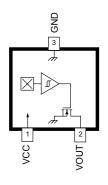
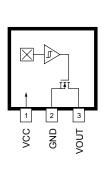
Chopper-Stabilized Unipolar Hall-Effect Switches

Package LH, 3-pin Surface Mount





Package UA, 3-pin SIP





ABSOLUTE MAXIMUM RATINGS

Supply Voltage, V _{CC}	28 V
Reverse-Supply Voltage, V _{RCC}	18 V
Reverse-Supply Current, I _{RCC}	2 mA
Output Off Voltage, V _{OUT}	28 V
Output Current, I _{OUTSINK} I	nternally Limited
Magnetic Flux Density, B	Unlimited
Operating Temperature	
Ambient, T _A , Range E	40°C to 85°C
Ambient, T _A , Range L	40°C to 150°C
Maximum Junction, T _{J(MAX)}	165°C
Storage Temperature, T _S	

The A3241 and A3242 integrated circuits are unipolar Hall-effect switches with digital outputs. These sensors are suited for operation over extended temperature ranges, up to +150°C. Superior high-temperature performance is made possible through an Allegro® patented dynamic offset cancellation, which reduces the residual offset voltage normally caused by device overmolding, temperature excursions, and thermal stress.

The A3241 and A3242 Hall-effect switches include the following on a single silicon chip: voltage regulator, Hall-voltage generator, small-signal amplifier, chopper stabilization, Schmitt trigger, and a short circuit protected open-drain output. Advanced BiCMOS wafer fabrication processing is used to take advantage of low-voltage requirements, component matching, very low input-offset errors, and small component geometries.

The integrated voltage regulator permits operation from 3.6 to 24 V. The unipolar family members operate with a sufficient south polarity field only, turning off in the absence of such a south polarity field.

The A3241 and A3242 are rated for operation between the ambient temperatures –40°C and 85°C for the E temperature range, and –40°C to 150°C for the L temperature range. The small geometries of the BiCMOS process allow these devices to be provided in ultrasmall packages. The package styles available provide magnetically optimized solutions for most applications. Package LH is an SOT23W, a miniature low-profile surface-mount package, while package UA is a three-lead ultramini SIP for through-hole mounting. Each package is available in a lead (Pb) free version, with 100% matte tin plated leadframes.

Features and Benefits

- Chopper stabilization
 - Superior temperature stability
- · Extremely low switchpoint drift
- · Insensitive to physical stress
- Reverse battery protection

- Output short circuit protection
- Solid state reliability
- Small size
- Robust EMC capability
- High ESD ratings (HBM)





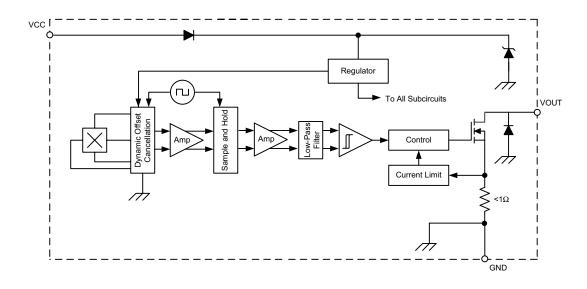
Chopper-Stabilized Unipolar Hall Effect Switches

Product Selection Guide

Part Number	Pb- free	Packing* Mounting		Ambient, T _A (°C)	B _{RP(MIN)} (G)	B _{OP(MAX)} (G)
A3241ELHLT	_	7 in real 2000 ninear/real	3-pin SOT23W surface mount		40	135
A3241ELHLT-T	Yes	7-in. reel, 3000 pieces/reel		-40 to 85		
A3241EUA	_	Dulk 500 pieces/bag	3-pin SIP through hole			
A3241EUA-T	Yes	Bulk, 500 pieces/bag				
A3241LLHLT	_	7 in real 2000 ninear/real	2 nin COT22W curface mount		40	
A3241LLHLT-T	Yes	7-in. reel, 3000 pieces/reel	3-pin SOT23W surface mount	40 to 150		
A3241LUA	_	Pulk 500 pioces/bag	3-pin SIP through hole	-40 to 150		
A3241LUA-T	Yes	Bulk, 500 pieces/bag				
A3242ELHLT	_	7 in real 2000 niceso/real	2 nin COT23W gurfage mount		- 110	200
A3242ELHLT-T	Yes	7-in. reel, 3000 pieces/reel	3-pin SOT23W surface mount	-40 to 85		
A3242EUA	_	Pulk 500 pioces/bag	2 pin SID through halo	-4 0 to 65		
A3242EUA-T	Yes	Bulk, 500 pieces/bag	3-pin SIP through hole			
A3242LLHLT	_	7 in real 2000 niceso/real	3-pin SOT23W surface mount	40 to 150		
A3242LLHLT-T	Yes	7-in. reel, 3000 pieces/reel				
A3242LUA	_	-40 to 150				
A3242LUA-T	Yes	Bulk, 500 pieces/bag	3-pin SIP through hole			

^{*}Contact Allegro for additional packing options.

