

### 10.7Gbps Laser Diode Drivers

## General Description

The MAX3930/MAX3931/MAX3932 are designed for direct modulation of laser diodes at data rates up to 10.7Gbps. They provide adjustable laser bias and modulation currents and are implemented using Maxim's second-generation in-house SiGe process.
The MAX3930 accepts differential CML clock and data input signals and includes $50 \Omega$ on-chip termination resistors. It delivers a 1 mA to 100 mA laser bias current and a 20 mA to 100 mA modulation current with a typical ( $20 \%$ to $80 \%$ ) 25ps rise time. An input data retiming latch can be used to reject input pattern-dependent jitter if a clock signal is available.
The MAX3931/MAX3932 have an alternate pad out with respect to the MAX3930. The MAX3931 includes the series damping resistor $\mathrm{RD}_{\mathrm{D}}$ on chip.
The MAX3930/MAX3931/MAX3932 also include an adjustable pulse-width control circuit to minimize laser pulse-width distortion.

## Applications

SONET OC-192 and SDH STM-64
Transmission Systems
Up to 10.7Gbps Optical Transmitters
Section Regenerators

Features

- Single +5 V or -5.2V Power Supply
- 108mA Supply Current
- Operates to 10.7Gbps
- $50 \Omega$ On-Chip Input Termination Resistors
- Programmable Modulation Current to 100 mA
- Programmable Laser Bias Current to 100mA
- 25ps Rise Time (MAX3930/MAX3932)
- Adjustable Pulse-Width Control
- Selectable Data Retiming Latch
- ESD Protection
- Internal Series Damping Resistor (MAX3931)

Ordering Information

| PART | TEMP RANGE | PIN-PACKAGE |
| :--- | :--- | :--- |
| MAX3930E/D | $-40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$ | Dice |
| MAX3931E/D | $-40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$ | Dice |
| MAX3932E/D | $-40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$ | Dice |
| MAX3932E/W | $-40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$ | Wafer |

Note: Dice are designed to operate over a $-40^{\circ} \mathrm{C}$ to $+120^{\circ} \mathrm{C}$ junction temperature (TJ) range but are tested and guaranteed at $T_{A}=+25^{\circ} \mathrm{C}$.

$\dagger$ Covered by U.S. Patent number 5,883,910.

### 10.7Gbps Laser Diode Drivers

## ABSOLUTE MAXIMUM RATINGS

Supply Voltage (VCC - VEE)..................................-0.5V to +6.0 V DATA+, DATA-, CLK+,
.$\left(\mathrm{V}_{\mathrm{TT}}-1.2 \mathrm{~V}\right)$ to the lower of
$\left(\mathrm{V}_{\mathrm{TT}}+1.2 \mathrm{~V}\right)$ or $\left(\mathrm{V}_{\mathrm{CC}}+0.5 \mathrm{~V}\right)$
$\overline{M O D E N}, \overline{R T E N}, V_{T T}$, BIASMON, MODMON,
PWC+, and PWC $\qquad$ .$\left(V_{E E}-0.5 \mathrm{~V}\right)$ to $\left(\mathrm{V}_{\mathrm{CC}}+0.5 \mathrm{~V}\right)$
MODN1, MODN2
$(\mathrm{VCC}-0.5 \mathrm{~V})$ to $(\mathrm{VCC}+0.5 \mathrm{~V})$
BIAS, MOD1, MOD2
. $\mathrm{V}_{\mathrm{EE}}+1 \mathrm{~V}$ ) to $\left(\mathrm{V}_{\mathrm{EE}}+1.5 \mathrm{~V}\right)$

MODSET and BIASSET .................... (VEE $-0.5 \mathrm{~V})$ to ( $\mathrm{V}_{\mathrm{EE}}+1.5 \mathrm{~V}$ )
Storage Temperature Range ............................. $55^{\circ} \mathrm{C}$ to $+150^{\circ} \mathrm{C}$
Operating Junction Temperature ...................... $-55^{\circ} \mathrm{C}$ to $+150^{\circ} \mathrm{C}$
Processing Temperature (die) ......................................... $+400^{\circ} \mathrm{C}$
Current into DATA+, DATA-, CLK+,
CLK- ( $\mathrm{V}_{\mathrm{TT}}=\mathrm{V}_{\mathrm{CC}}$ ).
-24 mA to +30.5 mA
Current into DATA+, DATA-, CLK+,
CLK- $\left(\mathrm{V}_{T T}=\mathrm{V}_{\mathrm{CC}}-1.3 \mathrm{~V}\right)$ $\qquad$ ..$-24 m A$ to $+24 m A$

