

PM7339



S/UNI-CDB

**SATURN USER NETWORK INTERFACE
FOR QUAD CELL DELINEATION BLOCK**

DATASHEET

PROPRIETARY AND CONFIDENTIAL

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1 FEATURES

- Quad cell delineation device operating up to a maximum rate of 52 Mbit/s.
- Provides a UTOPIA Level 2 compatible ATM-PHY Interface.
- Implements the Physical Layer Convergence Protocol (PLCP) for DS1 transmission systems according to the ATM Forum User Network Interface Specification and ANSI TA-TSY-000773, TA-TSY-000772, and E1 transmission systems according to the ETSI 300-269 and ETSI 300-270.
- Uses the PMC-Sierra PM4341 T1XC, PM4344 TQUAD, PM6341 E1XC, and PM6344 EQUAD T1 and E1 framer/line interface chips for DS1 and E1 applications.
- Provides programmable pseudo-random test pattern generation, detection, and analysis features.
- Provides integral transmit and receive HDLC controllers with 128-byte FIFO depths.
- Provides performance monitoring counters suitable for accumulation periods of up to 1 second.
- Provides an 8-bit microprocessor interface for configuration, control and status monitoring.
- Provides a standard 5 signal P1149.1 JTAG test port for boundary scan board test purposes.
- Low power 3.3V CMOS technology with 5V tolerant inputs.
- Available in a high density 256-pin SBGA package (27mm x 27mm).

The receiver section:

- Provides PLCP frame synchronization, path overhead extraction, and cell extraction for DS1 PLCP and E1 PLCP formatted streams.
- Provides a 50 MHz 8-bit wide or 16-bit wide Utopia FIFO buffer in the receive path with parity support, and multi-PHY (Level 2) control signals.

- Provides ATM framing using cell delineation. ATM cell delineation may optionally be disabled to allow passing of all cell bytes regardless of cell delineation status.
- Provides cell descrambling, header check sequence (HCS) error detection, idle cell filtering, header descrambling (for use with PPP packets), and accumulates the number of received idle cells, the number of received cells written to the FIFO, and the number of HCS errors.
- Provides a four cell FIFO for rate decoupling between the line, and a higher layer processing entity. FIFO latency may be reduced by changing the number of operational cell FIFOs.
- Provides programmable pseudo-random test-sequence detection (up to $2^{32}-1$ bit length patterns conforming to ITU-T O.151 standards) and analysis features.

The transmitter section:

- Provides a 50 MHz 8-bit wide or 16-bit wide Utopia FIFO buffer in the transmit path with parity support and multi-PHY (Level 2) control signals.
- Provides optional ATM cell scrambling, header scrambling (for use with PPP packets), HCS generation/insertion, programmable idle cell insertion, diagnostics features and accumulates transmitted cells read from the FIFO.
- Provides a four cell FIFO for rate decoupling between the line and a higher layer processing entity. FIFO latency may be reduced by changing the number of operational cell FIFOs.
- Provides an 8 kHz reference input for locking the transmit PLCP frame rate to an externally applied frame reference.
- Provides programmable pseudo-random test sequence generation (up to $2^{32}-1$ bit length sequences conforming to ITU-T O.151 standards). Diagnostic abilities include single bit error insertion or error insertion at bit error rates ranging from 10^{-1} to 10^{-7} .

Loopback features:

- Provides for diagnostic loopbacks and line loopbacks.

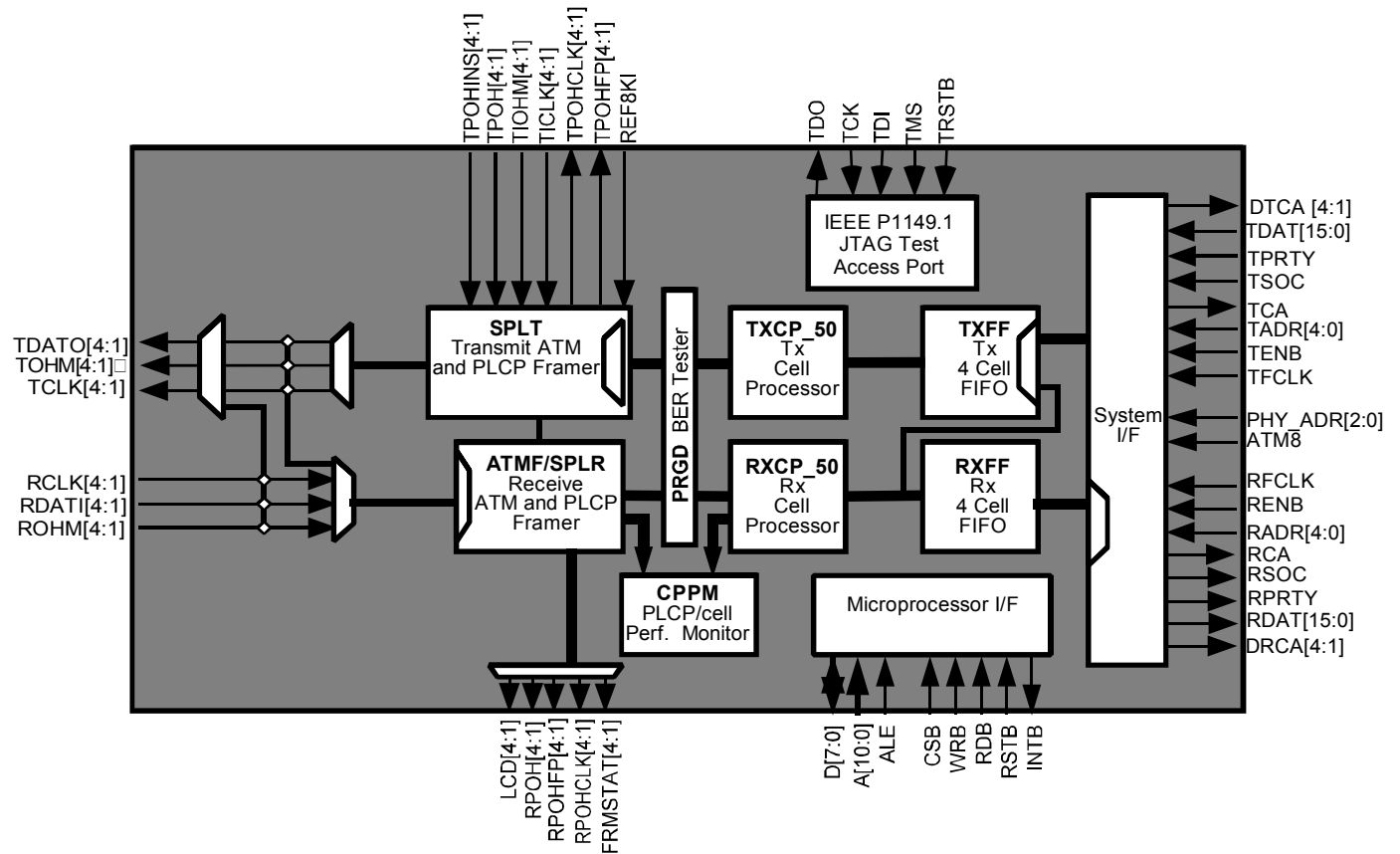
2 APPLICATIONS

- ATM Switches, Multiplexers, and Routers
- SMDS Switches, Multiplexers and Routers
- DSLAM
- Integrated Access Devices (IAD)

3 REFERENCES

1. ANSI T1.627 - 1993, "Broadband ISDN - ATM Layer Functionality and Specification".
2. ANSI T1.646 - 1995, "Broadband ISDN - Physical Layer Specification for User-Network Interfaces Including DS1/ATM".
3. ATM Forum - ATM User-Network Interface Specification, V3.1, October, 1995.
4. ATM Forum - "UTOPIA, An ATM PHY Interface Specification, Level 2, Version 1", June, 1995.
5. Bell Communications Research, TA-TSY-000773 - "Local Access System Generic Requirements, Objectives, and Interface in Support of Switched Multi-megabit Data Service" Issue 2, March 1990 and Supplement 1, December 1990.
6. ETS 300 269 Draft Standard T/NA(91)17 - "Metropolitan Area Network Physical Layer Convergence Procedure for 2.048 Mbit/s", April 1994.
7. ITU-T Recommendation O.151 - "Error Performance Measuring Equipment Operating at the Primary Rate and Above", October, 1992.
8. ITU-T Recommendation I.432 - "B-ISDN User-Network Interface - Physical Layer Specification", 1993
9. ITU-T Recommendation G.704 - "General Aspects of Digital Transmission Systems; Terminal Equipments - Synchronous Frame Structures Used At 1544, 6312, 2048, 8488 and 44 736 kbit/s Hierarchical Levels", July, 1995.

4 S/UNI-CDB BLOCK DIAGRAM



5 DATASHEET OVERVIEW

The PM7339 S/UNI-CDB is functionally equivalent to a PM7346 S/UNI-QJET placed in DS3/E3/J2 Frammer Bypass mode. The devices are software compatible and pin compatible. This datasheet provides a complete pin-out description for the S/UNI-CDB, as well as any differences between these devices. A software initialization sequence is required for the device to operate properly. This software initialization is described in section 10.1. For a complete functional and register description, please refer to the SUNI-QJET Datasheet, PMC-960835.

6 PIN DIAGRAM

The S/UNI-CDB is packaged in a 256-pin SBGA package having a body size of 27mm by 27mm and a pin pitch of 1.27 mm.

	20	19	18	17	16	15	14	13	12	11	10	9	8	7	6	5	4	3	2	1														
A	VSS	VSS	VSS	TDAT10I	TDAT14I	D[1]	D[5]	VSS	A[3]	A[7]	VSS	VSS	ALE	INTB	TRSTB	TOHM[4]	RCLK[4]	VSS	VSS	VSS	A													
B	VSS	VDD	VDD	TDAT8I	TDAT13I	D[0]	D[4]	A[0]	A[2]	A[6]	A[10]	WRB	TDO	TCK	TCLK[4]	TDAT0[3]	VDD	VDD	VSS		B													
C	VSS	VDD	VDD	TDAT7I	TDAT11I	TDAT15I	D[2]	D[6]	A[1]	A[5]	A[9]	CSB	RSTB	TMS	TDAT0[4]	ROHM[4]	TCLK[3]	VDD	VDD	VSS	C													
D	TDAT3I	TDAT4I	TDAT8I	NC	TDAT8I	TDAT12I	VDD	D[3]	D[7]	A[4]	VDD	RDB	TDI	VDD	RDAT4I	TOHM[3]	BIAS	TDAT0[2]	TCLK[2]	RDAT3I	D													
E	TFCLK	TDAT0I	TDAT2I	TDAT5I	BOTTOM VIEW												TOHM[2]	ROHM[3]	RDAT2I	ROHM[2]			E											
F	TADR[2]	TADR[3]	TADR[4]	TDAT11													RCLK[3]	RCLK[2]	TDAT0[1]	TOHM[1]														F
G	TSOC	TPRTY	TADR[1]	VDD													VDD	TCLK[1]	ROHM[1]	RCLK[1]														G
H	BIAS	TCA	TENB	TADR0I													RDAT11	VSS	NC	VSS														H
J	VSS	DTCA[2]	DTCA[3]	DTCA[4]													VSS	NC	NC	NC														J
K	VSS	DTCA[1]	PHY_ADR[2]	VDD													NC	VSS	VSS	NC														K
L	PHY_ADR[1]	PHY_ADR[0]	ATMB	DRC[4]													VDD	NC	NC	VSS														L
M	DRC[3]	DRC[2]	DRC[1]	RSOC													VSS	NC	NC	VSS														M
N	VSS	RCA	RENB	RADR[3]													NC	NC	NC	VSS														N
P	RFCLK	RADR[4]	RADR[2]	VDD													VDD	VSS	NC	NC														P
R	RADR[1]	RADR[0]	RPRTY	RDAT13I	NC	NC	NC	VSS														R												
T	RDAT15I	RDAT14I	RDAT12I	RDAT8I	FRMSTAT[2]	REFBK[1]	NC	NC														T												
U	RDAT11I	RDAT10I	RDAT8I	BIAS	RDAT6I	RDAT2I	VDD	TPOHCLK[4]	REFBK[0]	VDD	RPOHCLK[3]	TPOHNS[2]	RPOH[2]	VDD	TPOHCLK[1]	RPOHCLK[1]	BIAS	FRMSTAT[1]	FRMSTAT[3]	FRMSTAT[4]	U													
V	VSS	VDD	VDD	RDAT7I	RDAT3I	TCLK[4]	TPOHNS[4]	RPOH[4]	TIOHM[3]	TPOHCLK[3]	RPOH[3]	TIOHM[2]	TPOHCLK[2]	RPOHCLK[2]	TIOHM[1]	TPOHFP[1]	REFBK[0]	VDD	VDD	VSS		V												
W	VSS	VDD	VDD	RDAT5I	RDAT1I	TIOHM[4]	TPOHFP[4]	RPOHCLK[4]	TPOH[3]	TPOHNS[3]	LCD[3]	TCLK[2]	TPOH[2]	LCD[2]	TCLK[1]	TPOHNS[1]	RPOH[1]	VDD	VDD	VSS		W												
Y	VSS	VSS	VSS	RDAT4I	RDAT0I	TPOH[4]	LCD[4]	TCLK[3]	VSS	VSS	TPOHFP[3]	REFBK[0]	VSS	TPOHFP[2]	REFBK[2]	TPOH[1]	LCD[1]	VSS	VSS	VSS		Y												
	20	19	18	17	16	15	14	13	12	11	10	9	8	7	6	5	4	3	2	1														

7 PIN DESCRIPTION

Pin Name	Type	Pin No.	Function
TDATO[4]	Output	C6	Transmit Data (TDATO[4:1]). TDATO[4:1] contains the transmit data stream when the single-rail (unipolar) output format is enabled The TDATO[4:1] pin function selection is controlled by the TFRM[1:0] and the TUNI bits in the S/UNI-CDB Transmit Configuration Registers. TDATO[4:1] is updated on the falling edge of TCLK[4:1] by default, and may be configured to be updated on the rising edge of TCLK[4:1] through the TCLKINV bit in the S/UNI-CDB Transmit Configuration Registers. Finally, TDATO[4:1] can be updated on the rising edge of TICLK[4:1], enabled by the TICLK bit in the S/UNI-CDB Transmit Configuration Registers.
TDATO[3]		B4	
TDATO[2]		D3	
TDATO[1]		F2	

