{t_{PERIOD}}$$



## APPLICATION INFORMATION

### WIRING THE DIFFERENTIAL INPUT TO ACCEPT SINGLE ENDED LEVELS

Figure 1 shows how the differential input can be wired to accept single ended levels. The reference voltage  $V_{REF} \approx V_{CC}/2$  is generated by the bias resistors R1, R2 and C1. This bias circuit should be located as close as possible to the input pin. The ratio of R1 and R2 might need to be adjusted to position the  $V_{REF}$  in the center of the input voltage swing. For example, if the input clock swing is only 2.5V and  $V_{CC} = 3.3V$ ,  $V_{REF}$  should be 1.25V and  $R2/R1 = 0.609$ .

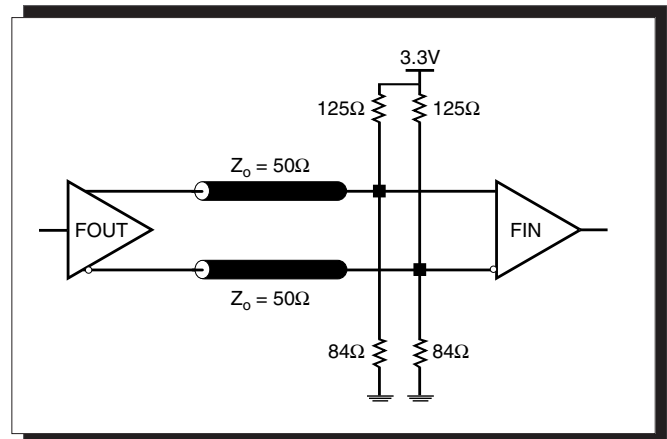
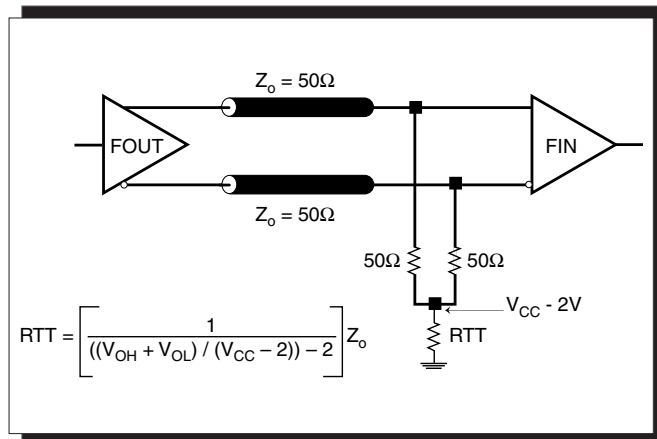


### TERMINATION FOR LVPECL OUTPUTS

The clock layout topology shown below is a typical termination for LVPECL outputs. The two different layouts mentioned are recommended only as guidelines.

FOUT and nFOUT are low impedance follower outputs that generate ECL/LVPECL compatible outputs. Therefore, terminating resistors (DC current path to ground) or current sources must be used for functionality. These outputs are designed to

drive 50Ω transmission lines. Matched impedance techniques should be used to maximize operating frequency and minimize signal distortion. Figures 2A and 2B show two different layouts which are recommended only as guidelines. Other suitable clock layouts may exist and it would be recommended that the board designers simulate to guarantee compatibility across all printed circuit and clock component process variations.

