

ICS843034

 $\mathsf{FEMTO}\mathsf{C}\mathsf{LOCKS}^\mathsf{TM}$

Multi-Rate 3.3V, 2.5V LVPECL Frequency Synthesizer

GENERAL DESCRIPTION



The ICS843034 is a general purpose, low phase noise LVPECL synthesizer which can generate frequencies for a wide variety of applications. The ICS843034 has a 4:1 input Multiplexer from which the following inputs can be selected: 1 differential input, 1 single-ended input, or two crystal

oscillators, thus making the device ideal for frequency translation or frequency generation. Each differential LVPECL output pair has an output divider which can be independently set so that two different frequencies can be generated. Additionally, each LVPECL output pair has a dedicated power supply pin so the outputs can run at 3.3V or 2.5V. The ICS843034 also supplies a buffered copy of the reference clock or crystal frequency on the single-ended REF_CLK pin which can be enabled or disabled (disabled by default). The output frequency can be programmed using either a serial or parallel programming interface.

The phase jitter of the ICS843034 is less than 1ps rms, making it suitable for use in Fibre Channel, SONET, and Ethernet applications.

Example applications include systems which must support both FEC and non FEC rates. In 10Gb Fibre Channel, for example, you can use a 25.5MHz crystal to generate a 159.375MHz reference clock, and then switch to a 20.544MHz crystal to generate 164.355MHz for 66/64 FEC. Other applications could include supporting both Ethernet frequencies and SONET frequencies in an application. When Ethernet frequencies are needed, a 25MHz crystal can be used and when SONET frequencies are needed, the input MUX can be switched to select a 38.88MHz Crystal.

FEATURES

- Dual differential 3.3V LVPECL outputs which can be set independently for either 3.3V or 2.5V
- 4:1 Input Mux:
 - 1 differential input
 - 1 single-ended input
 - 2 crystal oscillator interfaces
- CLK, nCLK pair can accept the following differential input levels: LVPECL, LVDS, LVHSTL, HCSL, SSTL
- TEST_CLK accepts LVCMOS or LVTTL input levels
- Output frequency range: 35MHz to 750MHz
- Crystal input frequency range: 12MHz to 40MHz
- VCO range: 560MHz to 750MHz
- Parallel or serial interface for programming feedback divider and output dividers
- RMS phase jitter at 333.33MHz, using a 22.222MHz crystal (12kHz to 20MHz): 0.80ps (typical)
- · Supply voltage modes:

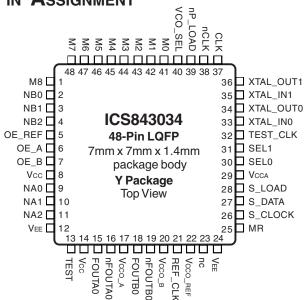
LVPECL outputs (core/outputs): 3.3V/3.3V

3.3V/2.5V

REF_CLK output (core/outputs): 3.3V/3.3V

- 0°C to 70°C ambient operating temperature
- Industrial temperature available upon request

PIN ASSIGNMENT



The Preliminary Information presented herein represents a product in prototyping or pre-production. The noted characteristics are based on initial product characterization. Integrated Circuit Systems, Incorporated (ICS) reserves the right to change any circuitry or specifications without notice.

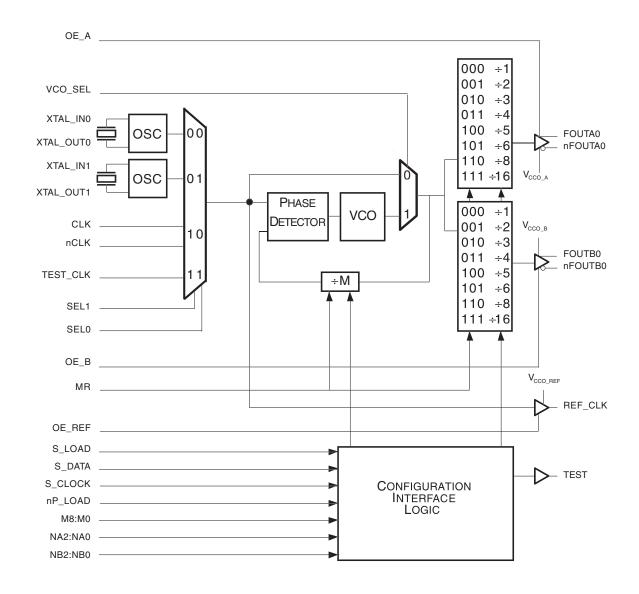


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Multi-Rate 3.3V, 2.5V LVPECL Frequency Synthesizer

BLOCK DIAGRAM





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Multi-Rate 3.3V, 2.5V LVPECL Frequency Synthesizer

FUNCTIONAL DESCRIPTION

NOTE: The functional description that follows describes operation using a 25MHz crystal. Valid PLL loop divider values for different crystal or input frequencies are defined in the Input Frequency Characteristics, Table 5, NOTE 1.

The ICS843034 features a fully integrated PLL and therefore requires no external components for setting the loop bandwidth. A fundamental crystal is used as the input to the onchip oscillator. The output of the oscillator is fed into the phase detector. A 25MHz crystal provides a 25MHz phase detector reference frequency. The VCO of the PLL operates over a range of 560MHz to 750MHz. The output of the M divider is also applied to the phase detector.

The phase detector and the M divider force the VCO output frequency to be M times the reference frequency by adjusting the VCO control voltage. Note that for some values of M (either too high or too low), the PLL will not achieve lock. The output of the VCO is scaled by a divider prior to being sent to each of the LVPECL output buffers. The divider provides a 50% output duty cycle.

The ICS843034 supports either serial or parallel programming modes to program the M feedback divider and N output divider. Figure 1 shows the timing diagram for each mode. In parallel mode, the nP_LOAD input is initially LOW. The data on the M, NA, and NB inputs are passed directly to the M divider and both N output dividers. On the LOW-to-HIGH transition of the nP_LOAD input, the data is latched and the M and N dividers remain loaded until the next LOW transition on nP_LOAD or until a serial event occurs. As a result, the M and Nx bits can be hardwired to set the M divider and Nx output divider to a specific default state that will automatically occur during power-up.

