

2.5-GHz Double-Balanced Mixer

Description

The U2795B is a 2.5-GHz mixer for WLAN and RF tele-communications equipment, e.g., DECT and PCN. The IC is manufactured using Atmel Wireless & Microcontrollers' advanced bipolar technology. A double-balanced approach was chosen to assure good isolation characteristics and a minimum of spurious products. The input and output are single ended, and their

characteristics are programmable. No output transformer or balun is required.

Electrostatic sensitive device. Observe precautions for handling.



Features

- Supply-voltage range: 2.7 to 5.5 V
- Single-ended output, no balun required
- Single-ended input for RF and LO
- Excellent isolation characteristics
- Power-down mode
- IP3 and compression point programmable
- 2.5-GHz operating frequency

Benefits

- Reduced system costs as only few external component (no balun) are required
- Stand-alone product
- 3-V operation reduces battery count and saves space

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Block Diagram

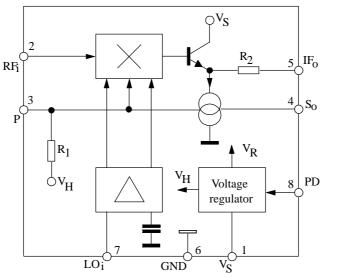


Figure 1. Block diagram

Ordering Information

Extended Type Number	Package	Remarks	
U2795B-MFP	SO8	Tube	
U2795B-MFPG3	SO8	Taped and reeled	

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Pin Description

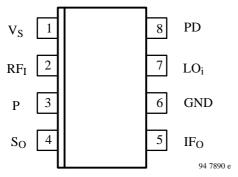


Figure 2. Pinning

Pin	Symbol	Function
1	V _S	Supply voltage
2	RFi	RF input
3	P	Progamming port IP3, CP
4	So	Output symmetry
5	IFO	IF output
6	GND	Ground
7	LOi	LO input
8	PD	Power down

Functional Description

Supply Voltage

The IC is designed for a supply-voltage range of 2.7 V to 5.5 V. As the IC is internally stabilized, the performance of the circuit is nearly independent of the supply voltage.

Input Impedance

The input impedance, Z_{RFi} , is about 700 Ω with an additional capacitive component. This condition provides the best noise figure in combination with a matching network.

3rd Order Intercept Point (IP3)

The voltage divider, $R_P \, / \, R_1$, determinates both the input and output intercept point, IIP3 and OIP3. If the value of R_P is infinite, the maximum value of IIP3 reachs about -4 dBm. The IP3/ R_P characteristics are shown in figure 3, and 4.

Output Impedance and Intercept Point

The output impedance is shown in figure 11. Both low output impedance and a high intercept point are defined to a high value of R_P.

Current Consumption, I_S

Depending on the chosen input and output conditions of the IC, the current consumption, I_{S_i} is between 4 mA and 10 mA. The current consumption in dependence of Rp is shown in figure 6.

Power Down

This feature provides extended battery lifetime. If this function is not used, Pin 8 has to be connected to V_S (Pin 1).

Output Symmetry

The symmetry of the load current can be matched and thus optimized for a given load impedance.

Absolute Maximum Ratings

Parameters		Symbol	Value	Unit	
Supply voltage	Pin 1	V_{S}	6	V	
Input voltage Pins 2, 3, 7 and 8		V _I	0 to V _S	V	
Junction temperature		Tj	125	°C	
Storage-temperature range		T _{stg}	-40 to +125	°C	

Thermal Resistance

Parameters		Symbol	Value	Unit	
Junction ambient	unction ambient SO 8		175	K/W	

Operating Range

Parameters	Symbol	Value	Unit
Supply-voltage range Pin 1	V_{S}	2.7 to 5.5	V
Ambient-temperature range	T _{amb}	-40 to +85	°C



Electrical Characteristics

 $V_S\!=\!3~V, f_{LOi}\!=\!1~GHz, IF\!=\!900~MHz, RF\!=\!100~MHz, R_P\!=\!\infty~,~system~impedance~Zo\!=\!50~\Omega, T_{amb}\!=\!25^{\circ}C, R_T\!=\!56~\Omega~, T_{amb}\!=\!25^{\circ}C, T_{amb}\!=\!25^{\circ}C,$

