

# DATA SHEET

## **TEA6824T**

**In Car Entertainment (ICE) car  
radio**

Preliminary specification  
Supersedes data of 1996 Nov 04  
File under Integrated Circuits, IC01

1997 May 05

## In Car Entertainment (ICE) car radio

TEA6824T

## FEATURES

- FM mixer for conversion from FM-IF<sub>1</sub> = 72.2 MHz to FM-IF<sub>2</sub> = 10.7 MHz
- AM mixer for conversion from AM-IF<sub>1</sub> = 10.7 MHz to AM-IF<sub>2</sub> = 450 kHz
- FM-IF gain stage
- Crystal oscillator providing mixer frequencies and references for IF-counter
- FM quadrature demodulator with automatic centre frequency adjustment and THD compensation
- AM-IF-amplifier and demodulator
- I<sup>2</sup>C-bus transceiver with interface to enable direct data transfer to radio front-end
- IF-count for AM and FM
- Analog-to-digital conversion of level voltage and multi-path information
- Reference frequency generation for PLL synthesizer.



## GENERAL DESCRIPTION

The TEA6824T together with the TEA6810T forms the receiver part of an AM/FM electronically tuned car radio providing FM MPX signal and AM AF signal for Car Digital Signal Processor (CDSP) applications.

## ORDERING INFORMATION

TYPE NUMBER	PACKAGE		
	NAME	DESCRIPTION	VERSION
TEA6824T	VSO56	plastic very small outline package; 56 leads; face down	SOT190-2

## In Car Entertainment (ICE) car radio

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## QUICK REFERENCE DATA

SYMBOL	PARAMETER	CONDITIONS	MIN.	TYP.	MAX.	UNIT
V <sub>DDA1</sub>	analog supply voltage 1 (+5 V; pin 5)	note 1	4.5	5.0	5.5	V
		operating range	4.75	5.0	5.25	V
I <sub>DDA1</sub>	analog supply current 1 (pin 5)	FM mode	17	21	25	mA
		AM mode	14	17	21	mA
I <sub>19</sub> + I <sub>20</sub>	total FM mixer output current (pins 19 and 20)		4.8	6.0	7.2	mA
I <sub>22</sub> + I <sub>23</sub>	total AM mixer output current (pins 22 and 23)		10	12	14	mA
V <sub>DDA2</sub>	analog supply voltage 2 (pin 28)	note 1	7.0	8.5	10	V
		operating range	8.1	8.5	8.9	V
I <sub>DDA2</sub>	analog supply current 2 (pin 28)	FM mode	2.4	3.0	3.6	mA
V <sub>DDA3</sub>	analog supply voltage 3 (+8.5 V; pin 56)	note 1	7.0	8.5	10	V
		operating range	8.1	8.5	8.9	V
I <sub>DDA3</sub>	analog supply current 3 (pin 56)	FM mode	19	25	31	mA
		AM mode	9.5	12	15	mA
V <sub>DDD</sub>	digital supply voltage 1 (+5 V; pin 5)	note 1	4.5	5.0	5.5	V
		operating range	4.75	5.0	5.25	V
I <sub>DDD</sub>	digital supply current (pin 52)	note 1	8	10	12	mA
$\frac{S+N}{N}$	signal plus noise-to-noise ratio	FM mode; $\Delta f = 22.5$ kHz at pin 47	66	75	–	dB
		AM mode; m = 0.3	54	60	–	dB
THD	total harmonic distortion	FM mode; $\Delta f = 75$ kHz	–	0.1	0.35	%
		AM mode	–	1.5	3.0	%
T <sub>amb</sub>	operating ambient temperature		–40	–	+85	°C

## Note

1. IC is operating; specified parameters may deviate from limits which are valid for operating range.

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## BLOCK DIAGRAM

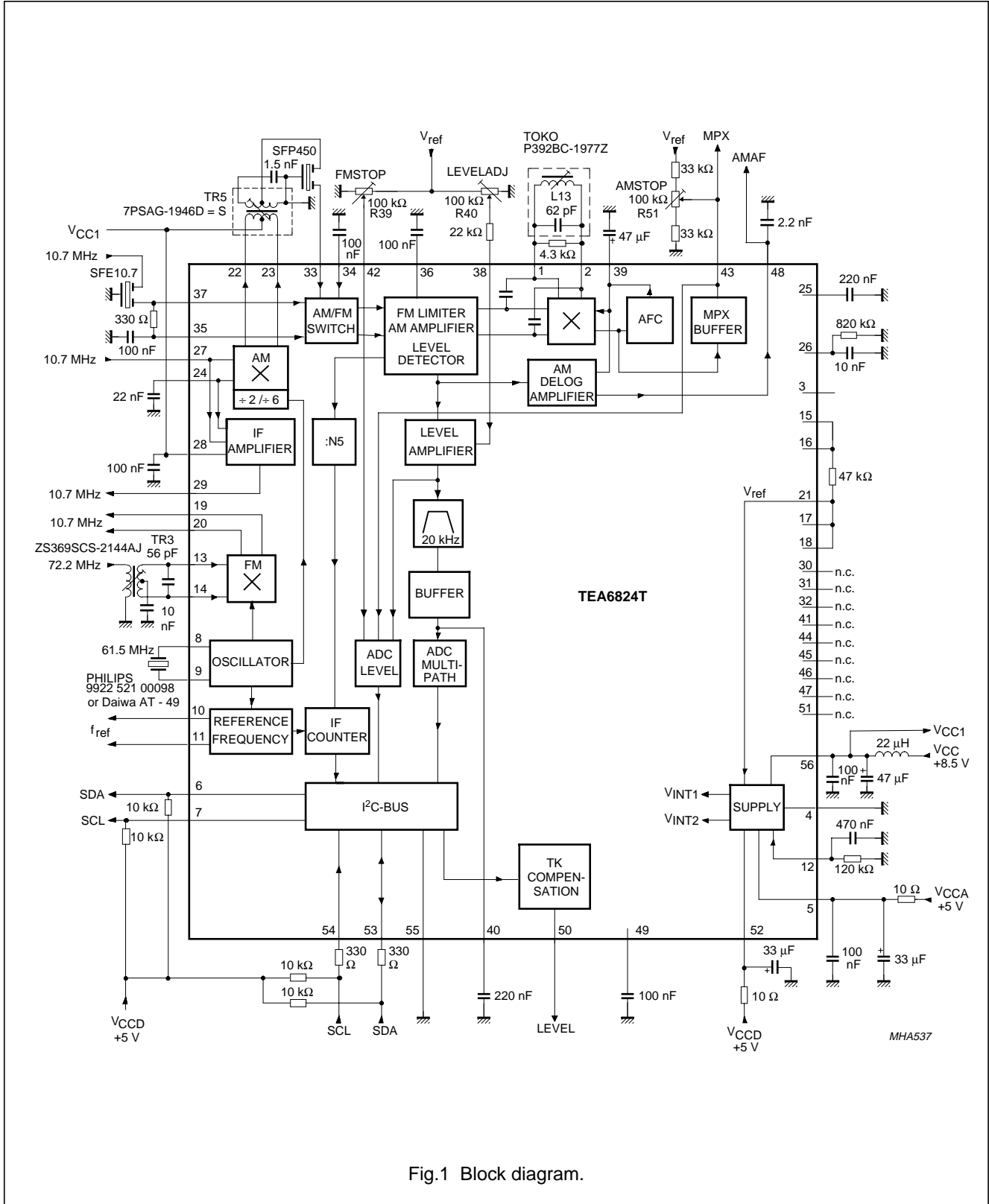


Fig.1 Block diagram.

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## PINNING

SYMBOL	PIN	DESCRIPTION
QDET1	1	demodulator tank 1
QDET2	2	demodulator tank 2
REFIN1	3	reference input 1
AGND	4	analog ground
V <sub>DDA1</sub>	5	analog supply voltage 1 (+5 V)
HFBUS1	6	HF bus 1 output; pull-up to +5 V
HFBUS2	7	HF bus 2 output; pull-up to +5 V
XTAL1	8	crystal oscillator 1
XTAL2	9	crystal oscillator 2
f <sub>ref(p)</sub>	10	PLL reference frequency output p
f <sub>ref(n)</sub>	11	PLL reference frequency output n
I <sub>ref</sub>	12	reference current input
FMIF1IN1	13	72 MHz FM-IF input 1
FMIF1IN2	14	72 MHz FM-IF input 2
REFIN2	15	reference input 2
REFIN3	16	reference input 3
REFIN4	17	reference input 4
REFIN5	18	reference input 5
FMIF2OUT1	19	FM mixer output 1
FMIF2OUT2	20	FM mixer output 2
V <sub>ref</sub>	21	reference voltage input
AMIF2OUT1	22	AM mixer output 1
AMIF2OUT2	23	AM mixer output 2
FMAMDEC	24	FM/AM 10.7 MHz decoupling input
DEC1	25	decoupling 1
DEC2	26	decoupling 2
FMAM10.7	27	FM/AM 10.7 MHz input
V <sub>DDA2</sub>	28	analog supply voltage 2