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

## DC ELECTRICAL CHARACTERISTICS—MAX3930

$\left(\mathrm{V}_{C C}-\mathrm{V}_{E E}=4.75 \mathrm{~V}\right.$ to $5.5 \mathrm{~V}, \mathrm{~T}_{\mathrm{A}}=-40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$. Typical values are at $\mathrm{V}_{\mathrm{CC}}-\mathrm{V}_{\mathrm{EE}}=5 \mathrm{~V}$, $\mathrm{I} \mathrm{IIAS}=50 \mathrm{~mA}, \mathrm{I}_{\mathrm{MOD}}=70 \mathrm{~mA}$, and $\mathrm{T}_{\mathrm{A}}=+25^{\circ} \mathrm{C}$, unless otherwise noted.)

| PARAMETER | SYMBOL | CONDITIONS | MIN | TYP | MAX | UNITS |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Power-Supply Voltage | VCC - VEE |  | 4.75 | 5 | 5.50 | V |
| Power-Supply Current | IcC | Excluding bias current and modulation current |  | 108 | 140 | mA |
| Single-Ended Input Resistance |  |  | 42.5 | 50 | 57.5 | $\Omega$ |
| Bias Current-Setting Range |  |  | 1 |  | 100 | mA |
| Bias Current-Setting Error |  | Bias current $=100 \mathrm{~mA}, \mathrm{~T}_{\mathrm{A}}=+25^{\circ} \mathrm{C}$ | -5 |  | +5 | \% |
|  |  | Bias current $=1 \mathrm{~mA}, \mathrm{~T}_{\text {A }}=+25^{\circ} \mathrm{C}$ | -10 |  | +10 |  |
| Bias Sensing Resistor | RBIAS |  | 2.7 | 3 | 3.3 | $\Omega$ |
| Bias Current Temperature Stability |  | IBIAS $=100 \mathrm{~mA}($ Note 1) | -480 |  | +480 | ppmo ${ }^{\circ} \mathrm{C}$ |
|  |  | IBIAS $=1 \mathrm{~mA}$ ( Note 1) |  | -200 |  |  |
| Bias Off-Current |  | BIASSET $\leq\left(\mathrm{V}_{\mathrm{EE}}+0.4 \mathrm{~V}\right)$ |  |  | 0.05 | mA |
| $\overline{\text { MODEN }}$ and $\overline{\text { RTEN }}$ Input High | $\mathrm{V}_{\mathrm{IH}}$ |  | $\begin{gathered} \mathrm{V}_{\mathrm{EE}}+ \\ 2 \end{gathered}$ |  |  | V |
| $\overline{\text { MODEN }}$ and $\overline{\text { RTEN }}$ Input Low | VIL |  |  |  | $\begin{gathered} \mathrm{V}_{\mathrm{EE}}+ \\ 0.8 \end{gathered}$ | V |
| Power-Supply Rejection Ratio | PSRR | $\mathrm{V}_{\mathrm{CC}}=4.75 \mathrm{~V}$ to 5.5V (Note 2) | 39.5 | 60 |  | dB |
| SIGNAL INPUT FOR $\mathrm{V}_{\text {TT }}=\mathrm{V}_{\mathbf{C c}}$ |  |  |  |  |  |  |
| Single-Ended Input (DC-Coupled) | VIS | At high |  | VCC |  | V |
|  |  | At low | $V_{\text {CC }}-1$ |  | $\begin{gathered} V_{C C}- \\ 0.15 \end{gathered}$ |  |
| Single-Ended Input (AC-Coupled) | VIS | At high | $\begin{aligned} & \text { VCC + } \\ & 0.075 \end{aligned}$ |  | $\begin{gathered} \text { VCC }+ \\ 0.4 \end{gathered}$ | V |
|  |  | At low | $\begin{gathered} \mathrm{V}_{\mathrm{CC}}- \\ 0.4 \end{gathered}$ |  | $\begin{aligned} & \text { VCC - } \\ & 0.075 \end{aligned}$ |  |
| Differential Input Swing (DC-Coupled) | VID |  | 0.3 |  | 2.0 | VP-P |
| Differential Input Swing (AC-Coupled) | $V_{\text {ID }}$ |  | 0.3 |  | 1.6 | VP-P |
| SIGNAL INPUT FOR $\mathrm{V}_{\mathrm{TT}}=\left(\mathrm{V}_{\mathrm{CC}}-1.3 \mathrm{~V}\right)$ |  |  |  |  |  |  |
| Input Common Mode | VICM |  |  | $\begin{array}{r} \hline \text { VCC - } \\ 1.3 \\ \hline \end{array}$ |  | V |

