# **Self Recovering Watchdog**

#### **Features**

- Self recovering watchdog function: reset goes active after the 1st timeout period, reset goes inactive again after the 2nd timeout period, repeated active reset signal until the system recovers
- Standard timeout period and power-on reset time (100 ms), externally programmable if required
- Unregulated DC monitoring (V<sub>IN</sub>) with 3 standard or programmable trigger voltages for: power-on reset initialization, advanced power-fail warning (SAVE), reset at power-down (RES)
- Regulated DC monitoring ( $V_{DD}$ ): power-on reset initialization enabled only if  $V_{DD} \ge 3.5 \text{ V}$
- Internal voltage reference
- Works down to 1.6 V supply voltage
- Push-pull or Open drain outputs
- Low current consumption
- Available for normal and extended temperature ranges
- DIP8 and SO8 package

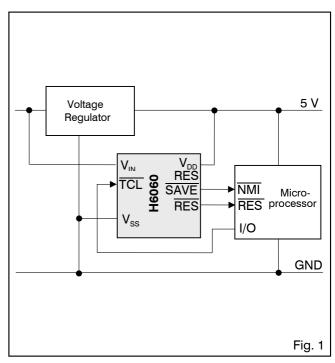
## **Description**

The H6060 is a monolithic low-power CMOS device combining a programmable timer and a series of voltage comparators on the same chip. The device is specially suited for watchdog functions such as microprocessor and supply voltage monitoring. If the µP system malfunctions, the watchdog will recover it by issuing repeated active reset signals. The voltage monitoring part provides double security by combining both the unregulated voltage (VIN) and the regulated voltage (VDD) monitoring simultaneously. The H6060 initializes the power-on reset after  $V_{\mbox{\tiny IN}}$  reaches  $V_{\mbox{\tiny SH}}$  (see table 4) and  $V_{\text{DD}}$  rises above 3.V. If  $V_{\text{IN}}$  drops below  $V_{\text{SL}}$  (see table 4), the H6060 gives an advanced warning signal for register saving and if the voltage drops further below V<sub>RL</sub> (see table 4), RES and RES go active. The H6060 functions at any supply voltage down to 1.6 V and is therefore particularly suited for start-up and shut-down control of microprocessor systems.

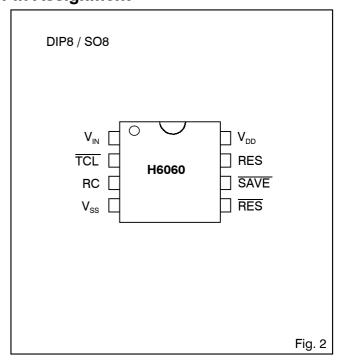
#### **Applications**

- Microprocessor and microcontroller systems
- Point of sales equipment
- Telecom products
- Automotive subsystems

## **Typical Operating Configuration**



## **Pin Assignment**





Units

°C

٧

٧

 $\mu F$ 

kΩ

+85

5.5

 $V_{\scriptscriptstyle DD}$ 



## **Absolute Maximum Ratings**

Parameter	Symbol	Conditions
Voltage V <sub>DD</sub> to V <sub>SS</sub>	V <sub>DD</sub>	-0.3  to  + 8  V
Voltage at any pin to V <sub>ss</sub>	$V_{MIN}$	- 0.3
Voltage at any pin to V <sub>DD</sub>		
(except V <sub>IN</sub> )	$V_{MAX}$	+ 0.3
Voltage at V <sub>IN</sub> to V <sub>ss</sub>	$V_{INMAX}$	+ 15 V
Current at any output	I <sub>MAX</sub>	± 10 mA
Storage temperature	T <sub>STO</sub>	–65+150 °C

Table 1

Stresses above these listed maximum ratings may cause permanent damage to the device. Exposure beyond specified operating conditions may affect device reliability or cause malfunction.

## **Handling Procedures**

This device has built-in protection against high static voltages or electric fields; however,

External resistance

Leakage < 1 μA

ing (see Fig. 15)

Parameter

Industrial

Supply voltage

Version 11,12

Table 2

Symbol Min. Typ Max.