SYMBOL	PIN	DESCRIPTION
FMIFAMPOUT	29	FM-IF-amplifier output
n.c.	30	not connected
n.c.	31	not connected
n.c.	32	not connected
AMIF2IN1	33	AM-IF <sub>2</sub> input 1
AMIF2IN2	34	AM-IF <sub>2</sub> input 2
FMIN2	35	FM limiter input 2
DCFEED	36	DC feed FM limiter
FMIN1	37	FM limiter input 1
LEVELADJ	38	level adjustment
C <sub>AFC</sub>	39	AFC capacitor
MPBUF	40	multi-path buffer time constant
n.c.	41	not connected
FMSTOP	42	FMSTOP adjustment
MPXOUT	43	FM demodulator MPX output
n.c.	44	not connected
n.c.	45	not connected
n.c.	46	not connected
n.c.	47	not connected
AMAFOUT	48	AM demodulator AF output
DEC3	49	decoupling 3
LEVELUNWEIG	50	level unweighted output
n.c.	51	not connected
V <sub>DDD</sub>	52	digital supply voltage (+5 V)
SDA	53	serial data input/output; pull-up to +5 V; I <sup>2</sup> C-bus
SCL	54	serial clock input; pull-up to +5 V; I <sup>2</sup> C-bus
DGND	55	digital ground
V <sub>DDA3</sub>	56	analog supply voltage 3 (+8.5 V)

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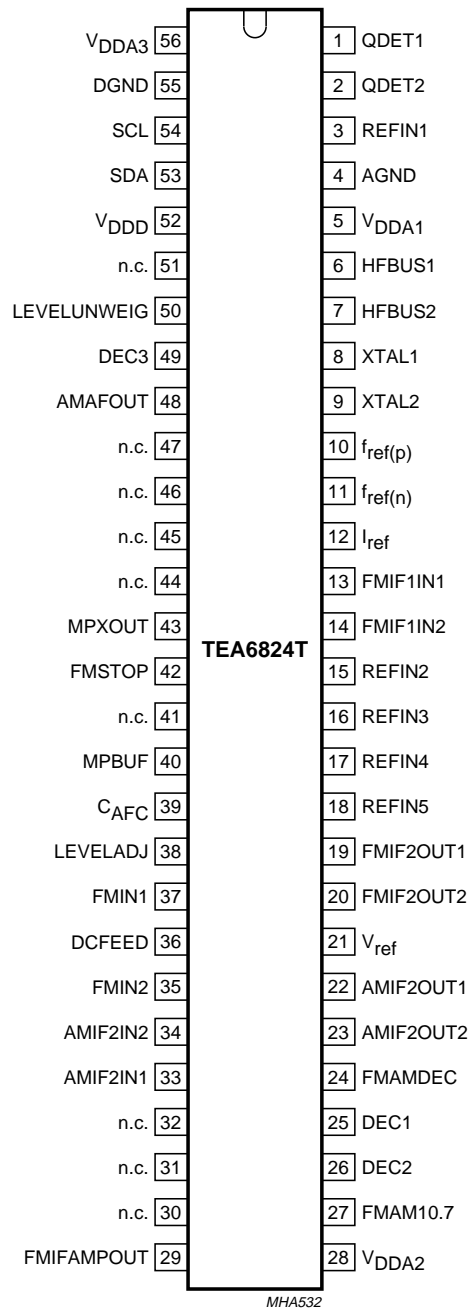


Fig.2 Pin configuration.

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**LIMITING VALUES**

In accordance with the Absolute Maximum Rating System (IEC 134).

SYMBOL	PARAMETER	CONDITIONS	MIN.	MAX.	UNIT
V <sub>DDA1</sub>	analog supply voltage 1 (pin 5)		-0.3	+6.5	V
V <sub>DDA2</sub>	analog supply voltage 2 (pin 28)		-0.3	+12	V
V <sub>DDA3</sub>	analog supply voltage 3 (pin 56)		-0.3	+12	V
V <sub>DDD</sub>	digital supply voltage (pin 52)		-0.3	+6.5	V
T <sub>stg</sub>	storage temperature		-55	+150	°C
T <sub>amb</sub>	operating ambient temperature		-40	+85	°C
V <sub>es</sub>	electrostatic handling	note 1			
	pins 8 and 9		-100	+100	V
	all other pins		-300	+300	V

**Note**

1. Charge device model class B: equivalent to discharging a 200 pF capacitor via a 0 Ω series resistor.

**CHARACTERISTICS**

V<sub>DDA3</sub> = V<sub>DDA2</sub> = 8.5 V (ref. AGND); V<sub>DDA1</sub> = 5 V (ref. AGND); V<sub>DDD</sub> = 5 V (ref. DGND); T<sub>amb</sub> = 25 °C; f<sub>mod</sub> = 1 kHz; Δf = 22.5 kHz; R<sub>g</sub> = 50 Ω; V<sub>37-35</sub> = 10 mV; with de-emphasis = 50 μs; unless otherwise specified.

SYMBOL	PARAMETER	CONDITIONS	MIN.	TYP.	MAX.	UNIT
<b>Current consumption</b>						
I <sub>DDA1</sub>	analog supply current 1 (pin 5)	FM mode	17	21	25	mA
		AM mode	14	17	21	mA
I <sub>DDA2</sub>	analog supply current 2 (pin 28)	FM mode	2.4	3.0	3.6	mA
I <sub>DDA3</sub>	analog supply current 3 (pin 56)	FM mode	19	25	31	mA
		AM mode	9.5	12	15	mA
I <sub>DDD</sub>	digital supply current (pin 52)		8	10	12	mA
I <sub>19</sub> + I <sub>20</sub>	total FM mixer output current pins 19 and 20		4.8	6.0	7.2	mA
I <sub>22</sub> + I <sub>23</sub>	total AM mixer output current pins 22 and 23	AM mode	10	12	14	mA
<b>FM-IF path; see Fig.3</b>						
V <sub>19-20max(p-p)</sub>	maximum output voltage pins 19 to 20 (peak-to-peak value)		12	14	–	V
I <sub>bias(19-20)</sub>	mixer bias current pins 19 to 20		2.4	3.0	3.6	mA
I <sub>LI(19-20)</sub>	mixer leakage current pins 19 to 20	in AM position	–	–	2	μA
I <sub>19IF2</sub> /V <sub>13-14IF1</sub>	conversion gain		1.65	1.9	2.2	mS
R <sub>i(13-14)</sub>	input resistance between pins 13 and 14		5	7	–	kΩ
C <sub>i(13-14)</sub>	input capacitance between pins 13 and 14		–	3	4.5	pF
R <sub>opt</sub>	optimum generator resistance		–	1.2	–	kΩ

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SYMBOL	PARAMETER	CONDITIONS	MIN.	TYP.	MAX.	UNIT
$R_{o(19-20)}$	output resistance between pins 19 and 20		15	20	–	k $\Omega$
$C_{o(19-4)}$ ; $C_{o(20-4)}$	output capacitance between pins 19 to 4 and pins 20 to 4		–	12	14	pF
IP3	third-order intermodulation		114	124	–	dB $\mu$ V
<b>Oscillator</b>						
$f_{osc}$	oscillator frequency		–	61.5	–	MHz
$\Delta f_{osc}$	oscillator frequency spread		–	–	250	Hz
$\Delta f_{osc}/\Delta T$	temperature dependence of oscillator frequency	crystal type Philips 9922 521 00098	–	$30 \times 10^{-6}$	–	K $^{-1}$
$R_1$	crystal motional resistance		–	–	70	$\Omega$
$C_0$	crystal shunt capacitance		–	–	5	pF
<b>FM-IF<sub>2</sub> amplifier</b>						
$V_{27-24max(rms)}$	maximum input voltage for 1 dB compression point between pins 27 and 24 (RMS value)		80	110	–	mV
$V_{29-4max(rms)}$	maximum output voltage between pins 29 and 4 (RMS value)		220	320	–	mV
$V_{29-4}/V_{27-24}$	amplifier gain between pins 29 and 4 and between pins 27 and 24	loaded with 330 $\Omega$ ; see Fig.4	9	12	15	dB
$\Delta V_{29-4}/V_{27-24}$	gain temperature dependence between pins 29 and 4 and between pins 27 and 24		–	0.05	–	dB/K
$R_{i(27-4)}$	input resistance between pins 27 and 4		300	330	360	$\Omega$
$C_{i(27-4)}$	input capacitance between pins 27 and 4		–	–	5	pF
$R_{o(29-4)}$	output resistance between pins 29 and 4		300	330	360	$\Omega$
$C_{o(29-4)}$	output capacitance between pins 29 and 4		–	–	5	pF
<b>FM-IF<sub>2</sub> limiter</b>						
$V_{o1-2(p-p)}$	limiter output voltage between pins 1 and 2 (peak-to-peak value)		500	700	–	mV
$V_{1-2}/V_{37-35}$	limiter gain between pins 1 and 2 and between pins 37 and 35	see Fig.5	–	80	–	dB
$C_{i(37-4)}$	input capacitance between pins 37 and 4		–	–	5	pF
$R_{o(1-2)}$	output resistance between pins 1 and 2		–	–	1.0	k $\Omega$
$C_{o(1-2)}$	output capacitance between pins 1 and 2		10	15	20	pF



