

# MS5534

# **BAROMETER MODULE**



- Integrated pressure sensor
- Pressure range 300-1100 mbar
- 15 Bit ADC
- 6 coefficients for a software compensation stored on-chip
- 3-wire serial interface
- 1 system clock line (32.768 kHz)
- Low voltage / low power

#### **DESCRIPTION**

The MS5534 is a SMD-hybrid device including a piezoresistive pressure sensor and an ADC-Interface IC. It provides a 16 Bit data word from a pressure- and temperature-dependent voltage. Additionally the module contains 6 readable coefficients for a highly accurate

software calibration of the sensor. MS5534 is a low-power, low-voltage device with automatic power down (ON/OFF) switching. A 3-wire interface will be used for all communications with a microcontroller.

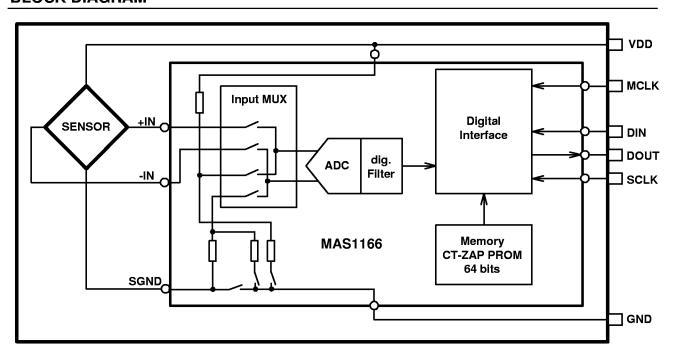
#### **FEATURES**

- 15 Bit ADC resolution
- Supply voltage 2.2 V to 3.6 V
- Low supply current
- -10°C to +60°C
- Small size
- No external components required

#### **APPLICATION**

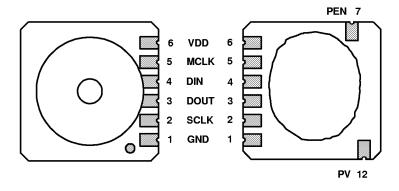
- 15 Bit ADC resolution pressure measurement and control systems
- Mobile altimeter/barometer systems
- Weather control systems
- Adventure or multi-mode watches

#### **BLOCK DIAGRAM**





#### **PIN CONFIGURATION**



# PIN DESCRIPTION

Pin Name	Pin	Type	Function
GND	1	G	Ground
SCLK	2	I	Serial Data Clock
DOUT	3	0	Data Output
DIN	4	I	Data Input
MCLK	5	I	Master Clock
VDD	6	Р	Positive Supply Voltage
PEN	7	I	Programming Enable
PV	12	N	Negative Programming Voltage

Note: Pins 7 (PEN) and 12 (PV) are only used by the manufacturer for calibration purposes.

# **ABSOLUTE MAXIMUM RATINGS**

Parameter	Symbol	Conditions	Min	Max	Unit
Supply Voltage	$V_{ extsf{DD}}$		-0.3	4	V
Overpressure	Р			4	bar
					abs
Storage Temperature	$T_{Stg}$		-20	+70	°C

**NOTE:** Power supply pins (VDD, GND) should be decoupled with a tantalum (47 $\mu$ F) capacitor to achieve best circuit performance.



#### RECOMMENDED OPERATING CONDITIONS

Parameter	Symbol	Conditions	Min	Тур	Max	Unit
Supply Voltage	$V_{DD}$		2.2	3.0	3.6	V
Supply Current, average (1) during conversion (2) standby (power down)	l <sub>avg</sub> ISC I <sub>ss</sub>	$V_{DD} = 3.0 \text{ V}$		5 1	0.5	μΑ mA μΑ
Operating pressure range	р		300		1100	mbar abs.
Operating temperature range	Ta		-10	+25	+60	~
Conversion time	$T_{conv}$	MCLK=32768Hz			35	ms
External clock signal (3)	MCLK		30000	3276 8	35000	Hz
Delay after conversion (4)	$T_{del}$	MCLK=32768Hz	0.1		100	ms
Duty cycle of MCLK			60/40	50/50	40/60	%
Serial Data Clock	SCLK				500	kHz

#### Notes:

Reliable operation requires protection of the pressure sensor from direct contact with light.