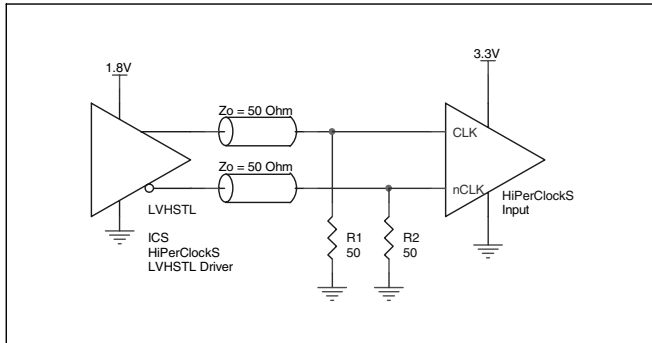




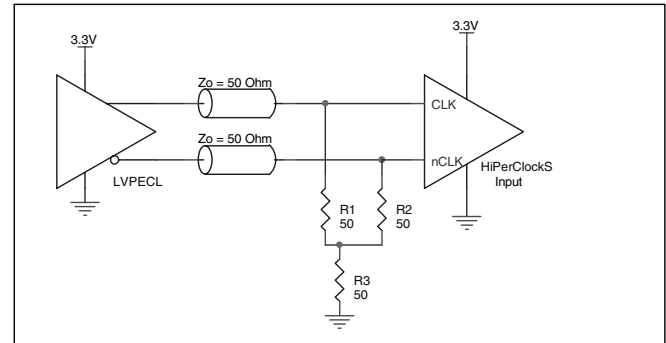
**DIFFERENTIAL CLOCK INPUT INTERFACE**

The CLK/nCLK accepts LVDS, LVPECL, LVHSTL, SSTL, HCSL and other differential signals. Both  $V_{SWING}$  and  $V_{OH}$  must meet the  $V_{PP}$  and  $V_{CMR}$  input requirements. Figures 3A to 3E show interface examples for the HiPerClockS CLK/nCLK input driven by the most common driver types. The input interfaces suggested

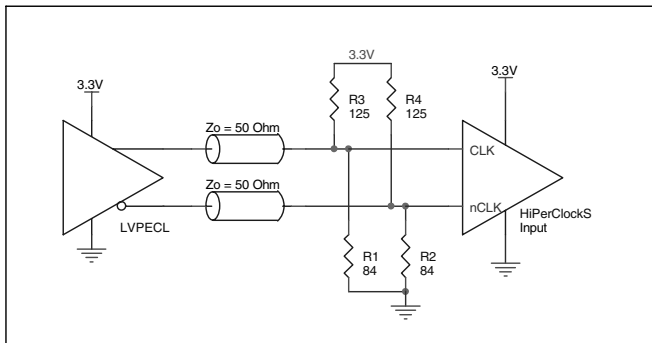
here are examples only. Please consult with the vendor of the driver component to confirm the driver termination requirements. For example in *Figure 3A*, the input termination applies for ICS HiPerClockS LVHSTL drivers. If you are using an LVHSTL driver from another vendor, use their termination recommendation.



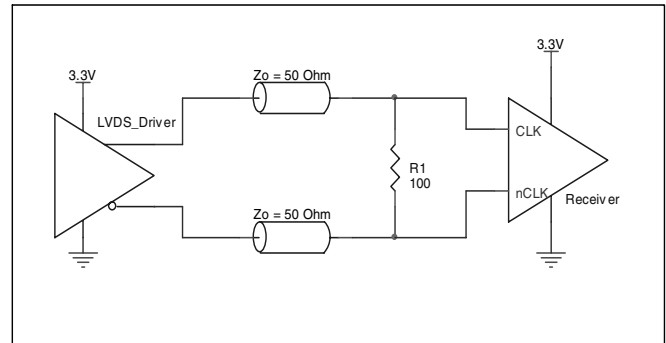
**FIGURE 3A. HiPerClockS CLK/nCLK INPUT DRIVEN BY ICS HiPerClockS LVHSTL DRIVER**



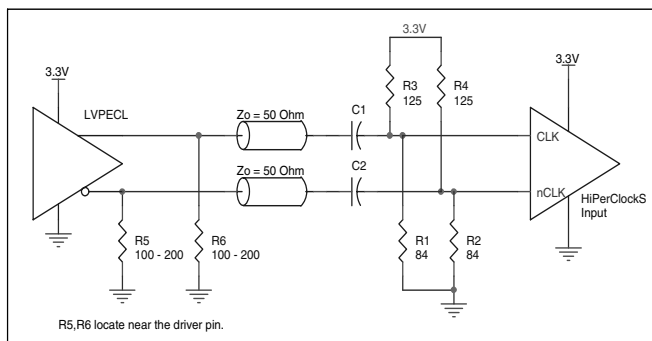
**FIGURE 3B. HiPerClockS CLK/nCLK INPUT DRIVEN BY 3.3V LVPECL DRIVER**