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SYMBOL	PARAMETER	CONDITIONS	MIN.	TYP.	MAX.	UNIT
<b>FM demodulator</b>						
$V_{43-4(\text{rms})}$	MPX output voltage between pins 43 and 4 (RMS value)		160	200	240	mV
$V_{37-35(\text{rms})}$	start of limiting voltage between pins 37 and 35 (RMS value)	$\alpha_{\text{AF}} = -3 \text{ dB}$	–	25	40	$\mu\text{V}$
$V_{37-35(\text{rms})}$	input voltage for signal plus noise-to-noise ratio (RMS value)	see Fig.6; $\frac{S+N}{N} = 26 \text{ dB}$ $\frac{S+N}{N} = 46 \text{ dB}$	–	30 70	45 100	$\mu\text{V}$
$\Delta V_{43\text{DC}}$	residual DC-offset voltage pin 43	$\Delta L_{\text{demod}} = \text{typical value};$ $10 \mu\text{V} < V_{37-35} < 80 \mu\text{V}$ $80 \mu\text{V} < V_{37-35} < 800 \text{ mV}$	– –	100 10	1000 30	mV mV
$V_{43\text{FM}}/V_{43\text{AM}}$	AM suppression pin 43	$\Delta f = 22.5 \text{ kHz};$ $f_{\text{modAM}} = 1 \text{ kHz};$ $m_{\text{AM}} = 30\%;$ $V_{37-35} = 3 \text{ to } 300 \text{ mV}$	50	60	–	dB
$\frac{\Delta V_{43-44\text{AFCdis}}}{\Delta V_{43-44\text{AFCactive}}}$	demodulator frequency control voltage (AFC) efficiency between pins 43 and 44 at 100 kHz detune from exact tuning		28	32	–	dB
$R_{o43}$	output resistance pin 43		–	–	1.5	$\text{k}\Omega$
$B_{43}$	AF bandwidth pin 43		200	–	–	kHz
$\frac{S+N}{N}$	signal plus noise-to-noise ratio		66	75	–	dB
THD	total harmonic distortion at pin 43	detuning $\leq 50 \text{ kHz};$ $\Delta f = 75 \text{ kHz}; f_{\text{mod}} = 1 \text{ kHz}$ without de-emphasis; $L_{\text{demod}} = \text{typical value};$ $V_{37-35} = 300 \mu\text{V to } 800 \text{ mV}$	–	0.1	0.35	%

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SYMBOL	PARAMETER	CONDITIONS	MIN.	TYP.	MAX.	UNIT
<b>Unweighted voltage level</b>						
$V_{50-4}$	unweighted voltage level	see Fig.7; $V_{38} = 2.52 \text{ V}$ ; $V_{37-35} \leq 2.5 \mu\text{V}$ $V_{37-35} = 1.0 \text{ mV}$	1.6 2.7	2.2 3.4	3.0 4.7	V V
$\Delta V_{50-4}/\Delta V_{37-35}$	slope of unweighted voltage level	$100 \mu\text{V} \leq V_{37-35(\text{rms})} < 300 \text{ mV}$ ; temperature compensation off temperature compensation on	0.75 0.6	0.9 0.75	1.05 0.9	V/20 dB V/20 dB
$\Delta V_{50-4}/(V_{50-4}\Delta T)$	temperature dependence	$V_{37-35} = 1 \text{ mV}$ ; temperature compensation off temperature compensation on	– –	5.0 2.0	– –	mV/VK mV/VK
$I_{\text{source(max)}}$	maximum output source current pin 50		0.3	–	–	mA
$I_{\text{sink(max)}}$	maximum output sink current pin 50		–50	–	–	$\mu\text{A}$
$R_{o50}$	output resistance pin 50		–	–	300	$\Omega$
ADJUSTMENT OF UNWEIGHTED VOLTAGE LEVEL AND $V_{\text{mutaml}}$ ; note 1						
$\Delta V_{50}$	adjusting range voltage pin 50	$V_{37-35(\text{rms})} = 10 \text{ mV}$	–1.8	–	+1.8	V
$V_{38-4}$	internal bias voltage		–	2.6	–	V
$\Delta V_{50-4}/\Delta V_{38-4}$	adjusting gain		–	–0.9	–	
$R_{i38}$	input resistance pin 38		–	80	–	k $\Omega$
<b>Multi-path detector</b>						
$f_{\text{MP}}$	multi-path detector band-pass centre frequency		–	20	–	kHz
$B_{\text{MP}}$	band-pass bandwidth		7	–	–	kHz
<b>Reference voltage</b>						
$V_{o(21-4)}$	output voltage	$I_{21} = -1 \text{ mA}$	4.5	5.1	5.7	V
$\Delta V_{21-4}/(V_{21-4}\Delta T)$	temperature dependence		–	3.3	–	mV/VK
$I_{o21}$	output current pin 21		–	–	1	mA
<b>AM-IF path; see Fig.9 and notes 2 and 3</b>						
$V_{22-23\text{max(p-p)}}$	maximum output voltage between pins 22 and 23 (peak-to-peak value)		12	15	–	V

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SYMBOL	PARAMETER	CONDITIONS	MIN.	TYP.	MAX.	UNIT
$I_{\text{bias}(22-23)}$	mixer bias current pins 22 and 23		5	6	7	mA
$I_{\text{LI}(22-23)}$	mixer leakage current pins 22 and 23	in FM position	–	–	2	$\mu\text{A}$
$I_{22\text{IF}_2}/V_{27-24\text{IF}_1}$	conversion gain		2.2	2.7	3.4	mS
$R_{i(27-24)}$	input resistance between pins 27 and 24		300	330	360	$\Omega$
$C_{i(27-24)}$	input capacitance between pins 27 and 24		–	5	8	pF
$R_{o(22-23)}$	output resistance between pins 22 and 23		10	20	–	k $\Omega$
$C_{o(22-23)}$	output capacitance between pins 22 and 23		–	5	10	pF
IP3	third order intermodulation		–	137	–	dB $\mu\text{V}$
AM DETECTOR; see notes 4 and 5						
$V_{48-4(\text{rms})}$	AF output level (RMS value)	$R_{L48} > 500 \text{ k}\Omega$ ; $100 \mu\text{V} \leq V_{33-34} \leq 300 \text{ mV}$	190	240	290	mV
$V_{33-34(\text{rms})}$	sensitivity voltage (RMS value)	$\frac{S+N}{N} = 26 \text{ dB}$	–	50	75	$\mu\text{V}$
		$\frac{S+N}{N} = 46 \text{ dB}$	–	200	400	$\mu\text{V}$
	AM-IF <sub>2</sub> minimum input voltage (RMS value)	THD $\leq$ 5%; m = 0.8	–	–	100	$\mu\text{V}$
	AM-IF <sub>2</sub> maximum input voltage (RMS value)	THD $\leq$ 5%; m = 0.8	800	–	–	mV
$R_{i(33-34)}$	IF <sub>2</sub> input resistance between pins 33 and 34		1.8	2.0	2.2	k $\Omega$
$C_{i(33-34)}$	IF <sub>2</sub> input capacitance between pins 33 and 34		–	10	15	pF
$R_{o48}$	output resistance pin 48		27	33	39	k $\Omega$
$C_{o48}$	output capacitance pin 48		–	–	10	pF
$\frac{S+N}{N}$	signal-plus-noise-to-noise ratio		54	60	–	dB
THD	total harmonic distortion	m = 0.8; $300 \mu\text{V} \leq V_{33-34} \leq 200 \text{ mV}$	–	1.5	3.0	%

**Notes to the Characteristics**

- For typical adjusting range see Fig.8.
- $f_{\text{IF}_1} = 10.7 \text{ MHz}$ ;  $f_{\text{IF}_2} = 450 \text{ kHz}$  for AM mixer.
- The AM oscillator signal is generated by division of the 61.5 MHz crystal oscillator. Two divider ratios programmable by the I<sup>2</sup>C-bus: division by 6 (AM-IF<sub>1</sub> = 10.7 MHz); division by 2 (AM-IF<sub>1</sub> = 30 MHz).
- For typical AM level curve see Fig.10.
- For AM detector;  $f_{\text{AMIF}_2} = 450 \text{ kHz}$ ;  $f_{\text{mod}} = 400 \text{ Hz}$ ; m = 30%.

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## ADCs CHARACTERISTICS FOR LEVEL AND MULTI-PATH VOLTAGES

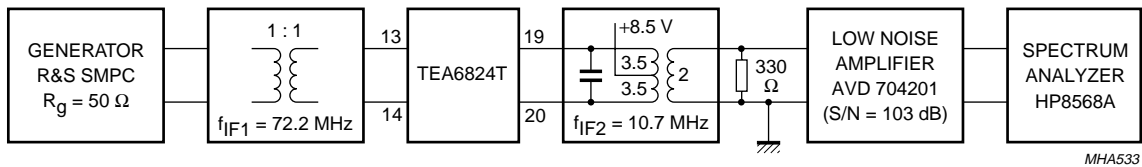
SYMBOL	PARAMETER	CONDITIONS	MIN.	TYP.	MAX.	UNIT
<b>ADC for FM level information; note 1</b>						
$\Delta V_{37-35}$	analog-to-digital conversion step size for $V_{37-35}$		2	4	8	dB/step
$\Delta V_{37-35}$	analog-to-digital conversion level range for $V_{37-35}$		43	56	69	dB
FM STOP						
$\Delta V_{\text{stop}}$	variation of stop level as function of $V_{42-4}$		–	16	–	dB/V
<b>ADC for AM level information; note 2</b>						
$\Delta V_{33-34}$	analog-to-digital conversion step size for $V_{33-34}$		2	4	8	dB/step
$\Delta V_{33-34}$	analog-to-digital conversion level range for $V_{33-34}$		43	56	69	dB
AM STOP						
$\Delta V_{\text{stop}}$	variation of stop level as a function of $V_{43-4}$		–	16	–	dB/V
<b>ADC for multi-path information; note 3</b>						
m	multi-path conversion	step 0	–	–	5	%
		step 1	–	15	–	%
		step 2	–	22	–	%
		step 3	–	28	–	%
		step 4	–	34	–	%
		step 5	–	40	–	%
		step 6	–	46	–	%
		step 7	–	52	–	%

## Notes

1. The FM level information  $V_{50-3}$  is analog-to-digital converted with 4 bits.
2. The AM level information  $V_{49-4}$  is analog-to-digital converted with 4 bits.
3. The multi-path information  $V_{40-4}$  is analog-to-digital converted with 3 bits covering an  $IF_2$  amplitude modulation range  $m \leq 0.6$ ;  $f_{\text{mod}} = 20$  kHz.