- 1. Under the assumption of one pressure measurement and one temperature measurement every second
- 2. During conversion the sensor will be switched on and off in order to reduce power consumption; the total on time within a conversion is about 2ms
- 3. It is strongly recommended that a crystal oscillator be used because the device is sensitive to clock jitter
- 4. T<sub>del</sub> is the time between the falling edge of DOUT (indicating the end of conversion) and the start of data read-out.

#### **ELECTRICAL CHARACTERISTICS**

# **♦** Digital inputs

(T=-10°C .. 60°C)

Parameter	Symbol	Conditions	Min	Тур	Max	Unit
Input High Voltage	V <sub>IH</sub>	$V_{DD} = 2.23.6V$	80% V <sub>DD</sub>		100% V <sub>DD</sub>	V
Input Low Voltage	V <sub>IL</sub>	V <sub>DD</sub> = 2.23.6 V	0% V <sub>DD</sub>		20% V <sub>DD</sub>	V
Signal Rise Time	t <sub>R</sub>			200		ns
Signal Fall Time	t <sub>f</sub>			200		ns



#### **ELECTRICAL CHARACTERISTICS**

# **◆** Digital outputs

 $(T=-10^{\circ}C...60^{\circ}C, V_{DD} = 2.2V..3.6V)$ 

Parameter	Symbol	Conditions	Min	Тур	Max	Unit
Output High Voltage	$V_{OH}$	I <sub>Source</sub> = 0.6 mA	80%		100%	V
			$V_{DD}$		$V_{ m DD}$	
Output Low Voltage	$V_{OL}$	$I_{Sink} = 0.6 \text{ mA}$	0%		20%	V
			$V_{DD}$		$V_{ m DD}$	
Signal Rise Time	t <sub>r</sub>			200		ns
Signal Fall Time	t <sub>f</sub>			200		ns

#### **◆ AD-converter**

Parameter	Symbol	Conditions	Min	Тур	Max	Unit
Resolution				15		bit
Conversion Time					35	ms
Accuracy					1	LSB

Note: The ADC output range is from 5,000 counts to 37,000 counts, thus requiring a 16 Bit output word.

# PRESSURE OUTPUT CHARACTERISTICS

With the calibration data provided by the MS5534 system (stored in the interface IC) it should be possible to obtain the following characteristics:

Parameter	Symbol	Conditions	Min	Тур	Max	Unit
Resolution			0.1			mbar
Accuracy		p = 7501100mbar	-0.5		+0.5	mbar
Maximum Error over Temperature		T <sub>a</sub> = -10+60 °C	-1		+1	mbar
Long-term Stability		6 month		2.5		mbar
Maximum Error over Supply Voltage		VDD = 2.23.6V	-0.3	0	+ 0.3	mbar/ V

<u>Note</u>: The long-term stability is based on reliability measurements with products comparable to MS5534.



#### TEMPERATURE OUTPUT CHARACTERISTICS

This temperature information is not required for most applications, but it is necessary to allow for temperature compensation of the pressure output. The reference temperature is 20 °C.

Parameter	Symbol	Conditions	Min	Тур	Max	Unit
Resolution			0.005		0.015	δ.
Accuracy		at reference temperature	-0.8		0.8	%
Maximum Error over Supply Voltage		VDD = 2.23.6V	-0.08	0	+ 0.08	%/V

## **FUNCTION**

# ◆ Pressure and temperature measurement

The main function of the MS5534 is to convert the uncompensated analog output voltage from the piezoresistive pressure sensor into a 16 Bit digital value, as well as providing a temperature-dependent voltage as 16 Bit digital value:

measured pressure (16 bit) "D1"measured temperature (16 bit) "D2"

As the output voltage of a pressure sensor is strongly dependent on temperature and shows high tolerances, it is necessary to compensate for these effects. Therefore, 6 sensor-specific coefficients are stored onto the MS5534 after calibration and testing at Micronas' production facilities. These 6 sensor-specific coefficients allow an accurate software compensation in the application; this compensation procedure must be performed externally in a microcontroller. The 6 compensation coefficients are