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## DC ELECTRICAL CHARACTERISTICS—MAX3930 (continued)

$\left(\mathrm{V}_{C C}-\mathrm{V}_{\mathrm{EE}}=4.75 \mathrm{~V}\right.$ to $5.5 \mathrm{~V}, \mathrm{~T}_{\mathrm{A}}=-40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$. Typical values are at $\mathrm{V}_{\mathrm{CC}}-\mathrm{V}_{\mathrm{EE}}=5 \mathrm{~V}$, $\mathrm{I}_{\mathrm{BIAS}}=50 \mathrm{~mA}, \mathrm{I}_{\mathrm{MOD}}=70 \mathrm{~mA}$, and $\mathrm{T}_{\mathrm{A}}=+25^{\circ} \mathrm{C}$, unless otherwise noted.)

| PARAMETER | SYMBOL | CONDITIONS | MIN | TYP | MAX | UNITS |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Single-Ended Input | VIS | At high | $\begin{aligned} & V_{C C}- \\ & 1.225 \end{aligned}$ |  | $\begin{gathered} \mathrm{V}_{\mathrm{CC}}- \\ 0.8 \end{gathered}$ | V |
|  |  | At low | $\begin{gathered} \hline \mathrm{V}_{\mathrm{CC}}- \\ 1.8 \end{gathered}$ |  | $\begin{aligned} & \mathrm{V}_{\mathrm{CC}}- \\ & 1.375 \end{aligned}$ |  |
| Differential Input Swing | VID |  | 0.3 |  | 2.0 | VP-P |

## DC ELECTRICAL CHARACTERISTICS—MAX3931/MAX3932

$\left(\mathrm{V}_{\mathrm{CC}}-\mathrm{V}_{\mathrm{EE}}=4.75 \mathrm{~V}\right.$ to $5.5 \mathrm{~V}, \mathrm{~T}_{\mathrm{A}}=-40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$. Typical values are at $\mathrm{V}_{\mathrm{CC}}-\mathrm{V}_{\mathrm{EE}}=5 \mathrm{~V}$, $\mathrm{I}_{\mathrm{BIAS}}=50 \mathrm{~mA}$, $\mathrm{I}_{\mathrm{MOD}}=70 \mathrm{~mA}$, and $\mathrm{T}_{\mathrm{A}}=+25^{\circ} \mathrm{C}$, unless otherwise noted.)

| PARAMETER | SYMBOL | CONDITIONS | MIN | TYP | MAX | UNITS |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Power-Supply Voltage | VCC - VEE |  | 4.75 | 5 | 5.50 | V |
| Power-Supply Current | IcC | Excluding bias current and modulation current |  | 108 | 140 | mA |
| Single-Ended Input Resistance |  |  | 42.5 | 50 | 57.5 | $\Omega$ |
| Bias Current Setting Range |  |  | 1 |  | 100 | mA |
| Bias Current Setting Error |  | Bias current $=100 \mathrm{~mA}, \mathrm{~T}_{\text {A }}=+25^{\circ} \mathrm{C}$ | -5 |  | +5 | \% |
|  |  | Bias current $=1 \mathrm{~mA}, \mathrm{~T}_{\mathrm{A}}=+25^{\circ} \mathrm{C}$ | -10 |  | +10 |  |
| Bias Sensing Resistor | RBIAS |  | 2.7 | 3 | 3.3 | $\Omega$ |
| Bias Current Temperature Stability |  | IBIAS $=100 \mathrm{~mA}($ Note 1) | -480 |  | +480 | ppm/ ${ }^{\circ} \mathrm{C}$ |
|  |  | IBIAS $=1 \mathrm{~mA}$ |  | -200 |  |  |
| Bias Off-Current |  | BIASSET $\leq\left(\mathrm{V}_{\mathrm{EE}}+0.4 \mathrm{~V}\right)$ |  |  | 0.05 | mA |
| $\overline{\text { MODEN }}$ and $\overline{\text { RTEN }}$ Input High | $\mathrm{V}_{\mathrm{IH}}$ |  | $\begin{gathered} \mathrm{V}_{\mathrm{EE}}+ \\ 2 \end{gathered}$ |  |  | V |
| $\overline{\text { MODEN }}$ and $\overline{\text { RTEN }}$ Input Low | VIL |  |  |  | $\begin{gathered} \mathrm{V}_{\mathrm{EE}}+ \\ 0.8 \end{gathered}$ | V |
| Power-Supply Rejection Ratio | PSRR | $\mathrm{V}_{\mathrm{CC}}=4.75 \mathrm{~V}$ to 5.5V (Note 3) | 39.5 | 60 |  | dB |
| SIGNAL INPUT |  |  |  |  |  |  |
| Single-Ended Input (DC-Coupled) | VIS | At high |  | VCC |  | V |
|  |  | At low | VCC - 1 |  | $\begin{gathered} \hline \mathrm{V}_{\mathrm{CC}}- \\ 0.15 \end{gathered}$ |  |
| Single-Ended Input (AC-Coupled) | VIS | At high | $\begin{aligned} & V_{C C}+ \\ & 0.075 \end{aligned}$ |  | $\begin{gathered} V_{C C}+ \\ 0.4 \end{gathered}$ | V |
|  |  | At low | $\begin{gathered} \mathrm{V}_{\mathrm{CC}}- \\ 0.4 \end{gathered}$ |  | $\begin{aligned} & V_{\text {CC }}- \\ & 0.075 \end{aligned}$ |  |
| Differential Input Swing (DC-Coupled) | $V_{\text {ID }}$ |  | 0.3 |  | 2.0 | VP-P |
| Differential Input Swing (AC-Coupled) | VID |  | 0.3 |  | 1.6 | VP-P |