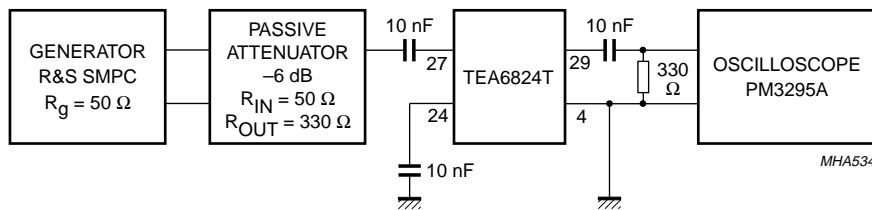
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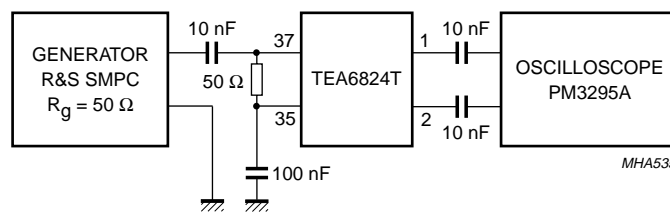
MHA533

Fig.3 Test circuit FM mixer.



MHA534

Fig.4 Test circuit IF-amplifier.

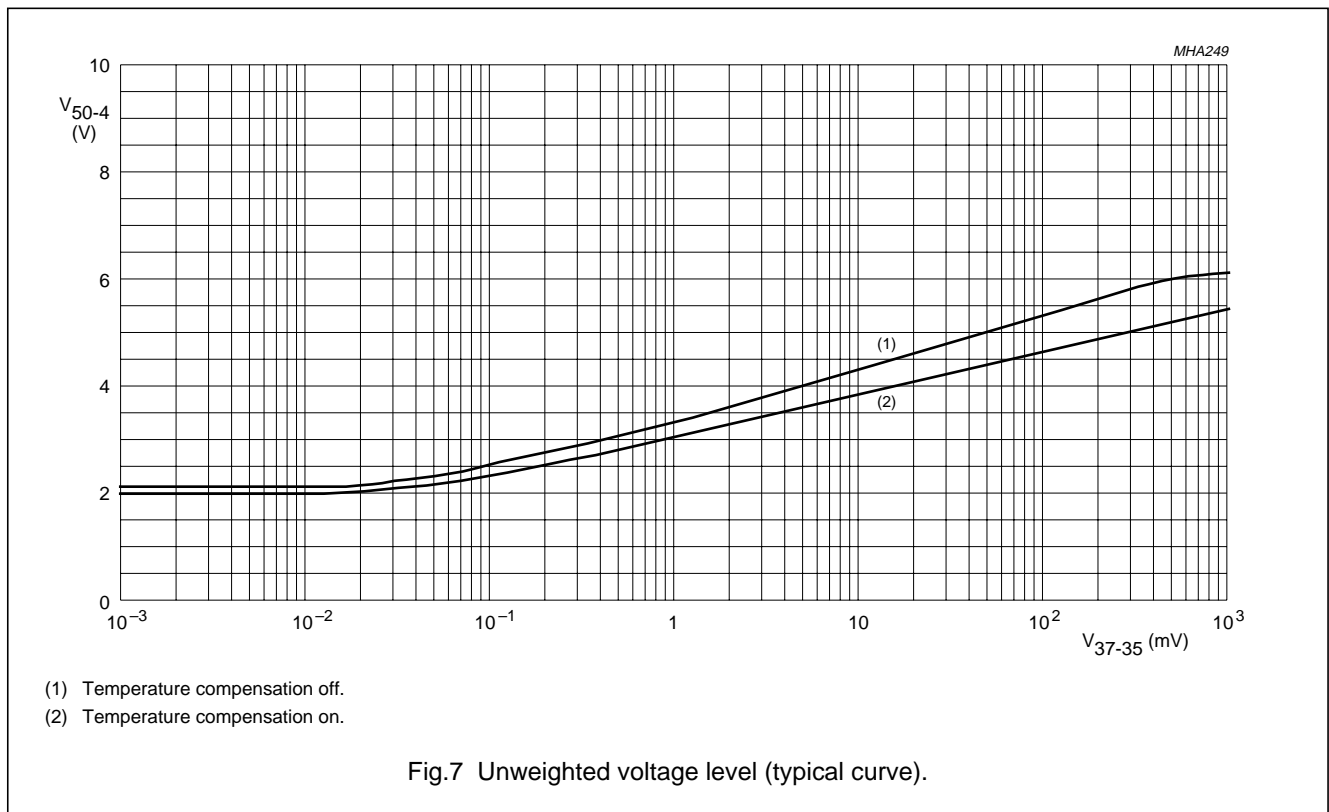
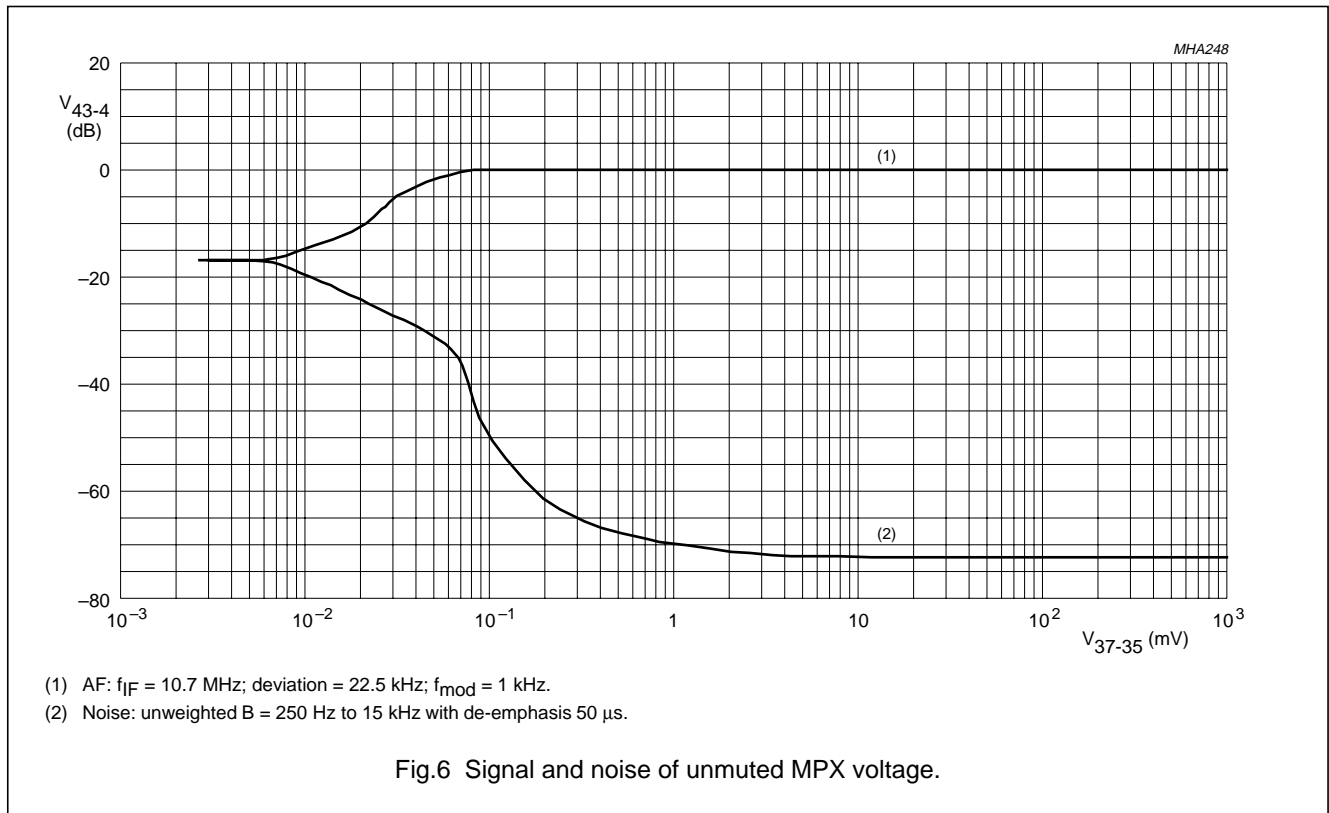


MHA535

Fig.5 Test circuit limiter gain.