"
2"
3"
<b>!</b> "
"C5"
3"
232



The principle of the required computation is described hereafter for a four point calibration. It is written as an easy-to-use and easy-to-program routine (only with addition, subtraction and multiplication):

#### **Basic system equations:**

```
1. dT(D2) = D2 - Tref
```

2. OFF (D2) = OFFT1 + TCO \* dT (D2)

3. SENS (D2) = SENST1 + TCS \* dT (D2)

4. PRESS (D1, D2) = (D1 \* SENS (D2)) - OFF (D2)

-> PRESS (D1, D2) is the calibrated pressure value in mbar

For the detailed equations, please refer to paragraph "Calculation of Pressure using Compensation Coefficients C1 to C5".

#### **◆** Measurement principle

For both pressure and temperature measurement the same ADC is used (sigma delta converter):

- for the pressure measurement, the differential output voltage from the pressure sensor is converted
- for the temperature measurement, the sensor bridge resistor is sensed and converted

During both measurements the sensor will only be switched on for a very short time in order to reduce power consumption.

The A/D converter has been optimised to work in the linear range (about 5,000 to 37,000) for all sensor variations in the specified range of temperature and pressure.

#### **♦** Calibration

The user of MS5534 does not need to calibrate the sensor, but it might be useful to understand how the compensation coefficients stored in the module have been calculated. Following is a description of the way in which compensation coefficients are calculated during calibration and testing at Micronas' production facilities.

The module consists of a piezoresistive sensor and a sensor interface IC. The Micronas 1 bar absolute pressure sensor used in this module has tolerances in various parameters such as:

- Offset voltage varies between -8mV/V and +8mV/V
- Sensitivity varies between 24 mV/(V·bar) and 35 mV/(V·bar)
- TCO (temperature coefficient of Offset) varies between -16uV/(V·°C) and +16uV/(V·°C)
- TCS (temperature coefficient of Sensitivity) varies between -1600ppm/°C and -2200ppm/°C

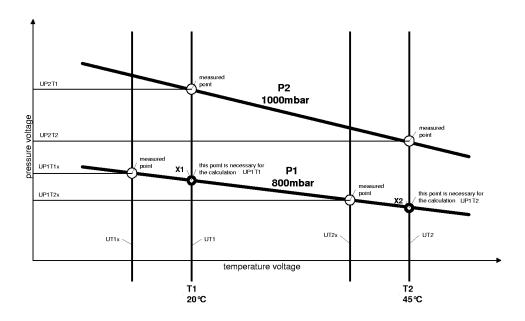


Also there are small tolerances and temperature dependencies in Gain and Offset of the A/D-converter.

The module is calibrated at two temperatures T1 =  $20 \, ^{\circ}$ C and T2 =  $45 \, ^{\circ}$ C, and two pressures P1 =  $800 \, \text{mbar}$  and P2 =  $1000 \, \text{mbar}$ .

Changes in pressure result in a change in temperature and vice versa (due to physics). This means, for example, that after the first calibration at 1000mbar/20 ℃ a change in pressure to 800mbar/20 ℃ will result not only in a pressure change but change the temperature as well, and it will take considerable time to stabilise to 20 ℃ again. Therefore the calibration algorithm had been optimised to accept tolerances in temperature from the reference.

The following graph shows a typical dependency of measured pressure value versus measured temperature for 2 different pressure values.



During calibration testing four pressure values UP2T1, UP2T2, UP1T1x and UP1T2x together with the equivalent temperature values UT1, UT2, UT1x and UT2x are measured. The "x" shall indicate that temperature differs slightly from the reference, caused by a change in pressure.

For the calculation of the sensitivity at T1=20 °C and T2=45 °C two imaginary points UP1T1 and UP1T2 are calculated as follows:

#### Calculation of UP1T1 and UP1T2

 $UP1T1 = (UP1T1x-UP1T2x)*(UT1-UT2x)/(UT1x-UT2x) + UP1T2x \\ UP1T2 = (UP1T1x-UP1T2x)*(UT2-UT2x)/(UT1x-UT2x) + UP1T2x$ 

These imaginary points are used to calculate the sensitivity.