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## AC ELECTRICAL CHARACTERISTICS—MAX3930/MAX3932

$\left(\mathrm{V}_{\mathrm{CC}}-\mathrm{V}_{\mathrm{EE}}=4.75 \mathrm{~V}\right.$ to $5.5 \mathrm{~V}, \mathrm{~V}_{\mathrm{TT}}=\mathrm{V}_{\mathrm{CC}}, \mathrm{T}_{\mathrm{A}}=-40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$. Typical values are at $\mathrm{V}_{\mathrm{CC}}-\mathrm{V}_{\mathrm{EE}}=5 \mathrm{~V}, \mathrm{I}_{\mathrm{MOD}}=70 \mathrm{~mA}$, and $\mathrm{T}_{\mathrm{A}}=+25^{\circ} \mathrm{C}$, unless otherwise noted.) (Note 3)

| PARAMETER | SYMBOL | CONDITIONS | MIN | TYP | MAX | UNITS |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Input Data Rates |  | NRZ |  | 10.7 |  | Gbps |
| Modulation Current Setting Range |  |  | 20 |  | 100 | mA |
| Modulation Current Setting Error |  | $20 \Omega$ load, $\mathrm{T}_{\mathrm{A}}=+25^{\circ} \mathrm{C}$ | -5 |  | +5 | \% |
| Modulation Sensing Resistor | RMOD |  | 2.7 | 3 | 3.3 | $\Omega$ |
| Modulation Current Temperature Stability |  |  | -480 |  | +480 | ppm $/{ }^{\circ} \mathrm{C}$ |
| Modulation Off-Current |  | MODSET $\leq\left(\mathrm{V}_{\mathrm{EE}}+0.4 \mathrm{~V}\right)$ |  |  | 0.1 | mA |
| Output Current Rise Time | tR | $\mathrm{Z}_{\mathrm{L}}=20 \Omega, 20 \%$ to 80\% (Note 4) |  | 25 | 35 | ps |
| Output Current Fall Time | $\mathrm{t}_{\mathrm{F}}$ | $\mathrm{Z}_{\mathrm{L}}=20 \Omega, 20 \%$ to 80\% (Note 4) |  | 29 | 36 | ps |
| Setup/Hold Time | tsu, thD | Figure 2 | 25 |  |  | ps |
| Pulse-Width Adjustment Range |  | (Note 4) | $\pm 25$ | $\pm 55$ |  | ps |
| Pulse-Width Stability |  | PWC+ and PWC- open (Note 4) |  |  | $\pm 13$ | ps |
| Pulse-Width Control Input Range |  | For PWC+ and PWC | $\begin{gathered} \mathrm{V}_{\mathrm{EE}}+ \\ 0 \end{gathered}$ | $\begin{gathered} \mathrm{V}_{\mathrm{EE}+}+ \\ 1 \end{gathered}$ | $\begin{gathered} \mathrm{V}_{\mathrm{EE}+}+ \\ 2 \end{gathered}$ | V |
| Overshoot |  | (Note 4) |  |  | 13 | \% |
| Driver Random Jitter |  |  |  | 0.75 | 1 | PSRMS |
| Driver Deterministic Jitter |  | (Note 5) |  | 6.7 | 21 | psp-P |
| Input Return Loss |  |  |  | 12 |  | dB |