Functional Block Diagram



Terminal List

Name	Description	Number		
	Description	Package LH	Package UA	
VCC	Connects power supply to chip	1	1	
VOUT	Output from circuit	2	3	
GND	Ground	3	2	



Chopper-Stabilized Unipolar Hall Effect Switches

OPERATING CHARACTERISTICS valid over full operating voltage and ambient temperature ranges, unless otherwise noted

Characteristic	Symbol	Test Conditions		Min.	Тур.	Max.	Units
Electrical Characteristics		·!		•			
Supply Voltage ¹	V _{CC}	Operating, T	Operating, T _J < 165°C		_	24	V
Output Leakage Current	I _{OUTOFF}	V _{OUT} = 24 V	, B < B _{RP}	_	_	10	μA
Output On Voltage	V _{OUT(SAT)}	I _{OUT} = 20 m/	I _{OUT} = 20 mA, B > B _{OP}		_	500	mV
Output Current Limit	I _{OM}	B > B _{OP}			_	60	mA
Power-On Time	t _{PO}	$V_{CC} > V_{CC(M)}$	/ _{CC} > V _{CC(MIN)}			50	μs
Chopping Frequency	f _c			_	200	_	kHz
Output Rise Time ²	t _r	R _{LOAD} = 820	0 Ω, C _S = 20 pF	_	_	1	μs
Output Fall Time ²	t _f	R _{LOAD} = 820	0 Ω, C _S = 20 pF	_	_	1	μs
Supply Current	I _{CCON}	B > B _{OP}		_	1.5	3.5	mA
	I _{CCOFF}	B < B _{RP}		_	1.5	3.5	mA
Reverse Battery Current	I _{RCC}	V _{RCC} = -18	V _{RCC} = -18 V		_	-2	mA
Supply Zener Clamp Voltage	V _{ZSupply}	$I_{CC} = 6.5 \text{ m/s}$	I _{CC} = 6.5 mA; T _A = 25°C		_	_	V
Supply Zener Current ³	I _{ZSupply}	V _S = 28 V		_	_	6.5	mA
Magnetic Characteristics ⁴						•	
On sunta Daint	int B _{OP}	A3241		50	95	135	G
Operate Point		A3242		120	150	200	G
Release Point	B _{RP}	A3241		40	70	110	G
		A3242		110	125	190	G
III at a sector	B _{HYS}	A3241	B B	10	25	42	G
Hysteresis		A3242 B _{OP} – B _{RP}		10	25	40	G

¹ Maximum voltage must be adjusted for power dissipation and junction temperature, see *Power Derating* section.

DEVICE QUALIFICATION PROGRAM

Contact Allegro for information.

EMC (Electromagnetic Compatibility) REQUIREMENTS Contact Allegro for information.



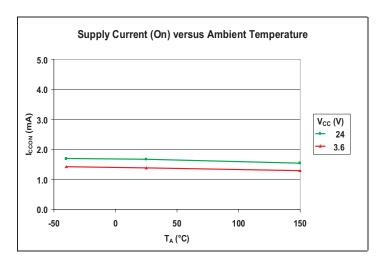
² C_S = oscilloscope probe capacitance.

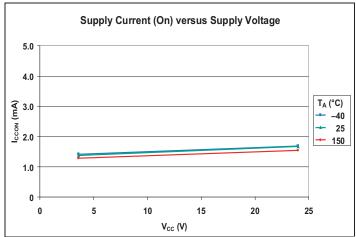
³ Maximum current limit is equal to the maximum I_{CC(MAX)} + 3 mA.