-40

1.6

0

10

precautions must be taken as for any other CMOS

operation can only occur when all terminal voltages are kept within the supply voltage range. Unused inputs must always be tied to a defined logic voltage level.

 $\mathsf{T}_{\mathsf{AI}}$ 

 $V_{DD}$ 

C1

R1

component. Unless otherwise specified,

**Operating Conditions** 

Operating temperature

Comparator input voltage

RC-oscillator programm-

External capacitance\*

Version 13, 14, 15, 16

#### **Electrical Characteristics**

 $V_{\text{DD}} = 5.0 \text{ V}, \, T_{\text{A}} = -40 \text{ to } +85 \, ^{\circ}\text{C}, \, \text{unless otherwise specified}$ 

Parameter	Symbol	Test Conditions	Min.	Тур.	Max.	Units
V <sub>DD</sub> activation threshold	V <sub>on</sub>	T <sub>A</sub> = 25 °C	3		3.5	V
V <sub>DD</sub> deactivation threshold	V <sub>OFF</sub>	T <sub>Δ</sub> = 25 °C		$V_{ON} - 0.3$		V
Supply current	I <sub>DD</sub>	RC open, TCL at V <sub>DD</sub> or V <sub>SS</sub>		80	140	μ <b>A</b>
Input V <sub>IN</sub> , TCL						
Leakage current	I <sub>IP</sub>	$V_{SS} \leq V_{IP} \leq V_{DD}$ ;				
		$T_A = 85  ^{\circ}C$		0.005	1	μΑ
Input current on pin V <sub>IN</sub>	I <sub>IN</sub>	Versions 11, 12; $V_{IN} = 10 \text{ V}$		100	180	μΑ
TCL input low level	V <sub>IL</sub>				8.0	V
TCL input high level	V <sub>IH</sub>		2.4			V
SAVE, RES, RES outputs						
Leakage currents	I <sub>OLK</sub>	Versions 11, 13, 15;				
		$V_{OUT} = V_{DD}$		0.05	1	μ <b>A</b>
Drive currents (all versions)	I <sub>OL</sub>	$V_{OI} = 0.4 \text{ V}$	3.2	8		mA
	I <sub>OL</sub>	$V_{DD} = 3.5 \text{ V}; V_{OL} = 0.4 \text{ V}$	2			mA
	I <sub>OL</sub>	$V_{DD} = 1.6 \text{ V}; V_{OL} = 0.4 \text{ V}$	80			μ <b>A</b>
Drive currents	I <sub>OH</sub>	$V_{OH} = 4.0 \text{ V}$	3.2	8		mA
(versions 12, 14,16) <sup>1)</sup>	I <sub>OH</sub>	$V_{DD} = 3.5 \text{ V}; V_{OH} = 2.8 \text{ V}$	2			mA
	I <sub>OH</sub>	$V_{DD} = 1.6 \text{ V}; V_{OH} = 1.2 \text{ V}$	80			μ <b>A</b>

<sup>1)</sup> Versions: 11, 13, 15 = open drain outputs; 12, 14, 16 = push-pull outputs

Table 3

#### **V**<sub>IN</sub> Surveillance

Voltage thresholds at  $T_A = 25~^{\circ}C$ 

Version 1)	Comparator	Input Resistance	Threshold			Thresholds	Ratio	
70101011	Reference	on V <sub>IN</sub> (R <sub>VIN</sub> )	$V_{SH}$	$\mathbf{V}_{\mathtt{SL}}$	V <sub>SL</sub> V <sub>RL</sub> Tole		3) Tolerance	
11, 12	$V_{DD}$	100kΩ	9.00	8.00	7.00 <sup>2)</sup>	±5%	±2%	
13, 14	$V_{DD}$	~100MΩ	2.25	2.00	1.75 <sup>2)</sup>	±5%	±2%	
15, 16	Band-gap reference	~100MΩ	2.00	1.95	1.90	±10%	±2%	

<sup>1)</sup> Versions: 11, 13, 15 = open drain ouputs; 12, 14, 16 = push-pull outputs