Parameters	Test Conditions / Pin	Symbol	Min.	Тур.	Max.	Unit
Supply voltage range	Pin 1	V _S	2.7		5.5	V
Typical supply-current range	Pin 1, depending on R _P	I_S	4		11	mA
Maximum supply current	Pin 1	I_S			13	mA
Conversion power gain	$R_L = 50 \ \Omega, R_T = \infty$ $R_L = 50 \ \Omega, R_T = 56 \ \Omega$	PG _C PG _C		9 4		dB dB
Operating frequencies						
RF _i frequency	Pin 2	RFi	10		2500	MHz
LO _i frequency	Pin 7	f _{LOi}	50		2500	MHz
IF _o frequency	Pin 5	f _{IFo}	50		2500	MHz
Isolation		1	I	II.		
LO spurious at R _{Fi}	Pin 7 to 2, $P_{iLO} = -10$ to 0 dBm	IS _{LO-RF}		-30		dBm
RF _i to LO _i	Pin 2 to 7, $P_{iRF} = -25 \text{ dBm}$	IS _{RF-LO}		35		dB
LO spurious at IF _o	Pin 7 to 5, $P_{iLO} = -10$ to 0 dBm	IS _{LO-IF}		-25		dBm
IF _o to LO _i	Pin 5 to 7	IS _{IF-LO}		30		dB
Output (IF)	1					
Output compression point	Pin 5	CPo		-10		dBm
Input (RF)	1	II.				
Input impedance	Pin 2	Z_{RFi}		700 0.8		$\Omega \ pF$
Input compression point	Pin 2	CPi		-14		dBm
3rd-order input intercept point	Pin 2	IIP3		-4		dBm
Input (LO)						
LO level	Pin 7	P _{iLO}		-6		dBm
Voltage standing wave ratio	(VSWR)	1	11			
Input LO	Pin 7	VSWR _{LOi}		<2		
Output IF	Pin 4	VSWR _{IFo}		<2		
Noise performance			I	ı		
Noise figure	$P_{iLO} = 0 \text{ dBm}, R_T = \infty$	NF		10		dB
Power-down mode			I			
Supply current	$\begin{array}{ccc} \text{Pin 1} & \text{V}_{\text{PD}} < 0.5 \text{ V} \\ \text{Pin 1} & \text{V}_{\text{PD}} = 0 \text{ V} \end{array}$	I _{SPD}		<5	30	μΑ
Power-down voltage		•				
"Power ON"	Pin 8 $V_S = 3.5 \text{ to } 5.5 \text{ V}$ $V_S = 2.7 \text{ to } 3.5 \text{ V}$	V _{PON}	V _S -0.5 V _S		V _S +0.5 V _S +0.5	V V
"Power DOWN"	Pin 8	V _{PDN}			1	V
Power-down current	Pin 8 Power ON Power DOWN	I _{PON} I _{PDN}		0.15 < 5		mA μA
	TOWEI DOWN	-FDN		` ` `		F

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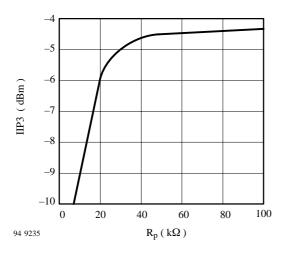


Figure 3. IIP3 versus resistor Rp, IF: 900 MHz

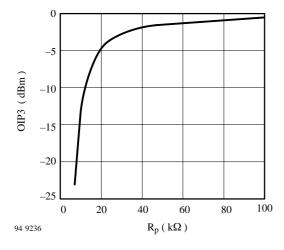


Figure 4. OIP3 versus resistor Rp, IF: 900 MHz

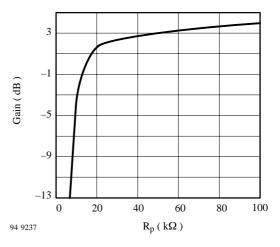


Figure 5. Gain versus resistor Rp, LO: 1030 MHz, level –10 dBm; RF: 130 MHz, –30 dBm, R_T = 56 Ω