Pin Name	Type	Pin No.	Function
TOHM[4] TOHM[3] TOHM[2] TOHM[1]	Output	A5 D5 E4 F1	<p>Transmit Overhead Mask (TOHM[4:1]). TOHM[4:1] indicates the position of overhead bits (non-payload bits) in the transmission system stream aligned with TDATO[4:1].</p> <p>When a PLCP formatted signal is transmitted, TOHM[4:1] is set to logic 1 once per transmission frame, and indicates the DS1 or E1 frame alignment.</p> <p>TOHM[4:1] is a delayed version of the TIOHM[4:1] input, and indicates the position of each overhead bit in the transmission frame. TOHM[4:1] is updated on the falling edge of TCLK[4:1].</p> <p>The TOHM[4:1] pin function selection is controlled by the TFRM[1:0] and the TUNI bits in the S/UNI-CDB Transmit Configuration Registers. TOHM[4:1] is updated on the falling edge of TCLK[4:1] by default, and may be enabled to be updated on the rising edge of TCLK[4:1]. This sampling is controlled by the TCLKINV bit in the S/UNI-CDB Transmit Configuration Registers. Finally, TOHM[4:1] can be updated on the rising edge of TICLK[4:1], enabled by the TICLK bit in the S/UNI-CDB Transmit Configuration Registers.</p>
TCLK[4] TCLK[3] TCLK[2] TCLK[1]	Output	B5 C4 D2 G3	<p>Transmit Output Clock (TCLK[4:1]). TCLK[4:1] provides the transmit direction timing. TCLK[4:1] is a buffered version of TICLK[4:1] and can be enabled to update the TDATO[4:1] and TOHM[4:1] outputs on its rising or falling edge.</p>

Pin Name	Type	Pin No.	Function
RDATI[4] RDATI[3] RDATI[2] RDATI[1]	Input	D6 D1 E2 H4	<p>Receive Data (RDATI[4:1]). RDATI[4:1] contains the data stream when the single-rail (unipolar) NRZ input format is enabled.</p> <p>The RDATI[4:1] pin function selection is controlled by the RFRM[1:0] bits in the S/UNI-CDB Configuration Registers. RDATI[4:1] is sampled on the rising edge of RCLK[4:1] by default, and may be enabled to be sampled on the falling edge of RCLK[4:1]. This sampling is controlled by the RCLKINV bit in the S/UNI-CDB Receive Configuration Registers.</p>
ROHM[4] ROHM[3] ROHM[2] ROHM[1]	Input	C5 E3 E1 G2	<p>Receive Overhead Mask (ROHM[4:1]). When a DS1 or E1 PLCP or ATM direct-mapped signal is received, ROHM[4:1] is pulsed once per transmission frame, and indicates the DS1 or E1 frame alignment relative to the RDATI[4:1] data stream. When an alternate frame-based signal is received, ROHM[4:1] indicates the position of each overhead bit in the transmission frame.</p> <p>The RLCV/ROHM[4:1] pin function selection is controlled by the RFRM[1:0] bits in the S/UNI-CDB Receive Configuration Registers, and the PLCPEN bit in the SPLR Configuration register. RLCV[4:1], and ROHM[4:1] are sampled on the rising edge of RCLK[4:1] by default, and may be enabled to be sampled on the falling edge of RCLK[4:1]. This sampling is controlled by the RCLKINV bit in the S/UNI-CDB Receive Configuration Registers.</p>
RCLK[4] RCLK[3] RCLK[2] RCLK[1]	Input	A4 F4 F3 G1	<p>Receive Clock (RCLK[4:1]). RCLK[4:1] provides the receive direction timing. RCLK[4:1] is the externally recovered transmission system baud rate clock that samples the RDATI[4:1] and RLCV/ROHM[4:1] inputs on its rising or falling edge.</p>

Pin Name	Type	Pin No.	Function
REF8KI	Input	T3	<p>Reference 8 kHz Input (REF8KI). The PLCP frame rate is locked to an external 8 kHz reference applied on this input . An internal phase-frequency detector compares the transmit PLCP frame rate with the externally applied 8 kHz reference and adjusts the PLCP frame rate.</p> <p>The REF8KI input must transition high once every 125 μs for correct operation. The REF8KI input is treated as an asynchronous signal and must be “glitch-free”. If the LOOPT register bit is logic 1, the PLCP frame rate is locked to the RPOHFP[x] signal instead of the REF8KI input.</p>
TPOHINS[4] TPOHINS[3] TPOHINS[2] TPOHINS[1]	Input	V14 W11 U9 W5	<p>Transmit Path Overhead Insertion (TPOHINS[4:1]). TPOHINS[4:1] controls the insertion of PLCP overhead octets on the TPOH[4:1] input. When TPOHINS[4:1] is logic 1, the associated overhead bit in the TPOH[4:1] stream is inserted in the transmit PLCP frame. When TPOHINS[4:1] is logic 0, the PLCP path overhead bit is generated and inserted internally. TPOHINS[4:1] is sampled on the rising edge of TPOHCLK[4:1].</p>
TPOH[4] TPOH[3] TPOH[2] TPOH[1]	Input	Y15 W12 W8 Y5	<p>Transmit PLCP Overhead Data (TPOH[4:1]). TPOH[4:1] contains the PLCP path overhead octets (Zn, F1, B1, G1, M1, M2, and C1) which may be inserted in the transmit PLCP frame. The octet data on TPOH[4:1] is shifted in order from the most significant bit (bit 1) to the least significant bit (bit 8). TPOH[4:1] is sampled on the rising edge of TPOHCLK[4:1].</p>

Pin Name	Type	Pin No.	Function
TCELL[4] TCELL[3] TCELL[2] TCELL[1]	Output	W14 Y10 Y7 V5	Transmit Cell Indication (TCELL[4:1]). TCELL[x] is valid when the TCELL bit in the S/UNI-CDB Misc. register (09BH, 19BH, 29BH, 39BH) is set. TCELL[x] pulses once for every cell (idle or assigned) transmitted. TCELL[x] is updated using timing derived from the transmit input clock (TICLK[x]), and is active for a minimum of 8 TICLK[x] periods (or 8 RCLK[x] periods if loop-timed).
TPOHCLK[4] TPOHCLK[3] TPOHCLK[2] TPOHCLK[1]	Output	U13 V11 V8 U6	Transmit PLCP Overhead Clock (TPOHCLK[4:1]). TPOHCLK[4:1] is active when PLCP processing is enabled. TPOHCLK[4:1] is nominally a 26.7 kHz clock for a DS1 PLCP frame and a 33.7 kHz clock for an E1 based PLCP frame. TPOHFP[4:1] is updated on the falling edge of TPOHCLK[4:1]. TPOH[4:1], and TPOHINS[4:1] are sampled on the rising edge of TPOHCLK[4:1].
TIOHM[4] TIOHM[3] TIOHM[2] TIOHM[1]	Input	W15 V12 V9 V6	Transmit Input Overhead Mask (TIOHM[4:1]). TIOHM[4:1] indicates the position of overhead bits when not configured for DS1 or E1 transmission system streams. TIOHM[4:1] is delayed internally to produce the TOHM[4:1] output. When configured for operation over a DS1 or an E1 transmission system sublayer, TIOHM[4:1] is not required, and should be set to logic 0. When configured for other transmission systems, TIOHM[4:1] is set to logic 1 for each overhead bit position. TIOHM[4:1] is set to logic 0 if the transmission system contains no overhead bits. TIOHM[4:1] is sampled on the rising edge of TICLK[4:1].

Pin Name	Type	Pin No.	Function
TICLK[4] TICLK[3] TICLK[2] TICLK[1]	Input	V15 Y13 W9 W6	Transmit Input Clock (TICLK[4:1]). TICLK[4:1] provides the transmit direction timing. TICLK[4:1] is the externally generated transmission system baud rate clock. It is internally buffered to produce the transmit clock output, TCLK[4:1], and can be enabled to update the TDATO[4:1] and TOHM[4:1] outputs on the TICLK[4:1] rising edge. The TICLK[4:1] maximum frequency is 52 MHz.
RPOHFP[4] RPOHFP[3] RPOHFP[2] RPOHFP[1]	Output	U12 Y9 Y6 V4	Receive PLCP Overhead Frame Position (RPOHFP[4:1]). RPOHFP[4:1] locates the individual PLCP path overhead bits in the receive overhead data stream, RPOH[4:1]. RPOHFP[4:1] is logic 1 while bit 1 (the most significant bit) of the path user channel octet (F1) is present in the RPOH[4:1] stream. RPOHFP[4:1] is updated on the falling edge of RPOHCLK[4:1]. RPOHFP[4:1] is available when the PLCPEN register bit is logic 1 in the SPLR Configuration Register.
RPOH[4] RPOH[3] RPOH[2] RPOH[1]	Output	V13 V10 U8 W4	Receive PLCP Overhead Data (RPOH[4:1]). RPOH[4:1] contains the PLCP path overhead octets (Zn, F1, B1, G1, M1, M2, and C1) extracted from the received PLCP frame when the PLCP layer is in-frame. When the PLCP layer is in the loss of frame state, RPOH[4:1] is forced to all ones. The octet data on RPOH[4:1] is shifted out in order from the most significant bit (bit 1) to the least significant bit (bit 8). RPOH[4:1] is updated on the falling edge of RPOHCLK[4:1].

Pin Name	Type	Pin No.	Function
RPOHCLK[4] RPOHCLK[3] RPOHCLK[2] RPOHCLK[1]	Output	W13 U10 V7 U5	Receive PLCP Overhead Clock (RPOHCLK[4:1]). RPOHCLK[4:1] is active when PLCP processing is enabled. The frequency of this signal depends on the selected PLCP format. RPOHCLK[4:1] is nominally a 26.7 kHz clock for a DS1 PLCP frame and a 33.7 kHz clock for an E1 based PLCP frame. RPOHFP[4:1] and RPOH[4:1] are updated on the falling edge of RPOHCLK[4:1].
LCD[4] LCD[3] LCD[2] LCD[1]	Output	Y14 W10 W7 Y4	Loss of Cell Delineation (LCD[4:1]). LCD[4:1] is an active high signal which is asserted while the ATM cell processor has detected a Loss of Cell Delineation defect.
FRMSTAT[4] FRMSTAT[3] FRMSTAT[2] FRMSTAT[1]	Output	U1 U2 T4 U3	Framer Status (FRMSTAT[4:1]). FRMSTAT[4:1] is an active high signal which can be configured to show when the PLCP framer has detected certain conditions. The FRMSTAT[4:1] outputs can be programmed via the STATSEL[2:0] bits in the S/UNI-CDB Configuration 2 Register to indicate: PLCP Loss of Frame, PLCP Out of Frame, AIS, and Loss of Signal. FRMSTAT[4:1] should be treated as a glitch free asynchronous signal.
ATM8	Input	L18	ATM Interface Bus Width Selection (ATM8). The ATM8 input pin determines whether the S/UNI-CDB works with a 8-bit wide interface (RDAT[7:0] and TDAT[7:0]) or a 16-bit wide interface (RDAT[15:0] and TDAT[15:0]). If ATM8 is set to logic 1, then the 8-bit wide interface is chosen. If ATM8 is set to logic 0, then the 16-bit wide interface is chosen.

Pin Name	Type	Pin No.	Function
TDAT[15]	Input	C15	<p>Transmit Cell Data Bus (TDAT[15:0]). This bus carries the ATM cell octets that are written to the selected transmit FIFO. TDAT[15:0] is sampled on the rising edge of TFCLK and is considered valid only when TENB is simultaneously asserted and the S/UNI-CDB has been selected via the TADR[4:2] and PHY_ADR[2:0] inputs.</p> <p>The S/UNI-CDB can be configured to operate with an 8-bit wide or 16-bit wide ATM data interface via the ATM8 input pin. When configured for the 8-bit wide interface, TDAT[15:8] are not used and should be tied to ground.</p>
TDAT[14]		A16	
TDAT[13]		B16	
TDAT[12]		D15	
TDAT[11]		C16	
TDAT[10]		A17	
TDAT[9]		B17	
TDAT[8]		D16	
TDAT[7]		C17	
TDAT[6]		D18	
TDAT[5]		E17	
TDAT[4]		D19	
TDAT[3]		D20	
TDAT[2]		E18	
TDAT[1]		F17	
TDAT[0]		E19	

Pin Name	Type	Pin No.	Function
TPRTY	Input	G19	<p>Transmit bus parity (TPRTY). The transmit parity (TPRTY) signal indicates the parity of the TDAT[15:0] or TDAT[7:0] bus. If configured for the 8-bit bus (via the ATM8 input pin), then parity is calculated over TDAT[7:0]. If configured for the 16-bit bus, then parity is calculated over TDAT[15:0].</p> <p>A parity error is indicated by a status bit and a maskable interrupt. Cells with parity errors are inserted in the transmit stream, so the TPRTY input may be unused.</p> <p>Odd or even parity selection is made using the TPTY register bit. TPRTY is sampled on the rising edge of TFCLK and is considered valid only when TENB is simultaneously asserted and the S/UNI-CDB has been selected via the TADR[4:0] and PHY_ADR[2:0] inputs.</p>
TSOC	Input	G20	<p>Transmit Start of Cell (TSOC). The transmit start of cell (TSOC) signal marks the start of cell on the TDAT bus. When TSOC is high, the first word of the cell structure is present on the TDAT bus. It is not necessary for TSOC to be present for each cell. An interrupt may be generated if TSOC is high during any word other than the first word of the cell structure. TSOC is sampled on the rising edge of TFCLK and is considered valid only when TENB is simultaneously asserted and the S/UNI-CDB has been selected via the TADR[4:2] and PHY_ADR[2:0] inputs.</p>

Pin Name	Type	Pin No.	Function
TENB	Input	H18	Transmit Multi-Phy Write Enable (TENB). The TENB signal is an active low input which is used along with the TADR[4:0] inputs to initiate writes to the transmit FIFOs. When sampled low using the rising edge of TFCLK, the word on the TDAT bus is written into the transmit FIFO selected by the TADR[4:0] address bus. When sampled high using the rising edge of TFCLK, no write is performed, but the TADR[4:0] address is latched to identify the transmit FIFO to be accessed. A complete 53 octet cell must be written to the transmit FIFO before it is inserted into the transmit stream. Idle cells are inserted when a complete cell is not available.
TADR[4] TADR[3] TADR[2] TADR[1] TADR[0]	Input	F18 F19 F20 G18 H17	Transmit Address (TADR[4:0]). The TADR[4:0] bus is used to select the FIFO (and hence port) that is written to using the TENB signal and the FIFO whose cell-available signal is visible on the TCA output. TADR[4:0] is sampled on the rising edge of TFCLK together with TENB. Note that the null-PHY address 1FH is an invalid address and will not be identified to any port on the S/UNI-CDB.