**FIGURE 3C. HiPerClockS CLK/nCLK INPUT DRIVEN BY 3.3V LVPECL DRIVER**



**FIGURE 3D. HiPerClockS CLK/nCLK INPUT DRIVEN BY 3.3V LVDS DRIVER**



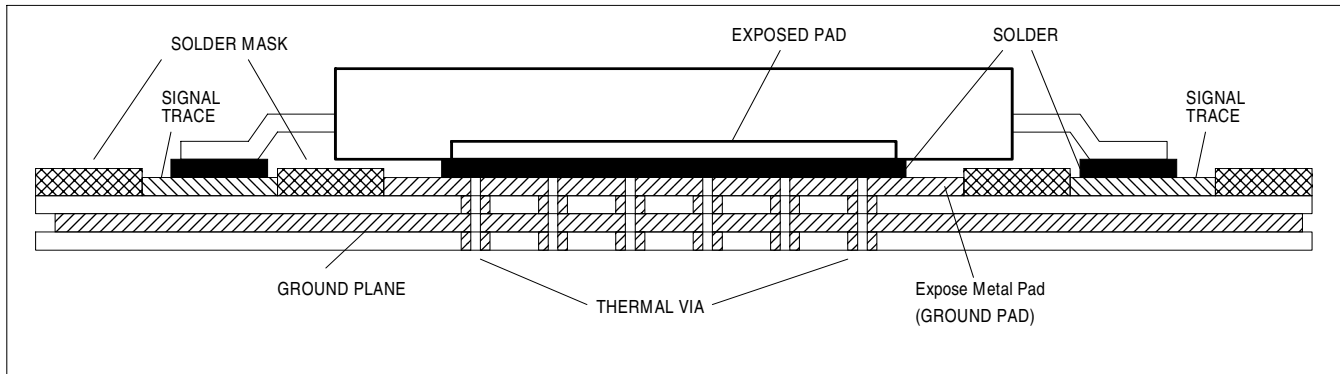
**FIGURE 3E. HiPerClockS CLK/nCLK INPUT DRIVEN BY 3.3V LVPECL DRIVER WITH AC COUPLE**



**THERMAL RELEASE PATH**

The expose metal pad provides heat transfer from the device to the P.C. board. The expose metal pad is ground pad connected to ground plane through thermal via. The exposed pad on the device to the exposed metal pad on the PCB is contacted through

solder as shown in *Figure 4*. For further information, please refer to the Application Note on Surface Mount Assembly of Amkor's Thermally /Electrically Enhance Leadframe Base Package, Amkor Technology.



**FIGURE 4. P.C. BOARD FOR EXPOSED PAD THERMAL RELEASE PATH EXAMPLE**



## POWER CONSIDERATIONS

This section provides information on power dissipation and junction temperature for the ICS8530I-01. Equations and example calculations are also provided.

### 1. Power Dissipation.

The total power dissipation for the ICS8530I-01 is the sum of the core power plus the power dissipated in the load(s). The following is the power dissipation for  $V_{CC} = 3.3V + 5\% = 3.465V$ , which gives worst case results.

**NOTE:** Please refer to Section 3 for details on calculating power dissipated in the load.

- Power (core)<sub>MAX</sub> =  $V_{CC\_MAX} * I_{EE\_MAX} = 3.465V * 115mA = 398.5mW$
- Power (outputs)<sub>MAX</sub> = **30mW/Loaded Output pair**  
If all outputs are loaded, the total power is  $16 * 30mW = 480mW$

**Total Power**<sub>MAX</sub> (3.465V, with all outputs switching) =  $398.5mW + 480mW = 878.5mW$

### 2. Junction Temperature.

Junction temperature,  $T_j$ , is the temperature at the junction of the bond wire and bond pad and directly affects the reliability of the device. The maximum recommended junction temperature for HiPerClockS™ devices is 125°C.

The equation for  $T_j$  is as follows:  $T_j = \theta_{JA} * Pd\_total + T_A$

$T_j$  = Junction Temperature

$\theta_{JA}$  = Junction-to-Ambient Thermal Resistance

$Pd\_total$  = Total Device Power Dissipation (example calculation is in section 1 above)

$T_A$  = Ambient Temperature

In order to calculate junction temperature, the appropriate junction-to-ambient thermal resistance  $\theta_{JA}$  must be used. Assuming a moderate air flow of 200 linear feet per minute and a multi-layer board, the appropriate value is 22.6°C/W per Table 6 below.

Therefore,  $T_j$  for an ambient temperature of 85°C with all outputs switching is:

$85°C + 0.879W * 22.6°C/W = 104.9°C$ . This is well below the limit of 125°C.

This calculation is only an example.  $T_j$  will obviously vary depending on the number of loaded outputs, supply voltage, air flow, and the type of board (single layer or multi-layer).

