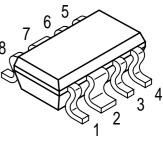
**BGC420** 

# Self-Biased BFP420

nfineon

- SIEGET<sup>®</sup>25- Technology
- Small SCT598-Package
- Control Pin For Switching The Device Off
- Current Easy Adjustable By An External Resistor
- Voltage Independent Current (2V 4.5V)

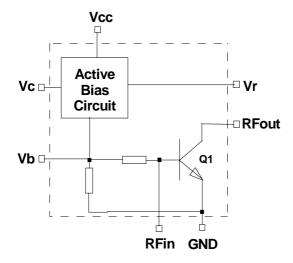


VPW05982

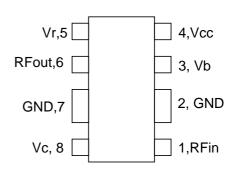
**ESD**: Electrostatic discharge sensitive device, observe handling precautions!

Туре	Marking	-	Pin Configuration (circuit Diagram)	Package
BGC420	42s	Q62702-G0092	see below	SCT598

### **Equivalent Circuit**



### Pin Connections, SCT598



Note: Top View

### Description

The BGC420 is a silicon self biased RF Transistor (Q1). It offers an adjustable collector current nearly independent from device voltage in the range from 2.0V to 4.5V. Additionally a control pin (Vc) for switching the device off is provided. The collector current can be adjusted by connecting a resistor (Rx) between Vcc and Vr.

#### Maximum Ratings

Parameter	Symbol		Unit
Device current	I <sub>CC</sub>	15	mA
Device voltage	Vcc	4.5	V
Total power dissipation, $T_s \leq 110^{\circ}C^{-1}$	P <sub>tot</sub>	68	mW
Control voltage	Vc	Vcc+0.5	V
Input Current for pin 1	Ir	380	μA
Junction temperature	T <sub>j</sub>	150	°C
Ambient temperature range	TA	-65+150	°C
Storage temperature range	T <sub>stg</sub>	-65+150	°C
Thermal Resistance			
Junction-soldering point <sup>1)</sup>	<b>R</b> th JS	<ul><li>≤ 270</li></ul>	K/W

1)T $_{S}$  is measured on the Ground lead at the soldering point to the pcb.

# Electrical Specifications (Measured in Test Fixture applying the circuit specified in Figure 1 with Rx=82 $\Omega$ ), Tc=25°C, Vcc=3V, I<sub>cc</sub>=7mA unless noted

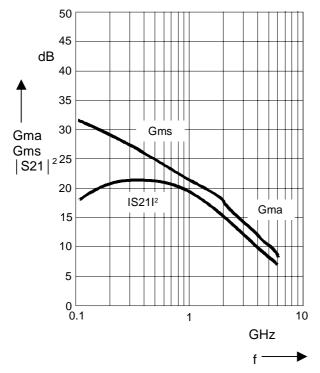
Symbol	Parameter		Unit	Min	Тур	Max
Gp	Power Gain $( S_{21} ^2)$	f=900MHz	dB	17.5	19	
		f=1.8GHz		14.5	16	
NF	Noise Figure (in 50 $\Omega$ System)	f=900MHz	dB		1.3	1.5
		f=1.8GHz			1.5	1.7
P <sub>-1dB</sub>	Output Power at 1dB Gain Compression	f=900MHz	dBm		1	
	(in 50Ω System)	f=1.8GHz			1	
IP <sub>3</sub>	Third Order Intercept Point	f=900MHz	dBm		15	
	(Output,Γ <sub>Opt</sub> )	f=1.8GHz			15	
RL <sub>in</sub>	Input Return Loss	f=900MHz	dB		7	
		f=1.8GHz			9	
RL <sub>out</sub>	Output Return Loss	f=900MHz	dB		4	
		f=1.8GHz			7	
t <sub>on</sub>	On Switching Time <sup>3)</sup>	On Switching Time <sup>3)</sup>			3.7	
t <sub>off</sub>	Off Switching Time <sup>3)</sup>		μs		2.5	
I <sub>leak</sub>	Leakage Current In Sleep Mode		μA		<10	
I <sub>VcOn</sub>	Controll Pin (Vc) Current in Active Mode <sup>2)</sup>		μA		35	
I <sub>VcOff</sub>	Controll Pin (Vc) Current in Sleep Mode <sup>2)</sup>		nA		-60	
V <sub>cmin</sub>	Minimum Voltage at Vc for Sleep Mode		V		V <sub>cc</sub> - 0.3V	
V <sub>cmax</sub>	Maximum Voltage at Vc for Active Mode		V		0V+0.3V	