Pin Name	Type	Pin No.	Function
TCA	Output	H19	<p>Transmit Multi-Phy Cell Available (TCA). The TCA signal indicates when a cell is available in the transmit FIFO for the port selected by TADR[4:0]. When high, TCA indicates that the corresponding transmit FIFO is not full and a complete cell may be written. When TCA goes low, it can be configured to indicate either that the corresponding transmit FIFO is near full or that the corresponding transmit FIFO is full. TCA will transition low on the rising edge of TFCLK which samples Payload byte 43 (TCALEVEL0=0) or 47 (TCALEVEL0=1) for the 8-bit interface (ATM8=1), or the rising edge of TFCLK which samples Payload word 19 (TCALEVEL0=0) or 23 (TCALEVEL0=1) for the 16-bit interface (ATM8=0) if the PHY being polled is the same as the PHY in use. To reduce FIFO latency, the FIFO depth at which TCA indicates "full" can be set to one, two, three or four cells. Note that regardless of what fill level TCA is set to indicate "full" at, the transmit cell processor can store 4 complete cells.</p> <p>TCA is tri-stated when either the null-PHY address (1FH) or an address not matching the address space set by PHY_ADR[2:0] is latched (by TFCLK) from the TADR[4:2] inputs.</p> <p>The polarity of TCA (with respect the the description above) is inverted when the TCAINV register bit is set to logic 1.</p>
TFCLK	Input	E20	<p>Transmit FIFO Write Clock (TFCLK). This signal is used to write ATM cells to the four cell transmit FIFOs. TFCLK cycles at a 52 MHz or lower instantaneous rate.</p> <p>Please note that the TFCLK input is not 5 V tolerant, it is a 3.3 V only input pin.</p>

Pin Name	Type	Pin No.	Function
DTCA[4] DTCA[3] DTCA[2] DTCA[1]	Output	J17 J18 J19 K19	<p>Direct Access Transmit Cell Available (DTCA[4:1]). These output signals indicate when a cell is available in the transmit FIFO for the corresponding port. When high, DTCA[x] indicates that the corresponding transmit FIFO is not full and a complete cell may be written. DTCA[x] can be configured to indicate either that the corresponding transmit FIFO is near full and can accept no more than four writes or that the corresponding transmit FIFO is full. DTCA[x] will thus transition low on the rising edge of TFLCK which samples Payload byte 43 (TCALEVEL0=0) or 47 (TCALEVEL0=1) for the 8-bit interface (ATM8=1), or the rising edge of TFCLK which samples Payload word 19 (TCALEVEL0=0) or 23 (TCALEVEL0=1) for the 16-bit interface (ATM8=0). To reduce FIFO latency, the FIFO depth at which DTCA[x] indicates "full" can be set to one, two, three or four cells. Note that regardless of what fill level DTCA[x] is set to indicate "full" at, the transmit cell processor can store 4 complete cells.</p> <p>The polarity of DTCA[x] (with respect the the description above) is inverted when the TCAINV register bit is set to logic 1.</p> <p>The DTCA[4:1] outputs can be used to support Utopia Direct Access mode.</p>

Pin Name	Type	Pin No.	Function
RDAT[15] RDAT[14] RDAT[13] RDAT[12] RDAT[11] RDAT[10] RDAT[9] RDAT[8] RDAT[7] RDAT[6] RDAT[5] RDAT[4] RDAT[3] RDAT[2] RDAT[1] RDAT[0]	Output	T20 T19 R17 T18 U20 U19 T17 U18 V17 U16 W17 Y17 V16 U15 W16 Y16	<p>Receive Cell Data Bus (RDAT[15:0]). This bus carries the ATM cell octets that are read from the receive ATM FIFO selected by RADR[4:0]. RDAT[15:0] is tri-stated when RENB is high. RDAT[15:0] is updated on the rising edge of RFCLK.</p> <p>The S/UNI-CDB can be configured to operate with an 8-bit wide or 16-bit wide ATM data interface via the ATM8 input pin. RDAT[15:8] will remain tri-stated if ATM8 is set to logic 1.</p> <p>RDAT[15:0] is tri-stated when either the null-PHY address (1FH) or an address not matching the address space set by PHY_ADR[2:0] is latched from the RADR[4:2] inputs when RENB is high.</p>
RPRTY	Output	R18	<p>Receive Parity (RPRTY). The receive parity (RPRTY) signal indicates the parity of the RDAT bus.</p> <p>The S/UNI-CDB can be configured to operate with an 8-bit wide or 16-bit wide ATM data interface via the ATM8 input pin. In the 8-bit mode, RPRTY reflects the parity of RDAT[7:0]. In the 16-bit mode, RPRTY reflects the parity of RDAT[15:0].</p> <p>Odd or even parity selection is made using the RXPTYP register bit.</p> <p>RPRTY is tri-stated when either the null-PHY address (1FH) or an address not matching the address space set by PHY_ADR[2:0] is latched from the RADR[4:2] inputs when RENB is high.</p>

Pin Name	Type	Pin No.	Function
RSOC	Output	M17	<p>Receive Start of Cell (RSOC). This signal marks the start of cell on the RDATA bus. RSOC marks the start of the cell on the RDATA bus.</p> <p>RSOC is tri-stated when either the null-PHY address (1FH) or an address not matching the address space set by PHY_ADR[2:0] is latched from the RADDR[4:0] inputs when RENB is high.</p>
RENB	Input	N18	<p>Receive Multi-Phy Read Enable (RENB). The RENB signal is used to initiate reads from the receive FIFOs. When sampled low using the rising edge of RFCLK, a byte is read (if one is available) from the receive FIFO selected by the RADDR[4:0] address bus and output on the RDATA bus. When sampled high using the rising edge of RFCLK, no read is performed and RDATA[15:0], RPRTY, and RSOC are tri-stated, and the address on RADDR[4:0] is latched to select the device or port for the next ATM FIFO access. RENB must operate in conjunction with RFCLK to access the FIFOs at a high enough rate to prevent FIFO overflows. The ATM layer device may de-assert RENB at anytime it is unable to accept another byte.</p>
RADDR[4] RADDR[3] RADDR[2] RADDR[1] RADDR[0]	Input	P19 N17 P18 R20 R19	<p>Receive Address (RADDR[4:0]). The RADDR[4:1] signal is used to select the FIFO (and hence port) that is read from using the RENB signal and the FIFO whose cell-available signal is visible on the RCA output. RADDR[4:0] is sampled on the rising edge of RFCLK together with RENB.</p> <p>Note that the null-PHY address 1FH is an invalid address and will not be identified to any port on the S/UNI-CDB.</p>

Pin Name	Type	Pin No.	Function
RCA	Output	N19	<p>Receive Multi-Phy Cell Available (RCA). The RCA signal indicates when a cell is available in the receive FIFO for the port selected by RADR[4:0]. RCA can be configured to be de-asserted when either zero or four bytes remain in the selected/addressed FIFO. RCA will thus transition low on the rising edge of RFCLK after Payload byte 48 (RCALEVEL0=1) or 43 (RCALEVEL0=0) is output for the 8-bit interface (ATM8=1), or after Payload word 24 (RCALEVEL0=1) or 19 (RCALEVEL0=0) is output for the 16-bit interface (ATM8=0) if the PHY being polled is the same as the PHY in use.</p> <p>RCA is tri-stated when either the null-PHY address (1FH) or an address not matching the address space set by PHY_ADR[2:0] is latched (by RFCLK) from the RADR[4:2] inputs.</p> <p>The polarity of RCA (with respect to the description above) is inverted when the RCAINV register bit is set to logic 1.</p>
RFCLK	Input	P20	<p>Receive FIFO Read Clock (RFCLK). This signal is used to read ATM cells from the receive FIFOs. RFCLK must cycle at a 52 MHz or lower instantaneous rate, but at a high enough rate to avoid FIFO overflows.</p> <p>Please note that the RFCLK input is not 5 V tolerant, it is a 3.3 V only input pin.</p>

Pin Name	Type	Pin No.	Function
DRCA[4] DRCA[3] DRCA[2] DRCA[1]	Output	L17 M20 M19 M18	<p>Direct Access Receive Cell Available (DRCA[4:1]). These output signals indicate when a cell is available in the receive FIFO for the corresponding port. DRCA[4:1] can be configured to be de-asserted when either zero or four bytes remain in the FIFO. DRCA[4:1] will thus transition low on the rising edge of RFCLK after Payload byte 48 (RCALEVEL0=1) or 43 (RCALEVEL0=0) is output for the 8-bit interface (ATM8=1), or after Payload word 24 (RCALEVEL0=1) or 19 (RCALEVEL0=0) is output for the 16-bit interface (ATM8=0).</p> <p>The DRCA[4:1] outputs can be used to support Utopia Direct Access mode.</p>
PHY_ADR[2] PHY_ADR[1] PHY_ADR[0]	Input	K18 L20 L19	<p>Device Identification Address (PHY_ADR[2:0]). The PHY_ADR[2:0] inputs are the most-significant bits of the address space which this S/UNI-CDB occupies. When the PHY_ADR[2:0] inputs match the TADR[4:2] or RADR[4:2] inputs, then one of the four quadrants (as determined by the TADR[1:0] or RADR[1:0] inputs) in this S/UNI-CDB is selected for transmit or receive ATM access.</p> <p>Note that the null-PHY address 1FH is an invalid address and will not be identified to any port on the S/UNI-CDB.</p>
CSB	Input	C9	<p>Active low Chip Select (CSB). This signal must be low to enable S/UNI-CDB register accesses. If CSB is not used, (RDB and WRB determine register reads and writes) then it should be tied to an inverted version of RSTB.</p>

Pin Name	Type	Pin No.	Function
WRB	Input	B8	Active low Write Strobe (WRB). This signal is pulsed low to enable a S/UNI-CDB register write access. The D[7:0] bus is clocked into the addressed register on the rising edge of WRB while CSB is low.
RDB	Input	D9	Active low Read Enable (RDB). This signal is pulsed low to enable a S/UNI-CDB register read access. The S/UNI-CDB drives the D[7:0] bus with the contents of the addressed register while RDB and CSB are both low.
D[7] D[6] D[5] D[4] D[3] D[2] D[1] D[0]	I/O	D12 C13 A14 B14 D13 C14 A15 B15	Bi-directional Data Bus (D[7:0]). The bi-directional data bus D[7:0] is used during S/UNI-CDB register read and write accesses.
A[10] A[9] A[8] A[7] A[6] A[5] A[4] A[3] A[2] A[1] A[0]	Input	B9 B10 C10 A11 B11 C11 D11 A12 B12 C12 B13	Address Bus (A[10:0]). The address bus A[10:0] selects specific registers during S/UNI-CDB register accesses.

Pin Name	Type	Pin No.	Function
RSTB	Input	C8	Active low Reset (RSTB). This signal is set low to asynchronously reset the S/UNI-CDB. RSTB is a Schmitt-trigger input with an integral pull-up resistor.
ALE	Input	A8	Address Latch Enable (ALE). The address latch enable (ALE) is active-high and latches the address bus A[10:0] when low. When ALE is high, the internal address latches are transparent. It allows the S/UNI-CDB to interface to a multiplexed address/data bus. ALE has an integral pull-up resistor.
INTB	Output	A7	Active low Open-Drain Interrupt (INTB). This signal goes low when an unmasked interrupt event is detected on any of the internal interrupt sources. Note that INTB will remain low until all active, unmasked interrupt sources are acknowledged at their source.
TCK	Input	B6	Test Clock (TCK). This signal provides timing for test operations that can be carried out using the IEEE P1149.1 test access port.
TMS	Input	C7	Test Mode Select (TMS). This signal controls the test operations that can be carried out using the IEEE P1149.1 test access port. TMS is sampled on the rising edge of TCK. TMS has an integral pull up resistor.
TDI	Input	D8	Test Data Input (TDI). This signal carries test data into the S/UNI-CDB via the IEEE P1149.1 test access port. TDI is sampled on the rising edge of TCK. TDI has an integral pull up resistor.

Pin Name	Type	Pin No.	Function
TDO	Output	B7	Test Data Output (TDO). This signal carries test data out of the S/UNI-CDB via the IEEE P1149.1 test access port. TDO is updated on the falling edge of TCK. TDO is a tri-state output which is inactive except when scanning of data is in progress.
TRSTB	Input	A6	Active low Test Reset (TRSTB). This signal provides an asynchronous S/UNI-CDB test access port reset via the IEEE P1149.1 test access port. TRSTB is a Schmitt triggered input with an integral pull up resistor. TRSTB must be asserted during the power up sequence. Note that if not used, TRSTB must be connected to the RSTB input.
BIAS	Input	H20 U17 D4 U4	+5V Bias (BIAS). When tied to +5V, the BIAS input is used to bias the wells in the input and I/O pads so that the pads can tolerate 5V on their inputs without forward biasing internal ESD protection devices. When tied to VDD, the inputs and bi-directional inputs will only tolerate input levels up to VDD.

Pin Name	Type	Pin No.	Function
VDD[1]	Power	B2	DC Power. The DC Power pins should be connected to a well-decoupled +3.3V DC supply.
VDD[2]		B3	
VDD[3]		B18	
VDD[4]		B19	
VDD[5]		C2	
VDD[6]		C3	
VDD[7]		C18	
VDD[8]		C19	
VDD[9]		D7	
VDD[10]		D10	
VDD[11]		D14	
VDD[12]		G4	
VDD[13]		G17	
VDD[14]		K17	
VDD[15]		L4	
VDD[16]		P4	
VDD[17]		P17	
VDD[18]		U7	
VDD[19]		U11	
VDD[20]		U14	
VDD[21]		V2	
VDD[22]		V3	
VDD[23]		V18	
VDD[24]		V19	
VDD[25]		W2	
VDD[26]		W3	
VDD[27]		W18	
VDD[28]		W19	

Pin Name	Type	Pin No.	Function
VSS[1]	Ground	A1	DC Ground. The DC Ground pins should be connected to GND.
VSS[2]		A2	
VSS[3]		A3	
VSS[4]		A9	
VSS[5]		A10	
VSS[6]		A13	
VSS[7]		A18	
VSS[8]		A19	
VSS[9]		A20	
VSS[10]		B1	
VSS[11]		B20	
VSS[12]		C1	
VSS[13]		C20	
VSS[14]		H1	
VSS[15]		H3	
VSS[16]		J4	
VSS[17]		J20	
VSS[18]		K2	
VSS[19]		K3	
VSS[20]		K20	
VSS[21]		L1	
VSS[22]		M1	
VSS[23]		M4	
VSS[24]		N1	
VSS[25]		N20	
VSS[26]		P3	
VSS[27]		R1	
VSS[28]		V1	
VSS[29]		V20	

Pin Name	Type	Pin No.	Function
VSS[30]	Ground	W1	DC Ground. The DC Ground pins should be connected to GND.
VSS[31]		W20	
VSS[32]		Y1	
VSS[33]		Y2	
VSS[34]		Y3	
VSS[35]		Y8	
VSS[36]		Y11	
VSS[37]		Y12	
VSS[38]		Y18	
VSS[39]		Y19	
VSS[40]		Y20	

Pin Name	Type	Pin No.	Function
NC	No connect	D17 H2 J1 J2 J3 K1 K4 L2 L3 M2 M3 N2 N3 N4 P1 P2 R2 R3 R4 T1 T2	No connect

Notes on Pin Description:

1. All S/UNI-CDB inputs and bi-directionals present minimum capacitive loading and operate at TTL logic levels.
2. All S/UNI-CDB outputs and bi-directionals have at least 3 mA drive capability. The data bus outputs, D[7:0], have 3 mA drive capability. The FIFO interface outputs, RDAT[15:0], RPRTY, RCA, DRCA[4:1], RSOC, TCA, and DTCA[4:1], have 12 mA drive capability. The outputs TCLK[4:1], TDATO[4:1], TOHM[4:1], TPOHFP[4:1], LCD[4:1], RPOH[4:1], RPOHCLK[4:1], and

RPOHFP[4:1] have 6 mA drive capability. All other outputs have 3 mA drive capability.