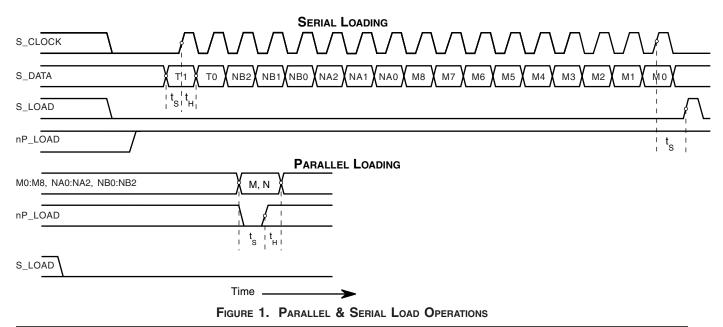
The TEST output is LOW when operating in the parallel input mode. The relationship between the VCO frequency, the crystal frequency and the M divider is defined as follows:

fVCO = fxtal x M

The M value and the required values of M0 through M8 are shown in Table 3B to program the VCO Frequency Function Table. Valid M values for which the PLL will achieve lock for a 25MHz reference are defined as $23 \le M \le 30$. The frequency out is defined as follows: FOUT = $\underline{\text{fVCO}}$ = fxtal x M

Serial operation occurs when nP_LOAD is HIGH and S_LOAD is LOW. The shift register is loaded by sampling the S_DATA bits with the rising edge of S_CLOCK. The contents of the shift register are loaded into the M divider and Nx output divider when S_LOAD transitions from LOW-to-HIGH. The M divide and Nx output divide values are latched on the HIGH-to-LOW transition of S_LOAD. If S_LOAD is held HIGH, data at the S_DATA input is passed directly to the M divider and Nx output divider on each rising edge of S_CLOCK. The serial mode can be used to program the M and Nx bits and test bits T1 and T0. The internal registers T0 and T1 determine the state of the TEST output as follows:

<u>T1</u>	<u>T0</u>	TEST Output
0	0	LOW
0	1	S_Data, Shift Register Output
1	0	Output of M divider
1	1	FOUTA0 same frequency





ICS843034

 $\mathsf{FEMTOCLOCKS^{\mathsf{TM}}}$

Multi-Rate 3.3V, 2.5V LVPECL Frequency Synthesizer

TABLE 1. PIN DESCRIPTIONS

Number	Name	Ty	уре	Description	
1, 41, 42, 43, 44, 45, 47, 48	M8, M0, M1, M2, M3, M4, M6, M7	Input	Pulldown	M divider input. Data latched on LOW-to-HIGH transition of nP_LOAD input. LVCMOS/LVTTL interface levels.	
2, 3	NB0, NB1	Input	Pullup	Determines output divider value as defined in Table 3C,	
4	NB2	Input	Pulldown	Function Table. LVCMOS/LVTTL interface levels.	
5	OE_REF	Input	Pulldown	Output enable. Controls enabling and disabling of REF_CLK output. LVCMOS/LVTTL interface levels.	
6	OE_A	Input	Pullup	Output enable. Controls enabling and disabling of FOUTA0, nFOUTA0 outputs. LVCMOS/LVTTL interface levels.	
7	OE_B	Input	Pullup	Output enable. Controls enabling and disabling of FOUTB0, nFOUTB0 outputs. LVCMOS/LVTTL interface levels.	
8, 14	V _{cc}	Power		Core supply pins.	
9, 10	NA0, NA1	Input	Pullup	Determines output divider value as defined in Table 3C,	
11	NA2	Input	Pulldown		
12, 24	V_{EE}	Power		Negative supply pins.	
13	TEST	Output		Test output which is ACTIVE in the serial mode of operation. Output driven LOW in parallel mode. LVCMOS/LVTTL interface levels.	
15, 16	FOUTA0, nFOUTA0	Output		Differential output for the synthesizer. LVPECL interface levels.	
17	V _{CCO_A}	Power		Output supply pin for FOUTA0, nFOUTA0.	
18, 19	FOUTB0, nFOUTB0	Output		Differential output for the synthesizer. LVPECL interface levels.	
20	V _{CCO_B}	Power		Output supply pin for FOUTB0, nFOUTB0.	
21	REF_CLK	Output		Reference clock output. LVCMOS/LVTTL interface levels.	
22	$V_{\text{CCO_REF}}$	Power		Output supply pin for REF_CLK.	
23	nc	Unused		No connect.	
25	MR	Input	Pulldown	Active High Master Reset. When logic HIGH, forces the internal dividers are reset causing the true outputs FOUTx to go low and the inverted outputs nFOUTx to go high. When logic LOW, the internal dividers and the outputs are enabled. Assertion of MR does not affect loaded M, N, and T values. LVCMOS/LVTTL interface levels.	
26	S_CLOCK	Input	Pulldown	Clocks in serial data present at S_DATA input into the shift register on the rising edge of S_CLOCK. LVCMOS/LVTTL interface levels.	
27	S_DATA	Input	Pulldown	Shift register serial input. Data sampled on the rising edge of S_CLOCK. LVCMOS/LVTTL interface levels.	
28	S_LOAD	Input	Pulldown	Controls transition of data from shift register into the dividers. LVCMOS/LVTTL interface levels.	
29	V _{CCA}	Power		Analog supply pin.	
30, 31	SEL0, SEL1	Input	Pulldown	Clock select inputs. LVCMOS/LVTTL interface levels.	
32	TEST_CLK	Input	Pulldown	Test clock input. LVCMOS/LVTTL interface levels.	
33, 34	XTAL_IN0, XTAL_OUT0	Input		Crystal oscillator interface. XTAL_IN0 is the input, XTAL_OUT0 is the output.	
35, 36	XTAL_IN1, XTAL_OUT1	Input		Crystal oscillator interface. XTAL_IN1 is the input, XTAL_OUT1 is the output.	