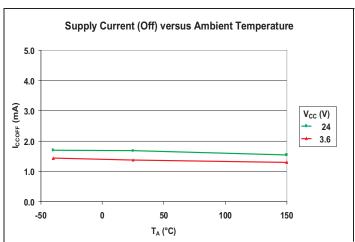
⁴ Magnetic flux density, B, is indicated as a negative value for north-polarity magnetic fields, and as a positive value for south-polarity magnetic fields. This so-called algebraic convention supports arithmetic comparison of north and south polarity values, where the relative strength of the field is indicated by the absolute value of B, and the sign indicates the polarity of the field (for example, a -100 G field and a 100 G field have equivalent strength, but opposite polarity).

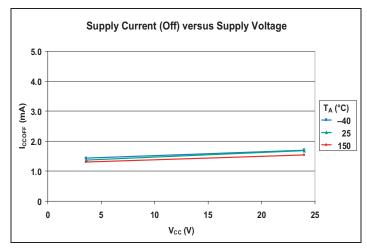
Chopper-Stabilized Unipolar Hall Effect Switches

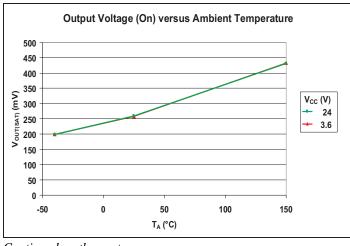
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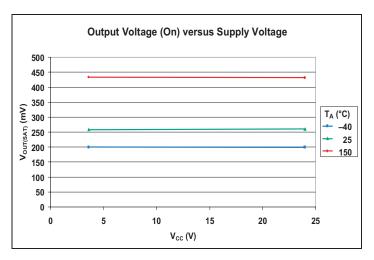






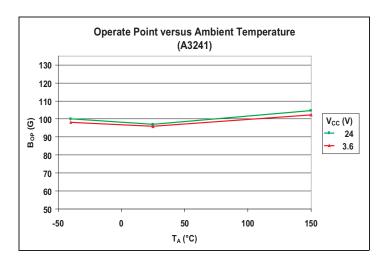


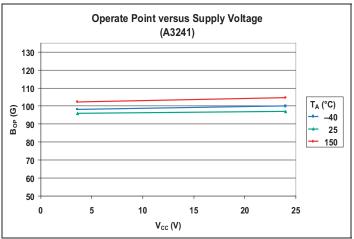


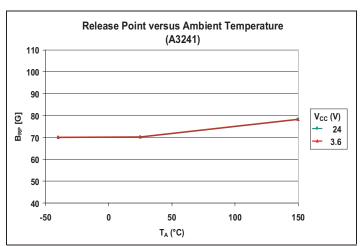


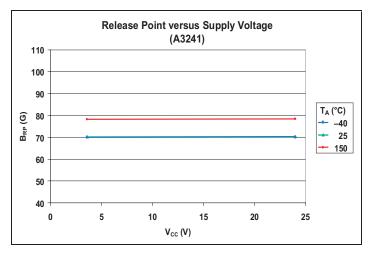
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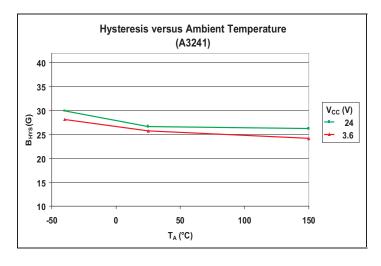
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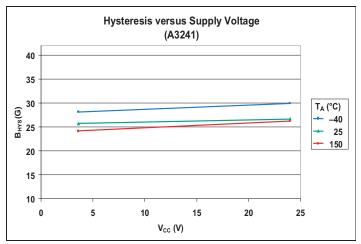