3. Inputs RSTB, ALE, TMS, TDI and TRSTB have internal pull-up resistors.
4. RSTB, TRSTB, TMS, TDI, TCK, REF8KI, TFCLK, RFCLK, TICLK[4:1], and RCLK[4:1] are schmitt trigger input pads.
5. RFCLK and TFCLK are 3.3 V only input pins – they are **not** 5 V tolerant. Connecting a 5 V signal to these inputs may result in damage to the part.
6. The VSS [42:1] ground pins are not internally connected together. Failure to connect these pins externally may cause malfunction or damage the S/UNI-CDB.
7. The VDD[28:1] power pins are not internally connected together. Failure to connect these pins externally may cause malfunction or damage the device. These power supply connections must all be utilized and must all connect to a common +3.3 V or ground rail, as appropriate.
8. During power-up and power-down, the voltage on the BIAS pin must be kept equal to or greater than the voltage on the VDD [28:1] pins, to avoid damage to the device.

8 FUNCTIONAL DESCRIPTION

8.1 SPLR PLCP Layer Receiver

The PLCP Layer Receiver (SPLR) Block integrates circuitry to support DS1 and E1 PLCP frame processing. The SPLR provides framing for PLCP based transmission formats.

The SPLR frames to DS1 and E1 based PLCP frames with maximum average reframe times of 635 μ s and 483 μ s respectively. Framing is declared (out of frame is removed) upon finding 2 valid, consecutive sets of framing (A1 and A2) octets and 2 valid and sequential path overhead identifier (POHID) octets. While framed, the A1, A2, and POHID octets are examined. OOF is declared when an error is detected in both the A1 and A2 octets or when 2 consecutive POHID octets are found in error. LOF is declared when an OOF state persists for more than 25 ms, 1 ms, 20 ms, or 1 ms for DS1 and E1 PLCP formats respectively. If the OOF events are intermittent, the LOF counter is decremented at a rate 1/10 (E1, DS1 PLCP) of the incrementing rate. LOF is thus removed when an in-frame state persists for more than 250 ms for a DS1 signal or 200 ms for an E1 signal. When LOF is declared, PLCP reframe is initiated.

When in frame, the SPLR extracts the path overhead octets and outputs them bit serially on output RPOH, along with the RPOHCLK and RPOHFP outputs. Framing octet errors and path overhead identifier octet errors are indicated as frame errors. Bit interleaved parity errors and far end block errors are indicated. The yellow signal bit is extracted and accumulated to indicate yellow alarms. Yellow alarm is declared when 10 consecutive yellow signal bits are set to logical 1; it is removed when 10 consecutive received yellow signal bits are set to logical 0. The C1 octet is examined to maintain nibble alignment with the incoming transmission system sublayer bit stream.

8.2 ATMF ATM Cell Delineator

The ATM Cell Delineator (ATMF) Block integrates circuitry to support HCS-based cell delineation for non-PLCP based transmission formats. The ATMF block accepts a bit serial cell stream from an upstream transmission system sublayer entity and performs cell delineation to locate the cell boundaries. For PLCP applications, ATM cell positions are fixed relative to the PLCP frame, but the ATMF still performs cell delineation to locate the cell boundaries.

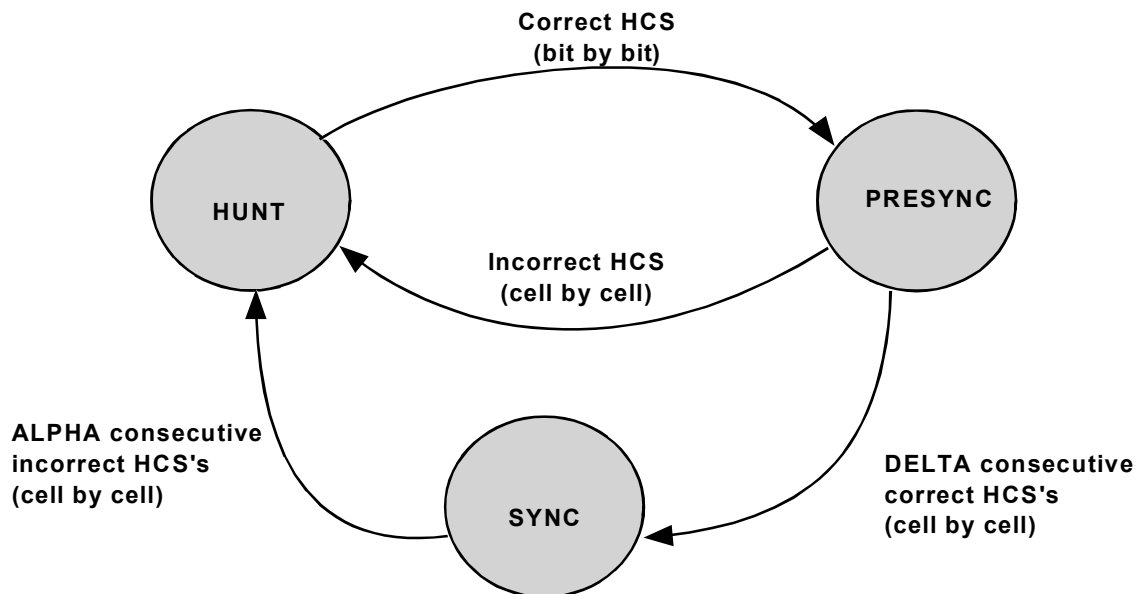
Cell delineation is the process of framing to ATM cell boundaries using the header check sequence (HCS) field found in the ATM cell header. The HCS is a CRC-8 calculation over the first 4 octets of the ATM cell header. When

performing delineation, correct HCS calculations are assumed to indicate cell boundaries.

The ATMF performs a sequential bit-by-bit, a nibble-by-nibble, or a byte-by-byte hunt for a correct HCS sequence. This state is referred to as the HUNT state. When receiving a bit serial cell stream from an upstream transmission system sublayer entity, the bit, nibble, or byte boundaries are determined from the location of the overhead.

When a correct HCS is found, the ATMF locks on the particular cell boundary and assumes the PRESYNC state. This state verifies that the previously detected HCS pattern was not a false indication. If the HCS pattern was a false indication then an incorrect HCS should be received within the next DELTA cells. At that point a transition back to the HUNT state is executed. If an incorrect HCS is not found in this PRESYNC period then a transition to the SYNC state is made. In this state synchronization is not relinquished until ALPHA consecutive incorrect HCS patterns are found. In such an event a transition is made back to the HUNT state. The state diagram of the cell delineation process is shown in Figure 1.

Figure 1 - Cell delineation State Diagram



The values of ALPHA and DELTA determine the robustness of the delineation method. ALPHA determines the robustness against false misalignments due to bit errors. DELTA determines the robustness against false delineation in the synchronization process. ALPHA is chosen to be 7 and DELTA is chosen to be 6

as recommended in ITU-T Recommendation I.432. These values result in a maximum average time to frame of 127 μ s for a DS3 stream carrying ATM cells directly mapped into the DS3 information payload.

Loss of cell delineation (LCD) is detected by counting the number of incorrect cells while in the HUNT state. The counter value is stored in the RXCP-50 LCD Count Threshold register. The threshold has a default value of 360 which results in an E1 application detection time of 77 ms and a DS1 application detection time of 100 ms. If the counter value is set to zero, the LCD output signal is asserted for every incorrect cell.

8.3 RXCP-50 Receive Cell Processor

The Receive Cell Processor (RXCP-50) Block integrates circuitry to support scrambled or unscrambled cell payloads, scrambled or unscrambled cell headers, header check sequence (HCS) verification, idle cell filtering, and performance monitoring.

The RXCP-50 operates upon a delineated cell stream. For PLCP based transmissions systems, cell delineation is performed by the SPLR. For non-PLCP based transmission systems, cell delineation is performed by the ATMF. Framing status indications from these blocks ensure that cells are not written to the RXFF while the SPLR is in the loss of frame state, or cells are not written to the RXFF while the ATMF is in the HUNT or PRESYNC states.

The RXCP-50 descrambles the cell payload field using the self synchronizing descrambler with a polynomial of $x^{43} + 1$. The header portion of the cells can optionally be descrambled also. Note that cell payload scrambling is enabled by default in the S/UNI-CDB as required by ITU-T Recommendation I.432, but may be disabled to ensure backwards compatibility with older equipment.

The HCS is a CRC-8 calculation over the first 4 octets of the ATM cell header. The RXCP-50 verifies the received HCS using the accumulation polynomial, $x^8 + x^2 + x + 1$. The coset polynomial $x^6 + x^4 + x^2 + 1$ is added (modulo 2) to the received HCS octet before comparison with the calculated result as required by the ATM Forum UNI specification, and ITU-T Recommendation I.432.

The RXCP-50 can be programmed to drop all cells containing an HCS error or to filter cells based on the HCS and the cell header. Filtering according to a particular HCS and the GFC, PTI, and CLP bits of the ATM cell header (the VCI and VPI bits must be all logic 0) is programmable through the RXCP-50 registers. More precisely, filtering is performed when filtering is enabled or when HCS errors are found when HCS checking is enabled. Otherwise, all cells are passed on regardless of any error conditions. Cells can be blocked if the HCS pattern is invalid or if the filtering 'Match Pattern' and 'Match Mask' registers are

programmed with a certain blocking pattern. ATM Idle cells are filtered by default. For ATM cells, Null cells (Idle cells) are identified by the standardized header pattern of 'H00, 'H00, 'H00 and 'H01 in the first 4 octets followed by the valid HCS octet.

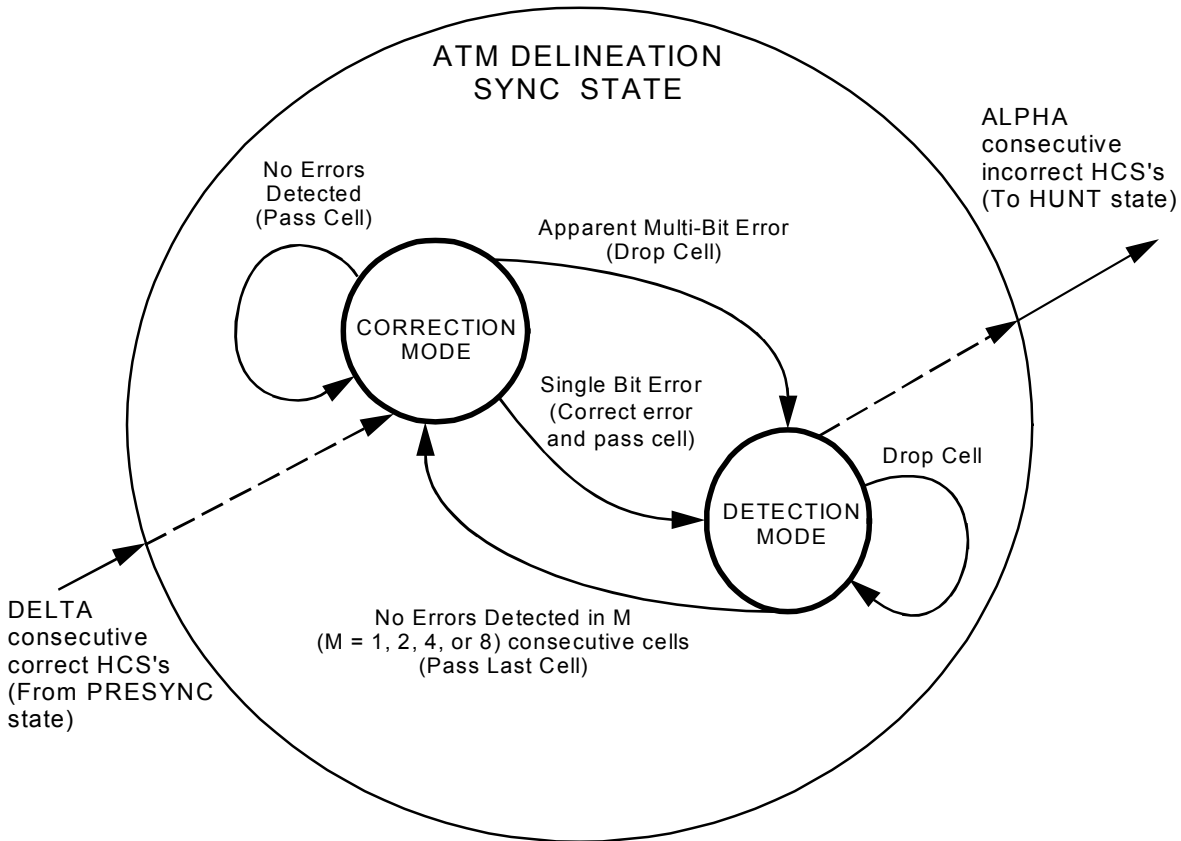
While the cell delineation state machine is in the SYNC state, the HCS verification circuit implements the state machine shown in Figure 2.

In normal operation, the HCS verification state machine remains in the 'Correction' state. Incoming cells containing no HCS errors are passed to the receive FIFO. Incoming single-bit errors are corrected, and the resulting cell is passed to the FIFO. Upon detection of a single-bit error or a multi-bit error, the state machine transitions to the 'Detection' state.

A programmable hysteresis is provided when dropping cells based on HCS errors. When a cell with an HCS error is detected, the RXCP-50 can be programmed to continue to discard cells until m (where $m = 1, 2, 4, 8$) cells are received with a correct HCS. The m^{th} cell is not discarded (see Figure 2). Note that the dropping of cells due to HCS errors only occurs while the ATMF is in the SYNC state.

Cell delineation can optionally be disabled, allowing the RXCP-50 to pass all data bytes it receives.

Figure 2 - HCS Verification State Diagram



8.4 RXFF Receive FIFO

The Receive FIFO (RXFF) provides FIFO management and the S/UNI-CDB receive cell interface. The receive FIFO contains four cells. The FIFO provides the cell rate decoupling function between the transmission system physical layer and the ATM layer.

In general, the management functions include filling the receive FIFO, indicating when the receive FIFO contains cells, maintaining the receive FIFO read and write pointers, and detecting FIFO overrun and underrun conditions.

The FIFO interface is "UTOPIA Level 2" compliant and accepts a read clock (RFCLK) and read enable signal (RENB). The receive FIFO output bus (RDAT[15:0]) is tri-stated when RENB is logic 1 or if the PHY device address (RADR[4:0]) selected does not match this device's address. The interface indicates the start of a cell (RSOC) and the receive cell available status (RCA and DRCA[4:1]) when data is read from the receive FIFO (using the rising edges

of RFCLK). The RCA (and DRCA[x]) status changes from available to unavailable when the FIFO is either empty (RCALEVEL0=1) or near empty (RCALEVEL0 is logic 0). This interface also indicates FIFO overruns via a maskable interrupt and register bits. Read accesses while RCA (or DRCA[x]) is a logic 0 will output invalid data.

8.5 CPPM Cell and PLCP Performance Monitor

The Cell and PLCP Performance Monitor (CPPM) Block interfaces directly to the SPLR to accumulate bit interleaved parity error events, framing octet error events, and far end block error events in saturating counters. When the PLCP framer (SPLR) declares loss of frame, bit interleaved parity error events, framing octet error events, far end block error events, header check sequence error events are not counted.

When an accumulation interval is signaled by a write to the CPPM register address space or to the S/UNI-CDB Identification, Master Reset, and Global Monitor Update register, the CPPM transfers the current counter values into holding registers and resets the counters to begin accumulating error events for the next interval. The counters are reset in such a manner that error events occurring during the reset period are not missed.