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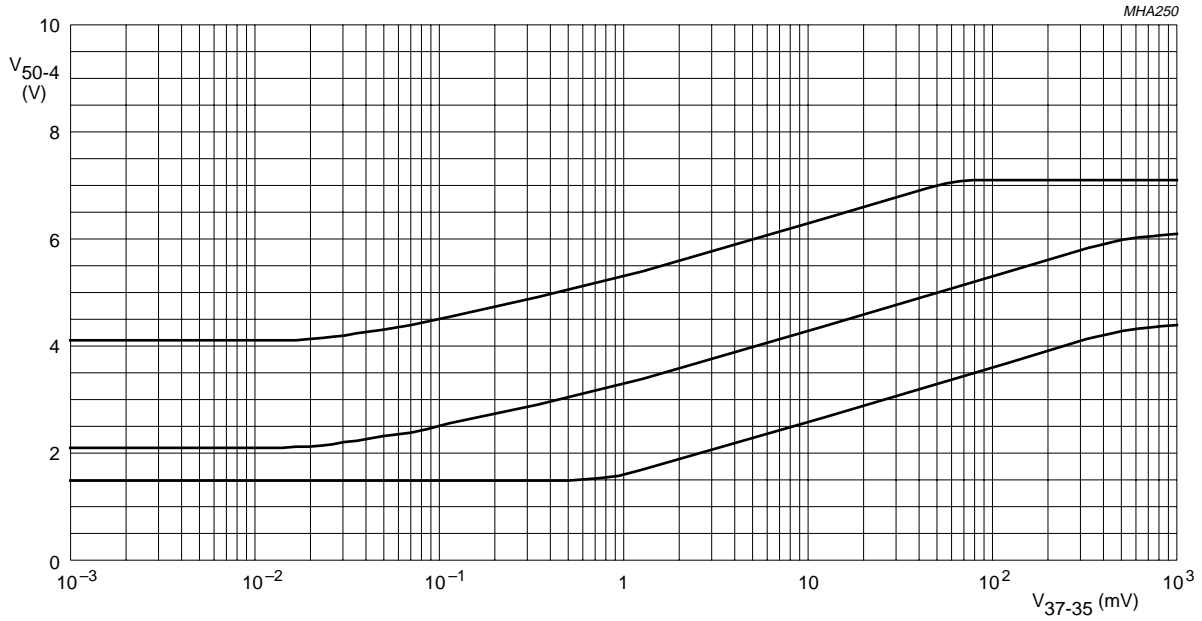


Fig.8 Adjustment range of unweighted voltage level.

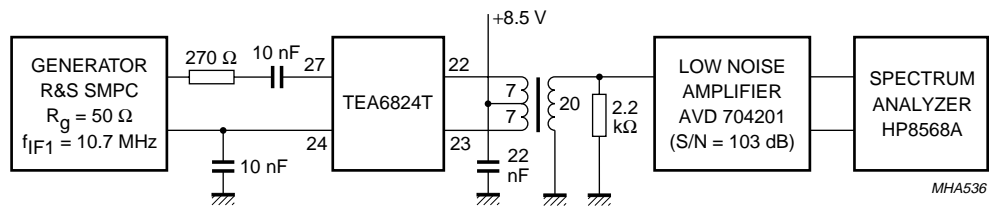
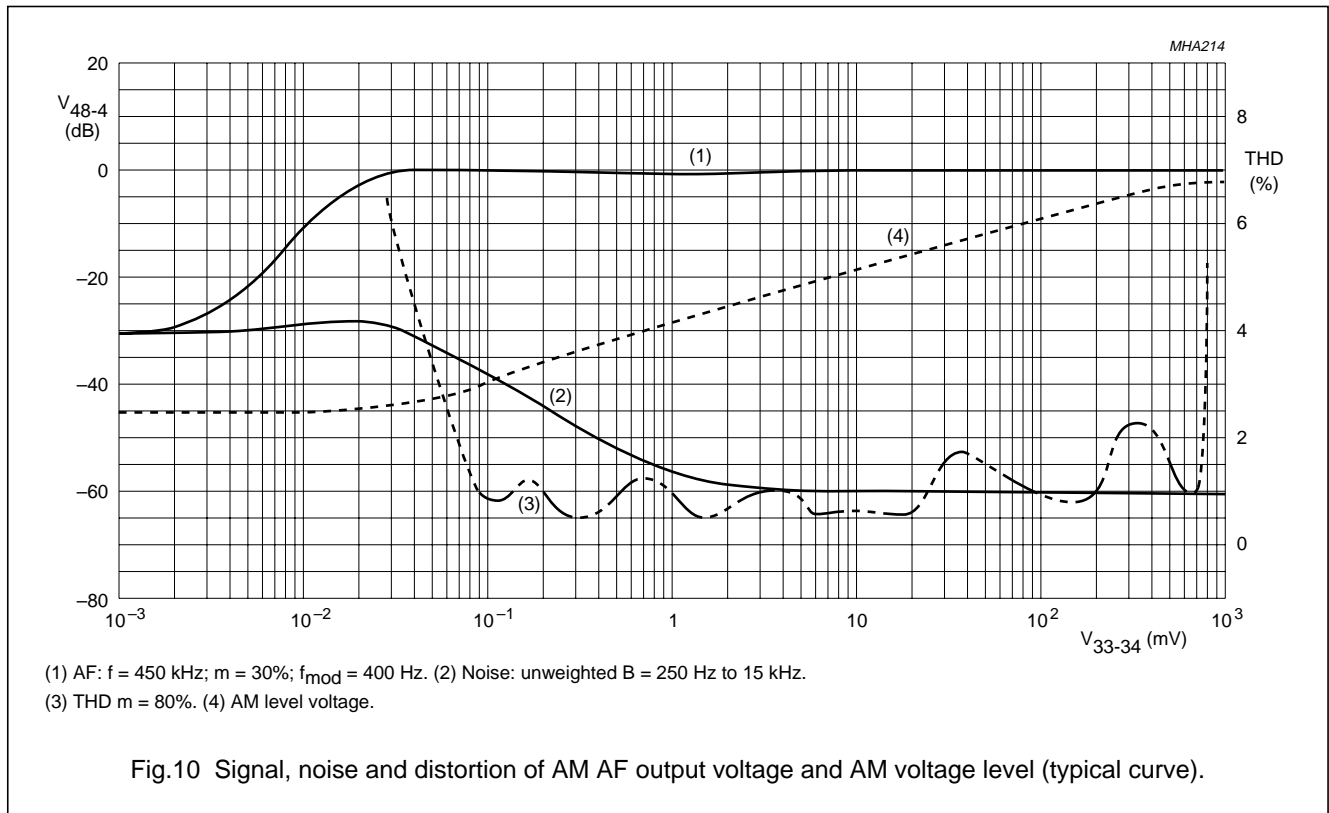


Fig.9 Test circuit AM mixer.

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**I<sup>2</sup>C-BUS AND I<sup>2</sup>C-BUS CONTROLLED FUNCTIONS**

**I<sup>2</sup>C- bus specification**

The standard I<sup>2</sup>C-bus specification is expanded by the following definitions.

Structure of the I<sup>2</sup>C-bus logic: slave transceiver with auto increment and expansion to switch a direct transfer of all transmissions to an output for the radio front-end IC TEA6810T.

Sub-addresses are not used.

**DATA TRANSFER FOR THE TEA6824T**

Data sequence:

- Address
- Byte 1
- Byte 2.

The data transfer may only be in this order. The transfer direction of the data bytes is defined by the LSB of the address.

The data becomes valid at the output of the internal latches with the acknowledge of each byte. A STOP condition after any byte can shorten transmission times.

When writing to the transceiver by using the STOP condition before completion of the whole transfer:

- The remaining bytes will contain the old information
- If the transfer of a byte was not completed, this byte is lost and the previous information is available.

**DATA TRANSFER TO AN OUTPUT OF THE FRONT-END IC**

A data bit in the transceiver of the TEA6824T enables or disables a direct transfer of all transmissions to an interface stage for the front-end IC.

For a transmission to the front-end IC the address and the data format of the front-end IC has to be used.

**Remark:** The pull-up resistors for the front-end interface (pins 6 and 7) should not be connected to the +5 V supply voltage of the front-end IC, otherwise a bus pull-down (pin 53) can occur during switching off the front-end supply when the interface stage is enabled.

**DATA TRANSFER TO THE IF IC**

Data transfer to the IF IC (TEA6824T) is independent of the state of interface stage for the front-end IC.



## In Car Entertainment (ICE) car radio

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**Table 1** Structure of the I<sup>2</sup>C-bus

DESCRIPTION	SPECIFICATION
Bus address of the TEA6824T	1100001X
Subaddress	not used
Hardware (pin) programmable address bits	not available
Default settings by power-on reset	data byte 1 bits 4 to 7 are set to logic 0; all other bits are random

**Table 2** Data to be received by the IC for data byte 1

BIT	DESCRIPTION	RESULT
0	not used	–
1	LSB reference frequency for synthesizer	–
2	reference frequency for synthesizer	–
3	MSB reference frequency for synthesizer	–
4	not used	–
5	not used	–
6	not used	–
7	I <sup>2</sup> C-bus to front-end ENABLED	bit 7 = 1
	I <sup>2</sup> C-bus to front-end DISABLED	bit 7 = 0

**Table 3** Reference frequency setting in byte 1; see Table 1

BIT 3	BIT 2	BIT 1	FREQUENCY SETTING
0	0	0	3 kHz
0	0	1	5 kHz
0	1	0	10 kHz
0	1	1	15 kHz
1	0	0	25 kHz
1	0	1	50 kHz
1	1	0	not defined
1	1	1	not defined

**Table 4** Data to be received by the IC for data byte 2

BIT	DESCRIPTION	RESULT	
0	AM/FM; AM mode	bit 0 = 0	
	AM/FM; FM mode	bit 0 = 1	
1	divider for AM mixer; divide-by-2	bit 1 = 0	
	divider for AM mixer; divide-by-6	bit 1 = 1	
2	measure time IF-count; 40 ms	bit 2 = 0	
	measure time IF-count; 4 ms	bit 2 = 1	
3	not used		
4	IF-prescaler division rate; divide-by-200	bit 4 = 0	
	IF-prescaler division rate; divide-by-25	bit 4 = 1	
5	not used		
6	temperature compensation of unweighted voltage level; temperature coefficient as specified in Chapter "Characteristics"	temperature compensation off	bit 6 = 0
		temperature compensation on	bit 6 = 1
7	not used		

**Table 5** Data to be transmitted by the IC for data byte 1; note 1

BIT	DESCRIPTION
0	bit 1 level information
1	bit 2 level information
2	MSB (bit 3) level information
3	LSB multi-path information
4	multi-path information
5	MSB multi-path information
6	not used
7	LSB (bit 0) level information

**Note**

1. The analog-to-digital conversion for multi-path and level will be performed during a transmission of any address to the I<sup>2</sup>C-bus.

