



## Implementing IIR Digital Filters

### INTRODUCTION

This application note describes the implementation of various digital filters using the PIC17C42, the first member of Microchip's 2nd generation 8-bit microcontrollers. The PIC17C42 is a very high speed 8-bit microcontroller with an instruction cycle time of 250ns (@ 16 MHz input clock). Even though PIC17C42 is an 8-bit device, it's high speed and efficient instruction set allows implementation of digital filters for practical applications. Traditionally digital filters are implemented using expensive Digital Signal Processors (DSPs). In a system the DSP is normally a slave processor being controlled by either an 8- or 16-bit microcontroller. Where sampling rates are not high (esp. in mechanical control systems), a single chip solution is possible using the PIC17C42.

This application note provides a few examples of implementing digital filters. Example code for 2nd order Infinite Impulse Response (IIR) filters is given. The following type of filters are implemented:

- Low Pass
- High Pass
- Band Pass
- Band Stop (notch) filter

This application note does not explain how to design a filter. Filter design theory is well established and is beyond the scope of this application note. It is assumed that a filter is designed according to the desired specifications. The desired digital filters may be designed using either standard techniques or using commonly available digital filter design software packages.

Finite Impulse Response (FIR) filters have many advantages over IIR filters, but are much more resource intensive (both in terms of execution time and RAM). On the other hand, IIR filters are quite attractive for implementing with the PIC17C42 resources. Especially where phase information is not so important, IIR filters are a good choice (FIR filters have a linear phase response). Of the various forms used for realizing digital filters (like, Direct form, Direct II form, Cascade form, Parallel, Lattice structure, etc.) the Direct II form is used in this application note. It is easy to understand and simple macros can be built using these structures.

### THEORY OF OPERATION

Digital filters in most cases assume the following form of relationship between the output and input sequences.

$$y(n) = - \sum_{i=0}^M a_i y(n - i) + \sum_{j=0}^N b_j x(n - j)$$

The above equation basically states that the present output is a weighted sum of the past inputs and past outputs. In case of FIR filters, the weighted constants  $a_i=0$  and in case of IIR filters, at least one of the  $a_i$  constant is non zero. In case of IIR, the above formula may be re written in terms of Z transform as:

$$H(z) = \frac{Y(z)}{X(z)} = \frac{\sum_{k=0}^M b_k z^{-k}}{1 + \sum_{k=1}^N a_k z^{-k}}$$

The above equation can further be rewritten in difference equation format as follows:

$$y(n) = - \sum_{i=1}^M a_i y(n - i) + \sum_{j=0}^N b_j x(n - j)$$

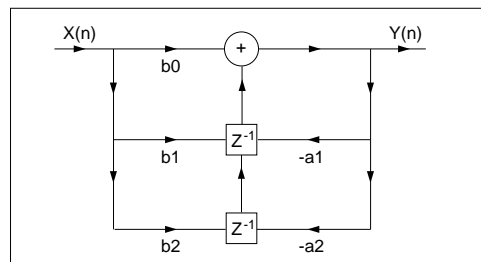
Realization of the above equation is called as the Direct Form II structure. For example, in case of a second order structure,  $M=N=2$ , gives the following difference equations :

$$d(n) = x(n) + a_1 d(n-1) + a_2 d(n-2) \quad (1)$$

$$y(n) = b_0 d(n) + b_1 d(n-1) + b_2 d(n-2) \quad (2)$$

The above difference equations may be represented as shown in Figure 1.

**FIGURE 1 - 2ND ORDER DIRECT FORM II STRUCTURE (TRANPOSED)**



# Implementing IIR Digital Filters

The structure as shown in Figure 1 may be cascaded to attain a higher order filter. For example, if two stages are cascaded together, a 4th Order IIR Filter is obtained. This way, the output of the 1st stage becomes the input to the second stage. Multiple order filters are thus implemented by cascading a 2nd order filter structure as shown in Figure 1.

## IMPLEMENTATION

A 4th order IIR Filter is implemented by cascading two structures shown in Figure 1. The output Y (output of each filter stage) is computed by direct implementation of Equations (1) & (2). Since each stage is similar algorithmically, it is implemented as a Macro using Microchip's Assembler/Linker for PIC17C42. This Macro (labelled "BIQUAD") is called twice for implementing a 4th order filter. The output of the 1st stage is directly fed to the input of the second stage without any scaling.

Scaling is required depending on the particular application. The user can modify the code very easily without any penalty on speed. Also, saturation arithmetic is not used. Overflows can be avoided by limiting the input sequence amplitude. All numbers are assumed to be 16 bits in Q15 format (15 decimal points, MSB is sign bit). Thus the user must scale and sign extend the input sequence accordingly. For example, if the input is from a 12-bit A/D converter, the user must sign extend the 12-bit input if bit 11 is a one.

The BIQUAD macro is a generic macro and can be used for all IIR filters whether it is Low Pass, High Pass, Band Pass or Band Stop. A general purpose 16x16 multiplier routine is also provided. This routine is implemented as a straight line code for speed considerations.

The 4th order IIR filter implemented is a Low Pass Filter with the specifications as shown in Table 1.

**TABLE 1 - FILTER CONSTANTS**

	BAND1	BAND2
Lower Band Edge	0.0	600 Hz
Upper Band Edge	500 Hz	1 KHz
Nominal Gain	1.0	0.0
Nominal Ripple	0.01	0.05
Maximum Ripple	0.00906	0.04601
Ripple in dB	0.07830	-26.75
Sampling Frequency = 2 KHz		

The Low Pass Filter is designed using a digital filter design package (DFDP by Atlanta Signal Processors Inc.). The filter package produces filter constants of the structure shown in Table 1. Table 2 shows the filter coefficients that are obtained for the above Low Pass filter specification.