8.6 PRGD Pseudo-Random Sequence Generator/Detector

The Pseudo-Random Sequence Generator/Detector (PRGD) block is a software programmable test pattern generator, receiver, and analyzer. Two types of test patterns (pseudo-random and repetitive) conform to ITU-T O.151.

The PRGD can be programmed to generate any pseudo-random pattern with length up to $2^{32}-1$ bits or any user programmable bit pattern from 1 to 32 bits in length. In addition, the PRGD can insert single bit errors or a bit error rate between 10^{-1} to 10^{-7} .

The PRGD can be programmed to check for the presence of the generated pseudo-random pattern. The PRGD can perform an auto-synchronization to the expected pattern, and generate interrupts on detection and loss of the specified pattern. The PRGD can accumulate the total number of bits received and the total number of bit errors in two saturating 32-bit counters. The counters accumulate over an interval defined by writes to the S/UNI-CDB Identification/Master Reset, and Global Monitor Update register (register 006H) or by writes to any PRGD accumulation register. When an accumulation is forced by either method, then the holding registers are updated, and the counters reset to begin accumulating for the next interval. The counters are reset in such a way that no events are missed. The data is then available in the

holding registers until the next accumulation. In addition to the two counters, a record of the 32 bits received immediately prior to the accumulation is available.

The PRGD may also be programmed to check for repetitive sequences. When configured to detect a pattern of length N bits, the PRGD will load N bits from the detected stream, and determine whether the received pattern repeats itself every N subsequent bits. Should it fail to find such a pattern, it will continue loading and checking until it finds a repetitive pattern. All the features (error counting, auto-synchronization, etc.) available for pseudo-random sequences are also available for repetitive sequences. Whenever a PRGD accumulation is forced, the PRGD stores a snapshot of the 32 bits received immediately prior to the accumulation. This snapshot may be examined in order to determine the exact nature of the repetitive pattern received by PRGD.

The pseudo-random or repetitive pattern can be inserted/extracted in the PLCP payload. It cannot be inserted into the ATM cell payload.

8.7 SPLIT SMDS PLCP Layer Transmitter

The SMDS PLCP Layer Transmitter (SPLT) Block integrates circuitry to support DS1 and E1 based PLCP frame insertion.

The SPLT automatically inserts the framing (A1, A2) and path overhead identification (POHID) octets and provides registers or automatic generation of the F1, B1, G1, M2, M1 and C1 octets.

Registers are provided for the path user channel octet (F1) and the path status octet (G1). The bit interleaved parity octet (B1) and the FEBE subfield are automatically inserted.

The DQDB management information octets, M1 and M2 are generated. The type 0 and type 1 patterns described in TA-TSY-000772 are automatically inserted. The type 1 page counter may be reset using a register bit in the SPLT Configuration register.

The PLCP transmit frame C1 cycle/stuff counter octet and the transmit stuffing pattern can be referenced to the REF8KI input pin. Alternately, a fixed stuffing pattern may be inserted into the C1 cycle/stuff counter octet. A looped timing operating mode is provided where the transmit PLCP timing is derived from the received timing. In this mode, the C1 stuffing is generated based on the received stuffing pattern as determined by the SPLR block. When DS1 or E1 PLCP format is enabled, the pattern 00H is inserted.

Stuff Length	C1(Hex)
17	3B
18	4F
19	75
20	9D
21	A7

The REF8KI input is provisioned to loop time the PLCP transmit frame to an externally applied 8 kHz reference.

The Zn, growth octets are set to 00H. The Zn octets may be inserted from an external device via the path overhead stream input, TPOH.

8.8 TXCP-50 Transmit Cell Processor

The Transmit Cell Processor (TXCP-50) Block integrates circuitry to support ATM cell payload scrambling, header check sequence (HCS) generation, and idle/unassigned cell generation.

The TXCP-50 scrambles the cell payload field using the self synchronizing scrambler with polynomial $x^{43} + 1$. The header portion of the cells may optionally also be scrambled. Note that cell payload scrambling may be disabled in the S/UNI-CDB, though it is required by ITU-T Recommendation I.432.

The HCS is generated using the polynomial, $x^8 + x^2 + x + 1$. The coset polynomial $x^6 + x^4 + x^2 + 1$ is added (modulo 2) to the calculated HCS octet as required by the ATM Forum UNI specification, and ITU-T Recommendation I.432. The resultant octet optionally overwrites the HCS octet in the transmit cell. When the transmit FIFO is empty, the TXCP-50 inserts idle/unassigned cells. The idle/unassigned cell header is fully programmable using five internal registers. Similarly, the 48 octet information field is programmed with an 8 bit repeating pattern using an internal register.

8.9 TXFF Transmit FIFO

The Transmit FIFO (TXFF) provides FIFO management and the S/UNI-CDB transmit cell interface. The transmit FIFO contains four cells. The FIFO depth may be programmed to four, three, two, or one cells. The FIFO provides the cell rate decoupling function between the transmission system physical layer and the ATM layer.

In general, the management functions include emptying cells from the transmit FIFO, indicating when the transmit FIFO is full, maintaining the transmit FIFO read and write pointers and detecting a FIFO overrun condition.

The FIFO interface is "UTOPIA Level 2" compliant and accepts a write clock (TFCLK), a write enable signal (TENB), the start of a cell (TSOC) indication, and the parity bit (TPRTY), and the ATM device address (TADR[4:0]) when data is written to the transmit FIFO (using the rising edges of TFCLK). The interface provides the transmit cell available status (TCA and DTCA[4:1]) which can transition from "available" to "unavailable" when the transmit FIFO is near full (when TCALEVEL0 is logic 0) or when the FIFO is full (when TCALEVEL0 is logic 1) and can accept no more writes. To reduce FIFO latency, the FIFO depth at which TCA and DTCA[x] indicates "full" can be set to one, two, three or four cells by the FIFODP[1:0] bits of TXCP-50 Configuration 2 register. If the programmed depth is less than four, more than one cell may be written after TCA or DTCA[x] is asserted as the TXCP-50 still allows four cells to be stored in its FIFO. This interface also indicates FIFO overruns via a maskable interrupt and register bit, but write accesses while TCA or DTCA[x] is logic 0 are not processed. The TXFF automatically transmits idle cells until a full cell is available to be transmitted.

8.10 JTAG Test Access Port

The JTAG Test Access Port block provides JTAG support for boundary scan. The standard JTAG EXTEST, SAMPLE, BYPASS, IDCODE and STCTEST instructions are supported. The S/UNI-CDB identification code is 073390CD hexadecimal.

8.11 Microprocessor Interface

The microprocessor interface block provides normal and test mode registers, and the logic required to connect to the microprocessor interface. The normal mode registers are required for normal operation, and test mode registers are used to enhance the testability of the S/UNI-CDB. The register set is accessed as follows:

Table 1 - Register Memory Map

Address				Register
000H	100H	200H	300H	S/UNI-CDB Configuration 1
001H	101H	201H	301H	S/UNI-CDB Configuration 2
002H	102H	202H	302H	S/UNI-CDB Transmit Configuration

Address				Register
003H	103H	203H	303H	S/UNI-CDB Receive Configuration
005H	105H	205H	305H	S/UNI-CDB Interrupt Status
006H				S/UNI-CDB Identification, Master Reset, and Global Monitor Update
	106H	206H	306H	S/UNI-CDB Reserved
007H	107H	207H	307H	S/UNI-CDB Clock Activity Monitor and Interrupt Identification
008H	108H	208H	308H	SPLR Configuration
009H	109H	209H	309H	SPLR Interrupt Enable
00AH	10AH	20AH	30AH	SPLR Interrupt Status
00BH	10BH	20BH	30BH	SPLR Status
00CH	10CH	20CH	30CH	SPLT Configuration
00DH	10DH	20DH	30DH	SPLT Control
00EH	10EH	20EH	30EH	SPLT Diagnostics and G1 Octet
00FH	10FH	20FH	30FH	SPLT F1 Octet
020H	120H	220H	320H	CPPM Reserved
021H	121H	221H	321H	CPPM Change of CPPM Performance Meter
022H	122H	222H	322H	CPPM BIP Error Count LSB
023H	123H	223H	323H	CPPM BIP Error Count MSB
024H	124H	224H	324H	CPPM PLCP Framing Error Event Count LSB
025H	125H	225H	325H	CPPM PLCP Framing Error Event Count MSB
026H	126H	226H	326H	CPPM PLCP FEBE Count LSB
027H	127H	227H	327H	CPPM PLCP FEBE Count MSB
028H-02FH	128H-12FH	228H-22FH	328H-32FH	CPPM Reserved
060H	160H	260H	360H	RXCP-50 Configuration 1
061H	161H	261H	361H	RXCP-50 Configuration 2

Address				Register
062H	162H	262H	362H	RXCP-50 FIFO/UTOPIA Control & Config
063H	163H	263H	363H	RXCP-50 Interrupt Enables and Counter Status
064H	164H	264H	364H	RXCP-50 Status/Interrupt Status
065H	165H	265H	365H	RXCP-50 LCD Count Threshold (MSB)
066H	166H	266H	366H	RXCP-50 LCD Count Threshold (LSB)
067H	167H	267H	367H	RXCP-50 Idle Cell Header Pattern
068H	168H	268H	368H	RXCP-50 Idle Cell Header Mask
069H	169H	269H	369H	RXCP-50 Corrected HCS Error Count
06AH	16AH	26AH	36AH	RXCP-50 Uncorrected HCS Error Count
06BH	16BH	26BH	36BH	RXCP-50 Received Cell Count LSB
06CH	16CH	26CH	36CH	RXCP-50 Received Cell Count
06DH	16DH	26DH	36DH	RXCP-50 Received Cell Count MSB
06EH	16EH	26EH	36EH	RXCP-50 Idle Cell Count LSB
06FH	16FH	26FH	36FH	RXCP-50 Idle Cell Count
070H	170H	270H	370H	RXCP-50 Idle Cell Count MSB
071H-07FH	171H-17FH	271H-27FH	371H-37FH	RXCP-50 Reserved
080H	180H	280H	380H	TXCP-50 Configuration 1
081H	181H	281H	381H	TXCP-50 Configuration 2
082H	182H	282H	382H	TXCP-50 Transmit Cell Status
083H	183H	283H	383H	TXCP-50 Interrupt Enable/Status
084H	184H	284H	384H	TXCP-50 Idle Cell Header Control
085H	185H	285H	385H	TXCP-50 Idle Cell Payload Control
086H	186H	286H	386H	TXCP-50 Transmit Cell Counter LSB
087H	187H	287H	387H	TXCP-50 Transmit Cell Counter
088H	188H	288H	388H	TXCP-50 Transmit Cell Counter MSB
089H-08FH	189H-18FH	289H-28FH	389H-38FH	TXCP-50 Reserved

Address				Register
09BH	19BH	29BH	39BH	S/UNI-CDB Misc.
0A0H	1A0H	2A0H	3A0H	PRGD Control
0A1H	1A1H	2A1H	3A1H	PRGD Interrupt Enable/Status
0A2H	1A2H	2A2H	3A2H	PRGD Length
0A3H	1A3H	2A3H	3A3H	PRGD Tap
0A4H	1A4H	2A4H	3A4H	PRGD Error Insertion
0A5H- 0A7H	1A5H- 1A7H	2A5H- 2A7H	3A5H- 3A7H	PRGD Reserved
0A8H	1A8H	2A8H	3A8H	PRGD Pattern Insertion Register #1
0A9H	1A9H	2A9H	3A9H	PRGD Pattern Insertion Register #2
0AAH	1AAH	2AAH	3AAH	PRGD Pattern Insertion Register #3
0ABH	1ABH	2ABH	3ABH	PRGD Pattern Insertion Register #4
0ACH	1ACH	2ACH	3ACH	PRGD Pattern Detector Register #1
0ADH	1ADH	2ADH	3ADH	PRGD Pattern Detector Register #2
0AEH	1AEH	2AEH	3AEH	PRGD Pattern Detector Register #3
0AFH	1AFH	2AFH	3AFH	PRGD Pattern Detector Register #4
0B0H- 0FFH	1B0H- 1FFH	2B0H- 2FFH	3B0H- 3FFH	S/UNI-CDB Reserved
400H				S/UNI-CDB Master Test Register
401H - 7FFH				Reserved for S/UNI-CDB Test

For all register accesses, CSB must be low.

9 NORMAL MODE REGISTER DESCRIPTION

Normal mode registers are used to configure and monitor the operation of the S/UNI-CDB. Normal mode registers (as opposed to test mode registers) are selected when A[10] is low.

Notes on Normal Mode Register Bits:

1. Writing values into unused register bits has no effect. However, to ensure software compatibility with future, feature-enhanced versions of the product, unused register bits must be written with logic zero. Reading back unused bits can produce either a logic one or a logic zero; hence, unused register bits should be masked off by software when read.
2. All configuration bits that can be written into can also be read back. This allows the processor controlling the S/UNI-CDB to determine the programming state of the block.
3. Writable normal mode register bits are cleared to logic zero upon reset unless otherwise noted.
4. Writing into read-only normal mode register bit locations does not affect S/UNI-CDB operation unless otherwise noted.
5. Certain register bits are reserved. These bits are associated with megacell functions that are unused in this application. To ensure that the S/UNI-CDB operates as intended, reserved register bits must only be written with the suggested logic levels. Similarly, writing to reserved registers should be avoided.
6. The S/UNI-CDB requires a software initialization sequence in order to guarantee proper device operation and long term reliability. Please refer to Section 10.1 of this document for the details on how to program this sequence.
7. All **reserved bits must be programmed** in order for device to function properly.

Register 000H, 100H, 200H, 300H: S/UNI-CDB Configuration 1

Bit	Type	Function	Default
Bit 7	R/W	8KREFO	1
Bit 6	R/W	DS27_53	1
Bit 5	R/W	TOCTA	0
Bit 4	R/W	Reserved4	0
Bit 3	R/W	LOOPT	0
Bit 2	R/W	LLOOP	0
Bit 1	R/W	DLOOP	0
Bit 0	R/W	Reserved0	0

Reserved0:

This reserved bit must be programmed to logic 0 for proper operation.

DLOOP:

The DLOOP bit controls the diagnostic loopback. When a logic 0 is written to DLOOP, diagnostic loopback is disabled. When a logic 1 is written to DLOOP, the transmit data stream is looped in the receive direction. The DLOOP should not be set to a logic 1 when either the LLOOP or LOOPT bit is a logic 1.

LLOOP:

The LLOOP bit controls the line loopback. When a logic 0 is written to LLOOP, line loopback is disabled. When a logic 1 is written to LLOOP, the stream received on RDATI and ROHM is looped to the TDATO and TOHM outputs. Note that the TDATO, TOHM, and TCLK outputs are referenced to RCLK when LLOOP is logic 1.

LOOPT:

The LOOPT bit selects the transmit timing source. When a logic 1 is written to LOOPT, the transmitter is loop-timed to the receiver. When loop timing is enabled, the receive clock (RCLK) is used as the transmit timing source. When a logic 0 is written to LOOPT, the transmit clock (TICK) is used as the transmit timing source. The nibble stuffing is derived from the REF8KI input, or is fixed internally. Setting the LOOPT bit disables the effect of the TICK and TXREF bits in the S/UNI-CDB Transmit Configuration and S/UNI-CDB

Configuration 2 registers (Reference: S/UNI-QJET Datasheet: PMC-960835) respectively, thereby forcing flow-through timing.