 $<sup>^{2)}</sup>$  at  $V_{DD} = 5 \text{ V}$ 

 $<sup>^{3)}</sup>$  Threshold ratio tolerance is defined as the tolerance of  $V_{\text{SH}}$  /  $V_{\text{SL}}$  and  $V_{\text{SL}}$  /  $V_{\text{RL}}.$ 



## **Timing Characteristics**

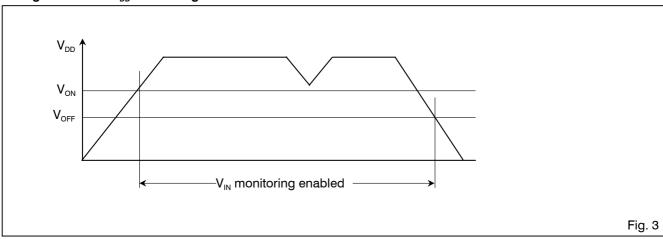
 $V_{\text{DD}} = 5.0 \text{ V}, \, T_{\text{A}} = -40 \,\,^{\circ}\text{C}$  to +85  $^{\circ}\text{C},$  unless otherwise specified

Parameter Symb		Test Conditions	Min.	Тур.	Max.	Units	
Propagation delays TCL to output pins V <sub>IN</sub> to output pins	T <sub>DIDO</sub>	Excluding debounce time T <sub>DB</sub>		250 4	500 10	ns μs	
Logic transition times on all output pins	T <sub>TR</sub>	Load 10 kΩ, 100 pF		30	100	ns	
Timeout period	$T_{TO}$	RC open, unshielded, T <sub>A</sub> = 25 °C RC open, unshielded (not tested)	60 45	100	160 200	ms ms	
T <sub>TCL</sub> input pulse width	T <sub>TCL</sub>		150			ns	
Power-on reset debounce	T <sub>DB</sub>			T <sub>TO</sub> /64		ms	
V <sub>IN</sub> low pulse	T <sub>VINL</sub>	Where debounce time T <sub>DB</sub> Is guaranteed	10			μs	