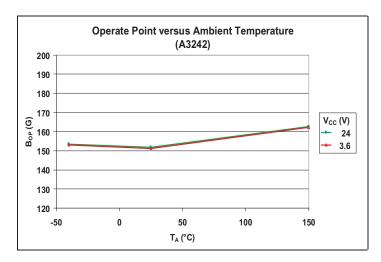


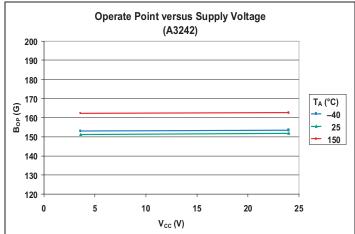


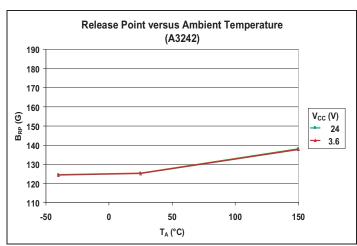


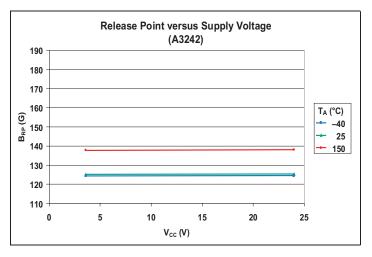


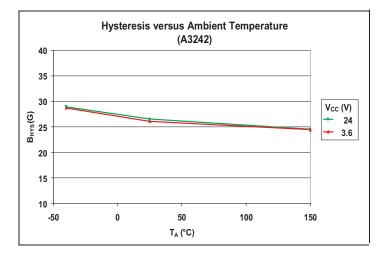
Chopper-Stabilized Unipolar Hall Effect Switches

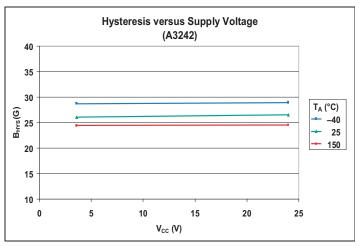










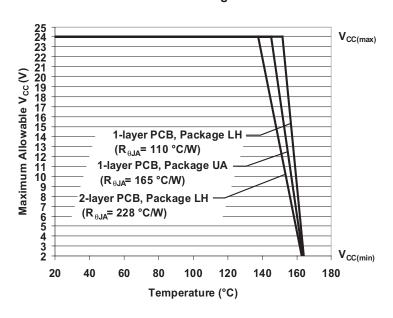


Chopper-Stabilized Unipolar Hall Effect Switches

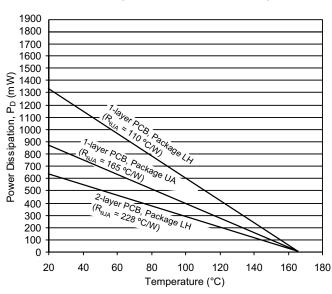
THERMAL CHARACTERISTICS may require derating at maximum conditions, see application information

Characteristic	Symbol	Test Conditions	Value	Units
Package Thermal Resistance	$R_{ heta JA}$	Package LH-3, 1-layer PCB with copper limited to solder pads	110	°C/W
		Package LH-3, 2-layer PCB with 0.926 in ² on each side, connected by thermal vias	228	°C/W
		Package UA, 1-layer PCB with copper limited to solder pads	165	°C/W

Power Derating Curve



Power Dissipation versus Ambient Temperature





Chopper-Stabilized Unipolar Hall Effect Switches

Functional Description

Operation

The output of these devices switches low (turns on) when a magnetic field (south polarity) perpendicular to the Hall sensor exceeds the operate point threshold, B_{OP}. After turn-on, the output voltage is V_{OUT(SAT)}. The output transistor is capable of sinking current up to the short circuit current limit, IOM, which is a minimum of 30 mA. When the magnetic field is reduced below the release point, B_{RP}, the device output goes high (turns off). The difference in the magnetic operate and release points is the hysteresis, B_{hvs}, of the device. This built-in hysteresis allows clean switching of the output even in the presence of external mechanical vibration and electrical noise.

Powering-on the device in the hysteresis region, less than B_{OP} and higher than B_{RP}, allows an indeterminate output state. The correct state is attained after the first excursion beyond B_{OP} or B_{RP}

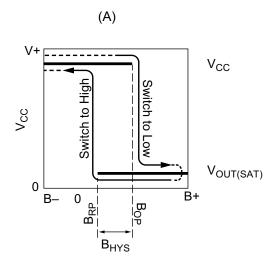
Applications

It is strongly recommended that an external bypass capacitor be connected (in close proximity to the Hall sensor) between the supply and ground of the device to reduce both external noise and noise generated by the chopper stabilization technique. As is shown in Panel B of figure 1, a 0.1 µF capacitor is typical.

Extensive applications information on magnets and Hall-effect sensors is available in:

- Hall-Effect IC Applications Guide, AN27701,
- Hall-Effect Devices: Gluing, Potting, Encapsulating, Lead Welding and Lead Forming, AN27703.1
- Soldering Methods for Allegro's Products SMT and Through-Hole, AN26009

All are provided in *Allegro Electronic Data Book*, AMS-702 and the Allegro Web site: www.allegromicro.com



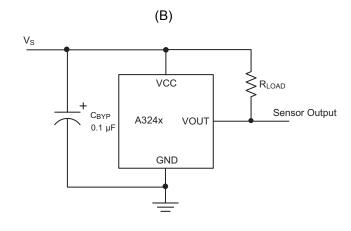


Figure 1: Switching Behavior of Unipolar Switches. In Panel A, on the horizontal axis, the B+ direction indicates increasing south polarity magnetic field strength, and the B- direction indicates decreasing south polarity field strength (including the case of increasing north polarity). This behavior can be exhibited when using a circuit such as that shown in panel B.