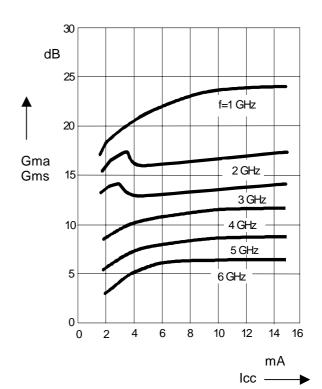
<sup>2)</sup> A positive sign denotes a current flowing form the Pin into the external circuit.

<sup>3)</sup> This values are valid for C2=1nF, C3=100pF and 220pF Coupling capacitors at RFin and RFout.



Power Gain versus Frequency Vcc=3V, Icc=5mA



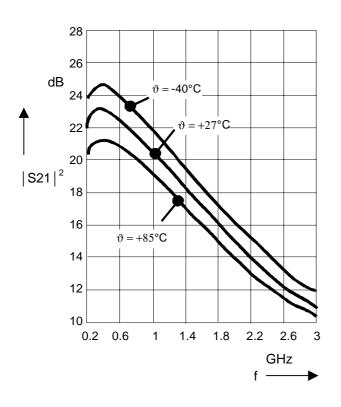


**Power Gain** 

Vcc=3V

versus Device Current

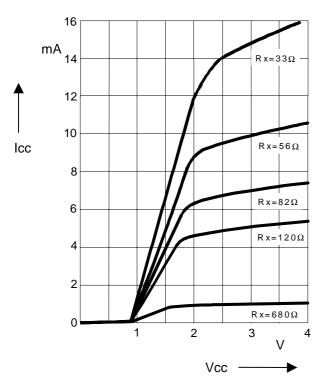
|S21|<sup>2</sup> versus Frequency and Temperature Vcc=3V, Icc=7mA



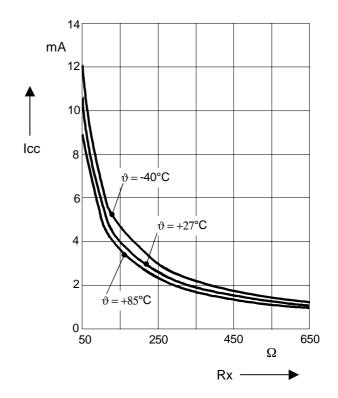
**High Frequency Products** 



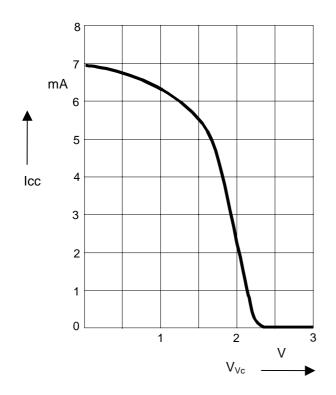
#### Device Current versus Device Voltage



Device Current versus Rx and Temperature Vcc=3V



Device Current versus Voltage at Vc Vcc=3V; Rx=82Ω





## **Typical Application**

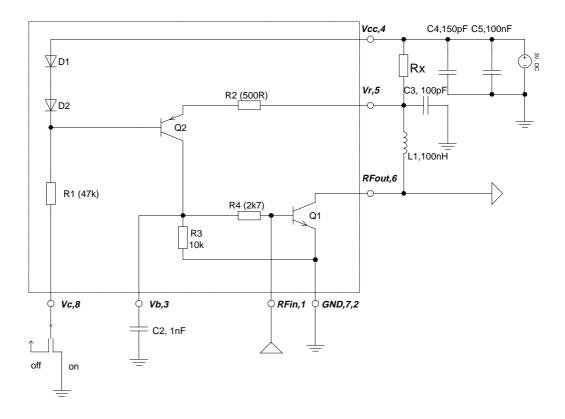


Figure 1. Typical Application and Internal Circuit

#### **Remarks:**

- 1) To provide low frequency stability C2 should be 10 times C3.
- 2) Be aware that also coupling capacitors determine the switching times.
- 3) The collector current at Q1 can be estimated by  $Ic=0.6V / Rx[\Omega]$ .
- 4) Place C2 as close to the device as possible.



