

# Design Idea DI-29

## DPA-Switch<sup>®</sup>

### 25 W Flyback DC-DC Converter



Application	Device	Power Output	Input Voltage	Output Voltage	Topology
DC-DC Converter	DPA425R	25 W	36 - 75 VDC	7 V	Flyback

### Design Highlights

- Extremely low component count
- High efficiency – 85% using Schottky rectifiers
- No current sense resistor or transformer required
- Accurate input under/over voltage meets ETSI standards
- Operates to zero load with no pre-load required
- Output overload, open loop and thermally protected
- 400 kHz operation minimizes size of magnetics

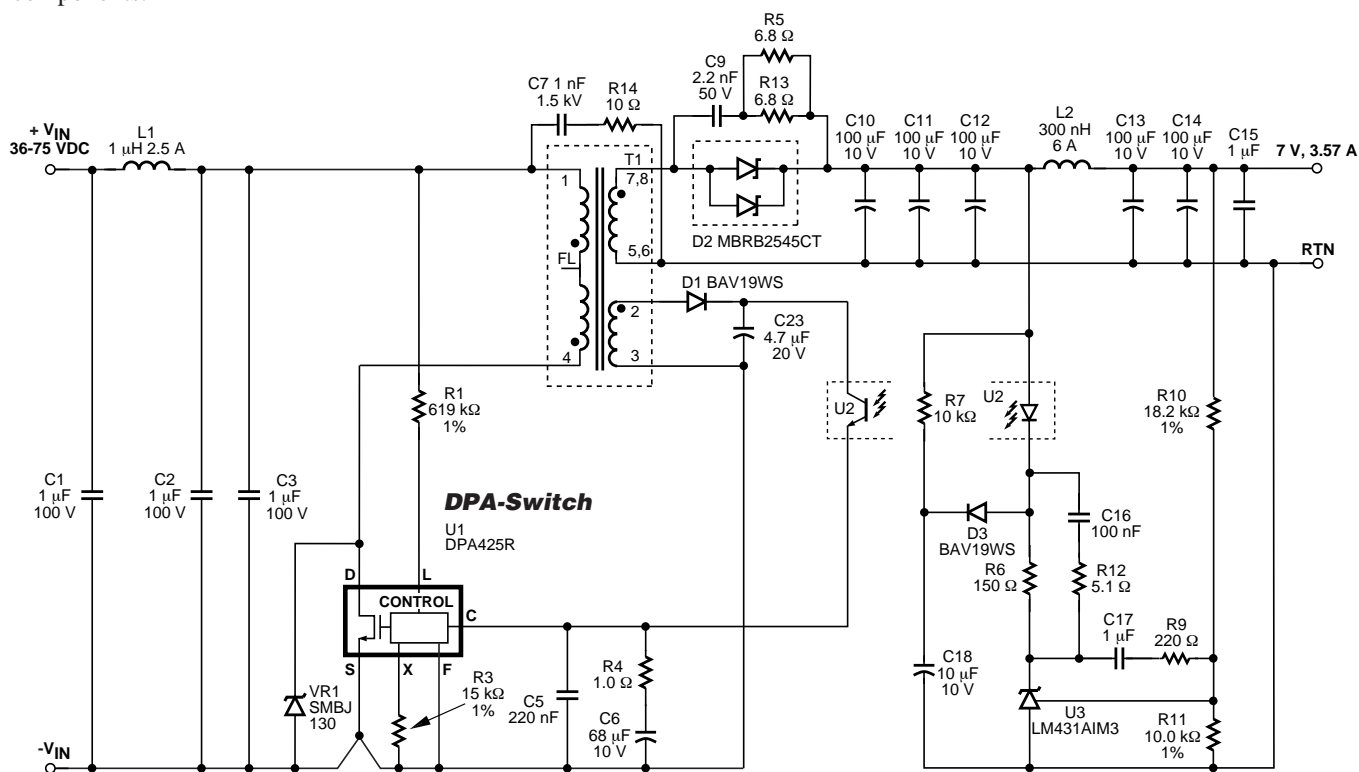
### Operation

DPA-Switch greatly simplifies the design compared to a discrete implementation. Resistor R1 programs the input under/over voltages to 33 V and 86 V, respectively. Including tolerances these thresholds guarantee the converter is operational between 36 V and 75 V, without the cost of additional line sense components.

Resistor R3 programs the internal current limit of the DPA425R to 50% of nominal. The larger DPA-Switch selection reduces conduction losses, raising efficiency without circuit changes or increased overload power.

Zener VR1 clamps leakage inductance spikes, keeping the DRAIN voltage below  $BV_{DSS}$ . The bias supply for U1 is provided from an auxiliary flyback transformer winding.

On the secondary, a snubber across D2 (C9, R5 and R13) limits the secondary leakage inductance spikes generated by diode reverse recovery. Inductor L2, C13 and C14 form a post-filter to reduce high frequency output switching ripple. A soft-finish network, C18, D3 and R7, eliminates output turn-on overshoot. The remaining components provide output voltage regulation and loop compensation.



PI-3012-120902

Figure 1. DPA-Switch Flyback DC-DC Converter Schematic.