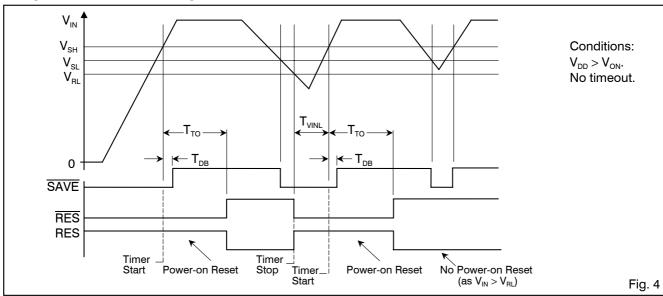
Table 5

## **Timing Waveforms**

## Voltage Reaction: $V_{\text{DD}}$ Monitoring

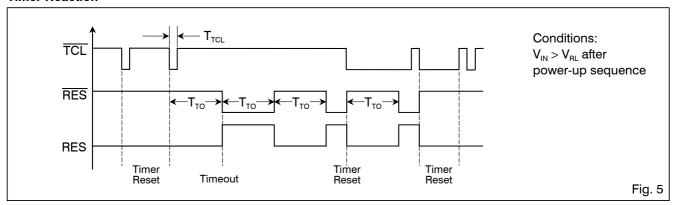


## **Voltage Reaction: VIN Monitoring**

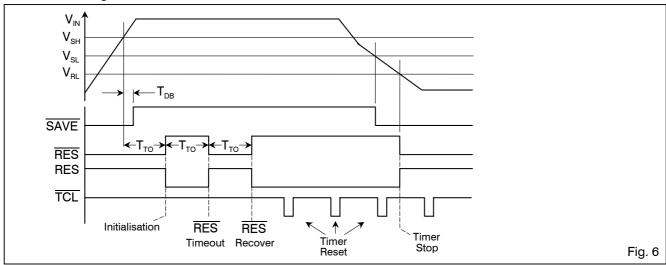




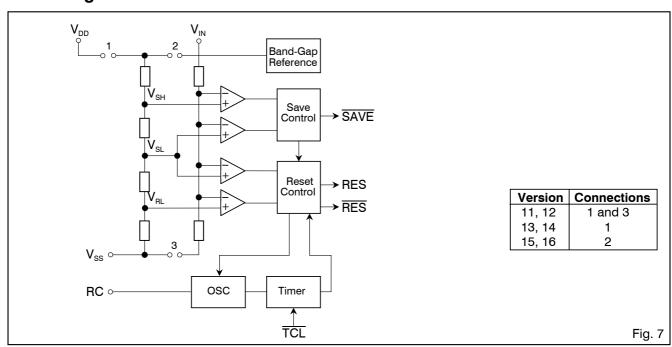
#### **Timer Reaction**



### **Combined Voltage and Timer Reaction**



## **Block Diagram**





## **Pin Description**

Pin	Name	Function
1	V <sub>IN</sub>	Voltage sense input
2	TCL	Timer clear input signal
3	RC	RC oscillator tuning input
4	$V_{ss}$	GND terminal
5	RES	Active low reset output
6	SAVE	Save output
7	RES	Active high reset output
8	$V_{DD}$	Positive supply voltage terminal

Table 6

### **Functional Description**

#### **Supply Lines**

The circuit is powered through the  $V_{DD}$  and  $V_{SS}$  pins. It monitors both its own  $V_{DD}$  supply and a voltage applied to the  $V_{IN}$  input.

#### **V<sub>DD</sub>** Monitoring

During power-up the  $V_{IN}$  monitoring is disabled and  $\overline{RES}$ , RES and  $\overline{SAVE}$  stay active low as long as  $V_{DD}$  is below  $V_{ON}$  (3.5 V). As soon as  $V_{DD}$  reaches the  $V_{ON}$  level, the state of the outputs depend on the watchdog timer and the voltage at  $V_{IN}$  relative to the thresholds (see Fig. 4). If the supply voltage  $V_{DD}$  falls back below  $V_{OFF}$  ( $V_{ON}-0.3$  V) the watchdog timer and the  $V_{IN}$  monitoring are disabled and the outputs  $\overline{RES}$ , RES and  $\overline{SAVE}$  become active. The  $V_{DD}$  line should be free of voltage spikes.

## V<sub>IN</sub> Monitoring

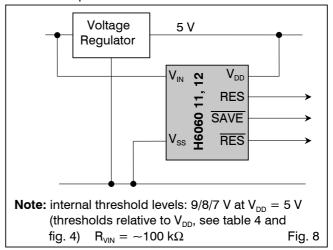
The analog voltage comparators compare the voltage applied to V<sub>IN</sub> (typically connected to the input of the voltage regulator) with the stabilized supply voltage  $V_{\text{DD}}$ (versions 11, 12, 13, 14) or with the bandgap voltage (versions 15, 16) (see Fig. 7). At power-up, when  $V_{DD}$ reached  $V_{ON}$  and  $V_{IN}$  reaches the  $V_{SH}$  level, the  $\overline{SAVE}$ output goes inactive, and the timer starts running, setting RES and RES in active after the time  $T_{TO}$  (see. Fig. 4). If  $V_{IN}$  falls below  $V_{SL}$  , the  $\overline{SAVE}$  output goes active and stays active until  $V_{\text{IN}}$  rises again above  $V_{\text{SH}}$  . If  $V_{\text{IN}}$  falls below the voltage  $V_{\text{RL}}$  ,  $\overline{\text{RES}}$  and RES will become active and the on-chip timer will stop. When  $V_{\mbox{\scriptsize IN}}$ rises again above V<sub>SH</sub>, the timer will initiate a power-up sequence. The RES and RES outputs may however be influenced independently of the voltage V<sub>IN</sub> by the timer action, see section "Combined Voltage and Timer Action". Monitoring the rough DC side of the regulator, as shown in Fig. 12, is the only way to have advanced warning of power-down. Spikes on  $V_{\mbox{\tiny IN}}$  should be filtered if they are likely to exceed the value ( $V_{\text{SL}} - V_{\text{RL}}$ ).