**BGC420** 

### Layout Proposal

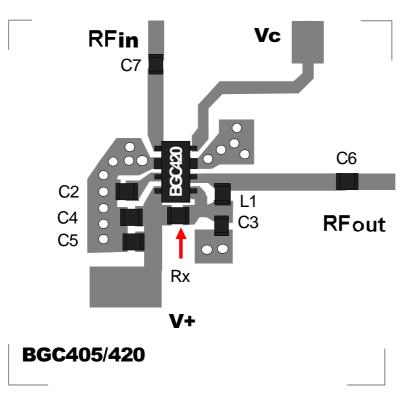


Figure 2. Layout Proposal

Component	Value	Comment	
L1	100nH	RFC	
C2	1nF	Compensation Capacitor for Low Frequency Stabilization	
C3	100pF	RFC	
C4	150pF	Blocking Capacitor	
C5	100nF	Blocking Capacitor	
C6	220pF	Coupling Capacitor	
C7	220pF	Coupling Capacitor	
Rx	82Ω	Current Adjust	
Substrate	h=0.5mm	$Fr4, \varepsilon_r = 4.5$	
BGC420			

This proposal demonstrates how to use the BGC420 as a *Self-Biased Transistor*. As for a discrete Transistor matching circuits have to be applied. A good starting point for various applications are the Application Notes provided for the BFP420.



### **SPICE Model**

The following SPICE Listing describes the circuit shown in figure 3. It is valid for low frequencies. For frequencies above 100MHz the parasitic circuit elements noted in figure 4 and table 1 should be added.

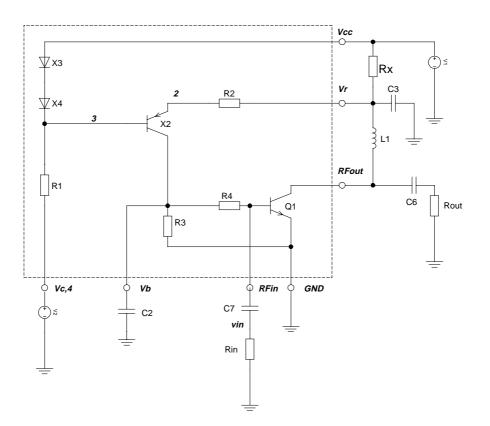


Figure. 3: Circuit used in the SPICE File

\* Preliminary SPICE Model for BGC420 (valid for frequencies below 100MHz)

\* SIEMENS HIGH FREQUENCY PRODUCTS \* Small Scale MMIC Design Group

```
.PARAM R=82
```

\*\* Analysis setup \*\*
\*.TRAN 2ns 15u 0 2n
.TEMP +27
.DC LIN V1 0V 4V 0.1V
\*.DC LIN V2 0V 3V 0.1V
.STEP PARAM R LIST 56 82 120 680
\* Voltage supply
V1 Vcc 0 DC 3.0V
V2 Vc 0 DC 0.0V
\*Vpul Vc 0 PULSE(0 3V 100ns 0 0 9us 1000m)