## AC ELECTRICAL CHARACTERISTICS—MAX3931

$\left(\mathrm{V}_{\mathrm{CC}}-\mathrm{V}_{E E}=4.75 \mathrm{~V}\right.$ to $5.5 \mathrm{~V}, \mathrm{~V}_{T T}=\mathrm{V}_{\mathrm{CC}}, \mathrm{T}_{\mathrm{A}}=-40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$. Typical values are at $\mathrm{V}_{\mathrm{CC}}-\mathrm{V}_{\mathrm{EE}}=5 \mathrm{~V}, \mathrm{I}_{\mathrm{MOD}}=70 \mathrm{~mA}$, and $\mathrm{T}_{\mathrm{A}}=+25^{\circ} \mathrm{C}$, unless otherwise noted.) (Note 3)

| PARAMETER | SYMBOL | CONDITIONS | MIN | TYP | MAX | UNITS |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Input Data Rates |  | NRZ |  | 10.7 |  | Gbps |
| Modulation Current Setting Range |  |  | 20 |  | 100 | mA |
| Modulation Current Setting Error |  | $20 \Omega$ load, $\mathrm{T}_{\mathrm{A}}=+25^{\circ} \mathrm{C}$ | -5 |  | +5 | \% |
| Modulation Sensing Resistor | RMOD |  | 2.7 | 3 | 3.3 | $\Omega$ |
| Output Series Resistance |  | RMOD1 in parallel with RMOD2 | 12.75 | 15 | 17.25 | $\Omega$ |
| Modulation Current Temperature Stability |  |  | -480 |  | +480 | ppm/ ${ }^{\circ} \mathrm{C}$ |
| Modulation Off-Current |  | MODSET $\leq\left(\mathrm{V}_{\mathrm{EE}}+0.4 \mathrm{~V}\right)$ |  |  | 0.1 | mA |
| Setup/Hold Time | tsu, thD | Figure 2 | 25 |  |  | ps |
| Pulse-Width Adjustment Range |  | (Note 4) | $\pm 25$ | $\pm 55$ |  | ps |
| Pulse-Width Stability |  | PWC+ and PWC- open (Note 4) |  |  | 13 | ps |
| Pulse-Width Control Input Range |  | For PWC+ and PWC- | $\begin{gathered} \mathrm{V}_{\mathrm{EE}}+ \\ 0 \end{gathered}$ | $\begin{gathered} \mathrm{V}_{\mathrm{EE}}+ \\ 1 \end{gathered}$ | $\begin{gathered} \mathrm{V}_{\mathrm{EE}}+ \\ 2 \end{gathered}$ | V |
| Input Return Loss |  |  |  | 12 |  | dB |

Note 1: Guaranteed by design and characterization.
Note 2: $\operatorname{PSRR}=20 \times \log (\Delta \mathrm{VCC} /(\Delta \mathrm{IMOD} \times 20 \Omega))$. $\mathrm{IMOD}=100 \mathrm{~mA}$
Note 3: Guaranteed by design and characterization using the circuit shown in Figure 1.
Note 4: Measured using a 10.7Gbps repeating 0000000011111111 pattern.
Note 5: Measured using a $10.7 \mathrm{Gbps} 2^{13}-1$ PRBS with eighty Os pattern.
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Pad Description