The combination of  $V_{\text{IN}}$  and  $V_{\text{DD}}$  monitoring provide high system security: if  $V_{\text{IN}}$  rises much faster than  $V_{\text{DD}}$ , then the device starts the power-on sequence only when  $V_{\text{DD}}$  reached  $V_{\text{ON}}$  (Fig. 11). Short circuits on the regulated supply voltage can be detected.

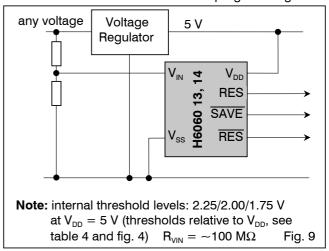
#### Voltage Thresholds on VIN

The H6060 is available with 3 different sets of thresholds:

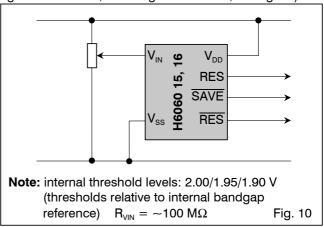
**Version 11, 12:** have an internal voltage divider for direct monitoring of the unregulated voltage without external components.



**Version 13, 14:** monitor the unregulated voltage and are ideal for programming of the  $V_{\text{IN}}$  voltage thresholds. Fixed resistor values can be used for programming.



**Version 15, 16:** monitor the regulated voltage. They are suited to applications where the unregulated voltage is not available. (The tolerance is  $\pm$  10%, see table 4. For tighter tolerances, trimming can be used, see fig. 10).





Monitoring of the unregulated voltage requires versions 11, 12, 13 and 14. These versions are based on the principle that  $V_{\text{DD}}$  rises with  $V_{\text{IN}}$  on power-up an  $V_{\text{DD}}$  holds up for a certain time after  $V_{\text{IN}}$  starts dropping on power-down. The versions 11 and 12 have a 100  $k\Omega$  nominal resistance from  $V_{\text{IN}}$  to  $V_{\text{SS}}$  (internal voltage divider). The versions 13, 14, 15 and 16 have high impedance  $V_{\text{IN}}$  inputs (see fig. 7 and table 4) for external threshold voltage programming by a voltage divider on pin  $V_{\text{IN}}$ . The levels obtained are proportional to the internal levels  $V_{\text{SH}}$ ,  $V_{\text{SL}}$  and  $V_{\text{RL}}$  on the chip itself (see Electrical Specifications).

#### **Timer Programming**

With pin RC unconnected, the on-chip RC oscillator together with its divider chain give a timeout  $T_{TO}$  of typically 100 ms. To program different  $T_{TO}$ , an approximation for calculating component values is given by the formula:

$$T_{TO} = \left[ 0.75 + \frac{(32 + C_1) \cdot 2}{5.5 + \frac{V_{DD} \cdot 1}{R_1}} \right] 8. \ 192$$

$$R_{1~\text{min.}}=10~\text{k}\Omega,\,C_{1~\text{max.}}=1~\mu\text{F}$$
 If  $R_{1}$  is in  $M\Omega$  and  $C_{1}$  in pF,  $T_{\text{TO}}$  will be in ms.

A resistor decreases and a capacitor increases the interval to timeout. Excellent temperature stability of  $T_{\text{TO}}$  can be achieved by using external components. A precise square wave of period  $2\times T_{\text{TO}}$  is generated at the outputs  $\overline{\text{RES}}$  and RES when  $\overline{\text{TCL}}$  is tied to either  $V_{\text{DD}}$  or  $V_{\text{SS}}.$  The oscillator and watchdog timer start running when both  $V_{\text{IN}}$  is greater than  $V_{\text{SH}}$  (see fig. 6) and  $V_{\text{DD}}$  is

greater than  $V_{ON}$  (see fig. 3).

They will remain running while both  $V_{IN}$  is greater than  $V_{RL}$  and  $V_{DD}$  is greater than  $V_{OFF}$  (see fig. 3).