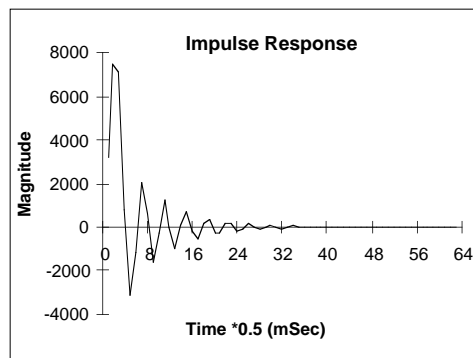
**TABLE 2 - FILTER COEFFICIENTS**

Co-efficients	a1	a2	b0	b1	b2
Stage 1	-0.133331	0.167145	0.285431	0.462921	0.285431
Stage 2	0.147827	0.765900	0.698273	0.499908	0.698273

The above filter co-efficients (5 per stage) are quantized to Q15 format (i.e they are multiplied by 32768) and saved in program memory (starting at label "\_coeff\_lpass"). The constants for both the stages are read into data memory using TLRD and TABLRD instructions in the Initialization Routine (labelled "initFilter"). The user may read the coefficients of only one stage at a time and save some RAM at the expense of speed.

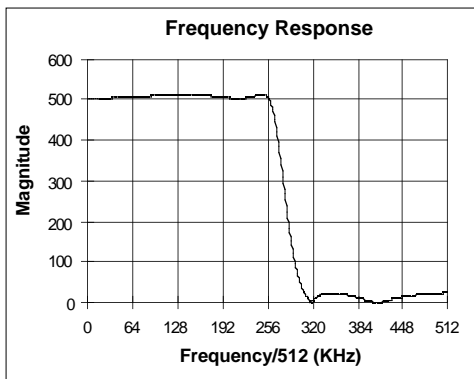
The sample 4th order Low Pass IIR Filter is tested by analyzing the impulse response of the filter. An impulse signal is fed as input to the filter. This is simulated by forcing the input to the filter by a large quantity (say 7F00h) on the first input sample, and the all zeros from the 2nd sample onwards. The output sequence is the filter's impulse response and is captured into the PICMASTER's (Microchip's Universal In-Circuit Emulator) real time trace buffer. This captured data from PICMASTER is saved to file and analyzed. Analysis was done using MathCad for Windows and DSP Analysis program from Burr-Brown (DSPLAY). The Fourier Transform of this Impulse response of the filter should display the Filter's frequency response, in this case being a Low Pass type. The plots of the impulse response and the frequency response are shown in figures below.

**FIGURE 2 - IMPULSE RESPONSE CAPTURED FROM PIC-MASTER**

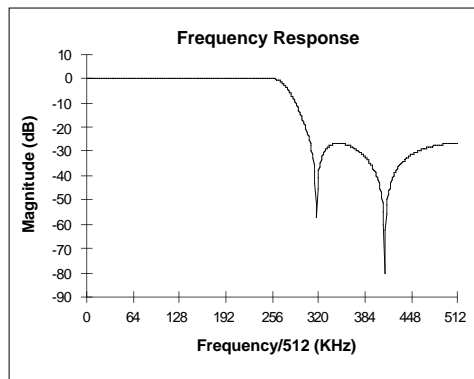


DSPLAY is a trademark of Burr-Brown  
 DFDP is a trademark of Atlanta Signal Processing Inc.  
 Windows is a trademark of Microsoft Corporation  
 MathCad is a registered trademark of MathSoft, Inc.

**FIGURE 3 - SPECTRUM COMPUTED FROM IMPULSE RESPONSE**



**FIGURE 4 - LOG MAGNITUDE SPECTRUM OF IMPULSE RESPONSE**



## PERFORMANCE

The resource requirements for filter implementations using PIC17C42 is given below. These numbers can be used to determine whether a higher order filter can be executed in real time. The same information may be used to determine the highest sampling rate possible.

## FILTER APPLICATIONS

Digital filters find applications in many areas especially involving processing of real world signals. In some applications like ABS systems in an automobile, digital filtering becomes a must. In this case elimination of noise (especially glitches and false readings of sensors) is very critical and thus the requirement of digital signal processing.

Digital filters are also needed in Process Control where notch filters and low pass filters are desired because the signals from sensors are transmitted over long lines, especially in a very noisy environment. In this case typically a notch filter (centering 50Hz or 60Hz) is used. In case of eliminating background noise, a band stop filter (e.g. 40Hz to 120Hz) is used. The sample code given in this application note can be used to design a feedback control system's digital compensator. For example, a typical DC Motor's digital compensator (like dead beat compensator) is of second order and has the same filter structure as is implemented in this application note.

**TABLE 3 - RESOURCE REQUIREMENTS**

Timing (Cycles)†	#of Filter Stages*775 + 16
Program Memory (locations)†	#of Filter Stages*68 + 290
RAM (File Registers)	#of Filter Stages*16+16

†: The above numbers do not include the initialization routine.

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**TABLE 4 - RESOURCE REQUIREMENTS**

Filter Order	Cycles	Real Time (@ 16 MHz)	Maximum Sampling	Program Memory†	RAM
2	791	197.75 uSec	5.05 KHz	358	32
4	1566	391.5 uSec	2.55 KHz	426	48
6	2341	585.25 uSec	1.7 KHz	494	64
8	3116	779.0 uSec	1.28 KHz	562	80
10	3891	972.75 uSec	1.0 KHz	630	96

†: The above numbers do not include the initialization routine.