| PAD |  | NAME | FUNCTION |
| :---: | :---: | :---: | :---: |
| MAX3930 | MAX3931/ MAX3932 |  |  |
| $\begin{gathered} 1,5,9,12, \\ 22,23,28, \\ 29 \end{gathered}$ | $\begin{gathered} 1,3,5,7, \\ 9,10,12, \\ 22,23,28, \\ 29 \end{gathered}$ | VCC | Power-Supply Voltage ( $\left.\mathrm{V}_{\mathrm{CC}}-\mathrm{V}_{\mathrm{EE}}=5 \mathrm{~V}\right)$. All pads must be connected to $\mathrm{V}_{\mathrm{CC}}$. |
| 2 | 2 | DATA+ | Noninverting Data Input. CML with on-chip termination resistor. |
| 3 | - | $V_{\text {TT }}$ | Terminating Voltage for Data Inputs |
| 4 | 4 | DATA- | Inverting Data Input. CML with on-chip termination resistor. |
| 6 | 6 | CLK+ | Noninverting Clock Input for Data Retiming. CML with on-chip termination resistor. |
| 7 | - | $V_{\text {TT }}$ | Terminating Voltage for Clock Inputs |
| 8 | 8 | CLK- | Inverting Clock Input for Data Retiming. CML with on-chip termination resistor. |
| $\begin{aligned} & 10,11,17 \\ & 18,21,32, \\ & 35,36,37 \end{aligned}$ | $\begin{gathered} 11,17,18, \\ 19,32,35 \\ 36,37 \end{gathered}$ | VEE | Power-Supply Voltage (VCC - $\mathrm{VEE}^{\text {E }}=5 \mathrm{~V}$ ) |
| 13 | 13 | $\overline{\text { RTEN }}$ | TTL/CMOS Data Retiming Input. Low for latched data, high for direct data. Internal $100 \mathrm{k} \Omega$ pullup to $\mathrm{V}_{\mathrm{Cc}}$. |
| 14 | 14 | PWC+ | Positive Input for Modulation Pulse-Width Adjustment. Connected to ground through Rpwc. |
| 15 | 15 | PWC- | Negative Input for Modulation Pulse-Width Adjustment. Connected to ground through Rpwc. |
| 16 | 16 | $\overline{\text { MODEN }}$ | TTL/CMOS Modulation Enable Input. Low for normal operation, high to switch modulation output off. Internal $100 \mathrm{k} \Omega$ pullup to $\mathrm{V}_{\mathrm{CC}}$. |
| 19 | 20 | MODMON | Modulation Current Monitor ( $\mathrm{V}_{\text {MODMON }}-\mathrm{V}_{\text {EE }}$ ) / RMOD $=1 \mathrm{IMOD}$ |
| 20 | 21 | MODSET | Modulation Current Set. Connected to the output of the external operational amplifier (see the Design Procedure section). |
| 24, 27 | 24, 27 | MODN2, <br> MODN1 | Complementary Laser Modulation Current Outputs. Connect to Vcc. |
| 25, 26 | 25, 26 | $\begin{aligned} & \text { MOD2, } \\ & \text { MOD1 } \end{aligned}$ | Laser Modulation Current Outputs |
| 30 | 30 | BIAS | Laser Bias Current Output |
| 31 | 31 | N.C. | No Connection. Leave unconnected. |
| 33 | 33 | BIASSET | Bias Current Set. Connected to the output of the external operational amplifier (see the Design Procedure section). |
| 34 | 34 | BIASMON | Bias Current Monitor (VBIASMON - Vee) / Rbias = IbIAS |

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Figure 1. Test Circuit


Figure 2. Required Input Signal, Setup/Hold Time Definition, and Output Polarity

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## Detailed Description

The MAX3930 laser driver consists of two main parts, a high-speed modulation driver and a laser-biasing block. The circuit operates from a single 5 V or -5.2 V supply. When operating from a 5 V supply, connect all $V_{C C}$ pins to 5 V and all $\mathrm{V}_{\mathrm{EE}}$ pins to ground. If operating from a -5.2 V supply, connect all $\mathrm{V}_{\text {EE }}$ pins to -5.2 V and all Vcc pins to ground. To eliminate pattern-dependent jitter on the input data signal, the device accepts a differential CML clock signal for data retiming. When $\overline{\text { RTEN }}$ is tied to a low potential, the input data is synchronized by the clock signal. When RTEN is tied high or left floating, the input data is transmitted directly to the output stage (retiming is disabled).
The output stage is composed of a high-speed differential pair and a programmable modulation current source with a maximum modulation current of 100 mA . The rise and fall times are typically 25 ps and 29ps, respectively.
The MAX3930/MAX3932 modulation output is optimized for driving a $20 \Omega$ load. The minimum voltage required at MOD is 1.55 V . To interface with a laser diode, a series damping resistor ( $R_{D}$ ) is required for impedance matching ( $R \mathrm{D}=15 \Omega$, assuming a laser resistance of $5 \Omega$; see Typical Application Circuit).
The MAX3931 output has an internal series damping resistor consisting of two parallel $30 \Omega$ resistors in series with the output. This simplifies interfacing with the laser diode. The MAX3931/MAX3932 have an alternate pad out with respect to MAX3930.
At the 10.7 Gbps data rate, any capacitive load at the cathode of a laser diode will degrade the optical output performance. Since the BIAS output is directly connected to the laser cathode, minimize the parasitic capacitance associated with this pad by using a ferrite bead (LB) to isolate the BIAS pin from the laser cathode.