#### Timer Clearing and RES/RES Action

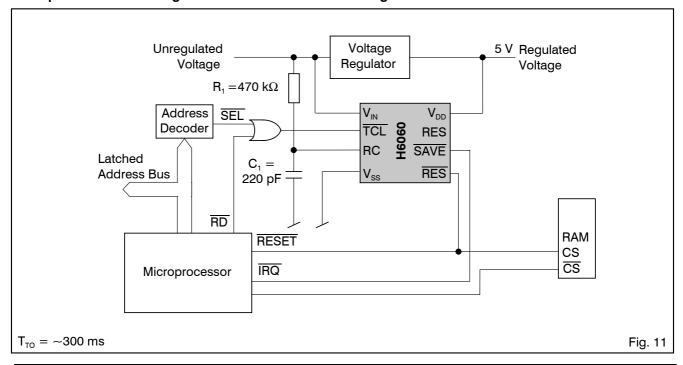
A negative edge or a negative pulse at the  $\overline{TCL}$  input for longer than 150 ns will reset the timer and set  $\overline{RES}$  and RES inactive. If a further  $\overline{TCL}$  signal edge or pulse is applied before  $T_{TO}$  timeout,  $\overline{RES}$  and RES will remain inactive and the timer will again be reset to zero (see fig. 5). If no  $\overline{TCL}$  signal is applied before the  $T_{TO}$  timeout,  $\overline{RES}$  and RES will start to generate square waves of period 2  $\times$   $T_{TO}$  starting with the inactive state. The watchdog will remain in this state until the next  $\overline{TCL}$  signal appears, or until a fresh power-up sequence.

#### **Combined Voltage and Timer Action**

The combination of voltage and timer actions is illustrated by the sequence of events shown in fig. 6. One timeout period after  $V_{IN}$  reaches  $V_{SH}$ , during powerup, RES and RES go inactive. A TCL pulse will have no effect until this power-on reset delay is completed. After completing the power-up sequence the watchdog timer starts acting. If no TCL pulse occurs, RES and RES go active after one timeout period  $T_{TO}$ . After each subsequent timeout period, without a timer clear pulse at TCL, RES and RES change polarity providing square wave signals. A TCL pulse clears the watchdog timer and causes RES and RES to go inactive. A voltage drop below the V<sub>RL</sub> level overrides the timer and immediately forces RES, RES and SAVE active. Any further TCL pulse has no effect until the next power-up sequence is completed.

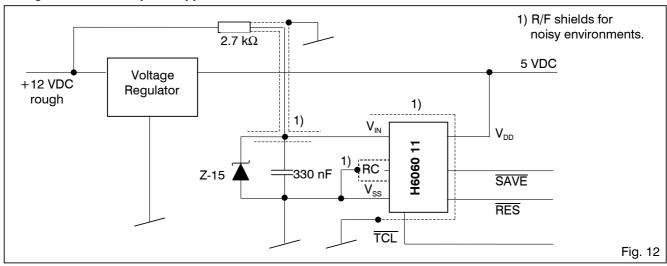
## **Typical Applications**

#### Microprocessor Watchdog with Power-On Reset and Voltage Monitor

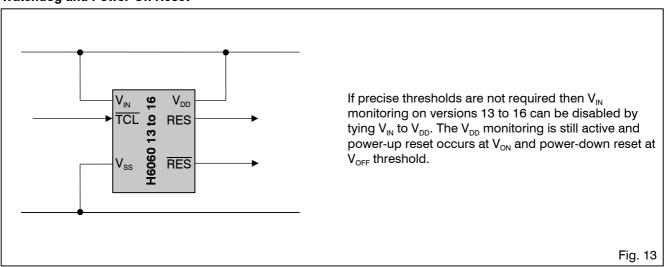




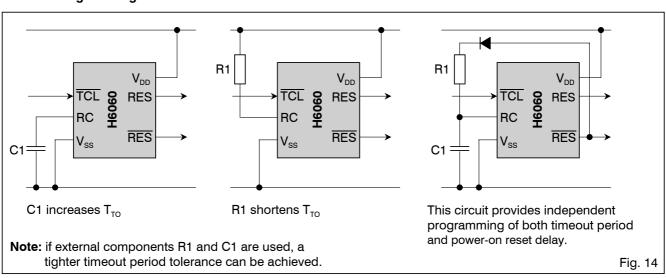
#### **Voltage Monitor with Spike Suppression**