# Calculation of Sensitivity and TC(Sensitivity)

SensT1 = (P2-P1)/(UP2T1-UP1T1) Sensitivity at T1 in mbar/PLSB

SensT2 = (P2-P1)/(UP2T2-UP1T2) Sensitivity at T2 in mbar/PLSB

mbar/PLSB/TLSB

#### Calculation of Offset and TC(Offset)

OffT1 = (UP1T1 - 7168)\*SensT1 - (P1 - 250mbar) Offset at T1 in mbar OffT2 = (UP1T2 - 7168)\*SensT2 - (P1 - 250mbar) Offset at T2 in mbar

TCO = (OffT2-OffT1)/(UT2-UT1) TC(Offset) in mbar/TLSB

#### Example 1:

P1 = 800mbar, P2=1000mbar, T1=20 °C, T2 = 45 °C

UP1T1x=22560 UP2T1=26155 UT1x=25927 (20 ℃+x) **UT1**=25923

(20℃)

UP1T2x=21881 UP2T2=25306 UT2x=29330 ( $45 \text{ }^{\circ}\text{C}+y$ ) **UT2**=29336 ( $45 \text{ }^{\circ}\text{C}$ )

→ UP1T1=(22560-21881)\*(25923-29330)/(25927-29330)+21881=22560.8

→ UP1T2=(22560-21881)\*(29336-29330)/(25927-29330)+21881=21879.8

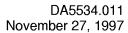
**SensT1** = (1000-800)/(26155-22560.8) = **0.0556452** mbar/PLSB SensT2 = (1000-800)/(25306-21879.8) = 0.0583737 mbar/PLSB **TCS** = (0.0583737 - 0.0556452)/(29336-25923) = **7.994\*10**<sup>-7</sup>mbar/PLSB/TLSB

**TCS** = (0.0583737 - 0.0556452)/(29336-25923) = **7.994\*10**<sup>-7</sup>mbar **OffT1** = (22560.8-7168)\*0.0556452-550 = **306.535** mbar OffT2 = (21879.8-7168)\*0.0583737-550 = 308.782 mbar

**TCO** = (308.782-306.535)/(29336-25923) = **6.584\*10**<sup>-4</sup>mbar/TLSB

The MS5534 has 64 Bit of PROM memory that store the compensation information in the form of coefficients that are calculated out of SensT1, OffT1,TCS,TCO and UT1 as follows:

C1 =  $(SensT1 * 2^{16} - 3072)* 2^3$  15 Bit C2 = OffT1 \* 2<sup>3</sup> 12 Bit C3 = TCS \* 2<sup>29</sup> 10 Bit C4 = TCO\* $2^{17} + 512$  10 Bit C5 =  $(UT1-20224)/ 2^3$  11 Bit C6 =  $10*(T2-T1)/(UT2-UT1)* 2^{10}$  -50 6 Bit





The calculations consist of adding offset and shift operations to obtain a maximum possible accuracy with a minimum of bits required to store the coefficients.



## Example 2 (taking the highlighted values from example1):

C1 = 
$$(0.0556452*2^{16} - 3072)*2^3$$
 = 4598  
C2 =  $306.535*2^3$  = 2452  
C3 =  $7.994*10^{-7}*2^{29}$  = 429  
C4 =  $6.584*10^{-4}*2^{17} + 512$  = 598  
C5 =  $(25923-20224)/2^3$  = 712  
C6 =  $10*(45-20)/(29336-25923)*2^{10}$ -50 = 25

For the exact location of the coefficients inside the PROM, please refer to Figure 2 in Paragraph "Serial Interface".

## ◆ Calculation of pressure using compensation coefficients C1 to C5

First the compensation coefficients C1 to C5 have to be read through the serial interface. This can be done one time after reset of the microcontroller that interfaces to the MS5534.