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ICS843034

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Number	Name	1	Гуре	Description
37	CLK	Input	Pulldown	Non-inverting differential clock input.
38	nCLK	Input	Pullup/ Pulldown	Inverting differential clock input.V _{cc} /2 default when left floating.
39	nP_LOAD	Input	Pulldown	Parallel load input. Determines when data present at M8:M0 is loaded into M divider, and when data present at NA2:NA0 and NB2:NB0 is loaded into the N output dividers. LVCMOS/LVTTL interface levels.
40	VCO_SEL	Input	Pullup	Determines whether synthesizer is in PLL or bypass mode. LVCMOS/LVTTL interface levels.
46	M5	Input	Pullup	M divider inputs. Data latched on LOW-to-HIGH transition of nP_LOAD input. LVCMOS/LVTTL interface levels.

NOTE: Pullup and Pulldown refer to internal input resistors. See Table 2, Pin Characteristics, for typical values.

TABLE 2. PIN CHARACTERISTICS

Symbol	Parameter		Test Conditions	Minimum	Typical	Maximum	Units
C _{IN}	Input Capacitance				4		рF
C _{PD}	Power Dissipation Capacitance	REF_CLK	V_{CC} , V_{CCA} , $V_{CCO_REF} = 3.465V$		TBD		pF
R _{PULLUP}	Input Pullup Resisto	r			51		kΩ
R _{PULLDOWN}	Input Pulldown Resistor				51		kΩ
R _{out}	Output Impedance	REF_CLK		5	7	12	Ω



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Multi-Rate 3.3V, 2.5V LVPECL Frequency Synthesizer

TABLE 3A. PARALLEL AND SERIAL MODE FUNCTION TABLE

			In	puts			Conditions
MR	nP_LOAD	М	N	S_LOAD	S_CLOCK	S_DATA	Oditations
Н	Х	Х	Х	Х	Х	Х	Reset. Forces outputs LOW.
L	L	Data	Data	Х	Х	Х	Data on M and N inputs passed directly to the M divider and N output divider. TEST output forced LOW.
L	1	Data	Data	L	×	Х	Data is latched into input registers and remains loaded until next LOW transition or until a serial event occurs.
L	Н	Х	Х	L	↑	Data	Serial input mode. Shift register is loaded with data on S_DATA on each rising edge of S_CLOCK.
L	Н	Х	Х	1	L	Data	Contents of the shift register are passed to the M divider and N output divider.
L	Н	Х	Х	\	L	Data	M divider and N output divider values are latched.
L	Н	Х	Х	L	Х	Х	Parallel or serial input do not affect shift registers.
L	Н	Х	Х	Н	1	Data	S_DATA passed directly to M divider as it is clocked.

NOTE: L = LOW H = HIGH

X = Don't care

 \uparrow = Rising edge transition \downarrow = Falling edge transition

TABLE 3B. PROGRAMMABLE VCO FREQUENCY FUNCTION TABLE

VCO Frequency	M Divide	256	128	64	32	16	8	4	2	1
(MHz)	M Divide	M8	M7	M6	M5	M4	М3	M2	M1	МО
575	23	0	0	0	0	1	0	1	1	1
•	•	•	•	•	•	•	•	•	•	•
700	28	0	0	0	0	1	1	1	0	0
•	•	•	•	•	•	•	•	•	•	•
750	30	0	0	0	0	1	1	1	1	0

NOTE 1: These M divide values and the resulting frequencies correspond to crystal or TEST_CLK input frequency of 25MHz.

TABLE 3C. PROGRAMMABLE OUTPUT DIVIDER FUNCTION TABLE

	Inputs		N Divider Value	Output Freq	uency (MHz)
*NX2	*NX1	*NX0	N Divider value	Minimum	Maximum
0	0	0	1	560	750
0	0	1	2	280	375
0	1	0	3	186.66	250
0	1	1	4	140	187.5
1	0	0	5	112	150
1	0	1	6	93.33	125
1	1	0	8	70	93.75
1	1	1	16	35	46.875

*NOTE: X denotes Bank A or Bank B



ICS843034

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Multi-Rate 3.3V, 2.5V LVPECL Frequency Synthesizer

ABSOLUTE MAXIMUM RATINGS

Supply Voltage, V_{CC} 4.6V

Inputs, V_{l} -0.5V to V_{CC} + 0.5V

Outputs, $V_{\rm o}$ (LVCMOS) -0.5V to $V_{\rm cco}$ + 0.5V

Outputs, I_{\odot} (LVPECL)

Continuous Current 50mA Surge Current 100mA

Package Thermal Impedance, θ_{JA} 47.9°C/W (0 Ifpm) Storage Temperature, T_{STG} -65°C to 150°C NOTE: Stresses beyond those listed under Absolute Maximum Ratings may cause permanent damage to the device. These ratings are stress specifications only. Functional operation of product at these conditions or any conditions beyond those listed in the *DC Characteristics* or *AC Characteristics* is not implied. Exposure to absolute maximum rating conditions for extended periods may affect product reliability.

 $\textbf{TABLE 4A. Power Supply DC Characteristics, } V_{\text{CC}} = V_{\text{CCA}} = 3.3 \text{V} \pm 5\%, V_{\text{CCO_A}} = V_{\text{CCO_B}} = 3.3 \text{V} \pm 5\% \text{ or } 2.5 \text{V} \pm 5\%, T_{\text{A}} = 0^{\circ}\text{C to } 70^{\circ}\text{C}$

Symbol	Parameter	Test Conditions	Minimum	Typical	Maximum	Units
V _{cc}	Core Supply Voltage		3.135	3.3	3.465	V
V _{CCA}	Analog Supply Voltage		3.135	3.3	3.465	V
V _{CCO_A} ,	Output Supply Voltage		3.135	3.3	3.465	V
V _{CCO_B}			2.375	2.5	2.625	V
V _{CCO_REF}	Output Supply	REF_CLK	3.135	3.3	3.465	V
I _{EE}	Power Supply Current			185		mA
I _{CCA}	Analog Supply Current			20		mA