#### **Watchdog and Power-On Reset**



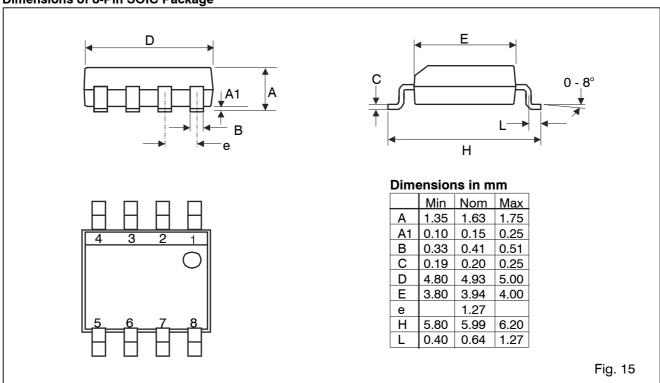
#### **External Programming of RC Oscillator**



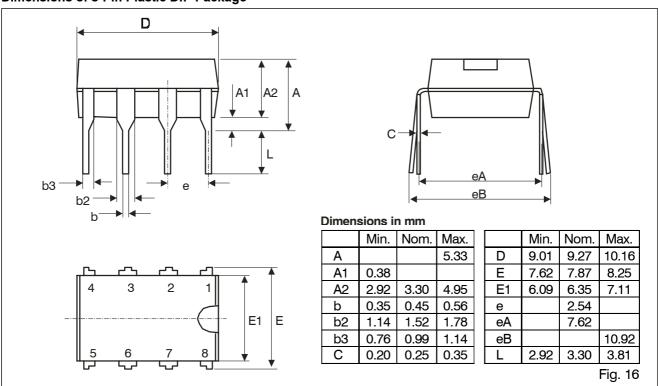


## **Package Information**

## **Dimensions of 8-Pin SOIC Package**



#### **Dimensions of 8-Pin Plastic DIP Package**





## **Ordering Information**

When ordering please specify complete part number.

Part Number	Version	Threshold (see Table 4)	Output	Package	Delivery Form	Package Marking (first line)	Temperature
H6060V11SO8A *				8-pin SOIC	Stick	606011	
H6060V11SO8B *	V11	8.00		8-pin SOIC	Tape&Reel	606011	
H6060V11DL8A *				8-pin plastic DIP	Stick	H606011	
H6060V13SO8A *				8-pin SOIC	Stick	606013	
H6060V13SO8B *	V13	2.00	Open drain	8-pin SOIC	Tape&Reel	606013	
H6060V13DL8A *				8-pin plastic DIP	Stick	H606013	
H6060V15SO8A				8-pin SOIC	Stick	606015	
H6060V15SO8B	V15	1.95		8-pin SOIC	Tape&Reel	606015	
H6060V15DL8A				8-pin plastic DIP	Stick	H606015	-40°C to +85°C
H6060V12SO8A *				8-pin SOIC	Stick	606012	
H6060V12SO8B *	V12	8.00		8-pin SOIC	Tape&Reel	606012	
H6060V12DL8A *				8-pin plastic DIP	Stick	H606012	
H6060V14SO8A				8-pin SOIC	Stick	606014	
H6060V14SO8B	V14	2.00	Push-pull	8-pin SOIC	Tape&Reel	606014	
H6060V14DL8A				8-pin plastic DIP	Stick	H606014	
H6060V16SO8A				8-pin SOIC	Stick	606016	
H6060V16SO8B	V16	1.95		8-pin SOIC	Tape&Reel	606016	
H6060V16DL8A				8-pin plastic DIP	Stick	H606016	

<sup>\* =</sup> non stock items. Might be available on request and upon minimum order quantity (please contact EM Microelectronic)

EM Microelectronic-Marin SA cannot assume any responsibility for use of any circuitry described other than entirely embodied in an EM Microelectronic-Marin SA product. EM Microelectronic-Marin SA reserves the right to change the circuitry and specifications without notice at any time. You are strongly urged to ensure that the information given has not been superseded by a more up-to-date version.

© 2002 EM-Microelectronic-Marin SA, 03/02, Rev. H/457

\_