**BGC420** 

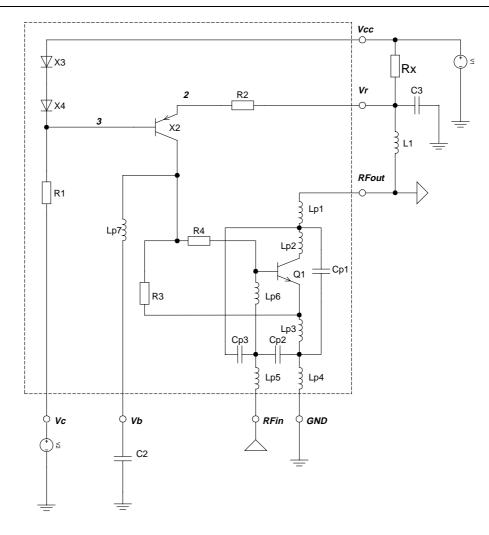
* Internal Re R1 3 Vc R2 Vr 2 R3 Vb 0 R4 Vb rfin	47k 500 10k	TC=-0.000 TC=-0.000	6,0.0000025	
* External Ro Rx Vcc Vr Rout vout 0 Rin vin 0	esistors {R} 50 50	TC=+0.000	050,0.0	
* External Ca	apacitors			
C2 Vb 0 C3 Vr 0 C7 rfin v C6 rfout v	1nF 100p in 220p vout 220p	F		
* Inductors	(external)			
Ll Vr rfou	ut 100n	Н		
* Transistors Q1 rfout r X2 2 3 Vb X3 Vcc 5 5 X4 5 3 3 0	rfin 0 0 5 0	BFP420 8PL18 2PL18 2PL18		
.PROBE				
.MODEL BFP424 + IS = 2.0044 + VAF = 28.33 + NE = 2.0513 + VAR = 19.74 + NC = 1.1724 + RBM = 8.573 + CJE = 1.804 + TF = 6.7663 + ITF = 0.003 + VJC = 0.813 + TR = 2.3244 + MJS = 0 + XTI = 3	5e-16 83 8 05 4 57 63e-15 1e-12 1 969 9e-09	BF = 72.534 $IKF = 0.487$ $BR = 7.8287$ $IKR = 0.691$ $RB = 3.4849$ $RE = 0.3111$ $VJE = 0.805$ $XTF = 0.421$ $PTF = 0$ $MJC = 0.302$ $CJS = 0$ $XTB = 0$ $FC = 0.7323$	31       ISE         .41       ISC         .41       ISC         .1       RC =         .1       MJE         .99       VTF         .232       XCJC         .232       XCJC         VJS       EG =	1.2432 = 1.9049e-14 1.3325 = 1.9237e-17 = 0.00072983 0.10105 = 0.46576 = 0.23794 = 2.3453e-13 = 0.3 = 0.75 1.11
* PNP: PL18	E B C Bul	k		
.SUBCKT 8PL Q1 993 Q2 94 Q3 94 RCEX 993 .ENDS	18 3 2 2 3 2 3 2 993 1	1 94 94 TL18 94 VSL18 94 LSL18 0.204	8 8 8 8	
.SUBCKT 2PL Q1 993 Q2 94 Q3 94 RCEX 993 .ENDS	18 3 2 2 3 2 3 2 993 1	1 94 94 TL18 94 VSL18 94 LSL18 0.816	8 2 8 2	

* * * * *					
. MODEL +IS +NE +BR +VAF +IKR +RBM +XTB +CJE +TF +ITF +ITF +MJC +CJS +PTF	TL18 = 2.914E-17 = 1.553E+00 = 2.869E+01 = 6.000E+01 = 2.474E-05 = 4.000E+01 = -6.000E-01 = 1.200E-14 = 7.600E-10 = 1.400E-02 = 3.760E-01 = 0.000E+00 = 0.000E+00	PNP NF ISE NC IKF RB RE EG VJE XTF CJC XCJC VJS FC	= 1.000E+00 = 6.923E-16 = 1.500E+00 = 1.676E-04 = 6.000E+01 = 2.597E+00 = 1.156E+00 = 4.900E-01 = 2.872E-01 = 4.700E-13 = 1.000E+00 = 7.500E-01 = 5.000E-01	BF NR ISC VAR IRB RC XTI MJE VTF VJC TR MJS	<pre>= 4.005E+02 = 1.000E+00 = 8.190E-15 = 2.214E+00 = 0.000E+00 = 4.000E+00 = 1.360E-01 = 1.000E+03 = 7.610E-01 = 0.000E+00 = 0.000E+00</pre>
* * * * *					
. MODEL +IS +NE +BR +VAF +IKR +RBM +XTB +CJE +TF +ITF +MJC +CJS +PTF	VSL18 = 1.630E-19 = 1.500E+00 = 1.000E+09 = 1.000E+02 = 1.000E+00 = 0.000E+00 = 0.000E+00 = 2.000E+00 = 1.000E+06 = 3.770E-01 = 0.000E+00 = 0.000E+00	PNP NF ISE NC IKF RB RE EG VJE XTF CJC XCJC VJS FC	= 1.000E+00 = 0.000E+00 = 2.000E+00 = 1.794E-04 = 0.000E+00 = 1.122E+00 = 6.800E-01 = 0.000E+00 = 1.950E-13 = 0.000E+00 = 7.500E-01 = 5.000E-01	BF NR ISC VAR IRB RC XTI MJE VTF VJC TR MJS	<pre>= 1.000E+09 = 1.000E+00 = 0.000E+00 = 1.700E+00 = 0.000E+00 = 3.000E+00 = 3.400E-01 = 1.000E+03 = 5.500E-01 = 0.000E+00 = 0.000E+00</pre>
***** .MODEL +IS +NE +BR +VAF +IKR +RBM +XTB +CJE +TF +ITF +MJC +CJS +PTF *****	LSL18 = 4.261E-17 = 1.500E+00 = 1.000E+09 = 6.000E+01 = 1.000E+00 = 0.000E+00 = 0.000E+00 = 1.000E-09 = 1.000E+00 = 3.000E-01 = 0.000E+00	PNP NF ISE NC IKF RB RE EG VJE XTF CJC XCJC VJS FC	<pre>= 1.000E+00 = 0.000E+00 = 2.000E+00 = 9.648E-05 = 0.000E+00 = 0.000E+00 = 1.158E+00 = 6.800E-01 = 0.000E+00 = 0.000E+00 = 0.000E+00 = 7.500E-01 = 5.000E-01</pre>	BF NR ISC VAR IRB RC XTI MJE VTF VJC TR MJS	<pre>= 1.000E+09 = 1.000E+00 = 0.000E+00 = 1.700E+00 = 0.000E+00 = 3.000E+00 = 3.400E-01 = 1.000E+03 = 4.600E-01 = 0.000E+00 = 0.000E+00</pre>