ICS843034

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MULTI-RATE 3.3V, 2.5V LVPECL FREQUENCY SYNTHESIZER

 $\textbf{TABLE 4B. LVCMOS/LVTTL DC Characteristics, } V_{\text{CC}} = V_{\text{CCA}} = 3.3 \text{V} \pm 5\%, V_{\text{CCO_A}} = V_{\text{CCO_REF}} = 3.3 \text{V} \pm 5\%, TA = 0^{\circ}\text{C to } 70^{\circ}\text{C}$

Symbol	Parameter		Test Conditions	Minimum	Typical	Maximum	Units
V _{IH}	Input High Vol	tage		2		V _{CC} + 0.3	V
V _{IM}	Input Mid Volta	age		V _{cc} /2 - 0.2V		$V_{cc}/2 + 0.2V$	V
V _{IL}	Input Low Volt	age		-0.3		0.8	V
I _{IH}	Input High Current	TEST_CLK, MR, SEL[1:0], OE_REF, S_CLOCK, S_DATA, S_LOAD, nP_LOAD, Nx2, M1:M4, M6:M8	$V_{\rm CC} = V_{\rm IN} = 3.465V$			150	μА
		Nx0, Nx1, M5, OE_A, OE_B, VCO_SEL	$V_{CC} = V_{IN} = 3.465V$			5	μΑ
I	Input Low Current	TEST_CLK, MR, SEL[1:0], OE_REF, S_CLOCK, S_DATA, S_LOAD, nP_LOAD, Nx2, M1:M4, M6:M8	$V_{CC} = 3.465V,$ $V_{IN} = 0V$	-5			μА
	Low Current	Nx0, Nx1, M5, OE_A, OE_B, VCO_SEL	$V_{CC} = 3.465V,$ $V_{IN} = 0V$	-150		V _{cc} + 0.3 V _{cc} /2 + 0.2V 0.8 150 4 0.4	μΑ
.,	Output	TEST; NOTE 1	V 0.0V.F0/	2.6			٧
V _{OH}	High Voltage	REF_CLK	$V_{CCO_REF} = 3.3V \pm 5\%$	V _{CCO_REF} - 0.3V		0.4	V
V _{OL}	Output Low Voltage	TEST; NOTE 1	$V_{CCO_REF} = 3.3V \pm 5\%$	_		0.5	V

NOTE 1: Output terminated with 50Ω to $V_{CCO_REF}/2$.



ICS843034

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Multi-Rate 3.3V, 2.5V LVPECL Frequency Synthesizer

 $\textbf{TABLE 4C. Differential DC Characteristics, } V_{\text{CC}} = V_{\text{CCA}} = 3.3 \text{V} \pm 5\%, V_{\text{CCO_A}} = V_{\text{CCO_B}} = 3.3 \text{V} \pm 5\% \text{ or } 2.5 \text{V} \pm 5\%, T_{\text{A}} = 0^{\circ}\text{C to } 70^{\circ}\text{C}$

Symbol	Parameter		Test Conditions	Minimum	Typical	Maximum	Units
	Input High Current	nCLK	$V_{IN} = V_{CC} = 3.465V$			150	μΑ
¹ _{IH}		CLK	$V_{IN} = V_{CC} = 3.465V$			150	μΑ
	Input Low Current	nCLK	$V_{IN} = 0V, V_{CC} = 3.465V$	-150			μΑ
' _{IL}		CLK	$V_{IN} = 0V, V_{CC} = 3.465V$	-5			μΑ
V _{PP}	Peak-to-Peak Input Voltage			0.15		1.3	V
V _{CMR}	Common Mode Inpu	ut Voltage; NOTE 1, 2		V _{EE} + 0.5		V _{cc} - 0.85	V

NOTE 1: For single ended applications, the maximum input voltage for CLK, nCLK is $V_{\rm cc}$ + 0.3V.

NOTE 2: Common mode voltage is defined as $V_{_{\rm IH}}$.

 $\textbf{TABLE 4D. LVPECL DC Characteristics, } V_{\text{CC}} = V_{\text{CCA}} = 3.3 \text{V} \pm 5\%, V_{\text{CCO_A}} = V_{\text{CCO_B}} = 3.3 \text{V} \pm 5\% \text{ or } 2.5 \text{V} \pm 5\%, T_{\text{A}} = 0^{\circ}\text{C to } 70^{\circ}\text{C}$

Symbol	Parameter	Test Conditions	Minimum	Typical	Maximum	Units
V _{OH}	Output High Voltage; NOTE 1		V _{cco} - 1.4		V _{cco} - 0.9	V
V _{OL}	Output Low Voltage; NOTE 1		V _{cco} - 2.0		V _{cco} - 1.7	V
V _{SWING}	Peak-to-Peak Output Voltage Swing		0.6		1.0	V

NOTE 1: Outputs terminated with 50 Ω to $V_{\text{CCO A.}}$ $V_{\text{CCO B}}$ - 2V.

Table 5. Input Frequency Characteristics, $V_{CC} = V_{CCA} = 3.3V \pm 5\%$, $V_{CCO_A} = V_{CCO_B} = 3.3V \pm 5\%$ or $2.5V \pm 5\%$, $T_A = 0^{\circ}C$ to $70^{\circ}C$

Symbol	Parameter		Test Conditions	Minimum	Typical	Maximum	Units
	Input Frequency	XTAL_IN0/XTAL_OUT0, XTAL_IN1/XTAL_OUT1		12		40	MHz
f _{IN}		CLK/nCLK, TEST_CLK		12		TBD	MHz
114		S_CLOCK				50	MHz
	. " Input Rise/Fall	TEST_CLK			TBD		ns
t _R /t _F	Time	S_LOAD, S_DATA, S_CLOCK			TBD		ns

NOTE: For the input crystal, CLK/nCLK and TEST_CLK frequency range, the M value must be set for the VCO to operate within the 560MHz to 750MHz range. Using the minimum input frequency of 12MHz, valid values of M are $47 \le M \le 62$. Using the maximum frequency of 40MHz, valid values of M are $14 \le M \le 18$.