**TABLE 6. THERMAL RESISTANCE  $\theta_{JA}$  FOR 48-PIN TQFP, E-PAD FORCED CONVECTION**

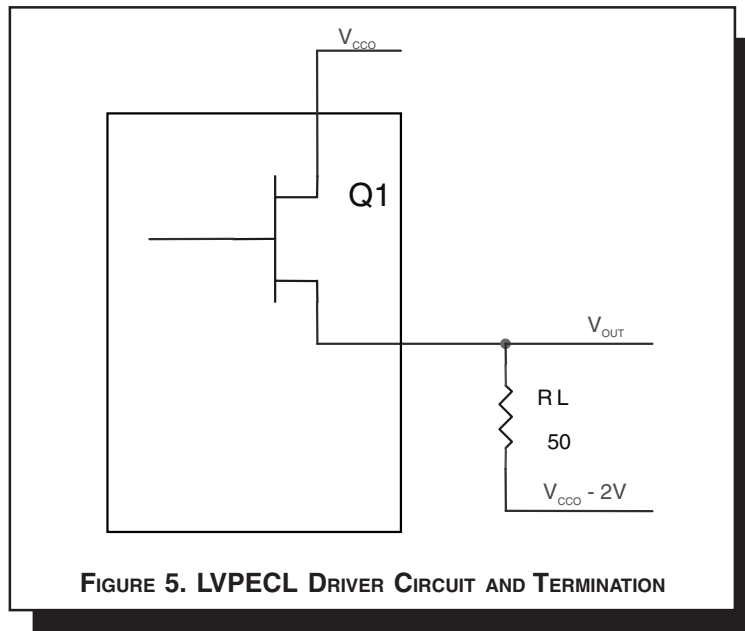
$\theta_{JA}$ by Velocity (Linear Feet per Minute)			
	0	200	500
Multi-Layer PCB, JEDEC Standard Test Boards	27.6°C/W	22.6°C/W	20.7°C/W



3. Calculations and Equations.

The purpose of this section is to derive the power dissipated into the load.

LVPECL output driver circuit and termination are shown in *Figure 5*.



**FIGURE 5. LVPECL DRIVER CIRCUIT AND TERMINATION**

To calculate worst case power dissipation into the load, use the following equations which assume a 50Ω load, and a termination voltage of  $V_{CCO} - 2V$ .

- For logic high,  $V_{OUT} = V_{OH\_MAX} = V_{CCO\_MAX} - 0.9V$

$$(V_{CCO\_MAX} - V_{OH\_MAX}) = 0.9V$$

- For logic low,  $V_{OUT} = V_{OL\_MAX} = V_{CCO\_MAX} - 1.7V$

$$(V_{CCO\_MAX} - V_{OL\_MAX}) = 1.7V$$

$Pd\_H$  is power dissipation when the output drives high.  
 $Pd\_L$  is the power dissipation when the output drives low.

$$Pd\_H = [(V_{OH\_MAX} - (V_{CCO\_MAX} - 2V))/R_L] * (V_{CCO\_MAX} - V_{OH\_MAX}) = [(2V - (V_{CCO\_MAX} - V_{OH\_MAX}))/R_L] * (V_{CCO\_MAX} - V_{OH\_MAX}) = [(2V - 0.9V)/50\Omega] * 0.9V = 19.8mW$$

$$Pd\_L = [(V_{OL\_MAX} - (V_{CCO\_MAX} - 2V))/R_L] * (V_{CCO\_MAX} - V_{OL\_MAX}) = [(2V - (V_{CCO\_MAX} - V_{OL\_MAX}))/R_L] * (V_{CCO\_MAX} - V_{OL\_MAX}) = [(2V - 1.7V)/50\Omega] * 1.7V = 10.2mW$$

Total Power Dissipation per output pair =  $Pd\_H + Pd\_L = 30mW$



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**PRELIMINARY**

**ICS8530I-01**

LOW SKEW, 1-TO-16

DIFFERENTIAL-TO-3.3V LVPECL FANOUT BUFFER

## RELIABILITY INFORMATION

TABLE 7.  $\theta_{JA}$  VS. AIR FLOW TABLE FOR 48 LEAD TQFP, E-PAD

$\theta_{JA}$ by Velocity (Linear Feet per Minute)			
	0	200	500
Multi-Layer PCB, JEDEC Standard Test Boards	27.6°C/W	22.6°C/W	20.7°C/W