Optional Input Data Retiming
To eliminate pattern-dependent jitter on the input data, a synchronous differential clock signal should be connected to the CLK+ and CLK- inputs, and the RTEN control input should be tied low. The input data is retimed on the rising edge of CLK + . If RTEN is tied high or left floating, the retiming function is disabled, and the input data is directly connected to the output stage. Leave CLK+ and CLK- open when retiming is disabled.

## Modulation Output Enable

The MAX3930/MAX3931/MAX3932 incorporate a modulation current enable input. When MODEN is low, the modulation outputs (MOD1, MOD2) are enabled. When MODEN is high, the modulation outputs (MOD1, MOD2)
are disabled. The typical laser enable time is 2 ns , and the typical disable time is 5 ns .

Pulse-Width Control
The pulse-width control circuit can be used to precompensate for laser pulse-width distortion. The differential voltage between PWC+ and PWC- adjusts the pulsewidth compensation.
When PWC+ and PWC- are left open, the pulse-width control circuit is automatically disabled.

Current Monitors
The MAX3930/MAX3931/MAX3932 feature a bias current monitor output (BIASMON) and a modulation current monitor output (MODMON). The voltage at BIASMON is equal to (lbias $\times$ Rbias) $+\mathrm{V}_{\mathrm{EE}}$, and the voltage at MODMON is equal to (IMOD $\times$ RMOD $)+$ VEE, where IBIAS represents the laser bias current, IMOD represents the modulation current, and RBIAS and RMOD are internal $3 \Omega$ ( $\pm 10 \%$ ) resistors. BIASMON and MODMON should be connected to the inverting input of an operational amplifier to program the bias and moduIation current (see Design Procedure).

## Design Procedure

When designing a laser transmitter, the optical output is usually expressed in terms of average power and extinction ratio. Table 1 gives relationships that are helpful in converting between the optical average power and the modulation current. These relationships are valid if the mark density and duty cycle of the optical waveform are 50\%.

## Programming the Modulation Current

For a desired laser average optical power, PAVG, and optical extinction ratio, $\mathrm{r}_{\mathrm{e}}$, the required modulation current can be calculated based on the laser slope efficiency, $\eta$, using the equations in Table 1.
To program the desired modulation current, connect the inverting input of an operational amplifier (such as the MAX480) to MODMON and connect the output to MODSET. Connect the positive op amp voltage supply to $\mathrm{V}_{C C}$ and the negative supply to $\mathrm{V}_{E E}$ (for 5 V operation, $\mathrm{V}_{\mathrm{CC}}=5 \mathrm{~V}$ and $\mathrm{V}_{E E}=$ ground; for -5.2 V operation, $\mathrm{V}_{C C}=$ ground and $\mathrm{V}_{E E}=-5.2 \mathrm{~V}$ ). The modulation current is set by connecting a reference voltage, VMOD, to the noninverting input of the operational amplifier. Refer to the IMOD vs. VMOD graph in the Typical Operating Characteristics to select the value of $\mathrm{V}_{\mathrm{MOD}}$ that corresponds to the required modulation current.

### 10.7Gbps Laser Diode Drivers

Table 1. Optical Power Relations

| PARAMETER | SYMBOL | RELATION |
| :---: | :---: | :---: |
| Average Power | Pavg | $\mathrm{PaVG}^{\text {a }}=\left(\mathrm{P}_{0}+\mathrm{P}_{1}\right) / 2$ |
| Extinction Ratio | $\mathrm{r}_{\mathrm{e}}$ | $\mathrm{r}_{\mathrm{e}}=\mathrm{P}_{1} / \mathrm{P}_{0}$ |
| Optical Power of a "1" | $\mathrm{P}_{1}$ | $\mathrm{P}_{1}=2 \mathrm{PaVGg}^{\text {e }} /\left(r_{\mathrm{e}}+1\right)$ |
| Optical Power of a "0" | P0 | $\mathrm{P}_{0}=2 \mathrm{P}_{\text {AVG }} /\left(\mathrm{r}_{\mathrm{e}}+1\right)$ |
| Optical Amplitude | Pp-P | $\begin{aligned} & P_{P-P}=P_{1}-P_{0}= \\ & 2 P_{\text {AVG }}\left(r_{e}-1\right) /\left(r_{e}+1\right) \end{aligned}$ |
| Laser Slope Efficiency | $\eta$ | $\eta=$ PP-P / IMOD |
| Modulation Current | Imod | $1 \mathrm{MOD}=$ PP-P / $\eta$ |

Note: Assuming a 50\% average input duty cycle and mark density.