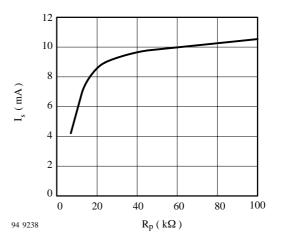


Figure 6. Supply current I_S versus resistor Rp



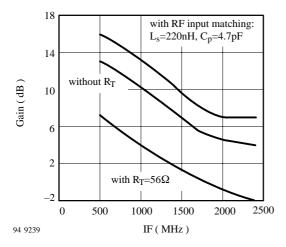


Figure 7. Gain versus IF output frequency, LO level: -6 dBm, RF: 130 MHz, -35 dBm; parameter: RF input termination

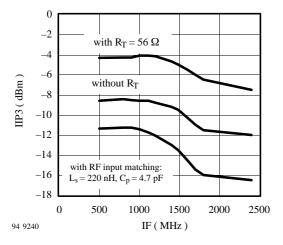


Figure 8. IIP3 versus IF output frequency, LO level: -6 dBm; RF: 130 MHz / 130.1 MHz, -35 dBm; parameter: RF input termination

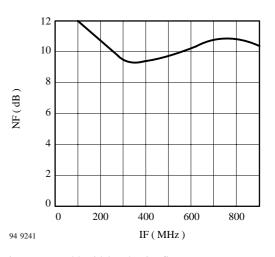


Figure 9. Double sideband noise figure versus IF output frequency; LO: 1000 MHz, level 0 dBm; no RF input matching, $$R_{\rm T}$$ left out

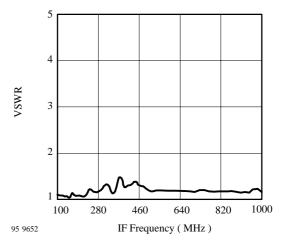
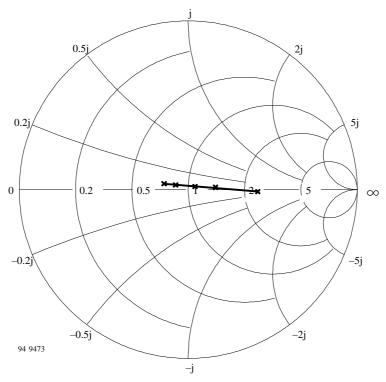


Figure 10. Typical VSWR frequency response of the IF output, $R_P = \infty$

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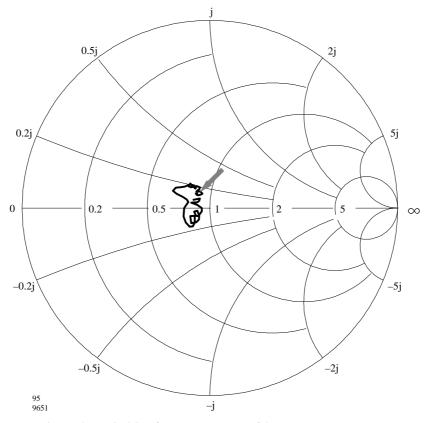


Figure 12. Typical S11 frequency response of the IF output, $R_P=\,\infty\,,$ IF frequency from 100 MHz to 1000 MHz, marker: 900 MHz



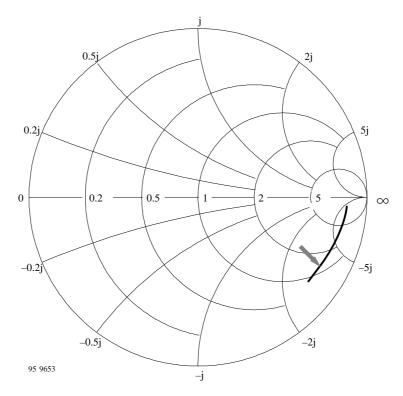


Figure 13. Typical S11 frequency response of the RF intput, $R_P=\infty$, $R_T=\infty$ RF frequency from 100 MHz to 1000 MHz, marker: 900 MHz