Chopper-Stabilized Unipolar Hall Effect Switches

Chopper Stabilization Technique

When using Hall-effect technology, a limiting factor for switchpoint accuracy is the small signal voltage developed across the Hall element. This voltage is disproportionally small relative to the offset that can be produced at the output of the Hall sensor. This makes it difficult to process the signal while maintaining an accurate, reliable output over the specified operating temperature and voltage ranges.

Chopper stabilization is a unique approach used to minimize Hall offset on the chip. The patented Allegro technique, namely Dynamic Quadrature Offset Cancellation, removes key sources of the output drift induced by thermal and mechanical stresses. This offset reduction technique is based on a signal modulationdemodulation process. The undesired offset signal is separated from the magnetic-field-induced signal in the frequency domain, through modulation. The subsequent demodulation acts as a modulation process for the offset, causing the magnetic-fieldinduced signal to recover its original spectrum at baseband, while the dc offset becomes a high-frequency signal. The magnetic-field-induced signal then can pass through a low-pass filter, while the modulated dc offset is suppressed. This configuration is illustrated in figure 2.

The chopper stabilization technique uses a 200 kHz high-frequency clock. For demodulation process, a sample and hold technique is used, where the sampling is performed at twice the chopper frequency (400 kHz). This high-frequency operation allows a greater sampling rate, which results in higher accuracy and faster signal-processing capability. This approach desensitizes the chip to the effects of thermal and mechanical stresses, and produces devices that have extremely stable quiescent Hall output voltages and precise recoverability after temperature cycling. This technique is made possible through the use of a BiCMOS process, which allows the use of low-offset, low-noise amplifiers in combination with high-density logic integration and sample-and-hold circuits.

The repeatability of magnetic-field-induced switching is affected slightly by a chopper technique. However, the Allegro highfrequency chopping approach minimizes the affect of jitter and makes it imperceptible in most applications. Applications that are more likely to be sensitive to such degradation are those requiring precise sensing of alternating magnetic fields; for example, speed sensing of ring-magnet targets. For such applications, Allegro recommends its digital sensor families with lower sensitivity to jitter. For more information on those devices, contact your Allegro sales representative.

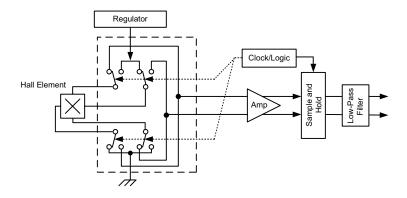


Figure 2. Chopper Stabilization Circuit (Dynamic Quadrature Offset Cancellation)



Chopper-Stabilized Unipolar Hall Effect Switches

Power Derating

The device must be operated below the maximum junction temperature of the device, $T_{J(max)}$. Under certain combinations of peak conditions, reliable operation may require derating supplied power or improving the heat dissipation properties of the application. This section presents a procedure for correlating factors affecting operating T_I. (Thermal data is also available on the Allegro MicroSystems Web site.)

The Package Thermal Resistance, $R_{\theta JA}$, is a figure of merit summarizing the ability of the application and the device to dissipate heat from the junction (die), through all paths to the ambient air. Its primary component is the Effective Thermal Conductivity, K, of the printed circuit board, including adjacent devices and traces. Radiation from the die through the device case, R_{AIC} , is relatively small component of $R_{\theta JA}$. Ambient air temperature, T_A, and air motion are significant external factors, damped by overmolding.