For the pressure measurement the microcontroller has to read the 15 Bit values for pressure (D1) and temperature (D2) via the serial interface in a loop (i.e. every 1 sec).

For the timing please refer to the next paragraph "Serial Interface".

The following text describes how to calculate the real pressure out of D1,D2 and C1 to C5.

# 1. Step -> Calculation of difference between actual temperature and calibration temperature T1

$$dUT = D2 - UT1$$
 where  $UT1 = 8*C5 + 20224$ 

Remark: dUT gives some indication about the temperature. It is around zero at T=20 ℃; the value is a positive number for higher temperatures and a negative number for lower temperatures.

#### 2. Step -> Calculation of Offset at actual temperature

OFF = 
$$(C2 + (C4-512) * dUT/ 2^{14})*4$$

#### 3. Step -> Calculation of Sensitivity at actual temperature

SENS = 
$$C1 + C3*dUT/2^{10} + 24576$$

#### 4. Step -> Calculation of the temperature compensated real pressure in mbar

$$X = SENS * (D1-7168) / 2^{14} - OFF$$

$$P = X * 100/2^5 + 250 * 100$$
 Pressure in mbar with 0.01mbar resolution  $P = X * 10/2^5 + 250 * 10$  Pressure in mbar with 0.1mbar resolution



#### ◆ Calculation of the actual temperature using compensation coefficient C6

 $TEMP = 200 + dUT * (C6+50)/2^{10}$ 

# **Example 3 (taking values from Example 1 and 2)**

```
D1 = 25000, D2 = 26848, C1 = 4598, C2 = 2452, C3 = 429, C4 = 598, C5 = 712, C6 = 25
dUT
            = 26848 - (8*712 + 20224)
                                                   = 928
            = (2452 + (598-512) * 928 / 2^{14}) * 4 = 9827
= 4598 + 429*928/2^{10} + 24576 = 29563
OFF
SENS
                                                  = 29563
            = 29563*(25000 - 7168)/2^{14} - 9827 = 22349
Χ
            = 22349*100/2^5 + 25000
Ρ
                                                   = 94841
                                                                  = 948.41 mbar
            = 200 + 928*(25+50)/2^{10}
TEMP
                                                   = 268
                                                                  = 26.8 °C
```

Remark: all calculations can be handled with 16 Bit variables. Intermediate results may be 32 Bit long where only the upper 16 Bit are significant.



#### ◆ Serial interface

The MS5534 communicates with microprocessors and other digital systems via a 3-wire serial interface as shown in Figure 1. The SCLK (Serial Clock) signal initiates the data transfer and synchronises the data transfer with each bit being transmitted on the falling edge of SCLK and each bit being received on the rising edge of SCLK. This signal will be generated by the microprocessor's system. The signal DOUT (Data Out) indicates the data transmission status and the conversion status. The digital data that is provided on the DOUT pin is the conversion result or the software calibration data. The selection of the output data is dependent on the data format of the DIN (Data Input) signal.

Following is a list of the possible output data instructions:

- Conversion start for pressure measurement and ADC-data-out "D1" (Figure1a)
- Conversion start for temperature measurement and ADC-data-out "D2"

(Figure1b)

- Calibration data read-out sequence for word 1 and word 3
   Calibration data read-out sequence for word 2 and word 4
   (Figure 1c)
   (Figure 1d)
- RESET sequence (Figure 1e)

Every communication starts with a data sequence on the DIN pin to adjust the MS5534 in the chosen state. Figure1 shows in detail the hole sequences for this communication. The MS5534 works without a "Chip select" signal, and this is the reason why the DIN signal begins with a start-sequence (3 Bit high) and ends with a stop-sequence (3 Bit low). Between both sequences are the adjustment bits (4 Bit for AD-converter start and measurement choice/ 6 Bit for calibration data read out and the selection address of the four 16 Bit blocks). It is possible to interrupt the data read-out sequence with a hold of the SCLK signal after the falling edge. The maximum break time between the falling edge of DOUT for the end of conversion indication and the start of data read-out is 100ms.

The RESET-sequence sets the internal control logic into the start-up state. Before start of the conversion, it is recommended to send this RESET-sequence.