## Key Design Points

- For nominal under-voltage set point  $V_{UV}$ :  
 $R1 = (V_{UV} - 2.35 \text{ V}) / 50 \mu\text{A}$ ,  $V_{OV} = (R1 \times 135 \mu\text{A}) + 2.5 \text{ V}$ .
- Zener VR1 voltage is 130 V to safely limit the DRAIN voltage below  $V_{DSS}$  of 200 V.
- Opto U2 should have a CTR of between 100% and 200% for optimum loop stability.
- Set resonance of L2 and C13 + C14 to beyond loop crossover frequency (typically 5% to 10% of switching frequency).
- Good layout practices should be followed:
  - Locate C5, C6 and R4 close to U1, with grounds returned to the SOURCE pin.
  - Primary return should be connected to the *DPA-Switch* tab, not the SOURCE pin.
  - Minimize the primary and secondary loop areas to reduce parasitic leakage inductance.

Transformer Parameters	
Core Material	PR1408 Siemens N87 Gap for 340 nH/T <sup>2</sup>
Bobbin	P1408 8 pin (B&B B-096 or equivalent)
Winding Details	Primary: 6T + 6T, 2 x 27 AWG Secondary: 3T, 4 x 25 AWG Bias: 6T, 32 AWG
Winding Order (pin numbers)	Primary (4-FL), tape, Bias (2-3), tape, Secondary (5,6-7,8), tape, Primary (FL-1), tape
Inductance	Primary: 49 $\mu\text{H} \pm 10\%$ , Leakage: 1 $\mu\text{H}$ (max)
Primary Resonant Frequency	3.8 MHz (minimum)

Table 1. Transformer Construction Information.

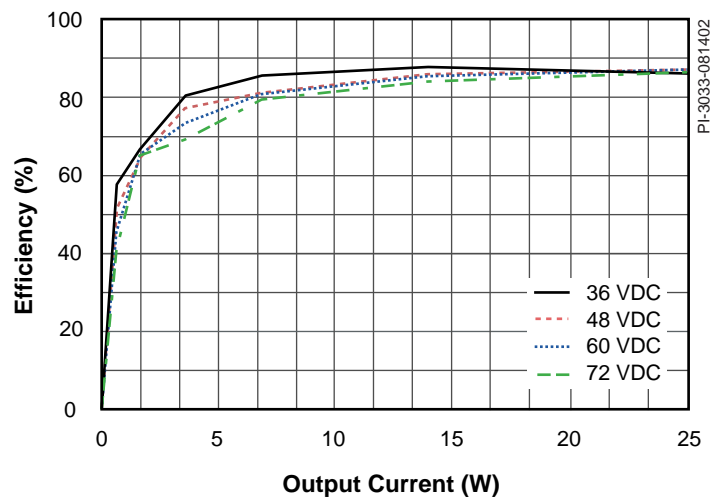


Figure 2. Efficiency vs. Output Power.

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### WORLD HEADQUARTERS AMERICAS

Power Integrations, Inc.  
San Jose, CA 95138 USA  
Customer Service:  
Phone: +1 408-414-9665  
Fax: +1 408-414-9765  
e-mail: [usasales@powerint.com](mailto:usasales@powerint.com)

### CHINA

Power Integrations International  
Holdings, Inc.  
China  
Phone: +86-755-8367-5143  
Fax: +86-755-8377-9610  
e-mail: [chinasales@powerint.com](mailto:chinasales@powerint.com)

### EUROPE & AFRICA

Power Integrations (Europe) Ltd.  
United Kingdom  
Phone: +44-1344-462-300  
Fax: +44-1344-311-732  
e-mail: [eurosales@powerint.com](mailto:eurosales@powerint.com)

### KOREA

Power Integrations  
International Holdings, Inc.  
Seoul, Korea  
Phone: +82-2-782-2840  
Fax: +82-2-782-4427  
e-mail: [koreasales@powerint.com](mailto:koreasales@powerint.com)

### SINGAPORE

Power Integrations, Singapore  
Republic of Singapore 308900  
Phone: +65-6358-2160  
Fax: +65-6358-2015  
e-mail: [singaporesales@powerint.com](mailto:singaporesales@powerint.com)

### JAPAN

Power Integrations, K.K.  
Keihin-Tatemono 1st Bldg.  
Japan  
Phone: +81-45-471-1021  
Fax: +81-45-471-3717  
e-mail: [japansales@powerint.com](mailto:japansales@powerint.com)

### APPLICATIONS HOTLINE

World Wide +1-408-414-9660

### TAIWAN

Power Integrations  
International Holdings, Inc.  
Taipei, Taiwan  
Phone: +886-2-2727-1221  
Fax: +886-2-2727-1223  
e-mail: [taiwansales@powerint.com](mailto:taiwansales@powerint.com)

### INDIA (Technical Support)

Innovatech  
Bangalore, India  
Phone: +91-80-226-6023  
Fax: +91-80-228-9727  
e-mail: [indiasales@powerint.com](mailto:indiasales@powerint.com)

### APPLICATIONS FAX

World Wide +1-408-414-9760