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**Table 6** Data to be transmitted by the IC for data byte 2

BIT	DESCRIPTION
0	LSB of the IF-counter
1	IF-counter
2	IF-counter
3	IF-counter
4	IF-counter
5	IF-counter
6	IF-counter
7	MSB of the IF-counter

## REFERENCED FREQUENCY GENERATION

**Table 7** Division ratios

DIVISION RATIO	REFERENCE FREQUENCY (kHz) <sup>(1)</sup>
20500	3
12300	5
6150	10
4100	15
2460	25
1230	50

**Note**

1. All specified frequencies are valid for a crystal oscillator frequency of 61.5 MHz.

**Table 8** Output signal of reference frequency divider

SYMBOL	PARAMETER	MIN.	TYP.	MAX.	UNIT
$V_{10-11(p-p)}$	differential output voltage between pins 10 and 11 (peak-to-peak value)	0.3	0.4	0.5	V
$V_{10-4(p-p)}$ ; $V_{11-4(p-p)}$	single-ended output voltage between pins 10 and 4 and between pins 11 and 4 (peak-to-peak value)	0.15	0.2	0.3	V
$C_{10-4}$ ; $C_{11-4}$	output capacitance between pins 10 and 4 and between pins 11 and 4	–	–	4	pF
$R_{10-52}$ ; $R_{11-52}$	output resistance between pins 10 and 52 and between pins 11 and 52	800	1000	1200	$\Omega$

## IF-COUNTER

**Table 9** IF-counter sensitivity

SYMBOL	PARAMETER	CONDITIONS	MIN.	UNIT
$V_{33-34}$	IF-counter sensitivity	AM mode; $m = 0$	200	$\mu V$
$V_{37-35}$	IF-counter sensitivity	FM mode	200	$\mu V$

AM and FM counting windows are 4 or 40 ms. AM counting resolution is 250 or 25 Hz. FM counting resolution is 5 kHz, 625 Hz, 50 kHz or 6.25 kHz. AM-IF-prescaler is divisible by 1. FM-IF-prescaler is divisible by 25 or 200.

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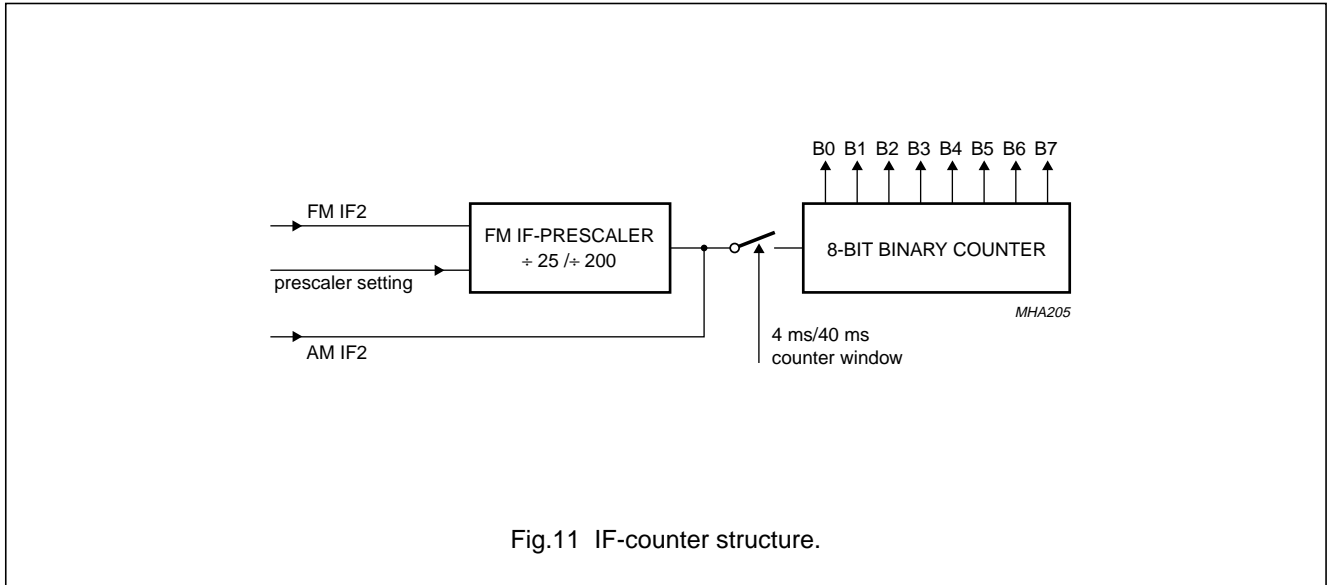


Fig.11 IF-counter structure.

IF-counter read out

IF-counter read out as a function of FM or AM position, counter window and prescaler setting

FM/AM	PRESCALER RATIO	WINDOW (ms)	f <sub>IF</sub> (kHz)	READ OUT	RESOLUTION (Hz/count)	RANGE (kHz)	
						MIN.	MAX.
FM	200	4	10700	D6H	50000	4300	17050
FM	200	40	10700	5CH	5000	10065	11335
FM	25	4	10700	B0H	6250	9906.25	11493.75
FM <sup>(1)</sup>	25	40	10700	E0H	625	10620.63	10779.38
AM	1	4	450	08H	250	418.25	481.75
AM <sup>(1)</sup>	1	40	450	50H	25	446.83	453.18

Note

- In position FM with a prescaler ratio of 25, counter window of 40 ms and in position AM with 40 ms counter window ambiguous counting results within the IF filter bandwidth are obtained. The counting range is 127 counts above and 127 counts below the nominal IF of 10.7 MHz for FM and 450 kHz for AM.

The IF-count windows are valid for a crystal oscillator frequency of 61.5 MHz.

The FM/AM switching is carried out by bit 0 of byte 2 of the received data of the IC.

The IF-counter operates continuously.

The IF-counter and window counter will be reset when the I<sup>2</sup>C-bus logic detects the address of the IC. This disables changes in the latches for the IF-counter, while reading this value. If the transmission to the front-end IC is disabled after the synthesizer loop of the TEA6810T front-end IC has locked to a new frequency, the IF-counter will be available after the set measuring time.

The IF-counter starts at 0. The IF-counter output are the 8 least significant bits of the counting result.

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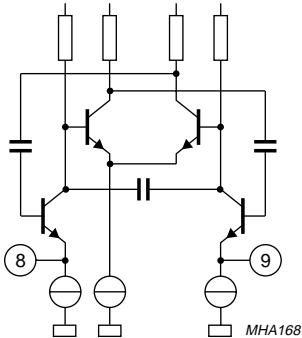
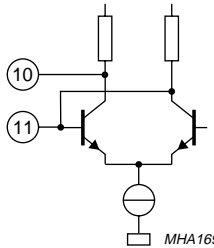
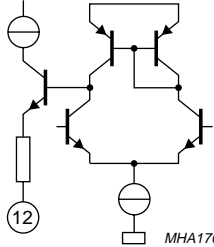
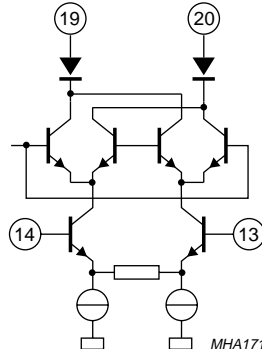
INTERNAL CIRCUITRY

Table 10 Equivalent pin circuits and pin voltages

PIN	SYMBOL	DC VOLTAGE (V)		EQUIVALENT CIRCUIT
		AM	FM	
1	QDET1	4.0	4.0	
2	QDET2	4.0	4.0	
3	REFIN1	0	0	
4	AGND	-	-	
5	V <sub>DDA1</sub>	5.0	5.0	
6	HFBUS1	5.0	5.0	
7	HFBUS2	5.0	5.0	

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PIN	SYMBOL	DC VOLTAGE (V)		EQUIVALENT CIRCUIT
		AM	FM	
8	XTAL1	4.1	4.1	
9	XTAL2	4.1	4.1	
10	$f_{ref(p)}$	4.9	4.9	
11	$f_{ref(n)}$	4.9	4.9	
12	$I_{ref}$	4.3	4.3	
13	FMIF1IN1	2.3	2.3	
14	FMIF1IN2	2.3	2.3	

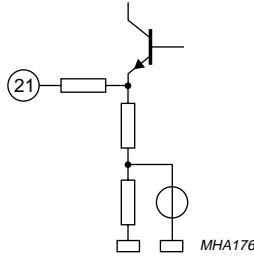
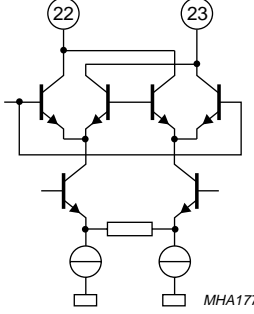
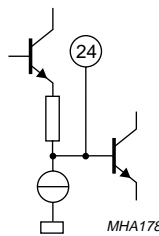
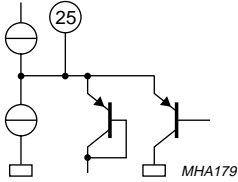
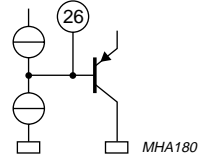
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PIN	SYMBOL	DC VOLTAGE (V)		EQUIVALENT CIRCUIT
		AM	FM	
15	REFIN2	4.9	4.9	
16	REFIN3	4.9	4.9	
17	REFIN4	4.9	4.9	
18	REFIN5	4.9	4.9	
19	FMIF2OUT1	8.5	8.5	
20	FMIF2OUT2	8.5	8.5	