TABLE 6. CRYSTAL CHARACTERISTICS

Parameter	Test Conditions	Minimum	Typical	Maximum	Units
Mode of Oscillation		Fundamental			
Frequency		12		40	MHz
Equivalent Series Resistance (ESR)				50	Ω
Shunt Capacitance				7	pF
Drive Level				1	mW



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Multi-Rate 3.3V, 2.5V LVPECL Frequency Synthesizer

Table 7A. AC Characteristics, $V_{CC} = V_{CCA} = V_{CCO~A} = V_{CCO~B} = 3.3V \pm 5\%$, Ta = 0°C to 70°C

Symbol	Parameter		Test Conditions	Minimum	Typical	Maximum	Units
F _{out}	Output Frequency			35		750	MHz
tjit(Ø)	Phase Jitter, RMS (Random); NOTE 1, 2		333.33MHz, Integration Range: 12kHz - 20MHz		0.80		ps
tjit(cc)	Cycle-to-Cycle	Jitter; NOTE 3, 4			TBD		ps
tsk(o)	Output Skew; NOTE 2, 4, 5		Measured @ the same Output Frequency		50		ps
+ /+	Output	LVPECL Outputs	20% to 80%	200		700	20
t _R /t _F	Rise/Fall Time	REF_CLK	20% 10 00%	200		700	ps
		M, N to nP_LOAD		5			ns
t _s	Setup Time	S_DATA to S_CLOCK		5			ns
		S_CLOCK to S_LOAD		5			ns
		M, N to nP_LOAD		5			ns
t _H	Hold Time	S_DATA to S_CLOCK		5			ns
		S_CLOCK to S_LOAD		5			ns
odc	Output Duty Cycle				50		%
t _{LOCK}	PLL Lock Time					1	ms

See Parameter Measurement Information section.

NOTE 1: Please refer to the Phase Noise Plot.

NOTE 2: Characterized with REF_CLK output disabled.

NOTE 3: Jitter perforance using XTAL inputs.

NOTE 4: This parameter is defined in accordance with JEDEC Standard 65.

NOTE 5: Defined as skew between outputs at the same supply voltage and with equal load conditions.

Measured at the output differential cross points.

 $\textbf{Table 7B. AC Characteristics, } V_{\text{CC}} = V_{\text{CCA}} = 3.3 \text{V} \pm 5\%, V_{\text{CCO_A}} = V_{\text{CCO_B}} = 2.5 \text{V} \pm 5\%, T_{\text{A}} = 0^{\circ}\text{C} \text{ to } 70^{\circ}\text{C}$

Symbol	Parameter		Test Conditions	Minimum	Typical	Maximum	Units
F _{OUT}	Output Frequer	ісу		35		750	MHz
<i>t</i> jit(Ø)	Phase Jitter, RMS (Random); NOTE 1, 2		333.33MHz, Integration Range: 12kHz - 20MHz		TBD		ps
tjit(cc)	Cycle-to-Cycle	Jitter; NOTE 3, 4			TBD		ps
tsk(o)	sk(o) Output Skew; NOTE 2, 4, 5		Measured @ the same Output Frequency		50		ps
+ /+	Output	LVPECL Outputs	20% to 80%	200		700	no
t _R /t _F	Rise/Fall Time	REF_CLK	20% 10 80%	200		700	ps
		M, N to nP_LOAD		5			ns
t _s	Setup Time	S_DATA to S_CLOCK		5			ns
		S_CLOCK to S_LOAD		5			ns
		M, N to nP_LOAD		5			ns
t _H	Hold Time	S_DATA to S_CLOCK		5			ns
		S_CLOCK to S_LOAD		5			ns
odc	Output Duty Cycle				50		%
t _{LOCK}	PLL Lock Time					1	ms

For notes, see Table 7A above.



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Multi-Rate 3.3V, 2.5V LVPECL Frequency Synthesizer

Table 7C. AC Characteristics, $V_{CC} = V_{CCA} = 3.3V \pm 5\%$, $V_{CCO_A} = 3.3V \pm 5\%$, $V_{CCO_B} = 2.5V \pm 5\%$, $V_{A} = 0^{\circ}$ C to 70°C or $V_{CC} = V_{CCA} = 3.3V \pm 5\%$, $V_{CCO_A} = 2.5V \pm 5\%$, $V_{CCO_B} = 3.3V \pm 5\%$

Symbol	Parameter		Test Conditions	Minimum	Typical	Maximum	Units
F _{OUT}	Output Frequer	су		35		750	MHz
<i>t</i> jit(∅)	Phase Jitter, RMS (Random); NOTE 1, 2		333.33MHz, Integration Range: 12kHz - 20MHz		TBD		ps
tjit(cc)	Cycle-to-Cycle	Jitter; NOTE 3, 4			TBD		ps
tsk(o)	Output Skew; NOTE 2, 4, 5		Measured @ the same Output Frequency		50		ps
+ /+		LVPECL Outputs	20% to 80%	200		700	ps
t_R/t_F		REF_CLK	20% 10 00%				
		M, N to nP_LOAD		5			ns
t _s	Setup Time	S_DATA to S_CLOCK		5			ns
		S_CLOCK to S_LOAD		5			ns
		M, N to nP_LOAD		5			ns
t _H	Hold Time	S_DATA to S_CLOCK		5			ns
		S_CLOCK to S_LOAD		5			ns
odc	Output Duty Cycle				50		%
t _{LOCK}	PLL Lock Time					1	ms

See Parameter Measurement Information section.

NOTE 1: Please refer to the Phase Noise Plot.

NOTE 2: Characterized with REF_CLK output disabled.

NOTE 3: Jitter perforance using XTAL inputs.

NOTE 4: This parameter is defined in accordance with JEDEC Standard 65.

NOTE 5: Defined as skew between outputs at the same supply voltage and with equal load conditions.

Measured at the output differential cross points.