Figure 2 shows the arrangement of the software calibration data in four data transmit blocks word 1 to word 4.

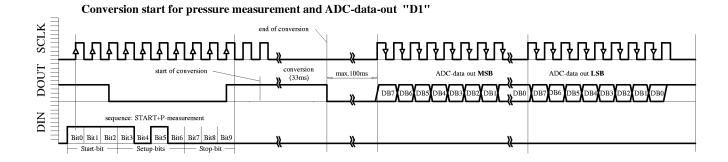


Figure1a



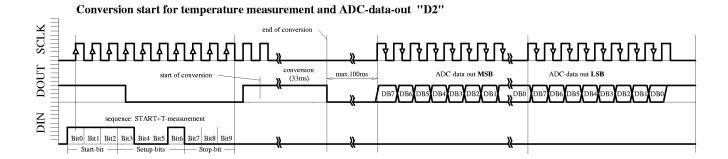


Figure1b

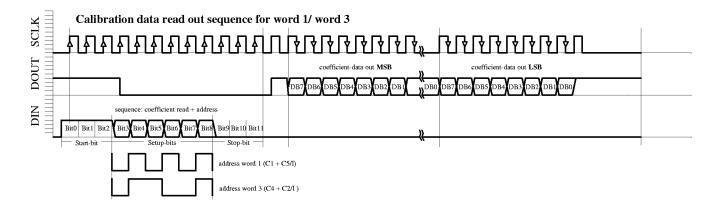


Figure1c

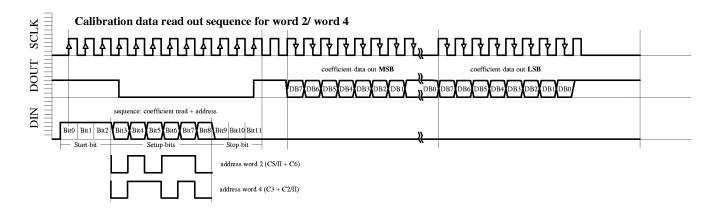


Figure1d



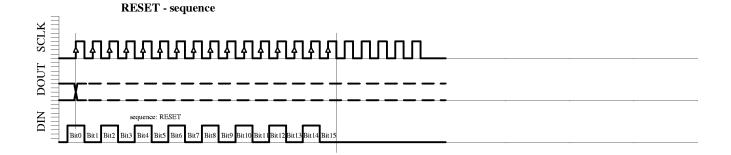


Figure1e

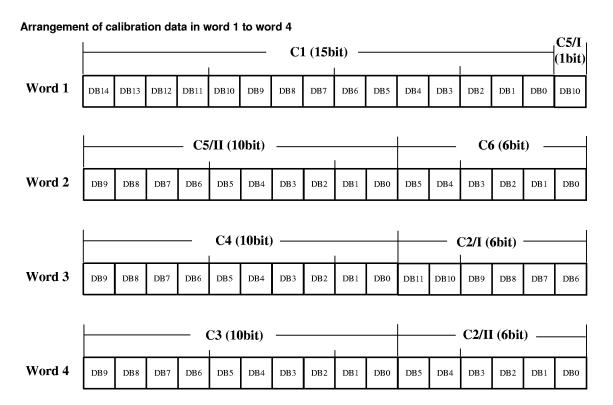


Figure 2



#### **APPLICATION INFORMATION**

#### GENERAL

The idea for this combination of a sensor with a direct adapted integrated circuit is to save other external components and to achieve very low power consumption. The main application field for this system includes portable devices with battery supply, but its high accuracy and resolution make it also suited for industrial and automotive applications. The possibility to compensate the sensor with a software allows the user to adapt it to his particular application. Communication between the MS5534 and the widely available microcontrollers is realised over an easy-to-use 3-wire serial interface. Customers may select which microcontroller system to be used, and there are no specific standard interface cells required, which may be of interest for specially designed 4 Bit-microcontroller applications.

The package outline of the module allows the use of a flexible PCB to connect it. This can be important for applications in watches and other special devices, and will also reduce mechanical stress on the device.