# Implementing IIR Digital Filters

## APPENDIX A: IIR.LST

MPASM B0.54

PAGE 1

Digital IIR Filter Using PIC17C42

```

                                TITLE  "Digital IIR Filter Using PIC17C42"
                                LIST   P=17C42, C=120, T=ON, R=DEC, N=0

;
include "17c42.h"

;
include "17c42.mac"

;
include "17c42iir.mac"
;*****
;
;                                PIC17C42 MACRO
;
;                                Macro For A Bi-Quad IIR Filter
;                                2nd order Direct Form (Transposed) Type
;
; Filter co-efficients B0 & B2 are assumed equal
;
; The difference equations for each cascade section is given by :
;     Y(n) = B0*D(n) + B1*D(n-1) + B2*D(n-2)
;     D(n) = X(n) - A1*D(n-1) - A2*D(n-2)
;     where X(n) = input sample, Y(n) = output of filter
;     and A1, A2, B0, B1, B2 are the Filter Co-efficients
;
; The above difference equations are only for 1 section of a
; 2nd order Direct_Form II Filter structure (IIR)
;
; NOTE :
;     It is possible to design the above structures
;     such that the co-efficients B0 = B2. If this is the
;     case,
;
;         Y(n) = B0*[D(n) + D(n-2)] = B2*[D(n) + D(n-2)]
;     This way, one multiplication can be avoided
;
; If a 4th order filter is to be implemented, the output of
; the 1st structure should be input to the 2nd cascade section
;
; Timing (WORST CASE) :
;
;                                59+4*179 = 775 Cycles
;                                (194 uS @ 16 Mhz)
;
; Program Memory :
;
;                                63 locations
;
;*****
;     The sample filters are desined so that B0=B2
;     This saves 1 multiplication
;
0001 B0_EQUALS_B2    equ    TRUE
;
;*****
; Parameters to BIQUAD Macro :
;     Filter Constants A1, A2, B0, B1, B2
;     & D(n), D(n-1), D(n-2), filter stage #
;
BIQUAD MACRO Ax1,Ax2,Bx0,Bx1,Dn,Dn_1,Dn_2,stage

;
; Compute Ax2*D(n-2)
```

```

;
MOVFP16 Dn_2,AARG          ; D(n-2) = multiplier
MOVFP16 Ax2,BARG          ; A2 = multiplicand
call    DblMult           ; (ACCd,ACCc) = A2*D(n-2)
;
; Add product to output of 1st section
; Save result in 32 bit Accumulator
;
ADD32   DPX,ACC
;
; Compute A1*D(n-1)
;
MOVFP16 Dn_1,AARG          ; AARG = D(n-2) = multiplier
MOVFP16 Ax1,BARG          ; BARG = A2 = multiplicand
call    DblMult           ; (ACCd,ACCc) = A1*D(n-1)
;
; Compute A1*D(n-1) + A2*D(n-2) + output of previous section
; multiplications already done, so simply perform a 32 bit add
; of previously obtained multiplication results
;
ADD32   DPX,ACC           ; ACC = A1*D(n-1)+A2*D(n-2)+(output of 1st
;
; save the upper 16 bits of D(n) from the 32 bit accumulator
; left shift the result by 1, to adjust the decimal point after
; a Q15*Q15 multiplication
;
rlcf    ACC+B1,w
rlcf    ACC+B2,w
movwf   Dn
rlcf    ACC+B3,w          ; decimal adjust ( mult by 2)
movwf   Dn+B1
;
; Compute B2 * [D(n) + D(n-2)]
;
if B0_EQUALS_B2
ADD16ACC Dn_2,Dn,AARG      ; AARG = Dn + D(n-2) = multiplier
MOVFP16 Bx0,BARG          ; BARG = A2 = multiplicand
call    DblMult           ; (ACCd,ACCc) = B2*[D(n)+D(n-2)]
MOVFP32 DPX,ACC

else
MOVFP16 Bx0,BARG
MOVFP16 Dn,AARG
call    DblMult           ; B0*D(n)
MOVFP32 DPX,ACC

MOVFP16 Bx2,BARG
MOVFP16 Dn_2,AARG
call    DblMult           ; B2*D(n-2)
ADD32   DPX,ACC
endif
;
; Shift down D(n-1) to D(n-2) after D(n-2) usage is no longer required.
; This way in the next iteration D(n-2) is equal to the present D(n-1)
;
movfp   Dn_1,AARG+B0
movfp   AARG+B0,Dn_2      ; Shift down D(n-1)
movfp   Dn_1+B1,AARG+B1
movfp   AARG+B1,Dn_2+B1  ; AARG = D(n-1) = multiplier
MOVFP16 Bx1,BARG          ; BARG = B1 = multiplicand
call    DblMult           ; (ACCd,ACCc) = B1*D(n-1)
;
; Compute Output Y = B1*D(n-1) + B2*D(n-2) + B0*D(n)
; = B1*D(n-1) + B0*[D(n) + D(n-2)]
; Since all multiplications are already done, simply perform a
; 32 bit addition
;
ADD32   DPX,ACC           ; ACC = B1*D(n-1) + B2*D(n-2) + B0*D(n)

```

# Implementing IIR Digital Filters

```
;
; Shift down D(n) to D(n-1) so that in the next iteration, the new
; D(n-1) is the present D(n)
;
MOV16   Dn,Dn_1           ; Shift down D(n) to D(n-1)
;
        ENDM

;
;*****
;                Second Order Direct Form IIR Filter
;
;
; In the code given below, a 4th order IIR Elliptic Lowpass Filter
; is implemented. Other order filters may be implemented by
; taking the following example code as a basis.
;
; The specifications of the filter are :
;
; Sampling Frequency = 2.0 Khz
;
; Filter Type = 4th Order Elliptic Lowpass Filter
;
;
; Lower Band Edge           Band1           Band2
; Upper Band Edge           500 Hz           1 Khz
; Nominal Gain               1.0             0.0
; Nominal Ripple             0.01            0.05
; Maximum Ripple             0.00906         0.04601
; Ripple in dB               0.07830         -26.75
;
; The Filter Co-efficients for the above specifications
; of the filter are computed as follows :
;
; 1st Section :
;
; A11 = -0.133331
; A12 = 0.167145
; B10 = 0.285431
; B11 = 0.462921
; B12 = 0.285431
;
; 2nd Section
;
; A21 = 0.147827
; A22 = 0.765900
; B20 = 0.698273
; B21 = 0.499908
; B22 = 0.698273
;
;
; Performance (WORST Case):
;
; Cycles   = #of Filter Stages*775 + 16
;           = 2*775+16 = 1566 Cycles
;           ( 391 uSec)
;           per each sample. Initialization
;           time after reset is not counted
;           Timing measured with B0_EQUALS_B2
;           set to TRUE (see BIQUAD Macro for
;           explanation
;
; Program Memory :
;
;           = 16+ # of FilterStages * (BIQUAD
;           + filter co-efficients)
;           + multiplier
;           = 16+2*(63+5)+274 = 421 locations
;           (excluding initialization)
;
;
; RAM usage = 48 file registers
; RAM usage/each additional stage = 16 file regs
;
```