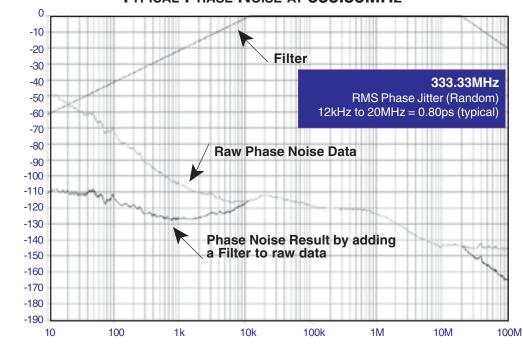
Noise Power dBc

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Typical Phase Noise at 333.33MHz



OFFSET FREQUENCY (Hz)

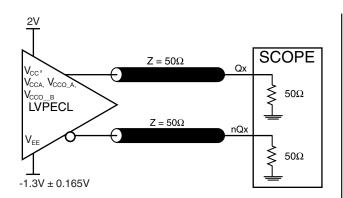


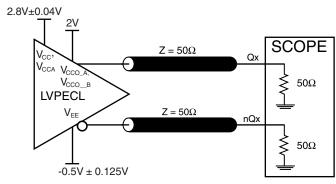
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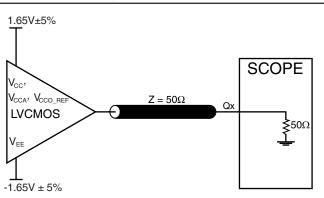
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PARAMETER MEASUREMENT INFORMATION

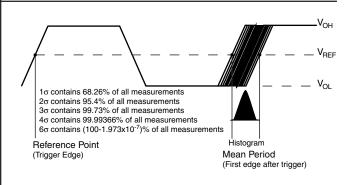




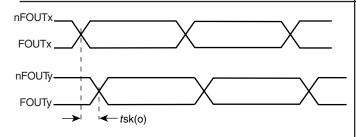
3.3V Core/3.3V OUTPUT LOAD AC TEST CIRCUIT FOUTAO/nFOUTAO, FOUTBO/nFOUTB0



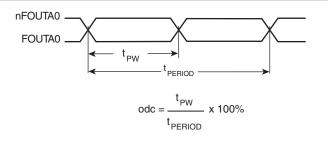
3.3V Core/2.5V OUTPUT LOAD AC TEST CIRCUIT FOUTA0/nFOUTA0, FOUTB0/nFOUTB0



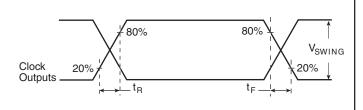
3.3VCore/3.3V REF_CLK OUTPUT LOAD AC TEST CIRCUIT



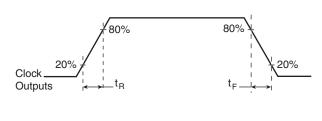
PERIOD JITTER



OUTPUT SKEW



OUTPUT DUTY CYCLE/OUTPUT PULSE WIDTH/PERIOD



LVPECL OUTPUT RISE/FALL TIME

LVCMOS OUTPUT RISE/FALL TIME

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Multi-Rate 3.3V, 2.5V LVPECL Frequency Synthesizer

APPLICATION INFORMATION

Power Supply Filtering Techniques

As in any high speed analog circuitry, the power supply pins are vulnerable to random noise. The ICS843034 provides separate power supplies to isolate any high switching noise from the outputs to the internal PLL. $V_{\rm CC}, V_{\rm CCA},$ and $V_{\rm CCO}$, should be individually connected to the power supply plane through vias, and bypass capacitors should be used for each pin. To achieve optimum jitter performance, power supply isolation is required. Figure 2 illustrates how a 10Ω resistor along with a $10\mu F$ and a $.01\mu F$ bypass capacitor should be connected to each $V_{\rm CCA}$ pin.

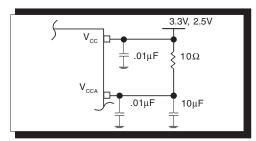
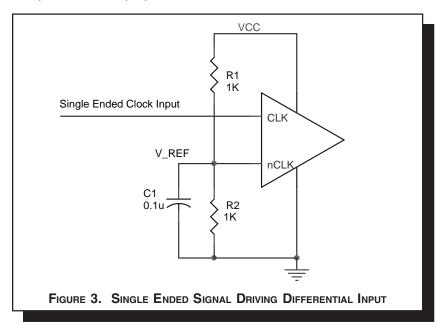


FIGURE 2. POWER SUPPLY FILTERING

WIRING THE DIFFERENTIAL INPUT TO ACCEPT SINGLE ENDED LVCMOS/LVTTL LEVELS

Figure 3 shows how the differential input can be wired to accept single ended levels. The reference voltage $V_REF = V_{cc}/2$ is generated by the bias resistors R1, R2 and C1. This bias circuit should be located as close as possible to the input pin. The ratio

of R1 and R2 might need to be adjusted to position the V_REF in the center of the input voltage swing. For example, if the input clock swing is only 2.5V and $V_{\rm CC}$ = 3.3V, V_REF should be 1.25V and R2/R1 = 0.609.





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DIFFERENTIAL CLOCK INPUT INTERFACE

The CLK /nCLK accepts LVDS, LVPECL, LVHSTL, SSTL, HCSL and other differential signals. Both V_{SWING} and V_{OH} must meet the V_{PP} and V_{CMR} input requirements. Figures 4A to 4D show interface examples for the HiPerClockS CLK/nCLK input driven by the most common driver types. The input interfaces suggested

here are examples only. Please consult with the vendor of the driver component to confirm the driver termination requirements. For example in *Figure 4A*, the input termination applies for ICS HiPerClockS LVHSTL drivers. If you are using an LVHSTL driver from another vendor, use their termination recommendation.