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PIN	SYMBOL	DC VOLTAGE (V)		EQUIVALENT CIRCUIT
		AM	FM	
21	V <sub>ref</sub>	5.1	5.1	
22	AMIF2OUT1	8.5	8.5	
23	AMIF2OUT2	8.5	8.5	
24	FMAMDEC	3.0	2.5	
25	DEC1	2.8 to 7.0	2.8 to 7.2	
26	DEC2	0.4	0.4 to 7.0	

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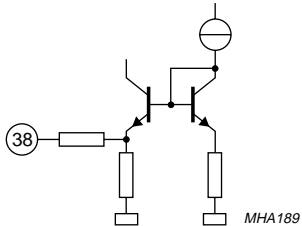
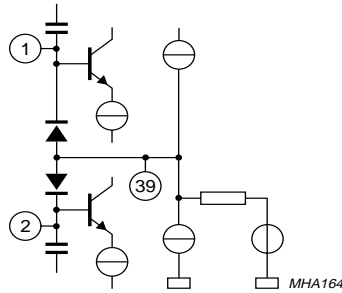
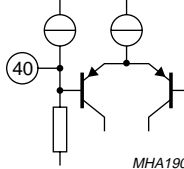
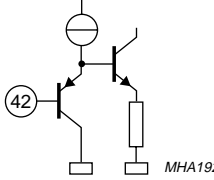
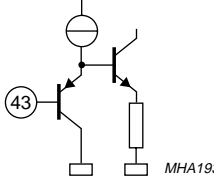
TEA6824T

PIN	SYMBOL	DC VOLTAGE (V)		EQUIVALENT CIRCUIT
		AM	FM	
27	FMAM10.7	3.0	2.5	
28	V <sub>DDA2</sub>	8.5	8.5	
29	FMIFAMPOUT	6.0	6.0	
30	n.c.			
31	n.c.			
32	n.c.			
33	AMIF2IN1	2.7	0.7	
34	AMIF2IN2	2.7	0.7	
35	FMIN2	0.7	2.7	
36	DCFEED	2.7	2.7	
37	FMIN1	0.7	2.7	



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PIN	SYMBOL	DC VOLTAGE (V)		EQUIVALENT CIRCUIT
		AM	FM	
38	LEVELADJ	2.6	2.6	
39	C <sub>AFC</sub>	1.0 to 2.2	1.0 to 7.0	
40	MPBUF	0.7 to 6.0	0.7 to 6.0	
41	n.c.			
42	FMSTOP	0 to 5.2	0 to 5.2	
43	MPXOUT	0 to 5.2	3.0	
44	n.c.			
45	n.c.			
46	n.c.			
47	n.c.			

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PIN	SYMBOL	DC VOLTAGE (V)		EQUIVALENT CIRCUIT
		AM	FM	
48	AMAFOUT	3.7	4.8	
49	DEC3	1.0 to 5.5	1.0 to 5.5	
50	LEVELUNWEIG	1.0 to 7.0	1.0 to 7.0	
51	n.c.			
52	V <sub>DDD</sub>	5.0	5.0	
53	SDA	5.0	5.0	
54	SCL	5.0	5.0	
55	DGND	0	0	
56	V <sub>DDA3</sub>	8.5	8.5	

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APPLICATION INFORMATION

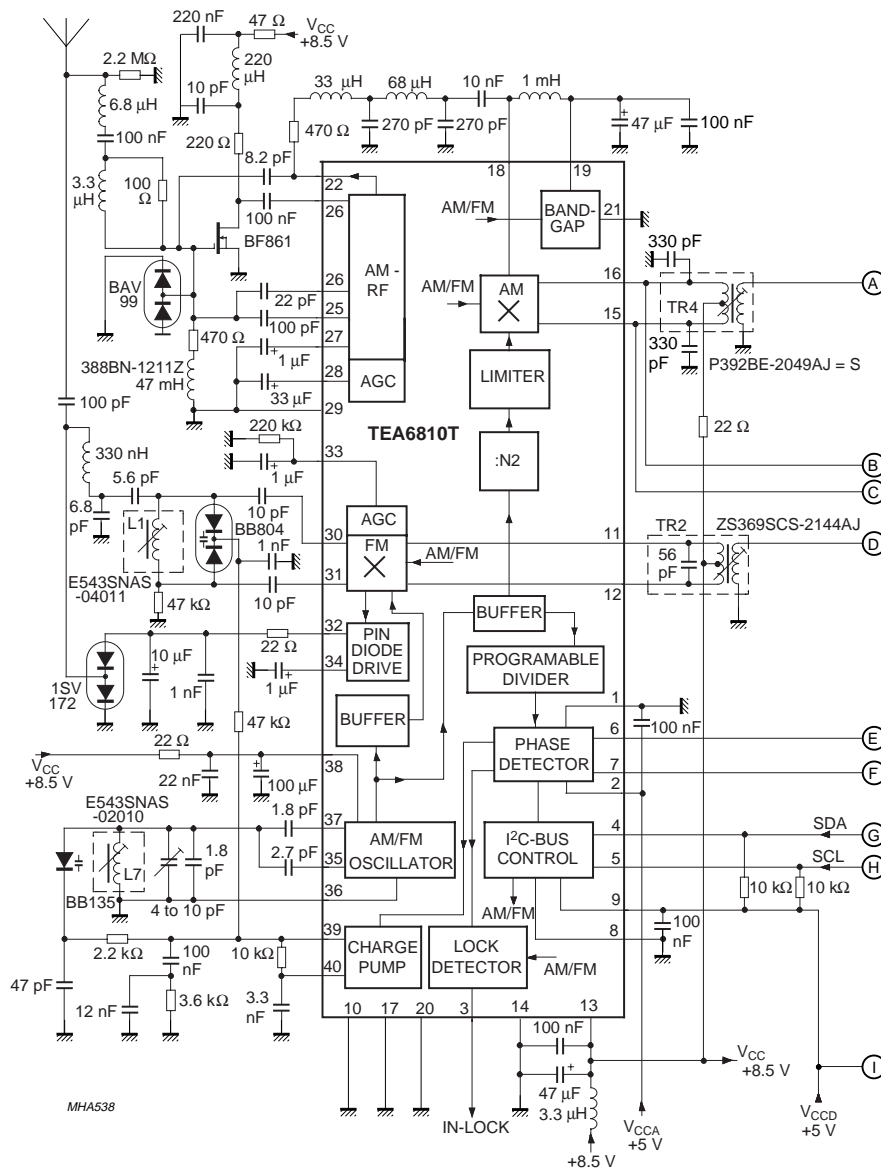


Fig.12 Application diagram (continued in Fig.13).

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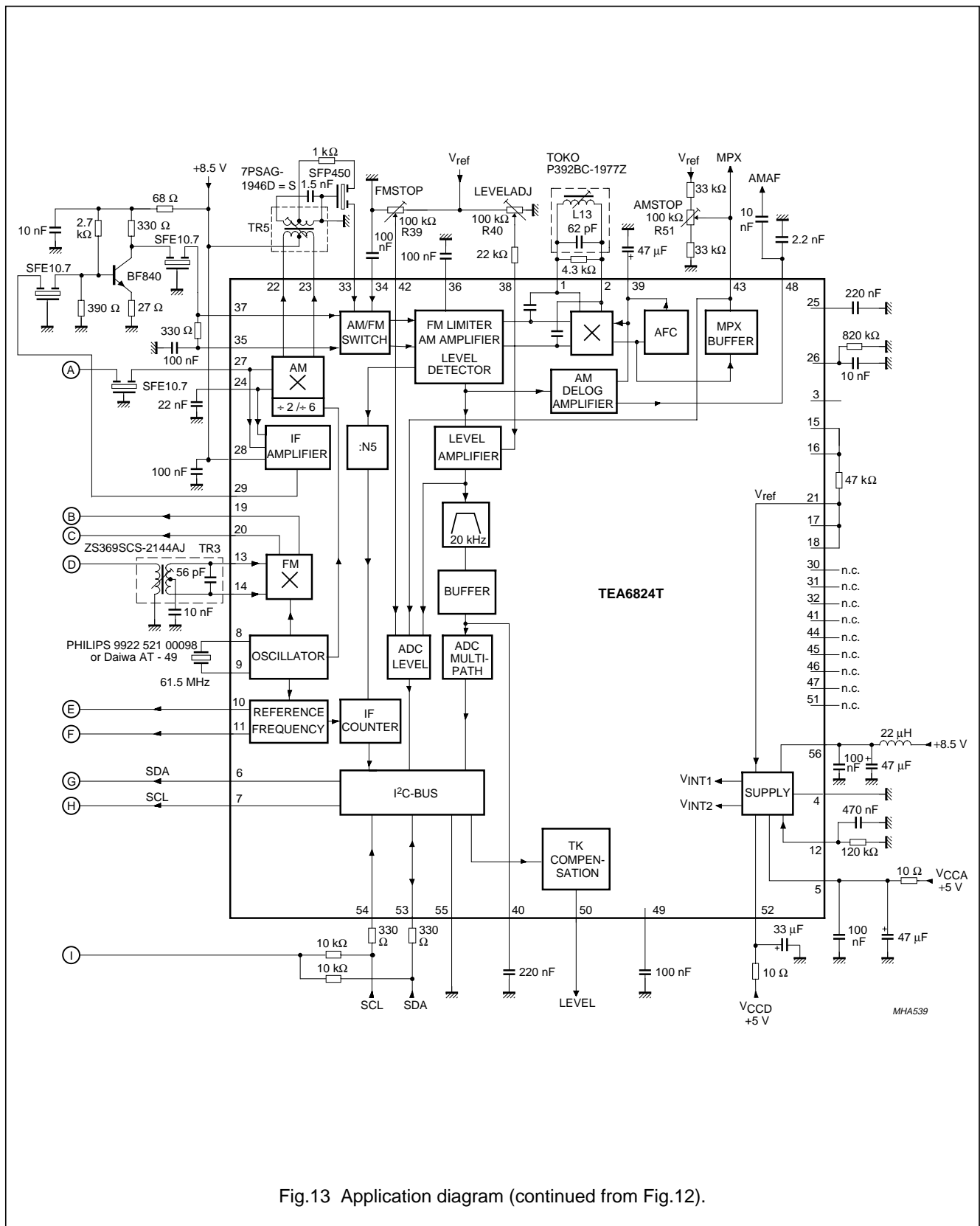


Fig.13 Application diagram (continued from Fig.12).