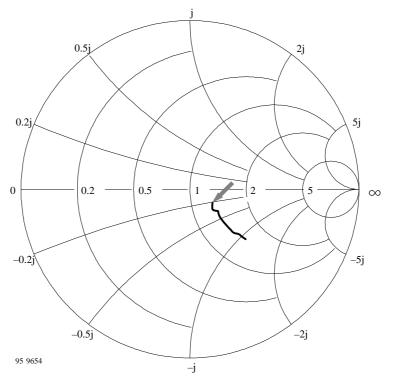


Figure 14. Typical S11 frequency response of the LO intput, $R_P=\,\infty\,,$ LO frequency from 100 MHz to 1000 MHz, marker: 900 MHz

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Application

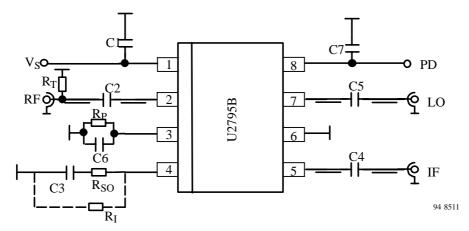


Figure 15.

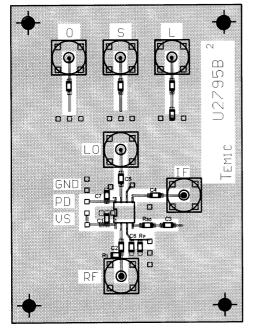
Part List			
C 1	10 nF		
C2, C3, C4, C5, C6, C7	100 pF		
*R _P			
	50-Ω Microstrip		
*R _{SO}	68 Ω		
	optional		
R _T	56 Ω		

If the part-list values are used, the PD settling time is < 20 μs . Using other values, time requirements in burst-mode applications have to be considered.

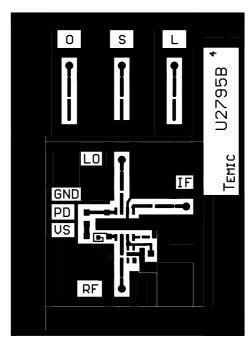
The values of R_{SO} and R_P depend on the input and output condition requirements. For R_{SO} , 68 Ω is recommended.

By means of the optional $R_{I,}$ the intercept and compression point can be slightly increased; values between 500 Ω and 1 $k\Omega$ are suitable. Please note that such modification will also increase the supply current.

Application Circuit (Evaluation Board)



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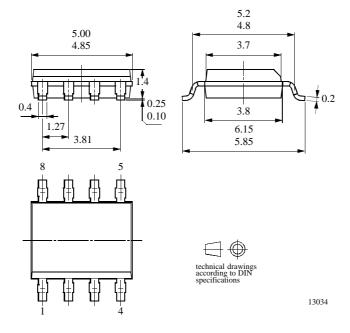
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Figure 16. Figure 17.



Package Information

Package SO8
Dimensions in mm



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Ozone Depleting Substances Policy Statement

It is the policy of Atmel Germany GmbH to

- 1. Meet all present and future national and international statutory requirements.
- Regularly and continuously improve the performance of our products, processes, distribution and operating systems with respect to their impact on the health and safety of our employees and the public, as well as their impact on the environment.

It is particular concern to control or eliminate releases of those substances into the atmosphere which are known as ozone depleting substances (ODSs).

The Montreal Protocol (1987) and its London Amendments (1990) intend to severely restrict the use of ODSs and forbid their use within the next ten years. Various national and international initiatives are pressing for an earlier ban on these substances.

Atmel Germany GmbH has been able to use its policy of continuous improvements to eliminate the use of ODSs listed in the following documents.

- 1. Annex A, B and list of transitional substances of the Montreal Protocol and the London Amendments respectively
- 2. Class I and II ozone depleting substances in the Clean Air Act Amendments of 1990 by the Environmental Protection Agency (EPA) in the USA
- 3. Council Decision 88/540/EEC and 91/690/EEC Annex A, B and C (transitional substances) respectively.

Atmel Germany GmbH can certify that our semiconductors are not manufactured with ozone depleting substances and do not contain such substances.

We reserve the right to make changes to improve technical design and may do so without further notice.

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Data sheets can also be retrieved from the Internet: http://www.atmel-wm.com

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