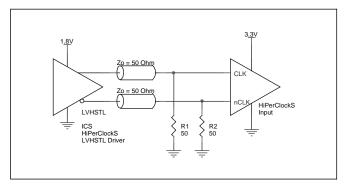


FIGURE 4A. HIPERCLOCKS CLK/nCLK INPUT DRIVEN BY ICS HIPERCLOCKS LVHSTL DRIVER

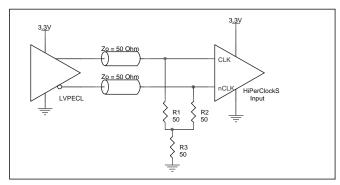


FIGURE 4B. HIPERCLOCKS CLK/nCLK INPUT DRIVEN BY 3.3V LVPECL DRIVER

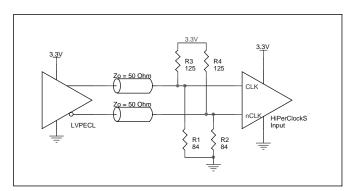


FIGURE 4C. HIPERCLOCKS CLK/nCLK INPUT DRIVEN BY 3.3V LVPECL DRIVER

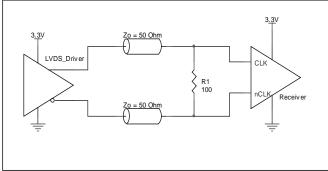


FIGURE 4D. HIPERCLOCKS CLK/nCLK INPUT DRIVEN BY 3.3V LVDS DRIVER



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TERMINATION FOR 2.5V LVPECL OUTPUT

Figure 5A and Figure 5B show examples of termination for 2.5V LVPECL driver. These terminations are equivalent to terminating 50Ω to $V_{\rm CC}$ - 2V. For $V_{\rm CC}$ = 2.5V, the $V_{\rm CC}$ - 2V is very close to

ground level. The R3 in *Figure 5B* can be eliminated and the termination is shown in *Figure 5C*.

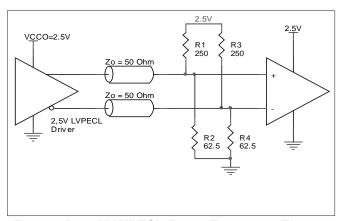


FIGURE 5A. 2.5V LVPECL DRIVER TERMINATION EXAMPLE

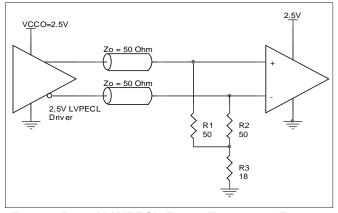


FIGURE 5B. 2.5V LVPECL DRIVER TERMINATION EXAMPLE

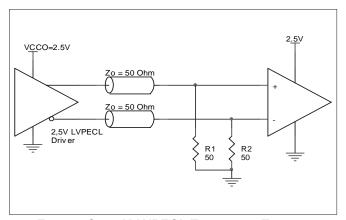


FIGURE 5C. 2.5V LVPECL TERMINATION EXAMPLE

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TERMINATION FOR 3.3V LVPECL OUTPUT

The clock layout topology shown below is a typical termination for LVPECL outputs. The two different layouts mentioned are recommended only as guidelines.

FOUTx and nFOUTx are low impedance follower outputs that generate ECL/LVPECL compatible outputs. Therefore, terminating resistors (DC current path to ground) or current sources must be used for functionality. These outputs are designed to

drive 50Ω transmission lines. Matched impedance techniques should be used to maximize operating frequency and minimize signal distortion. *Figures 6A and 6B* show two different layouts which are recommended only as guidelines. Other suitable clock layouts may exist and it would be recommended that the board designers simulate to guarantee compatibility across all printed circuit and clock component process variations.

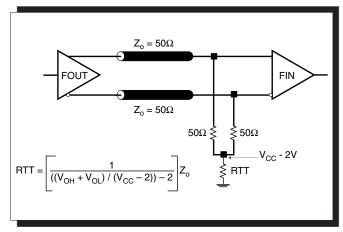


FIGURE 6A. LVPECL OUTPUT TERMINATION

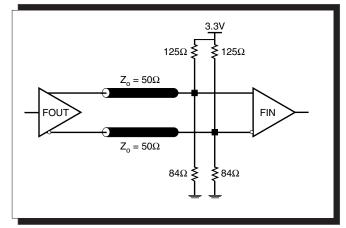
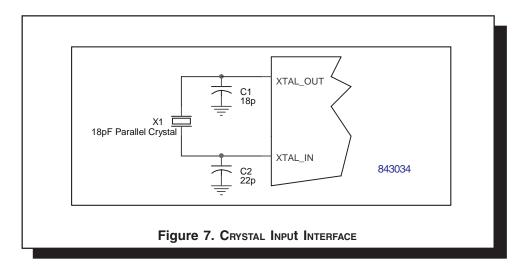


FIGURE 6B. LVPECL OUTPUT TERMINATION

CRYSTAL INPUT INTERFACE

The ICS843034 has been characterized with 18pF parallel resonant crystals. The capacitor values, C1 and C2, shown in *Figure 7* below were determined using a 25MHz, 18pF parallel resonant crystal and

were chosen to minimize the ppm error. The optimum C1 and C2 values can be slightly adjusted for different board layouts.





ICS843034

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Multi-Rate 3.3V, 2.5V LVPECL Frequency Synthesizer

POWER CONSIDERATIONS

This section provides information on power dissipation and junction temperature for the ICS843034. Equations and example calculations are also provided.

1. Power Dissipation.

The total power dissipation for the ICS843034 is the sum of the core power plus the power dissipated in the load(s). The following is the power dissipation for $V_{CC} = 3.3V + 5\% = 3.465V$, which gives worst case results.

NOTE: Please refer to Section 3 for details on calculating power dissipated in the load.

- Power (core)_{MAX} = V_{CC MAX} * I_{EE MAX} = 3.465V * 185mA = 641mW
- Power (outputs)_{MAX} = 30mW/Loaded Output pair
 If all outputs are loaded, the total power is 2 * 30mW = 60mW

Total Power MAX (3.465V, with all outputs switching) = 641mW + 60mW = 701mW

2. Junction Temperature.

Junction temperature, Tj, is the temperature at the junction of the bond wire and bond pad and directly affects the reliability of the device. The maximum recommended junction temperature for HiPerClockS TM devices is 125 $^{\circ}$ C.

The equation for Tj is as follows: $Tj = \theta_{1A} * Pd_total + T_A$

Tj = Junction Temperature

 $\theta_{1\Delta}$ = Junction-to-Ambient Thermal Resistance

Pd_total = Total Device Power Dissipation (example calculation is in section 1 above)

 T_{Δ} = Ambient Temperature

In order to calculate junction temperature, the appropriate junction-to-ambient thermal resistance $\theta_{\rm JA}$ must be used. Assuming a moderate air flow of 200 linear feet per minute and a multi-layer board, the appropriate value is 42.1°C/W per Table 8 below.