Reserved4:

This reserved bit must be programmed to logic 0 for proper operation.

TOCTA:

The TOCTA bit enables octet-alignment or nibble-alignment of the transmit cell stream to the transmission overhead when the arbitrary transmission format is chosen (TFRM[1:0] = 11 binary and SPLT Configuration register bit EXT = 1). When the arbitrary transmission format is chosen and TOCTA is set to logic 1, the ATM cell nibbles or octets are aligned to the arbitrary transmission format overhead boundaries (as set by the TIOHM input). Nibble alignment is chosen if the FORM[1:0] bits in the SPLT Configuration are set to 00. Byte alignment is chosen if these FORM[1:0] bits are set to any other value. The number of TICLK periods between transmission format overhead bit positions must be divisible by 4 (for nibble alignment) or 8 (for byte alignment). When TOCTA is set to logic 0, no octet alignment is performed, and there is no restriction on the number of TICLK periods between transmission format overhead bit positions.

DS27_53:

The DS27_53 bit is used to select between the long data structure (27 words in 16-bit mode and 53 bytes in 8-bit mode) and the short data structure (26 words in 16-bit mode and 52 bytes in 8-bit mode) on the ATM interface. When DS27_53 is set to logic one, the RXCP-50 and TXCP-50 blocks are configured to operate with the long data structure; when DS27_53 is set to logic zero, the RXCP-50 and TXCP-50 are configured to operate with the short data structure.

8KREFO:

The 8KREFO bit is used, in conjunction with the PLCPEN bit in the SPLR Configuration Register to select the function of the REF8KO/RPOHFP/RFPO/RMFPO[x] output pin. When PLCPEN is logic 1, the RPOHFP function will be selected and 8KREFO has no effect (note that RPOHFP is inherently an 8kHz reference). If PLCPEN is logic 0, then if 8KREFO is logic 1, then an 8kHz reference will be derived from the RCLK[x] signal and output on REF8KO. If 8KREFO and PLCPEN are both logic 0, then the RXMFPO register bit in the S/UNI-CDB Configuration 2 register (Reference: S/UNI-QJET Datasheet: PMC-960835) will select either the RFPO or RMFPO function.

Register 001H, 101H, 201H, 301H: S/UNI-CDB Configuration

Bit	Type	Function	Default
Bit 7	R/W	STATSEL[2]	0
Bit 6	R/W	STATSEL[1]	0
Bit 5	R/W	STATSEL[0]	0
Bit 4	R/W	Reserved4	0
Bit 3	R/W	Reserved3	0
Bit 2	R/W	Reserved2	0
Bit 1	R/W	Reserved1	0
Bit 0	R/W	Reserved0	0

Reserved0:

This reserved bit must be programmed to logic 0 for proper operation.

Reserved1:

This reserved bit must be programmed to logic 0 for proper operation.

Reserved2:

This reserved bit must be programmed to logic 0 for proper operation.

Reserved3:

This reserved bit must be programmed to logic 0 for proper operation.

Reserved4:

This reserved bit must be programmed to logic 0 for proper operation.

STATSEL[2:0]:

The STATSEL[2:0] bits are used to select the function of the FRMSTAT[4:1] output. The selection is shown in the following table:

Table 2 - STATSEL[2:0] Options

STATSEL[2:0]	FRMSTAT output pin indication function
000	Reserved
001	PLCP Loss of Frame

STATSEL[2:0]	FRMSTAT output pin indication function
010	Reserved
011	PLCP Out of Frame
100	Reserved
101	Reserved
110	Reserved
111	Reserved

Register 002H, 102H, 202H, 302H: S/UNI-CDB Transmit Configuration

Bit	Type	Function	Default
Bit 7	R/W	TXSETBIT[1]	0
Bit 6	R/W	TXSETBIT[0]	0
Bit 5	R/W	TXREF	0
Bit 4	R/W	TICLK	0
Bit 3	R/W	Reserved3	0
Bit 2	R/W	TCLKINV	0
Bit 1	R/W	TPOSINV	0
Bit 0	R/W	TNEGINV	0

TNEGINV:

The TNEGINV bit provides polarity control for outputs TOHM. When a logic 0 is written to TNEGINV, the TOHM output is not inverted. When a logic 1 is written to TNEGINV, the TOHM output is inverted. The TNEGINV bit setting does not affect the loopback data in diagnostic loopback.

TPOSINV:

The TPOSINV bit provides polarity control for output TDATO. When a logic 0 is written to TPOSINV, the TDATO output is not inverted. When a logic 1 is written to TPOSINV, the TDATO output is inverted. The TPOSINV bit setting does not affect the loopback data in diagnostic loopback.

TCLKINV:

The TCLKINV bit provides polarity control for output TCLK. When a logic 0 is written to TCLKINV, TCLK is not inverted and outputs TDATO and TOHM are updated on the falling edge of TCLK. When a logic 1 is written to TCLKINV, TCLK is inverted and outputs TDATO and TOHM are updated on the rising edge of TCLK.

Reserved3:

This reserved bit must be programmed to logic 0 for proper operation.

TICLK:

The TICLK bit selects the transmit clock used to update the TDATO and TOHM outputs. When a logic 0 is written to TICLK, the buffered version of the input transmit clock, TCLK, is used to update TDATO and TOHM on the

edge selected by the TCLKINV bit. When a logic 1 is written to TICLK, TDATO and TOHM are updated on the rising edge of TICLK, eliminating the flow-through TCLK signal. The TICLK bit has no effect if the LOOPT or LLOOP bit is a logic 1.

TXREF:

The TXREF register bit determines if TICLK[1] and TIOHM[1] should be used as the reference transmit clock and overhead pulse, respectively, instead of TICLK[X] and TIOHM[X]. If TXREF is set to a logic 1, then TICLK[1] and TIOHM[1] will be used as the reference transmit clock and overhead/frame pulse, respectively. If TXREF is set to a logic 0, then TICLK[X] and TIOHM[X] will be used as the reference transmit clock and overhead/frame pulse, respectively, for quadrant X. If loop-timing is enabled (LOPT = 1), the TXREF bit has no effect on the corresponding quadrant. Note that when TXREF is set to logic 1, the unused TICLK[x] and TIOHM[x] should be tied to power or ground, not left floating.

TXSETBIT[1:0]:

These bits must be programmed to logic 1 for proper operation.

Register 003H, 103H, 203H, 303H: S/UNI-CDB Receive Configuration

Bit	Type	Function	Default
Bit 7	R/W	RXSETBIT[1]	0
Bit 6	R/W	RXSETBIT[0]	0
Bit 5	R/W	Reserved5	0
Bit 4	R/W	Reserved4	0
Bit 3	R/W	Reserved3	0
Bit 2	R/W	RCLKINV	0
Bit 1	R/W	RPOSINV	0
Bit 0	R/W	RNEGINV	0

RNEGINV:

The RNEGINV bit provides polarity control for input ROHM. When a logic 0 is written to RNEGINV, the input ROHM is not inverted. When a logic 1 is written to RNEGINV, the input ROHM is inverted. The RNEGINV bit setting does not affect the loopback data in diagnostic loopback.

RPOSINV:

The RPOSINV bit provides polarity control for input RDATI. When a logic 0 is written to RPOSINV, the input RDATI is not inverted. When a logic 1 is written to RPOSINV, the input RDATI is inverted. The RPOSINV bit setting does not affect the loopback data in diagnostic loopback.

RCLKINV:

The RCLKINV bit provides polarity control for input RCLK. When a logic 0 is written to RCLKINV, RCLK is not inverted and inputs RDATI and ROHM are sampled on the rising edge of RCLK. When a logic 1 is written to RCLKINV, RCLK is inverted and inputs RDATI and ROHM are sampled on the falling edge of RCLK.

Reserved3:

This reserved bit must be programmed to logic 0 for proper operation

Reserved4:

This reserved bit must be programmed to logic 1 for proper operation

Reserved5:

This reserved bit must be programmed to logic 1 for proper operation.

RXSETBIT[1:0]:

These bits must be programmed to logic 1 for proper operation.

Register 008H, 108H, 208H, 308H: SPLR Configuration

Bit	Type	Function	Default
Bit 7	R/W	FORM[1]	0
Bit 6	R/W	FORM[0]	0
Bit 5	R/W	Reserved5	0
Bit 4	R/W	Reserved4	0
Bit 3	R/W	REFRAME	0
Bit 2	R/W	PLCPEN	0
Bit 1		Unused	X
Bit 0	R/W	EXT	0

EXT:

The EXT bit disables the internal transmission system sublayer timeslot counter from identifying DS1 and E1 overhead bits. The EXT bit allows transmission formats that are unsupported by the internal timeslot counter to be supported using the ROHM[x] input. When a logic 0 is written to EXT, input transmission system overhead (for DS1 and E1 formats) is indicated using the internal timeslot counter. This counter is synchronized to the transmission system frame alignment using the ROHM[x] (for DS1 or E1 ATM direct-mapped formats).

When a logic 1 is written to EXT, indications on ROHM[x] identify each transmission system overhead bit.

PLCPEN:

The PLCPEN bit enables PLCP framing. When a logic 1 is written to PLCPEN, PLCP framing is enabled. The PLCP format is specified by the FORM[1:0] bits in this register. When a logic 0 is written to PLCPEN, PLCP related functions in the SPLR block are disabled. PLCPEN must be programmed to logic 0 for arbitrary framing formats.

REFRAME:

The REFRAME bit is used to trigger reframing. When a logic 1 is written to REFRAME, the S/UNI-CDB is forced out of PLCP frame and a new search for frame alignment is initiated. Note that only a logic 0 to logic 1 transition of the REFRAME bit triggers reframing; multiple write operations are required to ensure such a transition.

Reserved4:

This reserved bit must be programmed to logic 0 for proper operation

Reserved5:

This reserved bit must be programmed to logic 0 for proper operation.

FORM[1:0]:

The FORM[1:0] bits select the PLCP frame format as shown below. These bits must be set to "11" if E1 direct mapped mode is being used (PLCPEN=0 and EXT=1).

Table 3 - SPLR FORM[1:0] Configurations

FORM[1]	FORM[0]	PLCP Framing Format
1	0	DS1
1	1	E1

Register 00CH, 10CH, 20CH, 30CH: SPLT Configuration

Bit	Type	Function	Default
Bit 7	R/W	FORM[1]	0
Bit 6	R/W	FORM[0]	0
Bit 5	R/W	M1TYPE	0
Bit 4	R/W	M2TYPE	0
Bit 3	R/W	Reserved3	0
Bit 2	R/W	PLCPEN	0
Bit 1		Unused	X
Bit 0	R/W	EXT	0

EXT:

The EXT bit disables the internal transmission system sublayer timeslot counter from identifying DS1 or E1 overhead bits. The EXT bit allows transmission formats that are unsupported by the internal timeslot counter and must be supported using the TIOHM[x] input. When a logic 0 is written to EXT, input transmission system overhead (for DS1 and E1 formats) is indicated using the internal timeslot counter. This counter flywheels to create the appropriate transmission system alignment. This alignment is indicated on the TOHM[x] output. When a logic 1 is written to EXT, indications on TIOHM[x] identify each transmission system overhead bit. These indications flow through the S/UNI-CDB and appear on the TOHM[x] output where they mark the transmission system overhead placeholder positions in the TDATO[x] stream. EXT should only be set to logic 1 if the TFRM[1:0] bits in the S/UNI-CDB Transmit Configuration register are both set to logic 1 and the arbitrary framing format is desired.

PLCPEN:

The PLCPEN bit enables PLCP frame insertion. When a logic 1 is written to PLCPEN, DS1, or E1 PLCP framing is inserted. The PLCP format is specified by the FORM[1:0] bits in this register. When a logic 0 is written to PLCPEN, PLCP related functions in the SPLT block are disabled. The PLCPEN bit must be set to logic 0 for arbitrary framing formats.

Reserved3:

This reserved bit must be programmed to logic 0 for proper operation.

M2TYPE:

The M2TYPE bit selects the type of code transmitted in the M2 octet. These codes are required in systems implementing the IEEE-802.6 DQDB protocol. When a logic 0 is written to M2TYPE, the fixed pattern type 0 code is transmitted in the M2 octet. When a logic 1 is written to M2TYPE, the 1023 cyclic code pattern (starting with B6 hexadecimal and ending with 8D hexadecimal) is transmitted in the M2 octet. Please refer to TA-TSY-000772, Issue 3 and Supplement 1, for details on the codes.

M1TYPE:

The M1TYPE bit selects the type of code transmitted in the M1 octet. These codes are required in systems implementing the IEEE-802.6 DQDB protocol. When a logic 0 is written to M1TYPE, the fixed pattern type 0 code is transmitted in the M1 octet. When a logic 1 is written to M1TYPE, the 1023 cyclic code pattern (starting with B6 hexadecimal and ending with 8D hexadecimal) is transmitted in the M1 octet. Please refer to TA-TSY-000772, Issue 3 and Supplement 1, for details on the codes.

FORM[1:0]:

When EXT = 0 and PLCPEN = 0, the FORM[1:0] bits and the TFRM[1:0] bits in the S/UNI-CDB Transmit Configuration register select the ATM direct-mapped transmission frame format as shown below. When EXT = 0 and PLCPEN = 1, the FORM[1:0] bits along with the TFRM[1:0] bits select the transmission and PLCP frame format as shown below. When EXT = 1 and TOCTA = 1, then the FORM[1:0] bits control the cell alignment with respect to the transmission overhead given on TIOHM[x] as shown below. The FORM bits have no effect if EXT = 1 and TOCTA = 0.

Table 4 - SPLT FORM[1:0] Configurations

FORM[1]	FORM[0]	PLCP or ATM direct-mapped Framing Format / Cell alignment
1	0	DS1 / byte
1	1	E1 / byte

10 OPERATION

10.1 Software Initialization Sequence

The S/UNI-CDB can come out of reset in a mode that consumes excess power. The device functionality is not altered except for excessive power consumption resulting excess heat dissipation which could lead to long term reliability problems.

The software initialization sequence in this section will put the S/UNI-CDB into a normal power consumption state should the device come out of reset in the excess power state. This reset sequence must be used to guarantee long term reliability of the device.

1. Reset the S/UNI-CDB.
2. Set IOTST (bit 2) in the Master Test Register to '1' (by writing 00000100 to register 400H).
3. Put the S/UNI-CDBReceive Cell Processor (RXCP) into test mode by writing:
 - 00000101 to test register 461H
 - 00000101 to test register 561H
 - 00000101 to test register 661H
 - 00000101 to test register 761H
4. Set S/UNI-CDB Receive Cell Processor block built in set test (BIST) controls signals by writing:
 - 01000000 to test register 462H
 - 01000000 to test register 562H
 - 01000000 to test register 662H
 - 01000000 to test register 762H
 - 10101010 to test register 463H
 - 10101010 to test register 563H
 - 10101010 to test register 663H

- 10101010 to test register 763H
5. Put the S/UNI-CDB Transmit Cell Processor (TXCP) into test mode by writing:
- 00000011 to test register 481H
- 00000011 to test register 581H
- 00000011 to test register 681H
- 00000011 to test register 781H
6. Set S/UNI-CDB Transmit Cell Processor block built in set test (BIST) controls signals by writing:
- 10000000 to test register 480H
- 10000000 to test register 580H
- 10000000 to test register 680H
- 10000000 to test register 780H
- 10101010 to test register 482H
- 10101010 to test register 582H
- 10101010 to test register 682H
- 10101010 to test register 782H
7. Toggle REF8KI (pin T3) signal at least eight times (this provides the clock to the RAM). REF8KI is the test clock used by the TXCP and RXCP blocks when in test mode.
8. Set IOTST (bit 2) in the Master Test register to '0' (by writing 00000000 to register 400H).
9. Resume normal device programming.