The effect of varying power levels (Power Dissipation, P_D), can be estimated. The following formulas represent the fundamental relationships used to estimate T_J, at P_D.

$$P_D = V_{IN} \times I_{IN} \tag{1}$$

$$\Delta T = P_D \times R_{\theta IA} \tag{2}$$

$$T_{I} = T_{A} + \Delta T \tag{3}$$

For example, given common conditions such as: $T_A = 25$ °C, $V_{CC} = 12 \text{ V}, I_{CC} = 1.5 \text{ mA}, \text{ and } R_{\theta \text{IA}} = 165 \text{ °C/W}, \text{ then:}$

$$P_D = V_{CC} \times I_{CC} = 12 \text{ V} \times 1.5 \text{ mA} = 18 \text{ mW}$$

$$\Delta T = P_D \times R_{\theta IA} = 18 \text{ mW} \times 165 \text{ }^{\circ}\text{C/W} = 3^{\circ}\text{C}$$

$$T_J = T_A + \Delta T = 25^{\circ}C + 3^{\circ}C = 28^{\circ}C$$

A worst-case estimate, P_{D(max)}, represents the maximum allowable power level ($V_{CC(max)}$, $I_{CC(max)}$), without exceeding $T_{J(max)}$, at a selected $R_{\theta JA}$ and T_A .

Example: Reliability for V_{CC} at T_A=150°C, package LH, using a low-K PCB.

Observe the worst-case ratings for the device, specifically: $R_{\theta JA} = 228 \text{ °C/W}, T_{J(max)} = 165 \text{ °C}, V_{CC(max)} = 24 \text{ V}, \text{ and}$ $I_{CC(max)} = 5 \text{ mA}.$

Calculate the maximum allowable power level, P_{D(max)}. First, invert equation 3:

$$\Delta T_{\text{max}} = T_{\text{I(max)}} - T_{\text{A}} = 165 \,^{\circ}\text{C} - 150 \,^{\circ}\text{C} = 15 \,^{\circ}\text{C}$$

This provides the allowable increase to T_I resulting from internal power dissipation. Then, invert equation 2:

$$P_{D(max)} = \Delta T_{max} \div R_{\theta JA} = 15^{\circ}C \div 228^{\circ}C/W = 65.8 \text{ mW}$$

Finally, invert equation 1 with respect to voltage:

$$V_{CC(est)} = P_{D(max)} \div I_{CC(max)} = 65.8 \text{ mW} \div 5 \text{ mA} = 13.2 \text{ V}$$

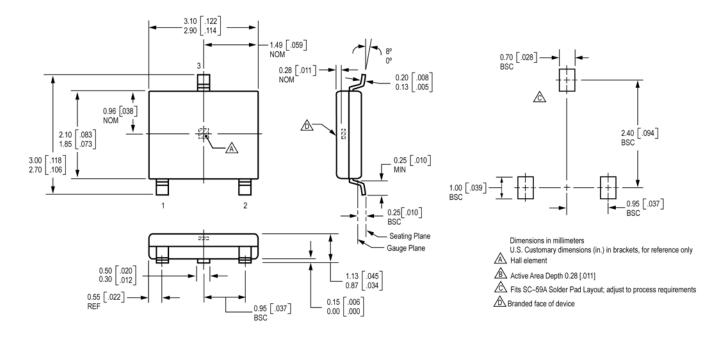
The result indicates that, at T_A, the application and device can dissipate adequate amounts of heat at voltages $\leq V_{CC(est)}$.

Compare $V_{CC(est)}$ to $V_{CC(max)}$. If $V_{CC(est)} \le V_{CC(max)}$, then reliable operation between $V_{CC(est)}$ and $V_{CC(max)}$ requires enhanced $R_{\theta JA}$. If $V_{CC(est)} \ge V_{CC(max)}$, then operation between $V_{CC(est)}$ and $V_{CC(max)}$ is reliable under these conditions.

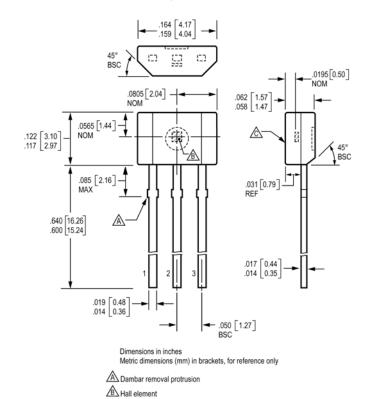


Chopper-Stabilized Unipolar Hall Effect Switches

Package LH, 3-Pin (SOT-23W)



Package UA, 3-Pin





Branded face of device

Chopper-Stabilized Unipolar Hall Effect Switches

The products described herein are manufactured under one or more of the following U.S. patents: 5,045,920; 5,264,783; 5,442,283; 5,389,889; 5,581,179; 5,517,112; 5,619,137; 5,621,319; 5,650,719; 5,686,894; 5,694,038; 5,729,130; 5,917,320; and other patents pending.

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