Storage of calibration data in the device will be done at module final test. Automatic test equipment is used to perform tests under pressure and temperature and to calculate individual coefficients for every sensor, which allows a highly accurate compensation.

Further, because the calibration data are stored in the MS5534, the user can save additional external memory in his microcontroller system (i.e. EEPROM).

#### ALTIMETER SYSTEM USING MS5534

MS5534 is a circuit that can be used in connection with a microcontroller in mobile altimeter applications. It is designed for low-voltage systems with a supply voltage of 3V, particularly in battery applications. The MS5534 is optimised for low current consumption as the AD-converter clock (MCLK) can use the 32.768kHz frequency of a standard watch crystal, which is supplied in most portable watch systems.

For applications in altimeter systems Micronas can deliver a simple formula based on a linear interpolation, where the number of interpolation points influences the accuracy of the formula.

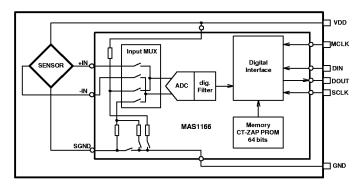


Figure 3: Block diagram of MS5534

Advantages of MS5534:

- easy layout (one side contact for flexible PCB)
- better reliability
- less test time
- universal to use, less development time
- high compensation accuracy because of individual sensor test
- lower price as a solution with sensor, amplifier,
- AD-converter and external parts
- easy-to-use 3 wire serial interface



# • APPLICATION EXAMPLES

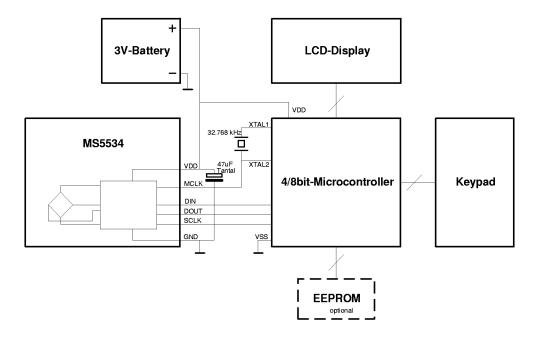
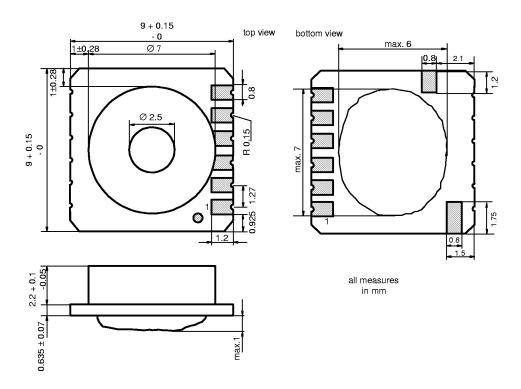


Figure 4: Demonstration of MS5534 in a mobile altimeter

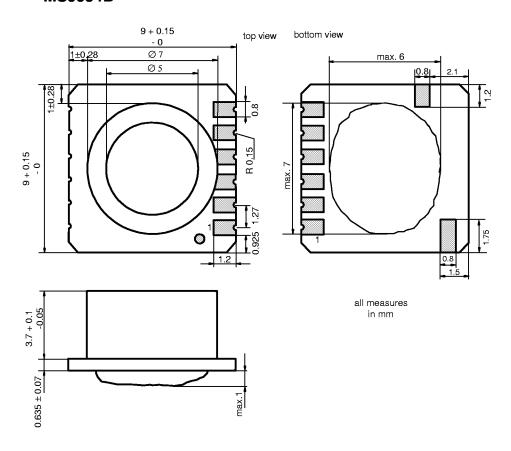


# **DEVICE PACKAGE OUTLINES**

# MS5534A



# MS5534B





#### **ORDERING INFORMATION**

Product Code	Product	Package	Comments
MS5534A	Barometer Module with plastic protection cap	SMD hybrid	standard version
MS5534B	Barometer Module with metal protection cap	SMD hybrid	

LOCAL DISTRIBUTOR					

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