# Implementing IIR Digital Filters

```

;
;
;           This time is less than 2 Khz (500 uSec),
;           which means real time filtering is possible
;
;*****
;*****
;
;
;           CBLOCK 0
0000 0004          B0,B1,B2,B3
;           ENDC
;*****
;
;           CBLOCK 0x18
0018 0004          DPX,DPX1,DPX2,DPX3          ; arithmetic accumulator
001C 0004          AARG,AARG1,BARG,BARG1      ; multiply arguments
;           ENDC
;
;           CBLOCK
0020 0002          Dn1, Dn1_Hi
0022 0002          Dn1_1, Dn1_1_Hi
0024 0002          Dn1_2, Dn1_2_Hi
0026 0000
0026 0002          Dn2, Dn2_Hi
0028 0002          Dn2_1, Dn2_1_Hi
002A 0002          Dn2_2, Dn2_2_Hi
;           ENDC
;
;           CBLOCK
002C 0002          A11, A11_Hi
002E 0002          A12, A12_Hi
0030 0002          B10, B10_Hi
0032 0002          B11, B11_Hi                ; 1st Section Filter Co-efficients
0034 0002          B12, B12_Hi
0036 0000
0036 0002          A21, A21_Hi
0038 0002          A22, A22_Hi
003A 0002          B20, B20_Hi
003C 0002          B21, B21_Hi                ; 2nd Section Filter Co-efficients
003E 0002          B22, B22_Hi
;           ENDC
;
;           CBLOCK
0040 0002          X, X1                      ; 16 bits of input stream
0042 0002          Y, Y1                      ; 16 bits of filter output
0044 0000
0044 0004          ACC, ACC1, ACC2, ACC3      ; 32 bit accumulator for computations
;           ENDC
;
;           .set      2
0002          FltStage      .set      2
000A          NumCoeff      equ      (5*FltStage) ; 5 Co-eff per stage
;
;           LPASS      .set      TRUE
0000          HPASS      .set      FALSE
0000          BPASS      .set      FALSE          ; select the desired filter type
0000          BSTOP      .set      FALSE
;
;           SIGNED      equ      TRUE          ; Set This To 'TRUE' for signed multi
0001          ;           and 'FALSE' for unsigned.
;
;*****
;           Test Program For Low Pass Filter
;*****
;           ORG      0x0000
;
;           start
0000 E00D          call      initFilter

```

# Implementing IIR Digital Filters

```
0001 B000          movlw    0x00
0002 0140          movwf    X
0003 B07F          movlw    0x7f          ; set initial Xn = X(0) = 0x7f00
0004 0141          movwf    X+B1          ; test for impulse response
;
NextPoint
0005 E022          call    IIR_Filter
0006 A442          tlwt     _LOW,Y
tracePoint
0007 AE43          tablwt   _HIGH,0,Y+B1
0008 0000          nop
0009 2940          clrfs   X          ; set X(n) = 0 , n <> 0
000A 2941          clrfs   X+B1        ; for simulating an Impulse
000B C005          goto    NextPoint
;
000C C00C          self    goto    self
;
;*****
;
initFilter
;
; At first read the Filter Co-efficients from Prog. Mem to Data RAM
;
;
;   if    LPASS
000D B0C1          movlw    _coeff_lpass
000E 010D          movwf    tblptrl
000F B001          movlw    page _coeff_lpass
0010 010E          movwf    tblptrh
;
;   endif
;   if    HPASS
;   movlw    _coeff_hpass
;   movwf    tblptrl
;   movlw    page _coeff_hpass
;   movwf    tblptrh
;
;   endif
;   if    BPASS
;   movlw    _coeff_bpas
;   movwf    tblptrl
;   movlw    page _coeff_bpas
;   movwf    tblptrh
;
;   endif
;   if    BSTOP
;   movlw    _coeff_bstop
;   movwf    tblptrl
;   movlw    page _coeff_bstop
;   movwf    tblptrh
;
;   endif
;
0011 B02C          movlw    A11
0012 0101          movwf    fsr0
0013 8404          bsf     _fs0
0014 8D04          bcf     _fsl          ; auto increment
;
; Read Filter Co-efficients from Program Memory
;
;
0015 B00A          movlw    NumCoeff
0016 A92C          tablr   _LOW,_INC,A11          ; garbage
NextCoeff
0017 A000          tlrd    _LOW,indf0
0018 AB00          tablr   _HIGH,_INC,indf0
0019 1700          decfsz wreg
001A C017          goto    NextCoeff
;
; Initialize "Dn"s to zero
;
001B B020          movlw    Dn1
```