11 TEST FEATURES DESCRIPTION

Simultaneously asserting (low) the CSB, RDB and WRB inputs causes all digital output pins and the data bus to be held in a high-impedance state. This test feature may be used for board testing.

Test mode registers are used to apply test vectors during production testing of the S/UNI-CDB. Test mode registers (as opposed to normal mode registers) are selected when A[10] is high.

Test mode registers may also be used for board testing. When all of the TSBs within the S/UNI-CDB are placed in test mode 0, device inputs may be read and device outputs may be forced via the microprocessor interface (refer to the section "Test Mode 0" for details).

In addition, the S/UNI-CDB also supports a standard IEEE 1149.1 five-signal JTAG boundary scan test port for use in board testing. All digital device inputs may be read and all digital device outputs may be forced via the JTAG test port.

Table 5 - Test Mode Register Memory Map

Address				Register
000H-3FFH				Normal Mode Registers
400H				Master Test Register
408H	508H	608H	708H	SPLR Test Register 0
409H	509H	609H	709H	SPLR Test Register 1
40AH	50AH	60AH	70AH	SPLR Test Register 2
40CH	50CH	60CH	70CH	SPLT Test Register 0
40DH	50DH	60DH	70DH	SPLT Test Register 1
40EH	50EH	60EH	70EH	SPLT Test Register 2
40FH	50FH	60FH	70FH	SPLT Test Register 3
460H	560H	660H	760H	RXCP-50 Test Register 0
461H	561H	661H	761H	RXCP-50 Test Register 1
462H	562H	662H	762H	RXCP-50 Test Register 2
463H	563H	663H	763H	RXCP-50 Test Register 3
464H	564H	664H	764H	RXCP-50 Test Register 4

Address				Register
465H	565H	665H	765H	RXCP-50 Test Register 5
480H	580H	680H	780H	TXCP-50 Test Register 0
481H	581H	681H	781H	TXCP-50 Test Register 1
482H	582H	682H	782H	TXCP-50 Test Register 2
483H	583H	683H	783H	TXCP-50 Test Register 3
484H	584H	684H	784H	TXCP-50 Test Register 4
485H	585H	685H	785H	TXCP-50 Test Register 5
4A0H	5A0H	6A0H	7A0H	PRGD Test Register 0
4A1H	5A1H	6A1H	7A1H	PRGD Test Register 1
4A2H	5A2H	6A2H	7A2H	PRGD Test Register 2
4A3H	5A3H	6A3H	7A3H	PRGD Test Register 3

Notes on Test Mode Register Bits:

1. Writing values into unused register bits has no effect. However, to ensure software compatibility with future, feature-enhanced versions of the product, unused register bits must be written with logic zero. Reading back unused bits can produce either a logic one or a logic zero; hence, unused register bits should be masked off by software when read.
2. Writable test mode register bits are not initialized upon reset unless otherwise noted.

Register 100H: S/UNI-CDB Master Test

Register 400H: S/UNI-CDB Master Test

Bit	Type	Function	Default
Bit 7		Unused	X
Bit 6	W	A_TM[9]	X
Bit 5	W	A_TM[8]	X
Bit 4	W	PMCTST	X
Bit 3	W	DBCTRL	0
Bit 2	R/W	IOTST	0
Bit 1	W	HIZDATA	0
Bit 0	R/W	HIZIO	0

This register is used to enable S/UNI-CDB test features. All bits, except PMCTST and A_TM[9:8], are reset to zero by a hardware reset of the S/UNI-CDB. The S/UNI-CDB Master Test register is not affected by a software reset (via the S/UNI-CDB Identification, Master Reset, and Global Monitor Update register (006H)).

HIZIO, HIZDATA:

The HIZIO and HIZDATA bits control the tri-state modes of the S/UNI-CDB. While the HIZIO bit is a logic one, all output pins of the S/UNI-CDB except the data bus and output TDO are held tri-state. The microprocessor interface is still active. While the HIZDATA bit is a logic one, the data bus is also held in a high-impedance state which inhibits microprocessor read cycles. The HIZDATA bit is overridden by the DBCTRL bit.

IOTST:

The IOTST bit is used to allow normal microprocessor access to the test registers and control the test mode in each TSB block in the S/UNI-CDB for board level testing. When IOTST is a logic one, all blocks are held in test mode and the microprocessor may write to a block's test mode 0 registers to manipulate the outputs of the block and consequentially the device outputs (refer to the "Test Mode 0 Details" in the "Test Features" section).

DBCTRL:

The DBCTRL bit is used to pass control of the data bus drivers to the CSB pin. When the DBCTRL bit is set to logic one and either IOTST or PMCTST are logic one, the CSB pin controls the output enable for the data bus. While the DBCTRL bit is set, holding the CSB pin high causes the S/UNI-CDB to

drive the data bus and holding the CSB pin low tri-states the data bus. The DBCTRL bit overrides the HIZDATA bit. The DBCTRL bit is used to measure the drive capability of the data bus driver pads.

PMCTST:

The PMCTST bit is used to configure the S/UNI-CDB for PMC's manufacturing tests. When PMCTST is set to logic one, the S/UNI-CDB microprocessor port becomes the test access port used to run the PMC manufacturing test vectors. The PMCTST bit is logically "ORed" with the IOTST bit, and can be cleared by setting CSB to logic one or by writing logic zero to the bit.

A_TM[9:8]:

The state of the A_TM[9:8] bits internally replace the input address lines A[9:8] respectively when PMCTST is set to logic 1. This allows for more efficient use of the PMC manufacturing test vectors.

11.1 JTAG Test Port

The S/UNI-CDB JTAG Test Access Port (TAP) allows access to the TAP controller and the 4 TAP registers: instruction, bypass, device identification and boundary scan. Using the TAP, device input logic levels can be read, device outputs can be forced, the device can be identified and the device scan path can be bypassed. For more details on the JTAG port, please refer to the Operations section.

Table 6 - Instruction Register

Length - 3 bits

Instructions	Selected Register	Instruction Codes, IR[2:0]
EXTEST	Boundary Scan	000
IDCODE	Identification	001
SAMPLE	Boundary Scan	010
BYPASS	Bypass	011
BYPASS	Bypass	100
STCTEST	Boundary Scan	101
BYPASS	Bypass	110
BYPASS	Bypass	111

Identification Register

Length - 32 bits

Version number - 2H

Part Number - 7346H

Manufacturer's identification code - 0CDH

Device identification - 273460CDH

Table 7 - Boundary Scan Register

Length - 198 bits

Pin/Enable	Register Bit	Cell Type	ID Bit	Pin/Enable	Register Bit	Cell Type	ID Bit
TDAT[15] ¹	0	IN_CELL	0	RX_OEB ⁴	66	OUT_CELL	(0)
TDAT[14]	1	IN_CELL	0	TICLK[4:1]	67:70	IN_CELL	(0)
TDAT[13]	2	IN_CELL	1	TIOHM[4:1]	71:74	IN_CELL	(0)
TDAT[12]	3	IN_CELL	0	TPOH[4:1]	75:78	IN_CELL	(0)
TDAT[11]	4	IN_CELL	0	TPOHINS[4:1]	79:82	IN_CELL	(0)
TDAT[10]	5	IN_CELL	1	TPOHCLK[4:1]	83:86	OUT_CELL	(0)
TDAT[9]	6	IN_CELL	1	TPOHFP[4:1]	87:90	OUT_CELL	(0)
TDAT[8]	7	IN_CELL	1	LCD[4:1]	91:94	OUT_CELL	(0)
TDAT[7]	8	IN_CELL	0	RPOH[4:1]	95:98	OUT_CELL	(0)
TDAT[6]	9	IN_CELL	0	RPOHCLK[4:1]	99:102	OUT_CELL	(0)
TDAT[5]	10	IN_CELL	1	RPOHFP[4:1]	103:106	OUT_CELL	(0)
TDAT[4]	11	IN_CELL	1	FRMSTAT[4:1]	107:110	OUT_CELL	(0)
TDAT[3]	12	IN_CELL	0	REF8KI	111	IN_CELL	(0)
TDAT[2]	13	IN_CELL	1	N/C	112:115		(0)
TDAT[1]	14	IN_CELL	0	N/C	116:119		(0)
TDAT[0]	15	IN_CELL	0	N/C	120:123		(0)
TFCLK	16	IN_CELL	0	N/C	124:127		(0)
TADR[4]	17	IN_CELL	1	N/C	128:131		(0)
TADR[3]	18	IN_CELL	1	VSS	132:135		(0)
TADR[2]	19	IN_CELL	0	VSS	136:139		(0)
TADR[1]	20	IN_CELL	0	RCLK[4:1]	140:143	IN_CELL	(0)
TADR[0]	21	IN_CELL	0	ROHM[4:1]	144:147	IN_CELL	(0)
TPRTY	22	IN_CELL	0	RDATI[4:1]	148:151	IN_CELL	(0)
TSOC	23	IN_CELL	0	TCLK[4:1]	152:155	OUT_CELL	(0)
TENB	24	IN_CELL	1	TOHM[4:1]	156:159	OUT_CELL	(0)
TCA	25	OUT_CELL	1	TDATO4:1]	160:163	OUT_CELL	(0)
TCA_OEB ²	26	OUT_CELL	0	INTB	164	OUT_CELL	(0)
DTCA[4]	27	OUT_CELL	0	RSTB	165	IN_CELL	(0)
DTCA[3]	28	OUT_CELL	1	WRB	166	IN_CELL	(0)
DTCA[2]	29	OUT_CELL	1	RDB	167	IN_CELL	(0)
DTCA[1]	30	OUT_CELL	0	ALE	168	IN_CELL	(0)
PHY_ADR[2]	31	IN_CELL	1	CSB	169	IN_CELL	(0)
PHY_ADR[1]	32	IN_CELL	(1)	A[10:0]	170:180	IN_CELL	(0)
PHY_ADR[0]	33	IN_CELL	(1)	D[7]	181	IO_CELL	(0)
ATM8	34	IN_CELL	(0)	DOENB [7] ⁵	182	OUT_CELL	(0)
DRCA[4]	35	OUT_CELL	(0)	D[6]	183	IO_CELL	(0)
DRCA[3]	36	OUT_CELL	(0)	DOENB[6] ⁵	184	OUT_CELL	(0)
DRCA[2]	37	OUT_CELL	(0)	D[5]	185	IO_CELL	(0)
DRCA[1]	38	OUT_CELL	(0)	DOENB [5] ⁵	186	OUT_CELL	(0)

RCA	39	OUT_CELL	(0)	D[4]	187	IO_CELL	(0)
RCA_OEB ³	40	OUT_CELL	(0)	DOENB [4] ⁵	188	OUT_CELL	(0)
RSOC	41	OUT_CELL	(0)	D[3]	189	IO_CELL	(0)
RENB	42	IN_CELL	(0)	DOENB [3] ⁵	190	OUT_CELL	(0)
RFCLK	43	IN_CELL	(0)	D[2]	191	IO_CELL	(0)
RADR[4]	44	IN_CELL	(0)	DOENB [2] ⁵	192	OUT_CELL	(0)
RADR[3]	45	IN_CELL	(0)	D[1]	193	IO_CELL	(0)
RADR[2]	46	IN_CELL	(0)	DOENB [1] ⁵	194	OUT_CELL	(0)
RADR[1]	47	IN_CELL	(0)	D[0]	195	IO_CELL	(0)
RADR[0]	48	IN_CELL	(0)	DOENB [0] ⁵	196	OUT_CELL	(0)
RPRTY	49	OUT_CELL	(0)	HIZ ⁶	197	OUT_CELL	(0)
RDAT[15:0]	50:65	OUT_CELL	(0)				

NOTES:

1. TDAT[15] is the first bit of the boundary scan chain.
2. TCA_OEB will set TCA to tri-state when set to logic 1. When set to logic 0, TCA will be driven.
3. RCA_OEB will set RCA to tri-state when set to logic 1. When set to logic 0, RCA will be driven.
4. RX_OEB will set RDAT[15:0], RPRTY, and RSOC to tri-state when set to logic 1. When set to logic 0, RDAT[15:0], RPRTY, and RSOC will be driven.
5. The DOENB signals will set the corresponding bidirectional signal (the one preceding the DOENB in the boundary scan chain — see note 1 also) to an output when set to logic 0. When set to logic 1, the bidirectional signal will be tri-stated.
6. HIZ will set all outputs not controlled by TCA_OEB, RCA_OEB, RX_OEB, and DOENB to tri-state when set to logic 1. When set to logic 0, those outputs will be driven.

Boundary Scan Cell Description

In the following diagrams, CLOCK-DR is equal to TCK when the current controller state is SHIFT-DR or CAPTURE-DR, and unchanging otherwise. The multiplexer in the center of the diagram selects one of four inputs, depending on the status of select lines G1 and G2. The ID Code bit is as listed in the Boundary Scan Register table located in the TEST FEATURES DESCRIPTION - JTAG Test Port section.