Therefore, Tj for an ambient temperature of 70°C with all outputs switching is:

 $70^{\circ}\text{C} + 0.701\text{W} * 42.1^{\circ}\text{C/W} = 99.5^{\circ}\text{C}$. This is well below the limit of 125°C .

This calculation is only an example. Tj will obviously vary depending on the number of loaded outputs, supply voltage, air flow, and the type of board (single layer or multi-layer).

Table 8. Thermal Resistance $\theta_{,ia}$ for 48-pin LQFP, Forced Convection

θ_{JA} by Velocity (Linear Feet per Minute)

 0
 200
 500

 Single-Layer PCB, JEDEC Standard Test Boards
 67.8°C/W
 55.9°C/W
 50.1°C/W

 Multi-Layer PCB, JEDEC Standard Test Boards
 47.9°C/W
 42.1°C/W
 39.4°C/W

NOTE: Most modern PCB designs use multi-layered boards. The data in the second row pertains to most designs.

ICS843034

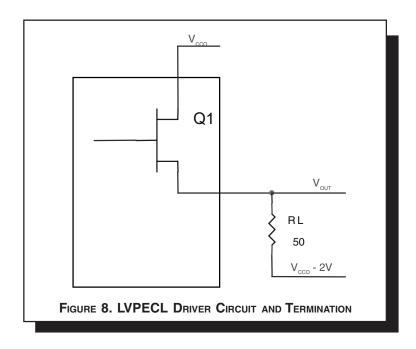
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3. Calculations and Equations.

The purpose of this section is to derive the power dissipated into the load.

LVPECL output driver circuit and termination are shown in Figure 8.



To calculate worst case power dissipation into the load, use the following equations which assume a 50Ω load, and a termination voltage of V $_{\text{CCO}}$ - 2V.

• For logic high,
$$V_{OUT} = V_{OH_MAX} = V_{CCO_MAX} - 0.9V$$

$$(V_{CCO_MAX} - V_{OH_MAX}) = 0.9V$$

• For logic low,
$$V_{OUT} = V_{OL_MAX} = V_{CCO_MAX} - 1.7V$$

$$(V_{CCO\ MAX} - V_{OL\ MAX}) = 1.7V$$

Pd_H is power dissipation when the output drives high.

Pd_L is the power dissipation when the output drives low.

$$Pd_H = [(V_{OH_MAX} - (V_{CCO_MAX} - 2V))/R_{L}] * (V_{CCO_MAX} - V_{OH_MAX}) = [(2V - (V_{CCO_MAX} - V_{OH_MAX}))/R_{L}] * (V_{CCO_MAX} - V_{OH_MAX}) = [(2V - 0.9V)/50\Omega] * 0.9V = 19.8mW$$

$$Pd_L = [(V_{OL_MAX} - (V_{CCO_MAX} - 2V))/R_{L}] * (V_{CCO_MAX} - V_{OL_MAX}) = [(2V - (V_{CCO_MAX} - V_{OL_MAX}))/R_{L}] * (V_{CCO_MAX} - V_{OL_MAX}) = [(2V - 1.7V)/50\Omega] * 1.7V = 10.2mW$$

Total Power Dissipation per output pair = Pd_H + Pd_L = 30mW



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RELIABILITY INFORMATION

Table 9. $\theta_{\text{JA}} \text{vs. Air Flow Table for 48 Lead LQFP}$

θ_{IA} by Velocity (Linear Feet per Minute)

	0	200	500
Single-Layer PCB, JEDEC Standard Test Boards	67.8°C/W	55.9°C/W	50.1°C/W
Multi-Layer PCB, JEDEC Standard Test Boards	47.9°C/W	42.1°C/W	39.4°C/W

NOTE: Most modern PCB designs use multi-layered boards. The data in the second row pertains to most designs.

TRANSISTOR COUNT

The transistor count for ICS843034 is: 11,748

ICS843034

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MULTI-RATE 3.3V, 2.5V LVPECL FREQUENCY SYNTHESIZER

PACKAGE OUTLINE - Y SUFFIX FOR 48 LEAD LQFP

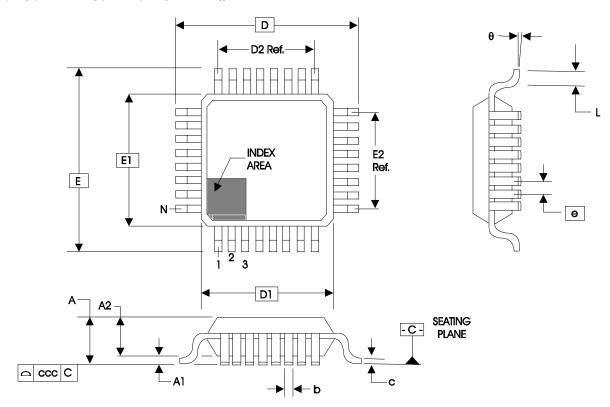


TABLE 10. PACKAGE DIMENSIONS

JEDEC VARIATION ALL DIMENSIONS IN MILLIMETERS						
CVMDOL	BBC					
SYMBOL	MINIMUM	NOMINAL	MAXIMUM			
N		48				
Α			1.60			
A1	0.05		0.15			
A2	1.35	1.40	1.45			
b	0.17	0.17 0.22 0.27				
С	0.09 0.20					
D	9.00 BASIC					
D1	7.00 BASIC					
D2	5.50 Ref.					
E	9.00 BASIC					
E1		7.00 BASIC				
E2		5.50 Ref.				
е	0.50 BASIC					
L	0.45	0.60	0.75			
θ	0°		7°			
ccc		0.08				

Reference Document: JEDEC Publication 95, MS-026



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TABLE 11. ORDERING INFORMATION

Part/Order Number	Marking	Package	Shipping Packaging	Temperature
ICS843034AY	ICS843034AY	48 Lead LQFP	tray	0°C to 70°C
ICS843034AYT	ICS843034AY	48 Lead LQFP	1000 tape & reel	0°C to 70°C

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