### TRANSISTOR COUNT

The transistor count for ICS8530I-01 is: 930



PACKAGE OUTLINE - Y SUFFIX FOR 48 LEAD TQFP, E-PAD

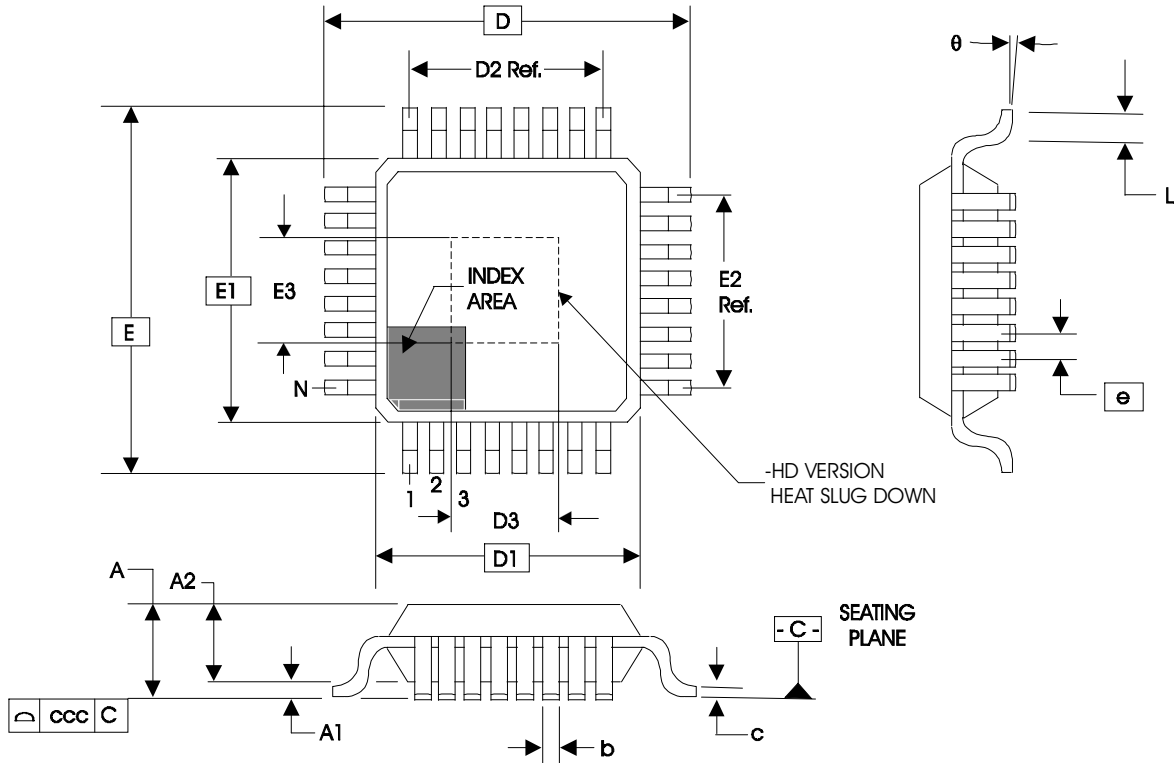


TABLE 8. PACKAGE DIMENSIONS

JEDEC VARIATION ALL DIMENSIONS IN MILLIMETERS			
SYMBOL	ABC - HD		
	MINIMUM	NOMINAL	MAXIMUM
N	48		
A	--	--	1.20
A1	0.05	--	0.15
A2	0.95	1.00	1.05
b	0.17	0.22	0.27
c	0.09		0.20
D	9.00 BASIC		
D1	7.00 BASIC		
D2	5.50 BASIC		
E	9.00 BASIC		
E1	7.00 BASIC		
E2	5.50 BASIC		
e	0.5 BASIC		
L	0.45	0.60	0.75
θ	0°		7°
ccc	--	--	0.08
D3 & E3	2.00		7.00

Reference Document: JEDEC Publication 95, MS-026





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**TABLE 9. ORDERING INFORMATION**

Part/Order Number	Marking	Package	Shipping Packaging	Temperature
ICS8530DYI-01	ICS8530DYI01	48 Lead TQFP, E-Pad	tray	-40°C to 85°C
ICS8530DYI-01T	ICS8530DYI01	48 Lead TQFP, E-Pad	1000 tape & reel	-40°C to 85°C

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