Figure 3 - Input Observation Cell (IN_CELL)

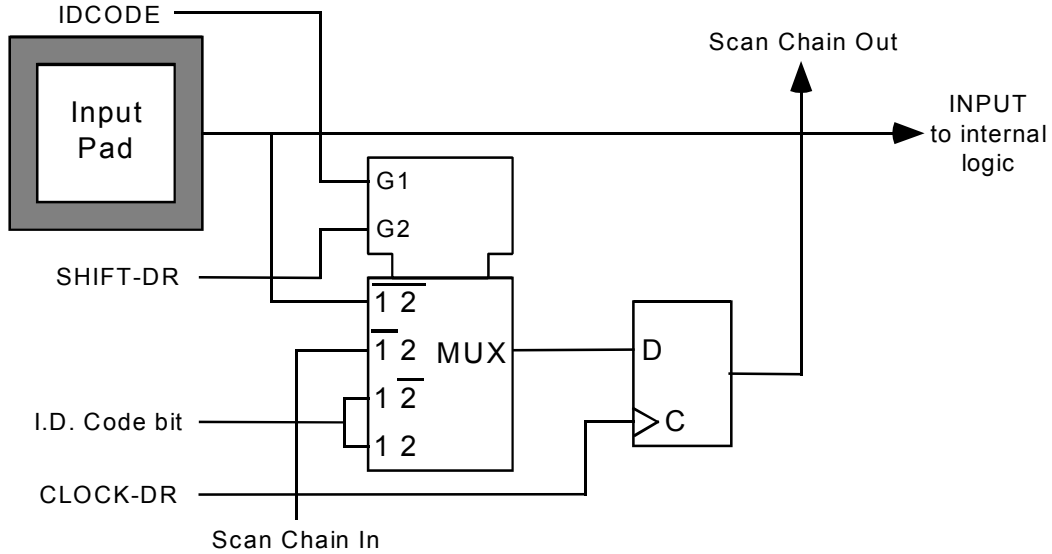


Figure 4 - Output Cell (OUT_CELL)

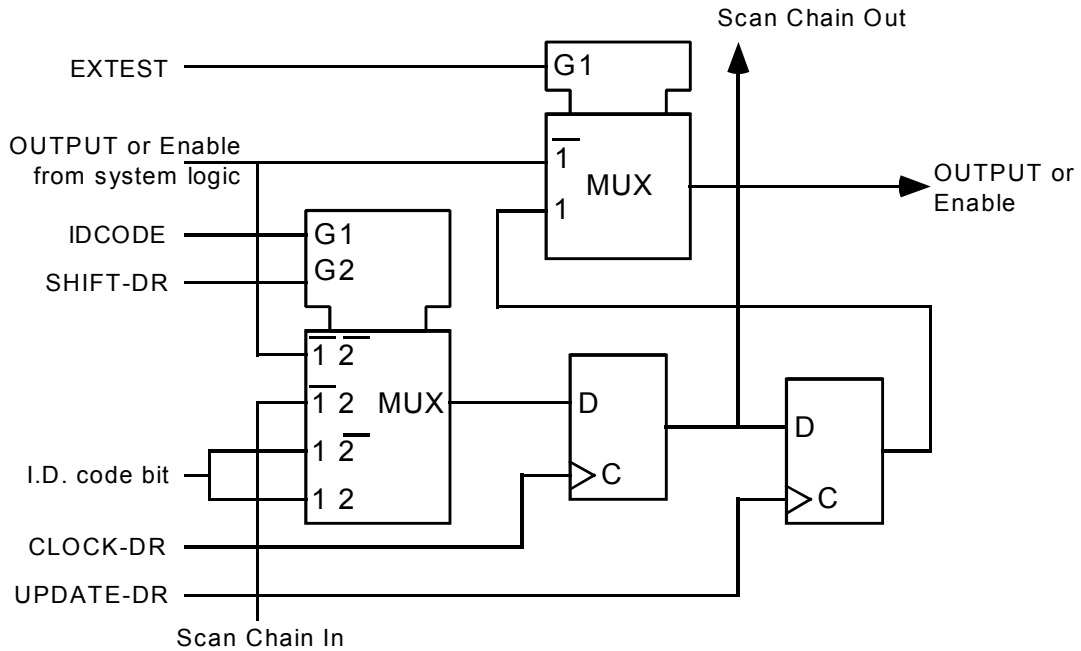


Figure 5 - Bi-directional Cell (IO_CELL)

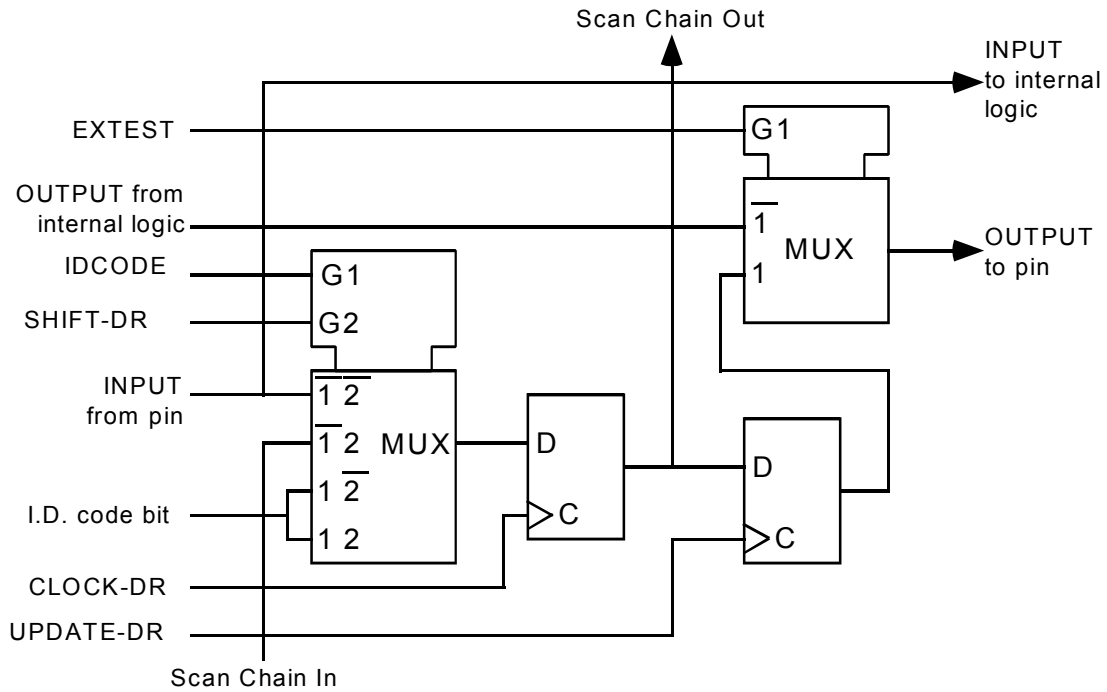
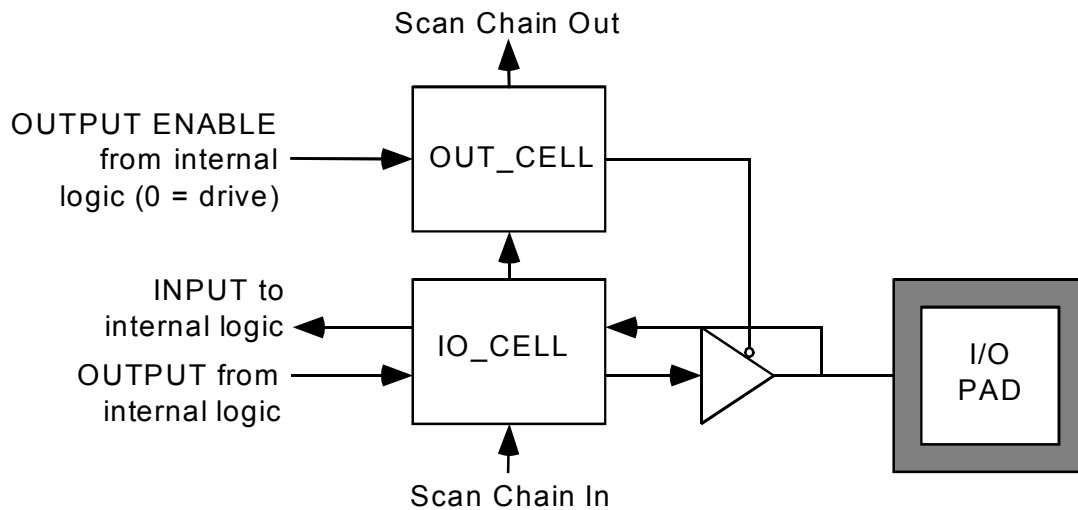


Figure 6 - Layout of Output Enable and Bi-directional Cells



12 D.C. CHARACTERISTICS

$T_C = -40^{\circ}\text{C}$ to $+85^{\circ}\text{C}$, $V_{DD} = 3.3\text{V} \pm 10\%$, $V_{DD} < \text{BIAS} < 5.5\text{V}$

(Typical Conditions: $T_C = 25^{\circ}\text{C}$, $V_{DD} = 3.3\text{V}$, $V_{BIAS} = 5\text{V}$)

Table 8 - DC Characteristics

Symbol	Parameter	Min	Typ	Max	Units	Conditions
VDD	Power Supply	2.97	3.3	3.63	Volts	
BIAS	5V Tolerant Bias	VDD	5.0	5.5	Volts	
I _{BIAS}	Current into 5V Bias		6.0		μA	V _{BIAS} = 5.5V
V _{IL}	Input Low Voltage	0		0.8	Volts	Guaranteed Input Low voltage.
V _{IH}	Input High Voltage	2.0		BIAS	Volts	Guaranteed Input High voltage.
V _{OL}	Output or Bi-directional Low Voltage		0.23	0.4	Volts	Guaranteed output Low voltage at VDD=2.97V and I _{OL} =maximum rated for pad. ^{4, 5, 6}
V _{OH}	Output or Bi-directional High Voltage	2.4	2.93		Volts	Guaranteed output High voltage at VDD=2.97V and I _{OH} =maximum rated current for pad. ^{4, 5, 6}
V _{T-}	Reset Input Low Voltage			0.8	Volts	Applies to RSTB, TRSTB, TICLK[4:1], RCLK[4:1], TFCLK, RFCLK, TCK, TDI, TMS, and REF8KI.
V _{T+}	Reset Input High Voltage	2.0			Volts	Applies to RSTB, TRSTB, TICLK[4:1], RCLK[4:1], TFCLK, RFCLK, TCK, TDI, TMS, and REF8KI.
V _{TH}	Reset Input Hysteresis Voltage		0.5		Volts	Applies to RSTB, TRSTB, TICLK[4:1], RCLK[4:1], TFCLK, RFCLK, TCK, TDI, TMS, and REF8KI.
I _{ILPU}	Input Low Current	-100	-60	-10	μA	V _{IL} = GND. ^{1, 3}
I _{IHPU}	Input High Current	-10	0	+10	μA	V _{IH} = VDD. ^{1, 3}
I _{IL}	Input Low Current	-10	0	+10	μA	V _{IL} = GND. ^{2, 3}
I _{IH}	Input High Current	-10	0	+10	μA	V _{IH} = VDD. ^{2, 3}
C _{IN}	Input Capacitance		6		pF	t _A =25°C, f = 1 MHz
C _{OUT}	Output Capacitance		6		pF	t _A =25°C, f = 1 MHz

Symbol	Parameter	Min	Typ	Max	Units	Conditions
C _{IO}	Bi-directional Capacitance		6		pF	t _A =25°C, f = 1 MHz
I _{DDOP2}	Operating Current		12.2	30	mA	VDD = 3.63V, Outputs Unloaded (T1/E1 PLCP mode)
I _{DDOP6}	Operating Current		258.3	330	mA	VDD = 3.63V, Outputs Unloaded (52 Mbit/s arbitrary framing format with ATM direct mapping)

Notes on D.C. Characteristics:

1. Input pin or bi-directional pin with internal pull-up resistor.
2. Input pin or bi-directional pin without internal pull-up resistor
3. Negative currents flow into the device (sinking), positive currents flow out of the device (sourcing).
4. The Utopia interface outputs, RDATA[15:0], RPRTY, RCA, DRCA[4:1], RSOC, TCA, and DTCA[4:1], have 12 mA drive capability.
5. The outputs TCLK[4:1], TDATA[4:1], TOHM[4:1], LCD [4:1], RPOH [4:1], RPOHCLK [4:1], and RPOHFP [4:1] have 6 mA drive capability.
6. The data bus outputs, D[7:0], and all outputs not specified above have 3 mA drive capability.
7. RFCLK and TFCLK are 3.3 V only input pins – they are **not** 5 V tolerant. Connecting a 5 V signal to these inputs may result in damage to the part.

13 ORDERING AND THERMAL INFORMATION

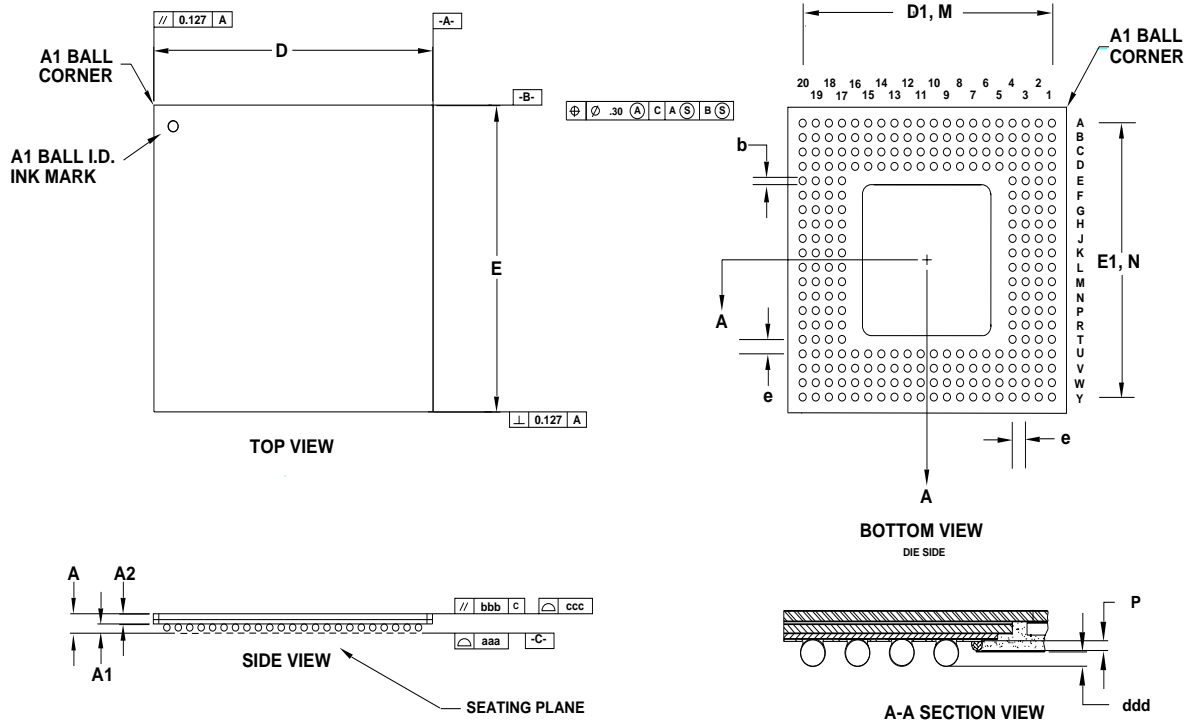
Table 9 - Packaging Information

PART NO	DESCRIPTION
PM7339	256-pin Ball Grid Array (SBGA)

Table 10 - Thermal Information

PART NO.	CASE TEMPERATURE	Theta Ja	Theta Jc
PM7339	-40°C to 85°C	19 °C/W	5 °C/W

14 MECHANICAL INFORMATION



- Notes: 1) ALL DIMENSIONS IN MILLIMETER.
 2) DIMENSION aaa DENOTES COPLANARITY
 3) DIMENSION bbb DENOTES PARALLEL
 4) DIMENSION ccc DENOTES FLATNESS

PACKAGE TYPE: 256 PIN THERMAL BALL GRID ARRAY															
BODY SIZE: 27 x 27 x 1.45 MM															
Dim.	A	A1	A2	D	D1	E	E1	M,N	e	b	aaa	bbb	ccc	ddd	P
Min.	1.32	0.56	0.76	26.90	24.03	26.90	24.03			0.60				0.15	0.20
Nom.	1.45	0.63	0.82	27.00	24.13	27.00	24.13	20x20	1.27	0.75				0.33	0.30
Max.	1.58	0.70	0.88	27.10	24.23	27.10	24.23			0.90	0.15	0.15	0.20	0.50	0.35

NOTES

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