# Implementing IIR Digital Filters

```

001C 0101          movwf    fsr0
001D B00C          movlw    6*FltStage

NextClr

001E 2900          clrfsz  indf0
001F 1700          decfsz  wreg
0020 C01E          goto    NextClr

;
0021 0002          return

;
;*****
;                               1st Cascade Section
;*****
;

IIR_Filter

;
; Compute  D(n) = X(n) + A1*D(n-1) + A2*D(n-2)
;           Since the filter constants are computed in Q15 format,
;           X(n) must be multiplied by 2**15 and then added to the
;           other terms.
;
; Move Input to accumulator after proper scaling
;
0022 8804          bcf     _carry
0023 1941          rrcfsz  X+B1
0024 1940          rrcfsz  X
0025 2900          clrfsz  wreg           ; Scale the input X
0026 1900          rrcfsz  wreg
0027 0145          movwf   ACC+B1
0028 6040          movf    X,wreg
0029 0146          movwf   ACC+B2
002A 6041          movf    X+B1,wreg
002B 0147          movwf   ACC+B3           ; ACC = scaled input : X*(2**15)

;
; 1st Biquad filter section
;
BIQUAD  A11,A12,B10,B11,Dn1,Dn1_1,Dn1_2,1

;
; Compute A12*D(n-2)
;

002C 7C24          MOVFP   Dn1_2+B0,AARG+B0       ; move Dn1_2(B0) to AARG(B0)
002D 7D25          MOVFP   Dn1_2+B1,AARG+B1       ; move Dn1_2(B1) to AARG(B1)

;

002E 7E2E          MOVFP   A12+B0,BARG+B0       ; move A12(B0) to BARG(B0)
002F 7F2F          MOVFP   A12+B1,BARG+B1       ; move A12(B1) to BARG(B1)

;

0030 E0AF          call    DblMult           ; (ACCd,ACCc) = A2*D(n-2)

;
; Add product to output of 1st section
; Save result in 32 bit Accumulator
;

0031 6018          MOVFP   DPX+B0,WREG           ; get lowest byte of DPX into w
0032 0F44          ADDWF   ACC+B0           ; add lowest byte of ACC, save in ACC(B0)
0033 6019          MOVFP   DPX+B1,WREG           ; get 2nd byte of DPX into w
0034 1145          ADDWFC  ACC+B1           ; add 2nd byte of ACC, save in ACC(B1)
0035 601A          MOVFP   DPX+B2,WREG           ; get 3rd byte of DPX into w
0036 1146          ADDWFC  ACC+B2           ; add 3rd byte of ACC, save in ACC(B2)

```

# Implementing IIR Digital Filters

```
0037 601B          MOVFP   DPX+B3,WREG          ; get 4th byte of DPX into w
0038 1147          ADDWFC  ACC+B3              ; add 4th byte of ACC, save in ACC(B3)

;
;   Compute A1*D(n-1)
;

0039 7C22          MOVFP   Dn1_1+B0,AARG+B0      ; move Dn1_1(B0) to AARG(B0)
003A 7D23          MOVFP   Dn1_1+B1,AARG+B1      ; move Dn1_1(B1) to AARG(B1)

003B 7E2C          MOVFP   A11+B0,BARG+B0      ; move A11(B0) to BARG(B0)
003C 7F2D          MOVFP   A11+B1,BARG+B1      ; move A11(B1) to BARG(B1)

003D E0AF          call    DblMult              ; (ACCd,ACCc) = A1*D(n-1)
;
;   Compute A1*D(n-1) + A2*D(n-2) + output of previous section
;   multiplications already done, so simply perform a 32 bit add
;   of previously obtained multiplication results
;

003E 6018          MOVFP   DPX+B0,WREG          ; get lowest byte of DPX into w
003F 0F44          ADDWF   ACC+B0              ; add lowest byte of ACC,save in ACC(B0)
0040 6019          MOVFP   DPX+B1,WREG          ; get 2nd byte of DPX into w
0041 1145          ADDWFC  ACC+B1              ; add 2nd byte of ACC, save in ACC(B1)
0042 601A          MOVFP   DPX+B2,WREG          ; get 3rd byte of DPX into w
0043 1146          ADDWFC  ACC+B2              ; add 3rd byte of ACC, save in ACC(B2)
0044 601B          MOVFP   DPX+B3,WREG          ; get 4th byte of DPX into w
0045 1147          ADDWFC  ACC+B3              ; add 4th byte of ACC, save in ACC(B3)

;
;
;   save the upper 16 bits of D(n) from the 32 bit accumulator
;   left shift the result by 1, to adjust the decimal point after
;   a Q15*Q15 multiplication
;

0046 1A45          rlc     ACC+B1,w
0047 1A46          rlc     ACC+B2,w
0048 0120          movwf  Dn1
0049 1A47          rlc     ACC+B3,w          ; decimal adjust ( mult by 2)
004A 0121          movwf  Dn1+B1

;
;   Compute B2 * [D(n) + D(n-2)]
;

        if B0_EQUALS_B2

004B 6024          movfp   Dn1_2+B0,wreg
004C 0E20          addwf  Dn1+B0,w
004D 011C          movwf  AARG+B0
004E 6025          movfp   Dn1_2+B1,wreg
004F 1021          addwfc  Dn1+B1,w
0050 011D          movwf  AARG+B1
0051 7E30          MOVFP   B10+B0,BARG+B0      ; move B10(B0) to BARG(B0)
0052 7F31          MOVFP   B10+B1,BARG+B1      ; move B10(B1) to BARG(B1)
0053 E0AF          call    DblMult              ; (ACCd,ACCc) = B2*[D(n)+D(n-2)]
0054 5844          MOVFP   DPX+B0,ACC+B0      ; move DPX(B0) to ACC(B0)
0055 5945          MOVFP   DPX+B1,ACC+B1      ; move DPX(B1) to ACC(B1)
0056 5A46          MOVFP   DPX+B2,ACC+B2      ; move DPX(B2) to ACC(B2)
0057 5B47          MOVFP   DPX+B3,ACC+B3      ; move DPX(B3) to ACC(B3)
```