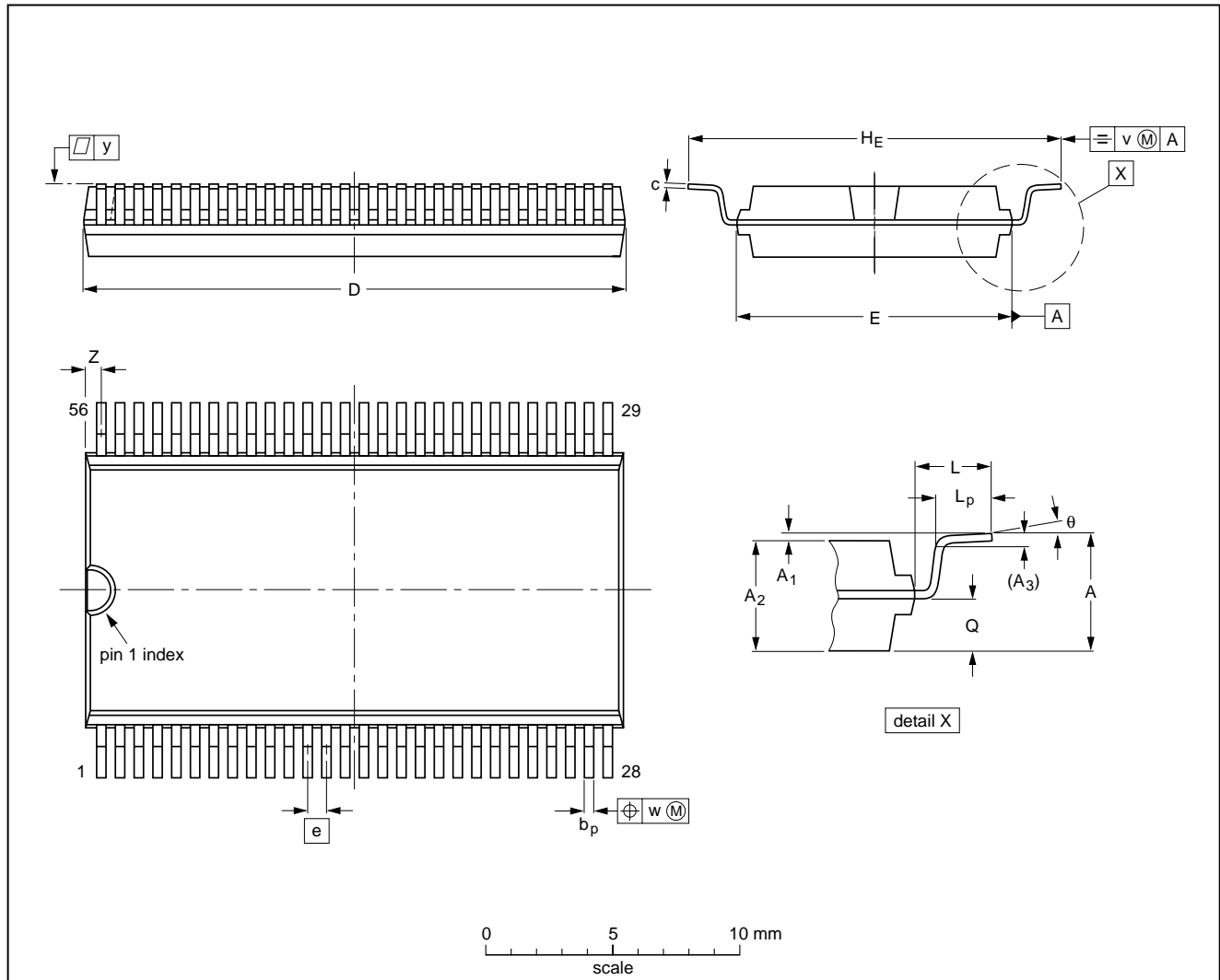
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PACKAGE OUTLINE

VSO56: plastic very small outline package; 56 leads; face down

SOT190-2



DIMENSIONS (inch dimensions are derived from the original mm dimensions)

UNIT	A max.	A <sub>1</sub>	A <sub>2</sub>	A <sub>3</sub>	b <sub>p</sub>	c	D <sup>(1)</sup>	E <sup>(2)</sup>	e	H <sub>E</sub>	L	L <sub>p</sub>	Q	v	w	y	Z <sup>(1)</sup>	θ
mm	3.3	0.3 0.1	3.0 2.8	0.25	0.42 0.30	0.22 0.14	21.65 21.35	11.1 11.0	0.75	15.8 15.2	2.25	1.6 1.4	1.45 1.30	0.2	0.1	0.1	0.90 0.55	7° 0°
inches	0.13	0.012 0.004	0.12 0.11	0.01	0.017 0.012	0.0087 0.0055	0.85 0.84	0.44 0.43	0.0295	0.62 0.60	0.089	0.063 0.055	0.057 0.051	0.008	0.004	0.004	0.035 0.022	

Note

1. Plastic or metal protrusions of 0.3 mm maximum per side are not included.
2. Plastic interlead protrusions of 0.25 mm maximum per side are not included.

OUTLINE VERSION	REFERENCES			EUROPEAN PROJECTION	ISSUE DATE
	IEC	JEDEC	EIAJ		
SOT190-2					96-04-02 97-08-11

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## In Car Entertainment (ICE) car radio

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### SOLDERING

#### Introduction

There is no soldering method that is ideal for all IC packages. Wave soldering is often preferred when through-hole and surface mounted components are mixed on one printed-circuit board. However, wave soldering is not always suitable for surface mounted ICs, or for printed-circuits with high population densities. In these situations reflow soldering is often used.

This text gives a very brief insight to a complex technology. A more in-depth account of soldering ICs can be found in our *"IC Package Databook"* (order code 9398 652 90011).

#### Reflow soldering

Reflow soldering techniques are suitable for all VSO packages.

Reflow soldering requires solder paste (a suspension of fine solder particles, flux and binding agent) to be applied to the printed-circuit board by screen printing, stencilling or pressure-syringe dispensing before package placement.

Several techniques exist for reflowing; for example, thermal conduction by heated belt. Dwell times vary between 50 and 300 seconds depending on heating method. Typical reflow temperatures range from 215 to 250 °C.

Preheating is necessary to dry the paste and evaporate the binding agent. Preheating duration: 45 minutes at 45 °C.

#### Wave soldering

Wave soldering techniques can be used for all VSO packages if the following conditions are observed:

- A double-wave (a turbulent wave with high upward pressure followed by a smooth laminar wave) soldering technique should be used.
- The longitudinal axis of the package footprint must be parallel to the solder flow.
- The package footprint must incorporate solder thieves at the downstream end.

During placement and before soldering, the package must be fixed with a droplet of adhesive. The adhesive can be applied by screen printing, pin transfer or syringe dispensing. The package can be soldered after the adhesive is cured.

Maximum permissible solder temperature is 260 °C, and maximum duration of package immersion in solder is 10 seconds, if cooled to less than 150 °C within 6 seconds. Typical dwell time is 4 seconds at 250 °C.

A mildly-activated flux will eliminate the need for removal of corrosive residues in most applications.

#### Repairing soldered joints

Fix the component by first soldering two diagonally-opposite end leads. Use only a low voltage soldering iron (less than 24 V) applied to the flat part of the lead. Contact time must be limited to 10 seconds at up to 300 °C. When using a dedicated tool, all other leads can be soldered in one operation within 2 to 5 seconds between 270 and 320 °C.

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## DEFINITIONS

<b>Data sheet status</b>	
Objective specification	This data sheet contains target or goal specifications for product development.
Preliminary specification	This data sheet contains preliminary data; supplementary data may be published later.
Product specification	This data sheet contains final product specifications.
<b>Limiting values</b>	
Limiting values given are in accordance with the Absolute Maximum Rating System (IEC 134). Stress above one or more of the limiting values may cause permanent damage to the device. These are stress ratings only and operation of the device at these or at any other conditions above those given in the Characteristics sections of the specification is not implied. Exposure to limiting values for extended periods may affect device reliability.	
<b>Application information</b>	
Where application information is given, it is advisory and does not form part of the specification.	

## LIFE SUPPORT APPLICATIONS

These products are not designed for use in life support appliances, devices, or systems where malfunction of these products can reasonably be expected to result in personal injury. Philips customers using or selling these products for use in such applications do so at their own risk and agree to fully indemnify Philips for any damages resulting from such improper use or sale

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Purchase of Philips I<sup>2</sup>C components conveys a license under the Philips' I<sup>2</sup>C patent to use the components in the I<sup>2</sup>C system provided the system conforms to the I<sup>2</sup>C specification defined by Philips. This specification can be ordered using the code 9398 393 40011.



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**NOTES**

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