.END

Infineon



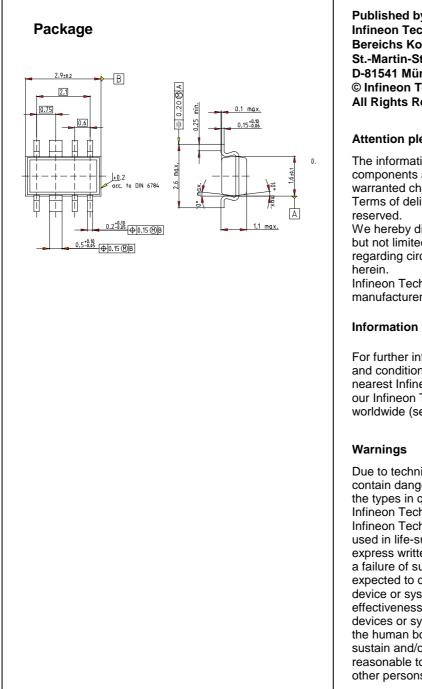




Element	Value		
Lp1	0.58nH		
Lp2	0.56nH		
Lp3	0.23nH		
Lp4	0.05nH		
Lp5	0.53nH		
Lp6	0.47nH		
Lp7	1nH		
Cp1	134fF		
Cp2	136fF		
Ср3	6.9fF		

Table 1. Parasitic circuit elements for frequencies above 100MHz





#### Published by Infineon Technologies AG i Gr., **Bereichs Kommunikation** St.-Martin-Strasse 76. D-81541 München © Infineon Technologies AG 1999 All Rights Reserved.

#### Attention please!

The information herein is given to describe certain components and shall not be considered as warranted characteristics.

Terms of delivery and rights to technical change

We hereby disclaim any and all warranties, including but not limited to warranties of non-infringement, regarding circuits, descriptions and charts stated

Infineon Technologies is an approved CECC manufacturer.

For further information on technology, delivery terms and conditions and prices please contact your nearest Infineon Technologies Office in Germany or our Infineon Technologies Representatives worldwide (see address list).

Due to technical requirements components may contain dangerous substances. For information on the types in question please contact your nearest Infineon Technologies Office. Infineon Technologies Components may only be

used in life-support devices or systems with the express written approval of Infineon Technologies, if a failure of such components can reasonably be expected to cause the failure of that life-support device or system, or to affect the safety or effectiveness of that device or system. Life support devices or systems are intended to be implanted in the human body, or to support and/or maintain and sustain and/or protect human life. If they fail, it is reasonable to assume that the health of the user or other persons may be endangered.