# Implementing IIR Digital Filters

```

else
    MOVFP16 B10,BARG
    MOVFP16 Dn1,AARG
    call    DblMult          ; B0*D(n)
    MOVFP32 DPX,ACC
    MOVFP16 Bx2,BARG
    MOVFP16 Dn1_2,AARG
    call    DblMult          ; B2*D(n-2)
    ADD32  DPX,ACC
endif

;
; Shift down D(n-1) to D(n-2) after D(n-2) usage is no longer required.
; This way in the next iteration D(n-2) is equal to the present D(n-1)
;
0058 7C22          movfp   Dn1_1,AARG+B0
0059 5C24          movfp   AARG+B0,Dn1_2          ; Shift down D(n-1)
005A 7D23          movfp   Dn1_1+B1,AARG+B1
005B 5D25          movfp   AARG+B1,Dn1_2+B1      ; AARG = D(n-1) = multiplier
005C 7E32          MOVFP   B11+B0,BARG+B0        ; move B11(B0) to BARG(B0)
005D 7F33          MOVFP   B11+B1,BARG+B1        ; move B11(B1) to BARG(B1)
005E E0AF          call    DblMult          ; (ACCd,ACCc) = B1*D(n-1)
;
; Compute Output Y = B1*D(n-1) + B2*D(n-2) + B0*D(n)
;                   = B1*D(n-1) + B0*[D(n) + D(n-2)]
; Since all multiplications are already done, simply perform a
; 32 bit addition
;
005F 6018          MOVFP   DPX+B0,WREG          ; get lowest byte of DPX into w
0060 0F44          ADDWF  ACC+B0          ; add lowest byte of ACC, save in ACC(B0)
0061 6019          MOVFP   DPX+B1,WREG          ; get 2nd byte of DPX into w
0062 1145          ADDWFC ACC+B1          ; add 2nd byte of ACC, save in ACC(B1)
0063 601A          MOVFP   DPX+B2,WREG          ; get 3rd byte of DPX into w
0064 1146          ADDWFC ACC+B2          ; add 3rd byte of ACC, save in ACC(B2)
0065 601B          MOVFP   DPX+B3,WREG          ; get 4th byte of DPX into w
0066 1147          ADDWFC ACC+B3          ; add 4th byte of ACC, save in ACC(B3)
;
; Shift down D(n) to D(n-1) so that in the next iteration, the new
; D(n-1) is the present D(n)
;
0067 6020          MOVFP   Dn1+B0,WREG          ; get byte of Dn1 into w
0068 0122          MOVWF  Dn1_1+B0          ; move to Dn1_1(B0)
0069 6021          MOVFP   Dn1+B1,WREG          ; get byte of Dn1 into w
006A 0123          MOVWF  Dn1_1+B1          ; move to Dn1_1(B1)
;
;
; 2nd Biquad filter section
;
BIQUAD  A21,A22,B20,B21,Dn2,Dn2_1,Dn2_2,2
;
; Compute A22*D(n-2)
;
006B 7C2A          MOVFP   Dn2_2+B0,AARG+B0        ; move Dn2_2(B0) to AARG(B0)
006C 7D2B          MOVFP   Dn2_2+B1,AARG+B1        ; move Dn2_2(B1) to AARG(B1) 006D 7E38
                                MOVFP   A22+B0,BARG+B0        ; move A22(B0) to BARG(B0)
006E 7F39          MOVFP   A22+B1,BARG+B1        ; move A22(B1) to BARG(B1)
006F E0AF          call    DblMult          ; (ACCd,ACCc) = A2*D(n-2)
;
; Add product to output of 1st section
; Save result in 32 bit Accumulator
;
0070 6018          MOVFP   DPX+B0,WREG          ; get lowest byte of DPX into w
0071 0F44          ADDWF  ACC+B0          ; add lowest byte of ACC, save in ACC(B0)
0072 6019          MOVFP   DPX+B1,WREG          ; get 2nd byte of DPX into w
0073 1145          ADDWFC ACC+B1          ; add 2nd byte of ACC, save in ACC(B1)
0074 601A          MOVFP   DPX+B2,WREG          ; get 3rd byte of DPX into w
0075 1146          ADDWFC ACC+B2          ; add 3rd byte of ACC, save in ACC(B2)
0076 601B          MOVFP   DPX+B3,WREG          ; get 4th byte of DPX into w