Figure 3. Optical Power Relations

## Programming the Bias Current

To program the desired laser bias current, connect the inverting input of an operational amplifier (such as the MAX480) to BIASMON, and connect the output to BIASSET. Connect the positive op amp voltage supply to $\mathrm{V}_{\mathrm{CC}}$ and the negative supply to $\mathrm{V}_{\mathrm{EE}}$ (for 5 V operation, $\mathrm{V}_{\mathrm{CC}}=5 \mathrm{~V}$ and $\mathrm{V}_{\mathrm{EE}}=$ ground; and for -5.2 V operation, $\mathrm{V}_{C C}=$ ground and $\mathrm{V}_{\mathrm{EE}}=-5.2 \mathrm{~V}$ ). The laser bias current is set by connecting a reference voltage, VBIAS, to the noninverting input of the operational amplifier. Refer to the IBIAS vs. VBIAS graph in the Typical Operating Characteristics to select the value of VBIAS that corresponds to the required laser bias current.

## Interfacing with Laser Diodes

Refer to Maxim Application Note HFAN-2.0, Interfacing Maxim Laser Drivers with Laser Diodes, for detailed information.

To minimize optical output aberrations caused by signal reflections at the electrical interface to the laser diode, a series damping resistor ( $\mathrm{RD}_{\mathrm{D}}$ ) is required (Figure 4). The MAX3930/MAX3932 modulation outputs are optimized for a $20 \Omega$ load; therefore, the series combination of $R_{D}$ and $R_{L}$ (where $R_{L}$ represents the laser diode resistance) should equal $20 \Omega$. Typical values for RD are $13 \Omega$ to $17 \Omega$. The MAX3931 includes an on-chip series damping resistor RD at $15 \Omega$ (Figure 5).
For best performance, a bypass capacitor (C), typically $0.01 \mu \mathrm{~F}$, should be placed as close as possible to the anode of the laser diode.
In some applications (depending on the laser diode parasitic inductance), an RF matching network at the laser cathode will improve the optical output.

## Applications Information

## Wire Bonding Die

For high current density and reliable operation, the MAX3930/MAX3931/MAX3932 use gold metalization. Make connections to the die with gold wire only, using ball-bonding techniques. Do not use wedge bonding. Die-pad size is $3.0 \mathrm{mil}(76 \mu \mathrm{~m})$ and $4.5 \mathrm{mil}(114 \mu \mathrm{~m})$. Die thickness is $8 \mathrm{mil}(203 \mu \mathrm{~m})$. Die size is $46 \mathrm{mil} \times 82 \mathrm{mil}$ ( $1.168 \mathrm{~mm} \times 2.083 \mathrm{~mm}$ ).

## Layout Considerations

To minimize inductance, keep the connections between the driver output and the laser diode as short as possible. Optimize the laser diode performance by placing a bypass capacitor as close as possible to the laser anode. Use good high-frequency layout techniques and multilayer boards with an uninterrupted ground plane to minimize EMI and crosstalk. Use controlled impedance lines for the clock and data inputs.

## Laser Safety and IEC 825

 Using the MAX3930/MAX3931/MAX3932 laser driver alone does not ensure that a transmitter design is compliant with IEC 825. The entire transmitter circuit and component selections must be considered. Customers must determine the level of fault tolerance required by their application, recognizing that Maxim products are not designed or authorized for use as components in systems intended for surgical implant into the body, for applications intended to support or sustain life, or for any other application where the failure of a Maxim product could create a situation where personal injury or death may occur.
### 10.7Gbps Laser Diode Drivers



* $V_{T T}$ IS INTERNALLY CONNECTED TO V ${ }_{C C}$ FOR MAX3932

Figure 4. MAX3930/MAX3932 Functional Diagram
10.7Gbps Laser Diode Drivers


Figure 5. MAX3931 Functional Diagram

### 10.7Gbps Laser Diode Drivers



Figure 6. MAX3930 Equivalent Input Circuit


Figure 8. MAX3931/MAX3932 Equivalent Input Circuit


Figure 7. MAX3930/MAX3932 Equivalent Output Circuit


Figure 9. MAX3931 Equivalent Output Circuit

### 10.7Gbps Laser Diode Drivers



Chip Information
TRANSISTOR COUNT: 1555
SUBSTRATE: SOI
PROCESS: BiPOLAR SILICON GERMANIUM
DIE THICKNESS: 8mil
$\qquad$