```

# Implementing IIR Digital Filters

```
0077 1147          ADDWFC  ACC+B3          ; add 4th byte of ACC, save in ACC(B3)
;
; Compute A1*D(n-1)
;

0078 7C28          MOVFP   Dn2_1+B0,AARG+B0      ; move Dn2_1(B0) to AARG(B0)
0079 7D29          MOVFP   Dn2_1+B1,AARG+B1      ; move Dn2_1(B1) to AARG(B1)

007A 7E36          MOVFP   A21+B0,BARG+B0      ; move A21(B0) to BARG(B0)
007B 7F37          MOVFP   A21+B1,BARG+B1      ; move A21(B1) to BARG(B1)

007C E0AF          call    DblMult          ; (ACCD,ACCc) = A1*D(n-1)
;
; Compute A1*D(n-1) + A2*D(n-2) + output of previous section
; multiplications already done, so simply perform a 32 bit add
; of previously obtained multiplication results
;

007D 6018          MOVFP   DPX+B0,WREG          ; get lowest byte of DPX into w
007E 0F44          ADDWFC  ACC+B0          ; add lowest byte of ACC, save in ACC(B0)
007F 6019          MOVFP   DPX+B1,WREG          ; get 2nd byte of DPX into w
0080 1145          ADDWFC  ACC+B1          ; add 2nd byte of ACC, save in ACC(B1)
0081 601A          MOVFP   DPX+B2,WREG          ; get 3rd byte of DPX into w
0082 1146          ADDWFC  ACC+B2          ; add 3rd byte of ACC, save in ACC(B2)
0083 601B          MOVFP   DPX+B3,WREG          ; get 4th byte of DPX into w
0084 1147          ADDWFC  ACC+B3          ; add 4th byte of ACC, save in ACC(B3)

;
;
; save the upper 16 bits of D(n) from the 32 bit accumulator
; left shift the result by 1, to adjust the decimal point after
; a Q15*Q15 multiplication
;

0085 1A45          rlcfc   ACC+B1,w          ; decimal adjust ( mult by 2)
0086 1A46          rlcfc   ACC+B2,w
0087 0126          movwf  Dn2
0088 1A47          rlcfc   ACC+B3,w
0089 0127          movwf  Dn2+B1

;
; Compute B2 * [D(n) + D(n-2)]
;

if      B0_EQUALS_B2

008A 602A          movfp   Dn2_2+B0,wreg
008B 0E26          addwf  Dn2+B0,w
008C 011C          movwf  AARG+B0
008D 602B          movfp   Dn2_2+B1,wreg
008E 1027          addwfc Dn2+B1,w
008F 011D          movwf  AARG+B1

0090 7E3A          MOVFP   B20+B0,BARG+B0      ; move B20(B0) to BARG(B0)
0091 7F3B          MOVFP   B20+B1,BARG+B1      ; move B20(B1) to BARG(B1)

0092 E0AF          call    DblMult          ; (ACCD,ACCc) = B2*[D(n)+D(n-2)]
0093 5844          MOVFP   DPX+B0,ACC+B0      ; move DPX(B0) to ACC(B0)
0094 5945          MOVFP   DPX+B1,ACC+B1      ; move DPX(B1) to ACC(B1)
0095 5A46          MOVFP   DPX+B2,ACC+B2      ; move DPX(B2) to ACC(B2)
0096 5B47          MOVFP   DPX+B3,ACC+B3      ; move DPX(B3) to ACC(B3)

else
```

# Implementing IIR Digital Filters

```

                                MOVFP16 B20,BARG
                                MOVFP16 Dn2,AARG
                                call   DblMult           ; B0*D(n)
                                MOVFP32 DPX,ACC
                                MOVFP16 Bx2,BARG
                                MOVFP16 Dn2_2,AARG
                                call   DblMult           ; B2*D(n-2)
                                ADD32  DPX,ACC
endif
;
; Shift down D(n-1) to D(n-2) after D(n-2) usage is no longer required.
; This way in the next iteration D(n-2) is equal to the present D(n-1)
;
0097 7C28      movfp   Dn2_1,AARG+B0
0098 5C2A      movfpf  AARG+B0,Dn2_2           ; Shift down D(n-1)
0099 7D29      movfpf  Dn2_1+B1,AARG+B1
009A 5D2B      movfpf  AARG+B1,Dn2_2+B1       ; AARG = D(n-1) = multiplier
009B 7E3C      MOVFPF  B21+B0,BARG+B0        ; move B21(B0) to BARG(B0)
009C 7F3D      MOVFPF  B21+B1,BARG+B1        ; move B21(B1) to BARG(B1)
009D E0AF      call    DblMult                 ; (ACCd,ACCc) = B1*D(n-1)
;
; Compute Output Y = B1*D(n-1) + B2*D(n-2) + B0*D(n)
;                   = B1*D(n-1) + B0*[D(n) + D(n-2)]
; Since all multiplications are already done, simply perform a
; 32 bit addition
;
009E 6018      MOVFPF  DPX+B0,WREG             ; get lowest byte of DPX into w
009F 0F44      ADDWF   ACC+B0                    ; add lowest byte of ACC, save in ACC(B0)
00A0 6019      MOVFPF  DPX+B1,WREG             ; get 2nd byte of DPX into w
00A1 1145      ADDWFC  ACC+B1                    ; add 2nd byte of ACC, save in ACC(B1)
00A2 601A      MOVFPF  DPX+B2,WREG             ; get 3rd byte of DPX into w
00A3 1146      ADDWFC  ACC+B2                    ; add 3rd byte of ACC, save in ACC(B2)
00A4 601B      MOVFPF  DPX+B3,WREG             ; get 4th byte of DPX into w
00A5 1147      ADDWFC  ACC+B3                    ; add 4th byte of ACC, save in ACC(B3)
;
; Shift down D(n) to D(n-1) so that in the next iteration, the new
; D(n-1) is the present D(n)
;
00A6 6026      MOVFPF  Dn2+B0,WREG             ; get byte of Dn2 into w
00A7 0128      MOVWF   Dn2_1+B0                ; move to Dn2_1(B0)
00A8 6027      MOVFPF  Dn2+B1,WREG             ; get byte of Dn2 into w
00A9 0129      MOVWF   Dn2_1+B1                ; move to Dn2_1(B1)
;
;
; The filter output is now computed
; Save the Upper 16 Bits of 32 bit Accumulator into Y after
; adjusting the decimal point
;
MOV16  ACC+B2,Y

00AA 6046      MOVFPF  ACC+B2+B0,WREG          ; get byte of ACC+B2 into w
00AB 0142      MOVWF   Y+B0                    ; move to Y(B0)
00AC 6047      MOVFPF  ACC+B2+B1,WREG          ; get byte of ACC+B2 into w
00AD 0143      MOVWF   Y+B1                    ; move to Y(B1)
;
00AE 0002      return                          ; Output Y(n) computed
;
;*****
; Set SIGNED/UNSIGNED Flag Before Including 17c42MPY.mac
;
include      "17c42MPY.mac"
;*****
; Low Pass Filter Co-efficients
;
;

```

# Implementing IIR Digital Filters

```

;
;                               ELLIPTIC    LOWPASS FILTER
;
;                               FILTER ORDER =    4
;                               Sampling frequency =    2.000 KiloHertz
;
;                               BAND    1        BAND    2
;
;
; LOWER BAND EDGE                .00000        .60000
; UPPER BAND EDGE                .50000        1.00000
; NOMINAL GAIN                    1.00000        .00000
; NOMINAL RIPPLE                  .01000        .05000
; MAXIMUM RIPPLE                  .00906        .04601
; RIPPLE IN dB                    .07830        -26.74235
;
; I      A(I,1)      A(I,2)      B(I,0)      B(I,1)      B(I,2)
;
; 1      -.133331    .167145    .285431    .462921    .285431
; 2      .147827    .765900    .698273    .499908    .698273
;
;
;_coeff_lpPASS                ; co-efficients for 1st Cascade
01C1 1111                    data    4369                ; -A11
01C2 EA9B                    data    -5477               ; -A12
01C3 2489                    data    9353                ; B10
01C4 3B41                    data    15169               ; B11
01C5 2489                    data    9353                ; B12
;
01C6 ED14                    data    -4844               ; co-efficients for 2nd Cascade
01C7 9DF7                    data    -25097              ; -A21
01C8 5961                    data    22881               ; -A22
01C9 3FFD                    data    16381               ; B20
01CA 5961                    data    22881               ; B21
;
;                               ; B22
;
;*****
;
;*****
; High Pass Filter Co-efficients
;
;
;                               ELLIPTIC    HIGHPASS FILTER
;
;                               FILTER ORDER =    4
;                               Sampling frequency =    2.000 KiloHertz
;
;                               BAND    1        BAND    2
;
;
; LOWER BAND EDGE                .00000        .50000
; UPPER BAND EDGE                .40000        1.00000
; NOMINAL GAIN                    .00000        1.00000
; NOMINAL RIPPLE                  .04000        .02000
; MAXIMUM RIPPLE                  .03368        .01686
; RIPPLE IN dB                    -29.45335       .14526
;
; I      A(I,1)      A(I,2)      B(I,0)      B(I,1)      B(I,2)
;
; 1      .276886    .195648    .253677    -.411407    .253677
; 2      -.094299    .780396    .678650    -.485840    .678650
;
;
;_coeff_hpPASS                ; co-efficients for 1st Cascade section
01CB DC8F                    data    -9073               ; -A11
01CC E6F5                    data    -6411               ; -A12
01CD 2079                    data    8313                ; B10
01CE CB57                    data    -13481              ; B11
01CF 2079                    data    8313                ; B12
;
;                               ; co-efficients for 2nd Cascade section

```

# Implementing IIR Digital Filters

```

01D0 0C12          data      3090          ; -A21
01D1 9C1C          data     -25572         ; -A22
01D2 56DE          data      22238         ; B20
01D3 C1D0          data     -15920         ; B21
01D4 56DE          data      22238         ; B22
;
;*****
;
;*****
; Band Pass Filter Co-efficients
;
;
;                               ELLIPTIC   BANDPASS FILTER
;
; FILTER ORDER =      4
; Sampling frequency =   2.000 KiloHertz
;
;                               BAND  1      BAND  2      BAND  3
;
; LOWER BAND EDGE           .00000      .30000      .90000
; UPPER BAND EDGE           .10000      .70000      1.00000
; NOMINAL GAIN               .00000      1.00000      .00000
; NOMINAL RIPPLE            .05000      .05000      .05000
; MAXIMUM RIPPLE            .03644      .03867      .03641
; RIPPLE IN dB              -28.76779      .32956      -28.77647
;
;
; I      A(I,1)      A(I,2)      B(I,0)      B(I,1)      B(I,2)
;
; 1      -.936676     .550568     .444000     -.865173     .444000
; 2      .936707     .550568     .615540     1.199402     .615540
;
;_coeff_bpas      ; co-efficients for 1st Cascade section
01D5 3BF2          data      30693/2         ; -A11
01D6 DCC4          data     -18041/2        ; -A12
01D7 1C6A          data      14549/2         ; B10
01D8 C8A1          data     -28350/2        ; B11
01D9 1C6A          data      14549/2         ; B12
;
;                               ; co-efficients for 2nd Cascade section
01DA C40D          data     -30694/2        ; -A21
01DB DCC4          data     -18041/2        ; -A22
01DC 2765          data      20170/2         ; B20
01DD 4CC3          data      39302/2         ; B21
01DE 2765          data      20170/2         ; B22
;
;*****
;
;*****
; Band Stop Filter Co-efficients
;
;
;                               ELLIPTIC   BANDSTOP FILTER
;
; FILTER ORDER =      4
; Sampling frequency =   2.000 KiloHertz
;
;                               BAND  1      BAND  2      BAND  3
;
; LOWER BAND EDGE           .00000      .45000      .70000
; UPPER BAND EDGE           .30000      .55000      1.00000
; NOMINAL GAIN               1.00000      .00000      1.00000
; NOMINAL RIPPLE            .05000      .05000      .05000
; MAXIMUM RIPPLE            .03516      .03241      .03517
; RIPPLE IN dB              .30015      -29.78523      .30027
;
;
;

```

# Implementing IIR Digital Filters

```

; I      A(I,1)      A(I,2)      B(I,0)      B(I,1)      B(I,2)
;
; 1      .749420     .583282     .392685     .087936     .392685
; 2      -.749390    .583282     1.210022     -.270935     1.210022
;
_coeff_bstop                ; co-efficients for 1st Cascade section
01DF D00A                   data    -24557/2      ; -A11
01E0 DAAC                   data    -19113/2      ; -A12
01E1 1922                   data    12868/2       ; B10
01E2 05A0                   data    2881/2        ; B11
01E3 1922                   data    12868/2       ; B12
                                ; co-efficients for 2nd Cascade section
01E4 2FF6                   data    24556/2       ; -A21
01E5 DAAC                   data    -19113/2      ; -A22
01E6 4D71                   data    39650/2       ; B20
01E7 EEA9                   data    -8878/2       ; B21
01E8 4D71                   data    39650/2       ; B22
;
;*****
                                END

Errors   :    0
Warnings :    0
```



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